Exam presentation

Assignment 2.1 and 3

Riccardo Miccini¹

¹Technical University of Denmark Digital Communication

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Overview

- Assignment 2
 - Eye diagrams
 - Q function
 - Matched filter
- Assignment 3
 - Link budget model
 - lacksquare SNR and P_E
 - Alternative modulations

Eye diagram

- Characteristics
- Impact of bandwidth

Eye diagram

- Plot composed by overlaying segments of different bit sequences
- Can be generated with an oscilloscope
- Shows effects of *inter-symbol interference*
- Provides a qualitative measure of the system performance

Eye diagram characteristics

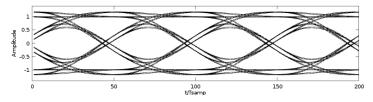


Figure: Eye diagram of baseband antipodal signal

- A Difference between high and low levels
- A_j Difference between A and the eye opening
- T_j Deviations from ideal timing
- T_b Bit time period



Eye diagram at different bandwidths

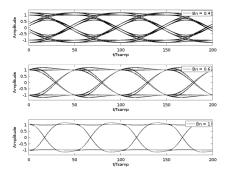


Figure: Eye diagram for normalized bandwidths 0.3, 0.7, 1.2

- Low BW: high amplitude and timing jitter
- High BW: no ISI, chances of higher noise

Q function

- Normal pdf
- Q function in relation to normal pdf
- Q function in relation to complementary error function
- Inverse Q function

Normal probability density function

- Continuous function representing likelihood of its argument
- Useful to analyze phenomena of unknown distribution due to central limit theorem
- General form:

$$\mathcal{N}(\mu, \sigma^2) = \frac{1}{\sigma\sqrt{2\pi}} e^{-\frac{(x-\mu)^2}{2\sigma^2}} \tag{1}$$

- μ Average value
- σ Standard deviation

Q function

- Represents the tail probability of $\varphi(x)$ (standard normal distribution)
- Definition of Q(x) through $\varphi(x)$:

$$Q(x) = \frac{1}{\sqrt{2\pi}} \int_x^\infty e^{-\frac{u^2}{2}} du \tag{2}$$

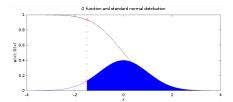


Figure: Relation between Q function and standard normal distribution

Q function and complementary error function

- Error function $\operatorname{erf}(x)$: probability of normally-distributed random variable X ($\mu=0,\ \sigma^2=\frac{1}{2}$) to be in the range [-x,x]
- Complementary error function $\operatorname{erfc}(x) = 1 \operatorname{erf}(x)$
- Definition of Q(x) through $\operatorname{erfc}(x)$:

$$Q(x) = \frac{1}{2}\operatorname{erfc}\left(\frac{x}{\sqrt{2}}\right) \tag{3}$$

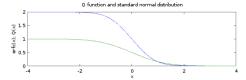


Figure: Q function (green) and complementary error function (blue)

Inverse Q function

 $lacksquare Q^{-1}(x)$ is the value u for which Q(u)=x

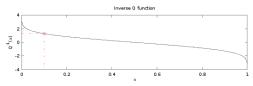


Figure: Inverse Q function

 \blacksquare If the argument x represents a bit error probability, $Q^{-1}(x)$ is proportional to the SNR

Matched filter



Link budget for satellite communication

- Link budget model
- SNR calculation
- $lacksquare P_E$ calculation
- Impact of of d, λ , B, and P_T
- Alternative modulation techniques: ASK and FSK

Link budget model

- A way of estimating the power of a received signal
- Takes into account all the gains and losses of transmitter, channel, and receiver

$$P_R = \left(\frac{\lambda}{4\pi d}\right)^2 \frac{P_T G_T G_R}{L_0};\tag{4}$$

In decibels:

$$P_{R,dB} = 20 \log_{10} \left(\frac{4\pi d}{\lambda} \right) + ERP_{dB} + G_{R,dB} - L_{0,dB};$$
 (5)



Link budget model variables

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(rac{4\pi d}{\lambda})^2 Free-space loss ERP=P_TG_T Effective radiated power G_R Gain of receiver antenna L_0 Other losses, link budget margin
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SNR calculation

Signal-to-noise ratio in decibels:

$$SNR_{dB} = P_{R,dB} - P_{int,dB} \tag{6}$$

 P_R Calculated using link budget model

 P_{int} Noise power, proportional to the receiver noise temperature and the transmission bandwidth

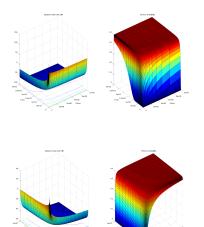
P_E calculation

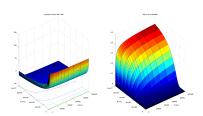
Bit error probability for BSFK transmissions:

$$P_E = Q\left(\sqrt{\frac{2E_b}{N_0}}\right) \tag{7}$$

- I ratio E_b/N_0 derived from $SNR = \rightarrow \frac{E_b}{N_0BT_b}$
- **2** For binary BPSK, $B=2/T_b$
- \blacksquare Factor BT_b is 2, or 3 dB

Impact of d, λ , B





- SNR and P_E are negatively affected by:
 - $\quad \blacksquare \ \, \mathsf{Higher} \,\, \mathsf{distance} \,\, d$
 - \blacksquare Lower wavelength λ
 - Wider bandwidth B (lower influence)



Impact of P_T

 \blacksquare P_E for P_T at 50 W, 5 W, and 500 mW:

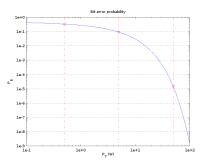


Figure: P_E over values of P_T

Alternative modulation: ASK

Amplitude-shift keying

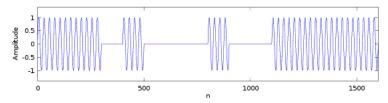


Figure: Bit stream modulated using ASK

- 0-bit represented as 0
- 1-bit represented as $A\cos(2\pi f_c t)$

Alternative modulation: ASK (P_E calculation)

- Correlation coefficients:
 - $\rho_{12} = \frac{1}{\sqrt{E_1 E_2}} \int_{-\infty}^{\infty} s_1(t) s_2(t) dt = 0$
 - $R_{12} = \frac{\sqrt{E_1 E_2}}{E_h} \rho_{12} = 0$
- SNR to E_b/N_0 : conversion factor $BT_b=2$
- Bit error probability:

$$P_E = Q\left(\sqrt{(1 - R_{12})\frac{E_b}{N_0}}\right) = Q\left(\sqrt{\frac{E_b}{N_0}}\right) \tag{9}$$

Alternative modulation: FSK

Frequency-shift keying

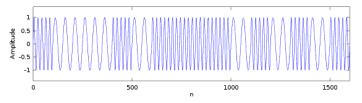


Figure: Bit stream modulated using FSK

- lacksquare 0-bit represented as $A\cos(\omega_c t)$
- 1-bit represented as $A\cos((\omega_c + \Delta\omega)t)$
- Assumptions: $\omega_c = \frac{2\pi n}{T}$ and $\Delta\omega = \frac{2\pi m}{T}$



Alternative modulation: FSK (P_E calculation)

- Correlation coefficient $R_{12}=\frac{\sqrt{E_1E_2}}{E_b}\rho_{12}=0$
- SNR to E_b/N_0 : conversion factor $BT_b=2.5$
- Bit error probability:

$$P_E = Q\left(\sqrt{(1 - R_{12})\frac{E_b}{N_0}}\right) = Q\left(\sqrt{\frac{E_b}{N_0}}\right) \tag{10}$$