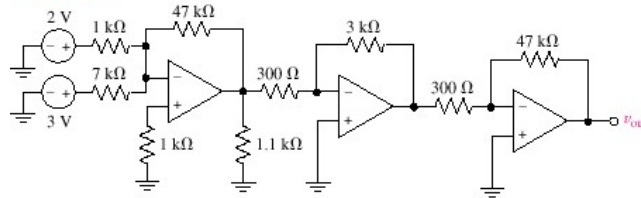


Example: For the cascaded op amp circuit shown in Fig. 6.40, compute the output voltage of each stage.

Figure 6.40



Solution: The output of the first op amp stage may be found by realising that the voltage at the non-inverting input (and hence the voltage at the inverting input) is 0, and writing a single nodal equation at the inverting input:

$$0 = \frac{0 - V_{\text{out}}|_{\text{stage 1}}}{47} + \frac{0 - 2}{1} + \frac{0 - 3}{7}$$

which leads to $V_{\text{out}}|_{\text{stage 1}} = -114.1V$

This voltage appears at the input of the second op amp stage, which has a gain of $3/0.3 = 10$.

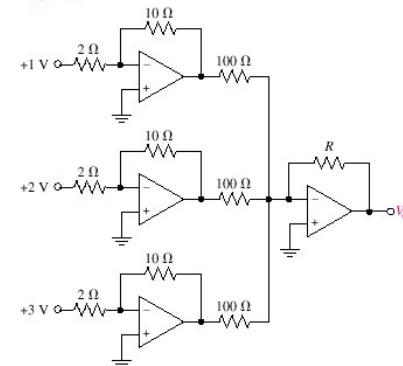
Thus, the output of the second op amp stage is $-10(-114.1) = 1141V$.

This voltage appears at the input of the final op amp stage, which has a gain of $47/0.3 = 156.7$.

Thus, the output of the circuit is $-156.7(1141) = -178.8kV$.

Example: Referring to the op amp circuit of Fig. 6.41, what value of R is required to obtain $V_{\text{out}} = 10V$?

Figure 6.41



Solution: The output of the top left stage is
 $-1(10/2) = -5$ V.

The output of the middle left stage is $-2(10/2) = -10$ V.

The output of the bottom right stage is $-3(10/2) = -15$ V.

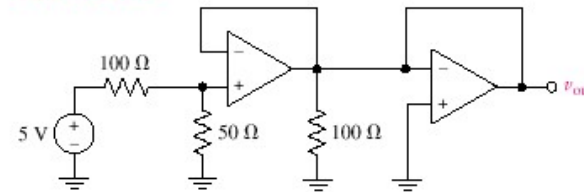
These three voltages are the input to a summing amplifier such that

$$V_{out} = -\frac{R}{100}(-5 - 10 - 15) = 10V$$

Solving, we find that $R = 33.33 \Omega$.

Example: Compute v_{out} for the two-stage op amp circuit of Fig. 6.42.

Figure 6.42



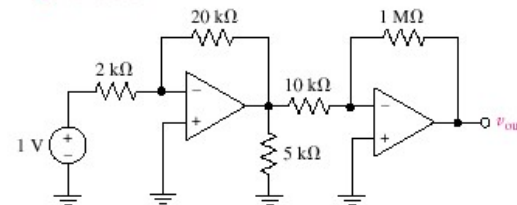
Solution: Stage 1 is configured as a voltage follower: the output voltage will be equal to the input voltage. Using voltage division, the voltage at the non-inverting input (and hence at the inverting input, as well), is

$$5 \frac{50}{100 + 50} = 1.667 \text{ V}$$

The second stage is wired as a voltage follower also, so
 $v_{out} = 1.667$ V.

Example: Compute v_{out} for the circuit of Fig. 6.38.

Figure 6.38

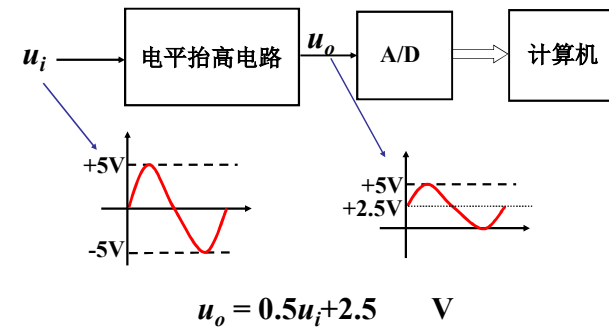


Solution: v_{out} of stage 1 is $(1)(-20/2) = -10 \text{ V}$.

v_{out} of stage 2 is $(-10)(-1000/10) = 1000 \text{ V}$

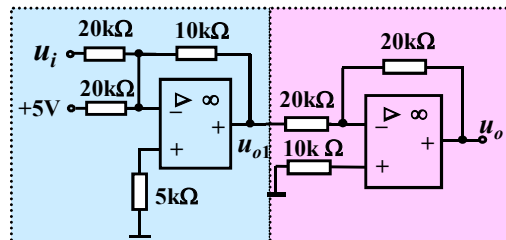
Note: in reality, the output voltage will be limited to a value less than that used to power the op amps.

例: A/D变换器要求其输入电压的幅度为 $0 \sim +5\text{V}$ ，现有信号变化范围为 $-5\text{V} \sim +5\text{V}$ 。试设计一电平抬高电路，将其变化范围变为 $0 \sim +5\text{V}$ 。



$$u_o = 0.5u_i + 2.5 \quad \text{V}$$

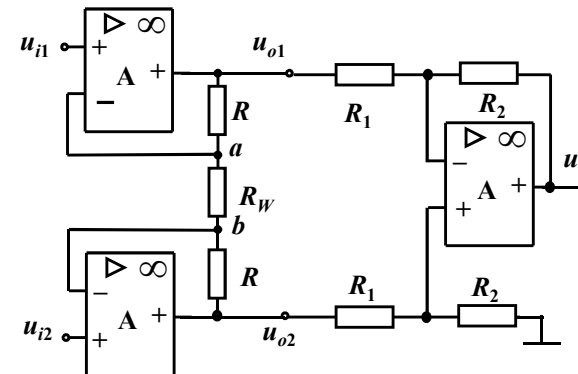
$$= 0.5(u_i + 5) \quad \text{V}$$

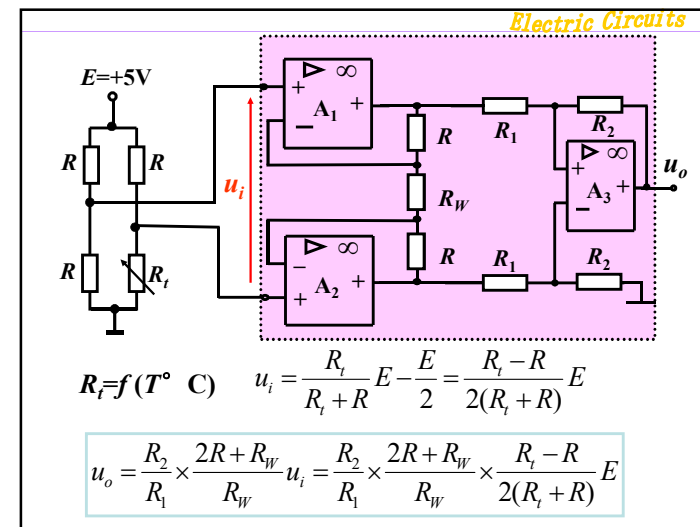
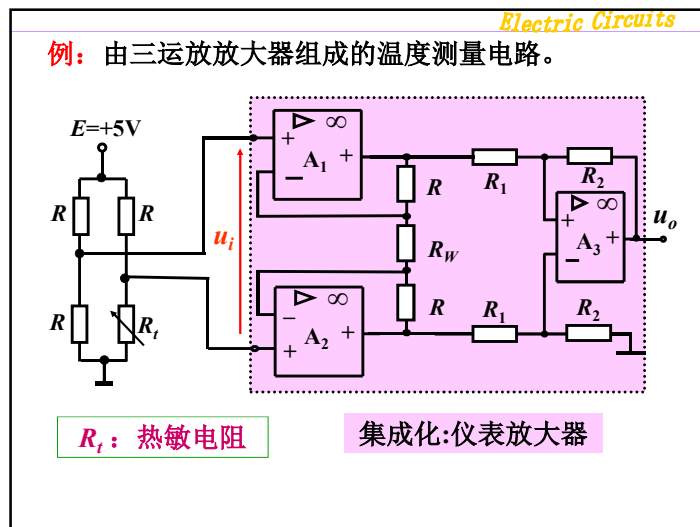
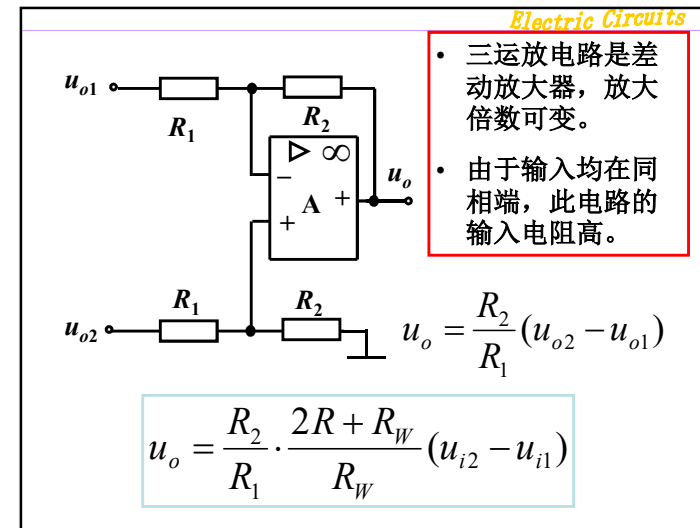
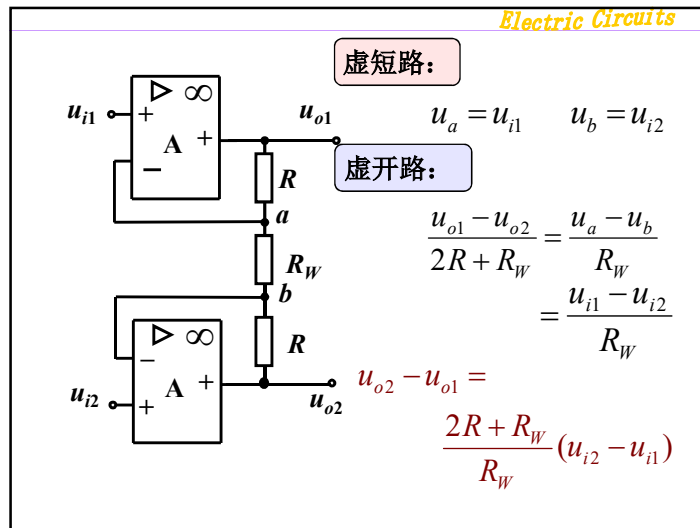


$$u_{o1} = -\frac{10}{20} \times (u_i + 5) = -0.5(u_i + 5)$$

$$u_o = -\frac{20}{20} \times u_{o1} = 0.5(u_i + 5)$$

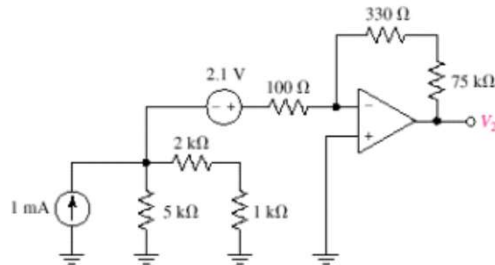
五、三运放电路



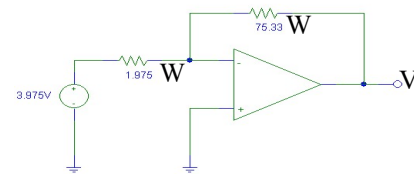


Example: For the circuit of Fig. 6.27, calculate the voltage V_2 .

Figure 6.27



Solution: A source transformation and some series combinations are well worthwhile prior to launching into the analysis. With $5 \text{ k}\Omega \parallel 3 \text{ k}\Omega = 1.875 \text{ k}\Omega$ and $(1 \text{ mA})(1.875 \text{ k}\Omega) = 1.875 \text{ V}$, we may redraw the circuit as

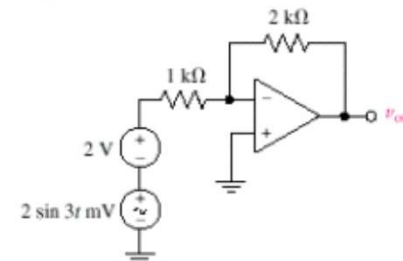


This is now a simple inverting amplifier with gain $-R_f/R_i = -75.33/1.875 = -38.14$.

Thus, $V_2 = -38.14(3.975) = -151.6 \text{ V}$.

Example: Find an expression for v_{out} in the circuit of Fig. 6.28, and evaluate it at $t = 3$ seconds.

Figure 6.28



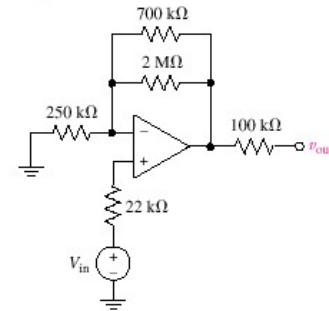
Solution: This is a simple inverting amplifier, so we may write

$$v_{\text{out}} = \frac{-2000}{1000}(2 + 2 \sin 3t) = -4(1 + \sin 3t) \text{ V}$$

$$v_{\text{out}}(t = 3 \text{ s}) = -5.648 \text{ V.}$$

Example: What value of V_{in} will lead to an output voltage of 18 V in the circuit of Fig.6.29?

Figure 6.29



Solution: We first combine the 2 MΩ and 700 kΩ resistors into a 518.5 kΩ resistor.

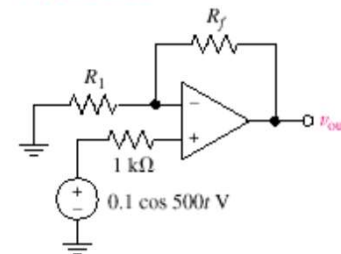
We are left with a simple non-inverting amplifier having a gain of

$$1 + 518.5 / 250 = 3.074. \text{ Thus,}$$

$$v_{\text{out}} = (3.074) v_{\text{in}} = 18 \text{ so } v_{\text{in}} = 5.856 \text{ V.}$$

Example: Choose R_1 and R_f in Fig. 6.30 to obtain $v_{\text{out}} = 23.7 \cos 500t$ volts.

Figure 6.30



Solution:

This is a simple non-inverting amplifier circuit, and so it has a gain of $1 + R_f/R_1$.

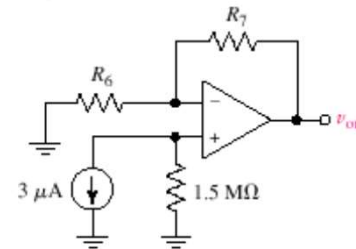
We want $v_{\text{out}} = 23.7 \cos 500t$ V when the input is $0.1 \cos 500t$ V, so a gain of $23.7/0.1 = 237$ is required.

One possible solution of many:

$$R_f = 236 \text{ k}\Omega \text{ and } R_1 = 1 \text{ k}\Omega.$$

Example: Derive an expression for v_{out} for the circuit of Fig. 6.31 without using source transformations.

Figure 6.31



Solution: Define a nodal voltage V_- at the inverting input, and a nodal voltage V_+ at the non-inverting input. Then,

$$\text{At the non-inverting input: } -3 \times 10^{-6} = \frac{V_+}{1.5 \times 10^6}$$

[1] Thus, $V_+ = -4.5$ V, and we therefore also know that $V_- = -4.5$ V.

$$\text{At the inverting input: } 0 = \frac{V_-}{R_6} + \frac{V_- - v_{\text{out}}}{R_7}$$

[2] Solving and making use of the fact that $V_- = -4.5$ V,

$$v_{\text{out}} = -\frac{R_7}{R_6}(-4.5) - 4.5 = -4.5 \left(\frac{R_7}{R_6} + 1 \right) \text{ V}$$