

Freq LT $\rightarrow H(s)$

p/z: if some p has ≥ 0 real part, unstable

\exists : no stability eff, changes I/O effect

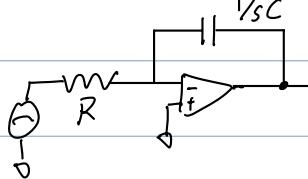
$H(s) \rightarrow$ impulse/step resp., natural/forced. z_i/z_s

$\hookrightarrow H(j\omega) \rightarrow$ bode plt

$$x(t) = A \sin(\omega t)$$

$$y(t) = |H(j\omega)| \sin(\omega t + \angle H(j\omega))$$

E. J

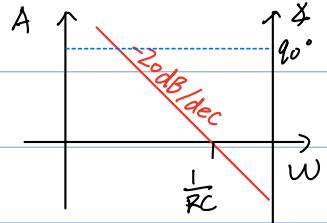


$$V_{out}(s) = 0 - \frac{1}{sC} \cdot V_{in} \cdot \frac{1}{R}$$

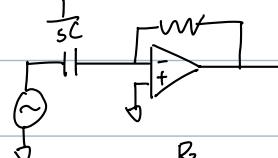
$$H(s) = -\frac{1}{sRC}$$

$$H(j\omega) = -\frac{j}{\omega RC}$$

$$p: s=0$$



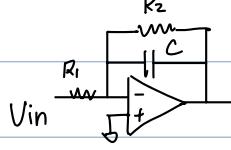
E. $\frac{d}{dx}$



$$H(s) = -sRC$$

$$+20 \text{ dB/dec}, -90^\circ$$

E. LPF

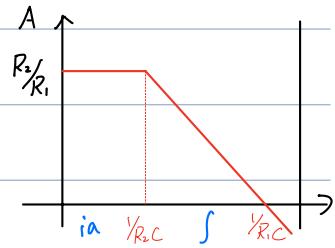


$$H(s) = -\frac{R_2}{R_1} \frac{1}{1+sR_2C}$$

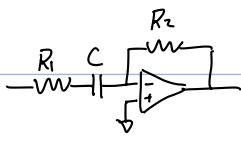
$$p: s = -\frac{1}{R_2C}$$

ia \rightarrow ideal J

$$-180^\circ \rightarrow -270^\circ (+90^\circ)$$



E. HPF



$$H(s) = -\frac{R_2}{R_1} \frac{sR_1C}{1+sR_1C}$$

ideal $\frac{d}{dx} \rightarrow$ ia

$$-90^\circ \rightarrow -180^\circ$$

1p/2

E. BPF (V)

$$H(s) = -\frac{R_2}{R_1} \frac{sR_1C_1}{1+sR_1C_1} \frac{1}{1+sR_2C_2}$$

2p/2

$$-90^\circ \rightarrow -180^\circ \rightarrow -270^\circ$$

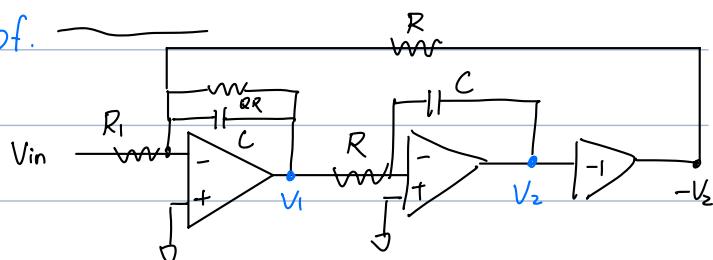
E. HW2 All comp. give same pole, but diff. zero pf.

$$\frac{V_{in}}{R} + \frac{V_1}{QR} + sCV_1 - \frac{V_2}{R} = 0$$

$$sCV_2 + \frac{V_1}{R} = 0$$

$$\begin{bmatrix} \frac{1}{QR} + sC & -\frac{1}{R} \\ \frac{1}{R} & sC \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} -\frac{1}{R_1} \\ 0 \end{bmatrix} V_{in}(s)$$

$$V_1(s) = \frac{-\frac{1}{R_1}V_{in}}{\left| \begin{bmatrix} \frac{1}{QR} + sC & -\frac{1}{R} \\ \frac{1}{R} & sC \end{bmatrix} \right|} = \frac{-\frac{1}{R_1}sCV_{in}}{sC(\frac{1}{QR} + sC) + \frac{1}{R^2}} = \frac{-\frac{R}{R_1}sRCV_{in}}{1 + s\frac{RC}{Q} + s^2R^2C^2}$$



$$V_2(s) = \underline{\underline{\underline{\quad}}}$$

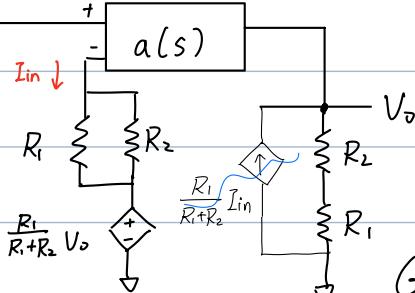
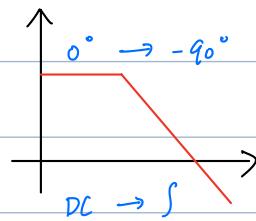
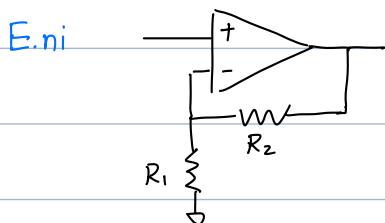
2p/2 BPF

2p 1p LPF

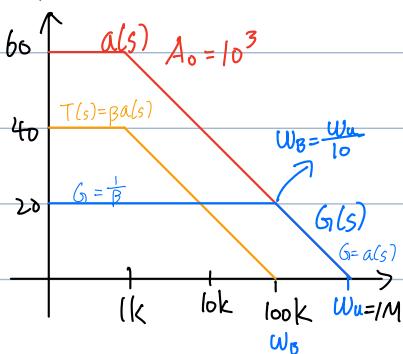
Opamp (bw)

Single pole $A(s) = \frac{A_0}{1 + \frac{s}{w_u/A_0}}$ → corner at $\frac{w_u}{A_0}$; unity at w_u

w_u : GBW (if 1st order up to w_u)



$A(\text{dB})$



$$\beta = \frac{1}{10}, T(s) = \beta_a(s)$$

$$G = \frac{1}{\beta} \frac{T(s)}{1 + T(s)}$$

$T(s) \gg 1, G = \frac{1}{\beta} \rightarrow \text{BW}\uparrow$

$T(s) < 1, G = a(s)$ (roll-off)

$$\begin{aligned} G(s) &= \frac{a(s)}{1 + \beta_a(s)} \\ &= \frac{A_0}{1 + s \frac{A_0}{w_u} + \beta A_0} \\ &= \frac{A_0}{(1 + \beta A_0) + s \frac{A_0}{w_u}} \\ &= \frac{A_0}{1 + \beta A_0} \frac{1}{1 + s \frac{1}{w_B}} \quad (w_B = w_u / A_0) \\ &\approx \frac{1}{\beta} \frac{1}{1 + s/w_B}, \quad w_B = \frac{w_u}{1/\beta} \end{aligned}$$

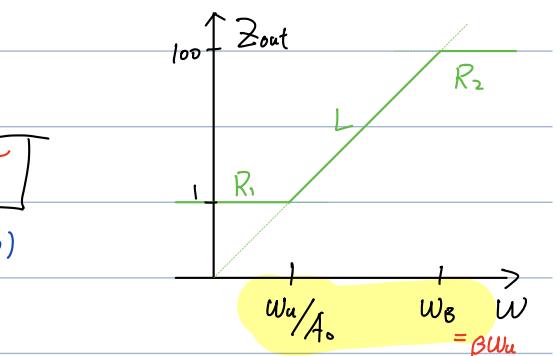
$$Z_{\text{out}}(s) = \frac{R_{\text{out}}}{1 + T(s)} = \frac{r_o || (R_1 + R_2)}{1 + T(s)}$$

Model w/ L ($Z \uparrow$ as $w \uparrow$) in sr.

$$E. R_{\text{out}} = 100\Omega : Z_{\text{out}} : (1\Omega \xrightarrow{\text{parallel}} 100\Omega)$$

$$| L = \frac{r_o}{w_B} = \frac{r_o}{\beta w_u} \approx 160 \mu H$$

If Z_L is cap → resonate :(



$$Z_{\text{in}}(s) = (R_1 || R_2) + r_{\text{in}} (1 + a(s) \beta)$$

only r_{in} !

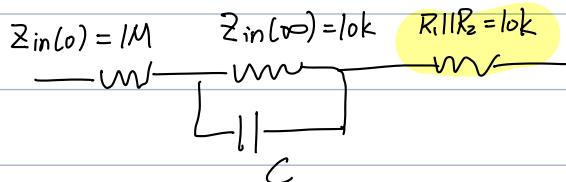
$$160 \cdot 100$$

$$E. R_{\text{in}} = 10k, R_1 || R_2 = 10k$$

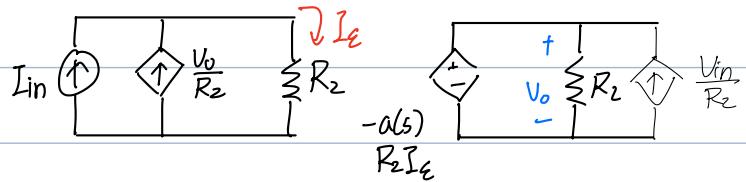
$$Z_{\text{in}}(s) = 10k + 10k (1 + a(s) \beta)$$

$$Z_{\text{in}}(0) = 1M\Omega, Z_{\text{in}}(\infty) = 20k$$

$$Z_c = \frac{1}{j\omega C} \quad C \approx 0.16 \text{ nF}$$



E. IA (sh-sh)



$$A(s) = \frac{V_o}{I_e} = -R_2 \alpha(s)$$

$$B = -\frac{1}{R_2}$$

$$T = BA = \alpha(s)$$

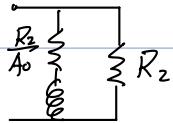
$$G = \frac{A}{1+\beta A}$$

If R_s is present in src \rightarrow G-BW tradeoff

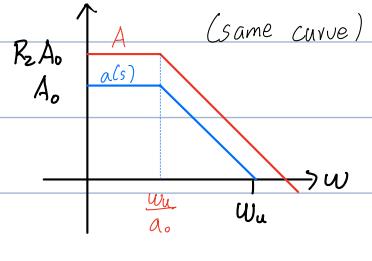
$$Z_{in} (\text{assume } r_{in} = \infty) = \frac{R_{in}}{1+T(s)} \quad (\text{Miller effect})$$

$$R_{in} = R_2$$

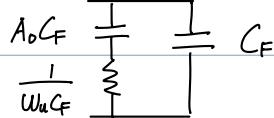
$$L_{eq} = \frac{R_2}{w_u}$$



$$Z = j\omega L$$



$$\text{If } Z_f = \frac{1}{sC_F}$$

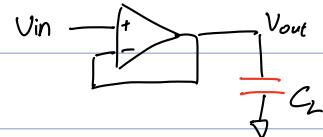


$$a(w_u) = 1$$

Zout ??

Stability

E unity buf. $\beta=1$, $w_B = 1 \text{ MHz}$

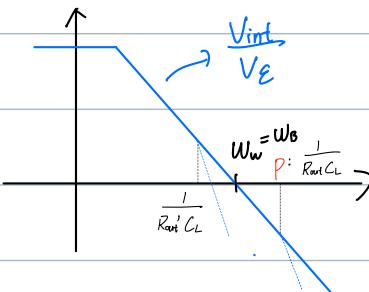
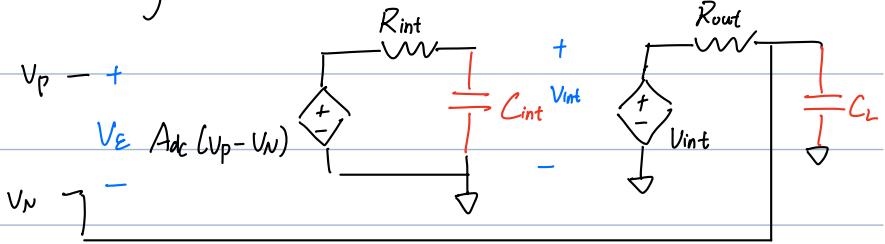


If $C_{out} \uparrow \rightarrow \text{ringing!}$ More severe if $C_L \uparrow$

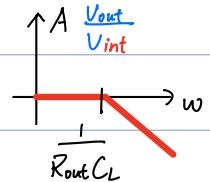
If $w_B \uparrow \rightarrow \text{ringing!}$ (faster, but more ring)

Two poles: C_{int} and C_L

two stages to isolate



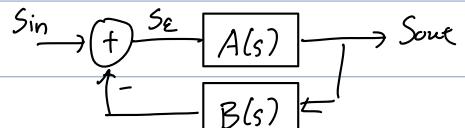
additional pole at $w = \frac{1}{R_{out} C_L}$



$$G(s) = \frac{A(s)}{1 + \beta(s)A(s)} \approx \begin{cases} \frac{1}{\beta(s)} & \text{for } A\beta \gg 1 \\ A(s) & \text{for } A\beta \ll 1 \end{cases}$$

(fine if $A\beta \neq -1$)

Let $A(s) = \frac{N(s)}{D(s)}$ $\Rightarrow G(s) = \frac{N(s)}{D(s) + \beta N(s)}$ (assume $\beta(s) = \beta$)



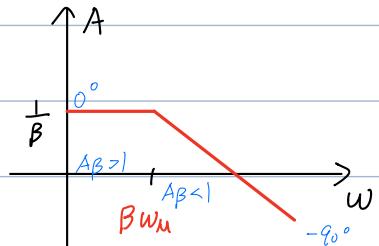
zero of $A(s) \rightarrow \infty$ of $G(s)$

pole of $A(s) \rightarrow ??$ matters for stability

1st order

$$\text{E. } \beta(s) = \beta, A(s) = \frac{w_u}{s} \rightarrow G(s) = \frac{w_u}{s + \beta w_u}$$

$$P = -\beta w_u < 0 \rightarrow \text{stable}$$



Second order E. $A(s) = \frac{A_0}{(s/w_{p1} + 1)(s/w_{p2} + 1)}$

$$G(s) = \frac{A(s)}{1 + \beta A(s)} = \frac{\frac{A_0}{(s/w_{p1} + 1)(s/w_{p2} + 1)}}{\frac{(s/w_{p1} + 1)(s/w_{p2} + 1)}{(s/w_{p1} + 1)(s/w_{p2} + 1)}} = \frac{A_0 w_{p1} w_{p2}}{s^2 + (w_{p1} + w_{p2})s + (1 + A_0\beta) w_{p1} w_{p2}}$$

$$G(s) \equiv \frac{A_0}{s^2 + (\frac{w_u}{Q})s + w_u^2}$$

Let $w_u^2 = (1 + A_0\beta) w_{p1} w_{p2}$

$$Q = \frac{\sqrt{(1 + A_0\beta) w_{p1} w_{p2}}}{w_{p1} + w_{p2}}$$

If $w_{p2} \gg w_{p1}, A\beta \gg 1, Q \approx \sqrt{\frac{A_0 \beta w_{p1}}{w_{p2}}} \equiv \sqrt{\frac{w_u}{w_{p2}}}$

We typically want $Q < 1$, so $w_{p2} > w_u$

(worst case unity)

$Q = 0.5$: critically damped

$$w_u = 0.25 w_{p2}$$

under

Freq domain criteria stability given $\beta(s)A(s)$

$G(s) = \frac{A(s)}{1 + \beta(s)A(s)}$ → p at $A(s)\beta(s) = -1$, stable if all poles in LHP

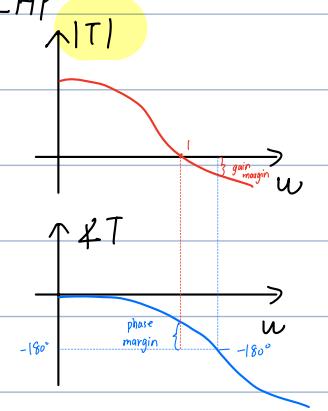
1. ω_n where $\chi[A(j\omega_n)B(j\omega_n)] = -180^\circ \Rightarrow |A(j\omega_n)B(j\omega_n)| < 1$

At phase crossover, $|T| < 1$ (gain margin $\frac{1}{|T|} > 1$)

2. ω_u where $|A(j\omega_u)B(j\omega_u)| = 1 \Rightarrow \chi(\text{---}) > -180^\circ$

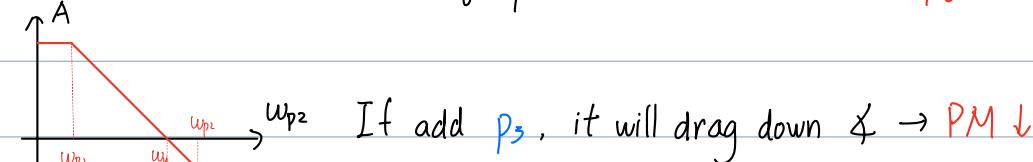
At gain crossover, $\chi T > -180^\circ$ (phase margin $180^\circ + \chi T > 0$)

if too close to margin, risk unstable

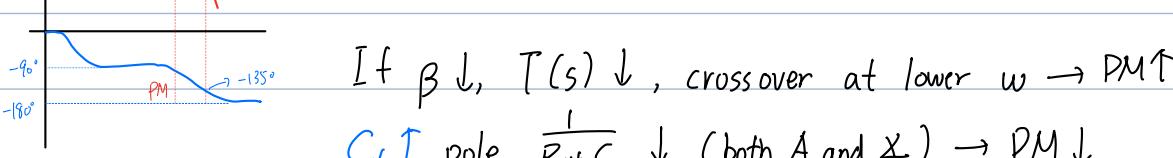


ω_n and ω_u need to be uniquely defined. Else need Nyquist test

E.



If add p_s , it will drag down $\chi \rightarrow PM \downarrow$



Aim for $PM > 45^\circ, 60^\circ \rightarrow \text{no ringing}$, critically damped

Need $\omega_u = A_0 \omega_{p1} < \omega_{p2} < \omega_{ps}$ (higher p no interfere)

if $\beta < 1$, $T(s) = \beta A(s) \downarrow \rightarrow \omega_b \downarrow \rightarrow pm \uparrow$

constant GBW & increasing C_L

$\omega_{p2} \downarrow : C$

$C_{L1} < C_{L2}$

$T = \frac{V_{FB}}{V_E}$

$\omega_u < \omega_{p1}$

$\omega_u < \omega_{p2}$

$\omega_u < \omega_{ps}$

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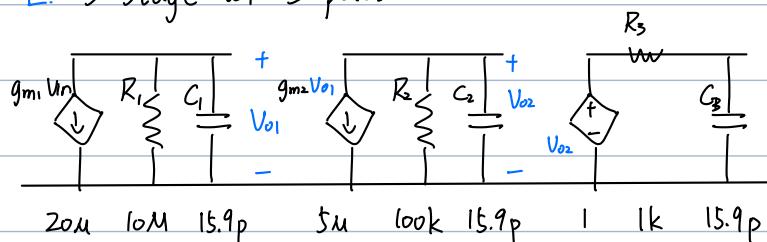
$\omega_u < \omega_{ps}$

Compensation

change $\beta(s)$ if opamp given

change $A(s)$ if designing opamp

E. 3-stage w/ 3 poles



$$DC \quad A_1 = -200$$

$$A_2 = -500$$

$$A_3 = 1$$

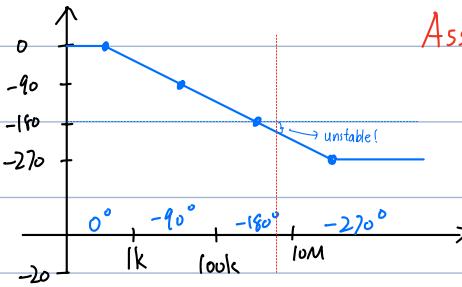
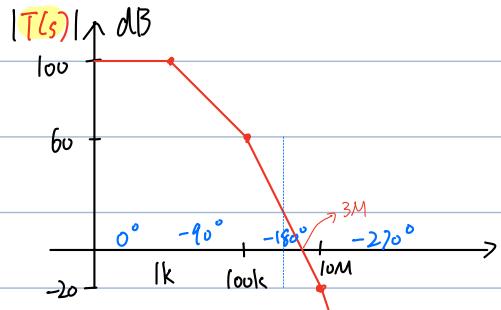
$$A = 100k \text{ (100 dB)}$$

$$\omega_p(\text{Hz}) \quad -1\text{k}\text{Hz}$$

$$-100\text{k}$$

$$-10M$$

$$BW = 1\text{kHz}$$



Sln. (can't change cap) Need $w_u = w_{p2} \rightarrow 45^\circ \text{ pm}$

1. E Reduce gain @ p_2 to 0dB \rightarrow reduce A by 60 dB

: (Less benefit from feedback E. accuracy, Z_{in}/Z_{out})

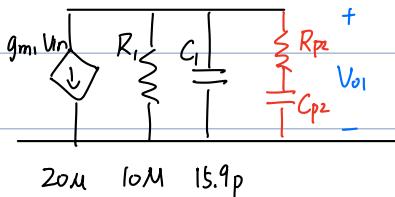
2-E. add Cap. (dominant pole compensation) Make $C_1 = 15.9 \text{nF} \rightarrow p_1 = 1 \text{Hz}$ (pull p, low)

$$w_u = w_{p2} = 100 \text{kHz} \rightarrow w_{p1} = \frac{100 \text{k}}{10^{100/20 \text{dB}}} = 1 \text{Hz}$$

pm ↑, bw ↓ : (

* Moving p_2 won't help. Need all other poles beyond $w_u = A_0 w_d$

3 p/z compensation Add z to cancel p_2 . (R dom. below p ; C above).



$$W_{p1} = -\frac{1}{R(C_1 + C_{p2})} \quad R_{p2} \approx 0 \quad \text{Assume } R_{p2} \ll R_1$$

$$W_z = -\frac{1}{R_{p2} C_{p2}}$$

$$W_{p4} = -\frac{1}{R_{p2} C_1} \quad R_1 \approx \infty$$

$$W_{p1} \ll W_{p2}$$

Also add zeroes to cancel poles

Problem: if p/z misalign \rightarrow (more overshoot, -)

E. $R_{p2} = 500k$, $C_{p2} = 500p$

$$W_4 = G_B W = W_1 a_0 \rightarrow \frac{1}{R_{p2} C_1} = a_0 \frac{1}{R_1 (C_{p2} + C_1)} \quad \left. \begin{array}{l} \\ \end{array} \right\}$$

$$W_z = \frac{1}{R_{p2} C_{p2}} = W_2 \quad \left. \begin{array}{l} \\ \end{array} \right\}$$

$$C_{p2} = \sqrt{\frac{a_0 C_1}{R_1 W_2}} = 500p$$

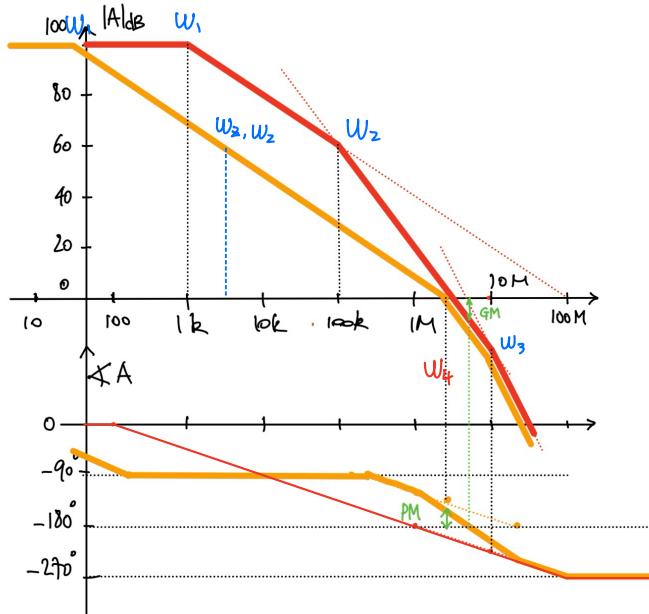
$$R_{p2} = \frac{1}{W_2 C_{p2}} = 3.2k$$

$$W_1 \doteq 31 \text{ Hz}$$

$$W_4 \doteq 31 \text{ MHz} \rightarrow 45^\circ \text{ pm}$$

idea: cancel w_2 w/ w_z . Replace w/ w_4 @ unity

Caveat if p/z misaligned \rightarrow problem



	$P_1 \text{ (rad/s)}$	P_2	$G_B W$	$P_3 \dots$	A_o'
Reduce A_o	$\frac{1}{R_1 C_1}$	$\frac{1}{R_2 C_2}$	$P_2 = \frac{1}{R_2 C_2}$	$P_3 \dots$	$A_o \frac{w_1}{w_2}$
dom. pole	$\frac{1}{R_1 C_1} \uparrow$	$\frac{1}{R_2 C_2}$			
p/z comp.	$\frac{1}{R_1 (C_1 + C_{p2})}$	$\frac{1}{R_{p2} C_{p2}} = W_2$	$\frac{1}{R_{p2} C_1} = W_4$		
Miller	\uparrow	\downarrow	$\frac{gm_i}{C_C}$		

Miller (sh-sh feedback)

if comp put across gain element, $Z \uparrow$

$$V_{cc} = V_{o1} - g_{m2} R_2 V_{o1} = (1 + g_{m2} R_2) V_{o1}$$

$$W_1 = \frac{1}{R_1(C_c + C_{eq})} = \frac{1}{R_1 A_2 C_c} \quad (\downarrow)$$

$$GBW = \frac{1}{R_1 A_2 C_c} (g_{m1} R_1 A_2) = \frac{g_{m1}}{C_c}$$

$$g_{m2 eq} = g_{m2} - s C_c$$

$$\hookrightarrow RHP \text{ } \exists @ W_{z2} = \frac{g_{m2}}{C_c}$$

$$\frac{V_{o2}(s)}{V_{in}(s)} = \frac{g_{m1} R_1 g_{m2} R_2 (1 - \frac{s}{g_{m2} C_c})}{1 + s(L(C_1 + C_c(1 + g_{m2} R_2))R_1) + s^2(R_1 R_2 L(C_2 + C_c)C_1 + C_2 C_c)}$$

$$= \frac{N(s)}{1 + as + bs^2} = \frac{N(s)}{(1 + \frac{s}{\omega_1})(1 + \frac{s}{\omega_2})}$$

$$\text{if } \omega_1 \ll \omega_2 \quad \frac{1}{\omega_1} = R_1(C_c(1 + g_{m2} R_2) + C_1) + R_2(C_c + C_2)$$

$$\omega_2 = \frac{R_1(C_c(1 + g_{m2} R_2))}{R_1 R_2 ((C_1 + C_c)C_1 + C_2 C_c)} \approx \frac{g_{m2} C_c}{C_1 C_2 + C_1 C_c + C_2 C_c}$$

if $C_2 \gg C_c, C_2 > C_1$,

$$\omega_1' \approx \omega_1 \frac{C_1}{g_{m2} R_2 C_c}$$

$$\omega_2' \approx g_{m2} R_2 \omega_2 \quad (C_2, C_c \gg C_1) \uparrow$$

$C_c = 0$	$C_c \text{ large}$
$\frac{1}{R_1 C_1}$	$\frac{1}{R_1 (C_1 + g_{m2} R_2 C_c)}$
$\frac{1}{R_2 C_2}$	$\frac{g_{m2}}{C_1 + C_2 + \frac{C_1 C_c}{C_2}}$
RHP \exists	X

P_2 up !!! (pole splitting) \rightarrow pm \uparrow

additional \exists LHP \exists $0^\circ \rightarrow +90^\circ$ help pm

leading 90°

RHP \exists $0^\circ \rightarrow -90^\circ ?$ hurt pm

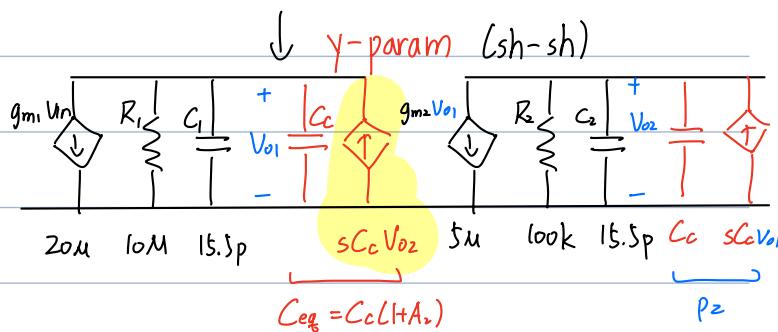
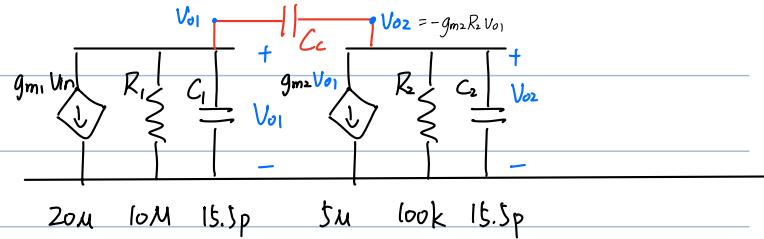
lagging 90°

Need push \exists high, or make it LHP

If add large C directly, slow RC

but Miller, $RC \downarrow$

If add R_C , less overshoot / oscillation (improved PM)



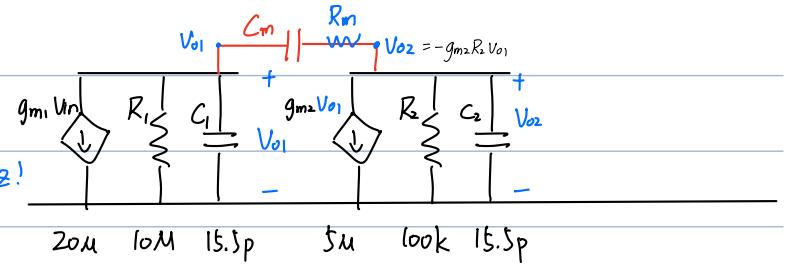
RHP zero compensation

Add $C_m - R_m$, avoid C_c feedfwd ($sC_c V_{o1}$)

$$g_{m2, eq} = g_{m2} - \frac{sC_m}{1+sR_m C_m} = g_{m2} \frac{1+sC_m(R_m - 1/g_{m2})}{1+sR_m C_m} \rightarrow z!$$

If $R_m > \frac{1}{g_{m2}}$ → no RHP z

(LHP helps)



↓ γ-param (sh-sh)

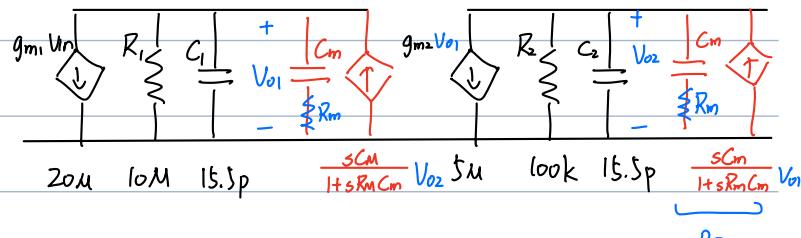
V_{cm} is a new state var. → 3p

if $g_{m1}R_1, g_{m2}R_2 \gg 1, C_m \gg C_1, C_2 \gg C_1$,

$$W_{p1} \approx \frac{1}{R_1 g_{m2} R_2 C_m} \quad \left. \begin{array}{l} \\ \end{array} \right\} \text{same}$$

$$W_{p2} \approx \frac{g_{m2} C_m}{C_1 C_2 + C_1 C_m + C_2 C_m}$$

$$W_{p3} \approx \frac{1}{R_m C_1} \quad \rightarrow \text{new}$$



Pz

Pz