

## EE247 Lecture 27

- Term project student presentations
- Acknowledgements
- Examples of systems utilizing analog-digital interface circuitry (not part of final exam - self study)

## Administrative

- Office hours on Frid. Dec. 12<sup>th</sup>, 3 to 4:30pm @ 477 Cory
- No office hours Thurs. Dec. 11<sup>th</sup>
- Questions can be asked via email

## Term Project Presentations

- **Ping-Chen & James**
- **Rikky & Chintan**
- **Jiash & Maryam**
- **Nam-Seog & Jungdong**
- **Kyoo Hyun & Kwangmo**
- **Lingkai & Thura**
- **Lauren & Mervin**
- **Abhinav & Jason**
- **Lu Ye**

## Acknowledgements

- The course notes for EE247 are based on numerous sources including:
  - Prof. P. Gray's EE290 course
  - Prof. B. Boser's EE247 course notes
  - Prof. B. Murmann's Nyquist ADC notes
  - Fall 2004 & '05 & '06 & '07 EE247 class feedback
  - Last but not least, Fall 2008 EE247 class
    - The instructor would like to thank the class of 2008 for their enthusiastic & active participation!

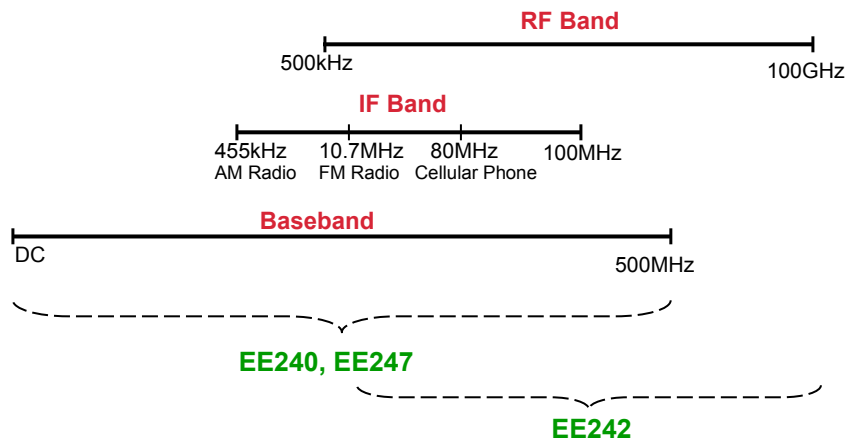
## Material Covered in EE247

- Filters
  - Continuous-time filters
    - Biquads & ladder type filters
    - Opamp-RC, Opamp-MOSFET-C, gm-C filters
    - Automatic frequency tuning
  - Switched capacitor (SC) filters
- Data Converters
  - D/A converter architectures
  - A/D converter
    - Nyquist rate ADC- Flash, Interpolating & Folding, Pipeline ADCs,....
    - Self-calibration techniques
    - Oversampled converters

## Systems Including Analog-Digital Interface Circuitry (Not Included in Final Exam)

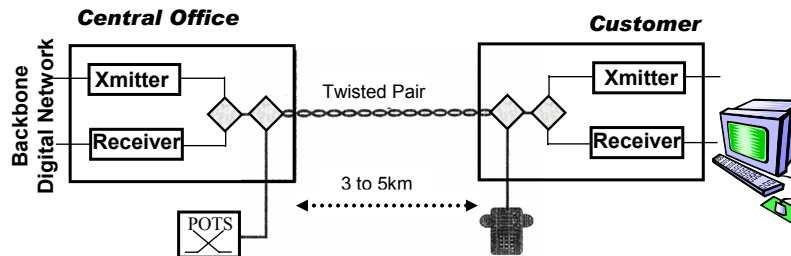
- Wireline communications
  - Telephone related (DSL, ISDN, CODEC)
  - Television circuitry (Cable modems, TV tuners...)
  - Ethernet (10/1Gigabit, 10/100BaseT...)
- Wireless
  - Cellular telephone (CDMA, Analog, GSM....)
  - Wireless LAN (Blue tooth, 802.11a/b/g.....)
  - Radio (analog & digital), Television
- Disk drives
- Fiber-optic systems

## E.E. Circuit Course vs. Frequency Range



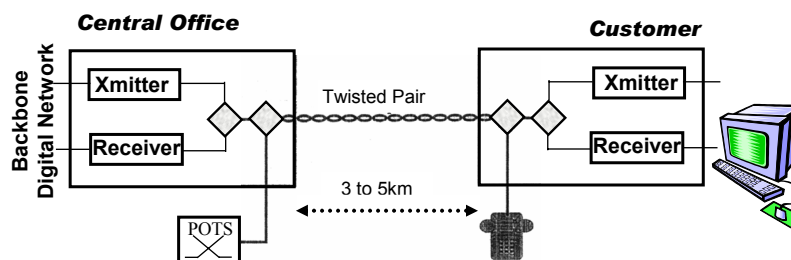
## Wireline Communications Telephone Based

## Data Transmission Over Existing Twisted-Pair Phone Lines



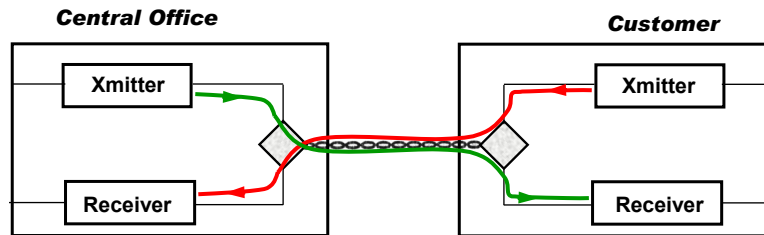
- Data transmitted over existing phone lines covering distances close to 3.5 miles
  - Voice-band MODEMs (up to 56Kb/s)
  - ISDN (160Kb/s)
  - HDSL, SDSL,.....
  - ADSL (up to 8Mb/s)

## Data Transmission Over Twisted-Pair Phone Lines ISDN (U-Interface) Transceiver

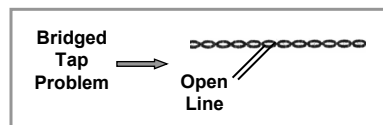


- Full duplex transmission (RX & TX signals sent simultaneously)
- 160kbit/sec baseband data (80kHz signal bandwidth)
- Standardized line code 2B1Q (4 level code 3:1:-1:-3)
- Max. desired loop coverage 18kft (~36dB signal attenuation)
- Final required BER (bit-error-rate)  $10^{-7} \rightarrow$  (min. SNDR=27dB)

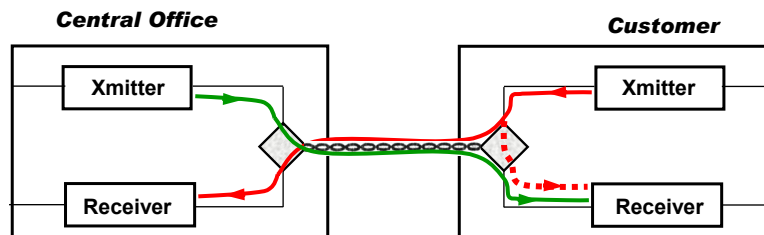
## ISDN (U-Interface) Transceiver Echo Problem



- Transformer coupling to line
  - For a perfectly matched system → no leakage of TX signal into RX path
  - Unfortunately, system has poor matching + complicating factor of bridged-taps

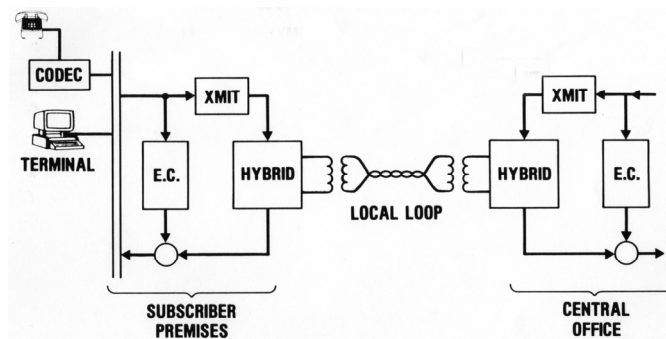


## ISDN (U-Interface) Transceiver Echo Problem



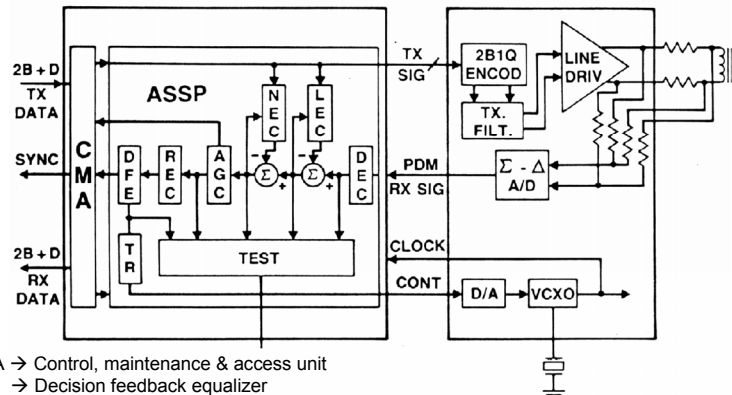
- System full duplex transmission → RX & TX signals sent simultaneous (& at the same frequency band)
  - Leakage of TX signal to RX path (echo)
  - Worst case → echo could be **30dB** higher compared to the received signal!!

## ISDN (U-Interface) Transceiver Echo Cancellation



- Echo cancellation performed in the digital domain
  - Typically echo cancellation performed by transversal adaptive digital filter
  - Any non-linearity incurred by the analog circuitry makes echo canceller significantly more complex
  - Desirable to have high linearity analog circuitry (75dB range)

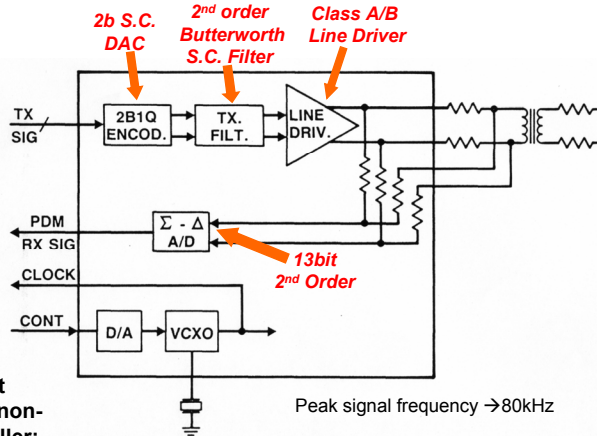
## Simplified Transceiver Block Diagram



CMA → Control, maintenance & access unit  
 DFE → Decision feedback equalizer  
 DEC → Decimation filter  
 REC → Reconstruction filter  
 LEC & NEC → Linear/non-linear echo-canceller

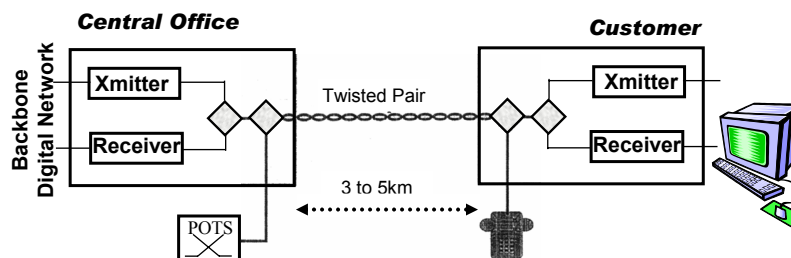
Ref: H. Khorramabadi, et. al "An ANSI standard ISDN transceiver chip set," *IEEE International Solid-State Circuits Conference*, vol. XXXII, pp. 256 - 257, February 1989

## Analog Front-End



To avoid stringent requirements for non-linear echo canceller:  
→ high linearity analog circuitry needed (~ 75dB)

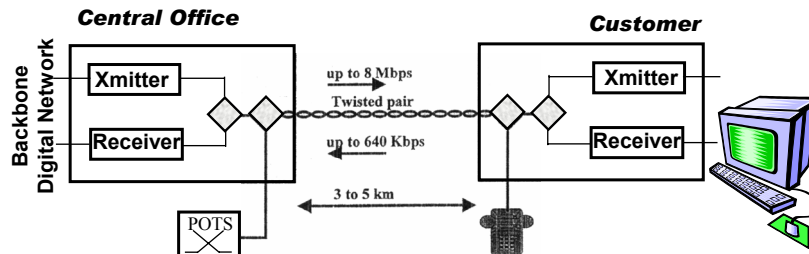
## Data Transmission Over Twisted-Pair Phone Lines DSL (Digital Subscriber Loop)



- HDSL & SDSL more like ISDN @ higher frequencies
  - Full duplex transmission with RX & TX signals on the same frequency band



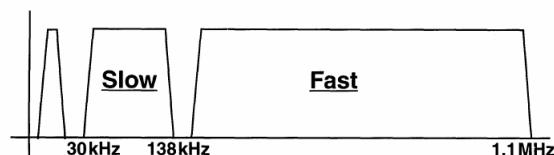
## Data Transmission Over Twisted-Pair Phone Lines ADSL (Asymmetric Digital Subscriber Loop)



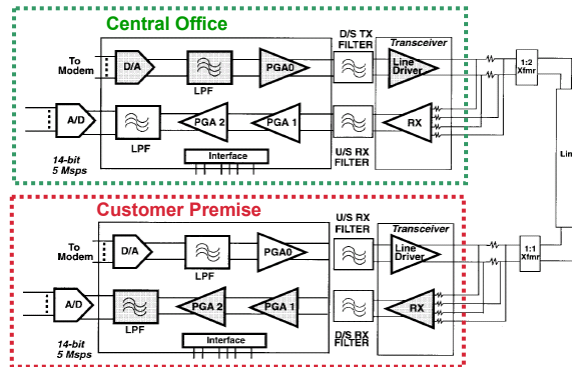
- In USA mostly ADSL → FDM (frequency division multiplex)
  - Signal from CO to customer on a different band compared to customer to CO
    - Echo cancellation can be performed by simple filtering
  - Data rates up to 8Mbps (much higher compared to ISDN)

## ADSL Signal Characteristics

- Main difference compared to ISDN: TX & RX signals on different frequency bands
  - Downstream (*fast*, from CO to customer) 138kHz to 1.1MHz
  - Upstream (*slow*, from customer to CO) 30kHz to 138kHz
    - Echo cancellation much easier
- More severe signal attenuation at high frequencies (1MHz DSL v.s. 80kHz ISDN)

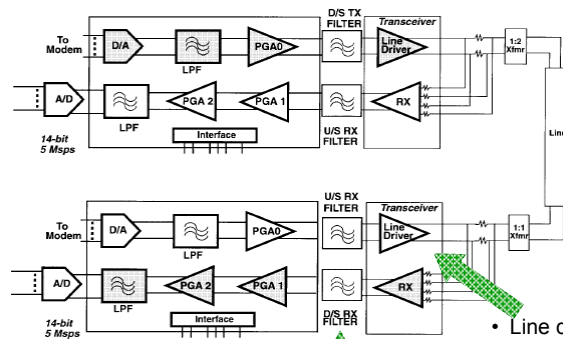


## Typical ADSL Analog Front-End



- ADC 16/14b with 14bit linearity, pipeline with auto. calibration @ 5Ms/s
  - DAC 16/14b with 14bit linearity, with auto. calibration
  - On-chip filters 3<sup>rd</sup> to 4<sup>th</sup> order LPF with  $f_c$  1.1MHz for downstream and 138kHz upstream (typically continuous-time type filters with on-chip frequency tuning)
- Ref: D.S. Langford, et al, "A BiCMOS Analog Front-End Circuit for an FDM-Based ADSL System,"  
*IEEE Journal of Solid State Circuits*, Vol. 33, No. 9, pp. 1383-1393, Dec. 1998.

## Typical ADSL Analog Front-End



# Wireless Communication Circuits

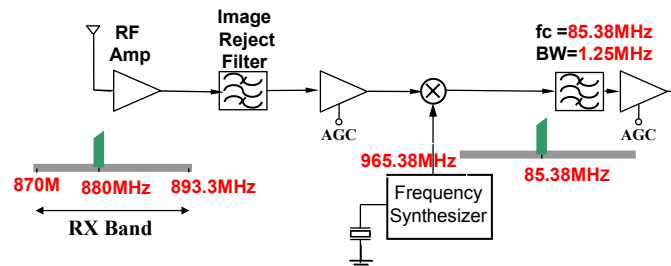
## Wireless Circuits

- Differ from wired comm. circuits
  - Includes RF circuitry + IF circuitry + baseband circuits (three different frequency ranges)
  - Signal scenarios in wireless receivers more challenging
  - Requirement for received signal BER in the order of  $10^{-3}$  for voice-only → (min. SNR ~ 9dB)

The diagram illustrates a software-defined radio receiver architecture. It starts with an antenna connected to an RF Amp. The signal then passes through an Image Reject Filter, followed by an AGC (Automatic Gain Control) block. The output of the AGC is multiplied (indicated by a circle with an 'X') by a signal from a Frequency Synthesizer, which provides a local oscillator signal  $f_2$ . The multiplier's output is then filtered to produce the final signal  $f_c = f_2 - f_1$ . Spectral plots are shown at various stages: the input signal  $f_1$ , the intermediate signals  $f_2 - f_1$  and  $f_2 + f_1$ , and the final output signal  $f_c = f_2 - f_1$ .

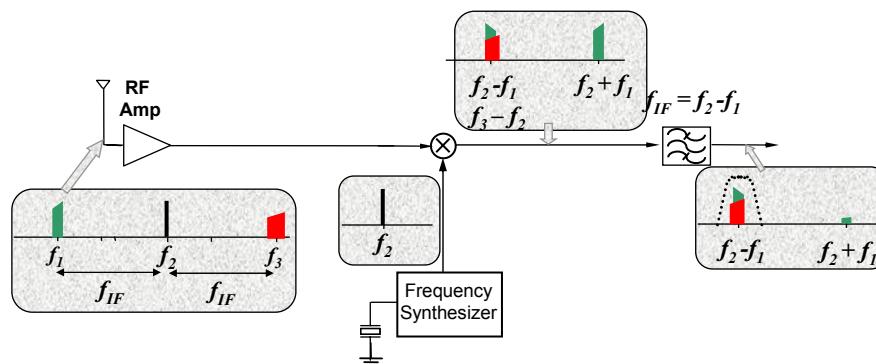
- © 2008 H. K. Page 24

## RF Superheterodyne Receiver Example: CDMA Receiver



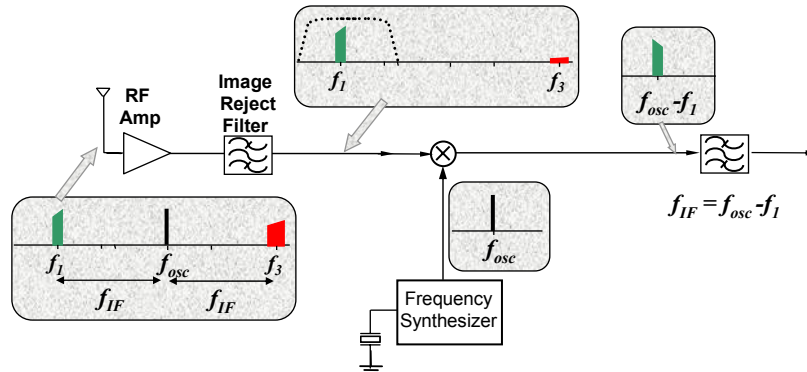
- Received frequency is mixed down to a fixed IF frequency and then filtered with a bandpass filter

## Why Image Reject Filter?



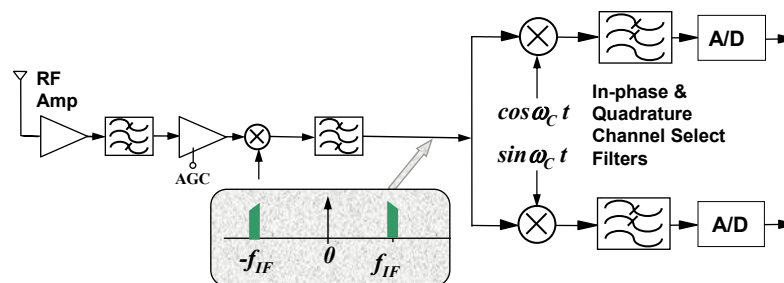
- Any signal @ the image frequency of the RX signal with respect to Osc. frequency will fall on the desired RX signal and cause impairment

## Why Image Reject Filter?



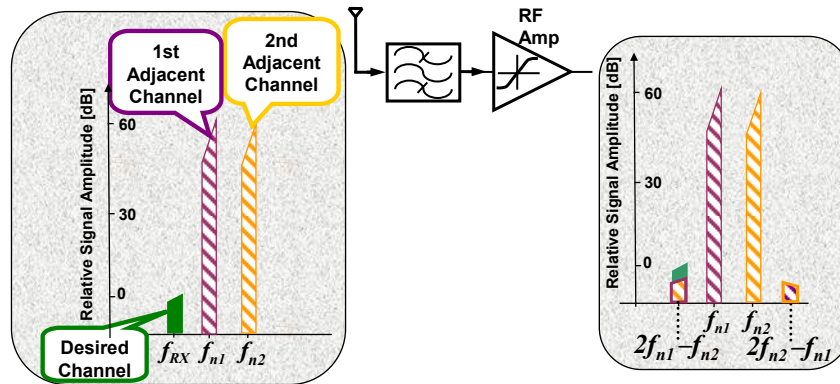
- Image reject filter attenuate signals out of the RX band
- Typically, image reject filters are ceramic or LC type filters

## Quadrature Downconversion



- In systems with phase or freq. modulation, since signal is not symmetric around  $f_{IF}$ , directly converting down to baseband corrupts the sidebands  
→ Quadrature downconversion overcomes this problem

## Effect of Adjacent Channels

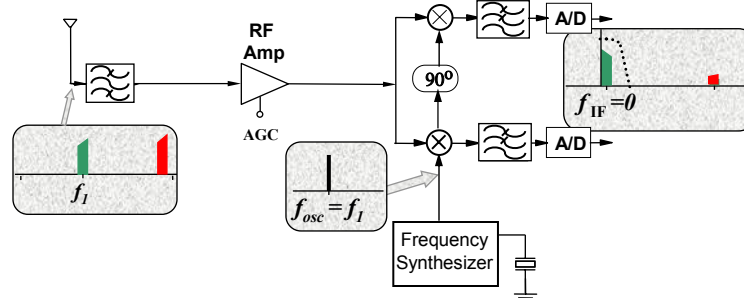


- Adjacent channels can be as much as **60dB** higher compared to the desired RX signal!
- Linearity of stages prior and including channel selection filters extremely important

## Effect of Adjacent Channels

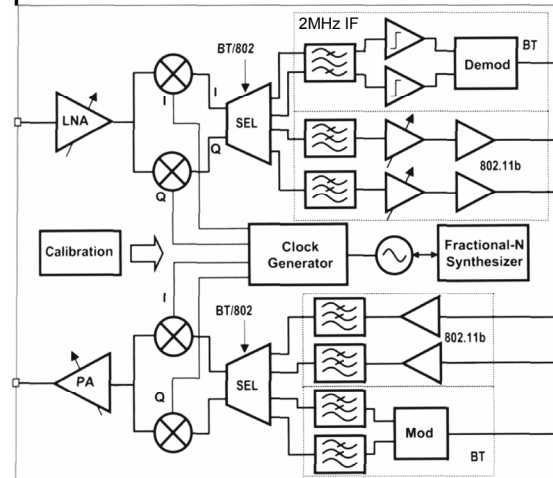
- Due to existence of large unwanted signals & limited dynamic range for the front-end circuitry:
  - Can not amplify the signal up front due to linearity issues
  - Need to allocate amplification/filtering numbers to RX blocks carefully
  - Can only amplify when unwanted signals are filtered adequately
  - System design critical with respect to tradeoffs affecting:
    - Gain
    - Linearity
    - Power dissipation
    - Chip area

## Homodyne (Direct to Baseband) Receivers



- No intermediate frequency, signal mixed directly down to baseband
- Almost all of the filtering performed at baseband
  - Higher levels of integration possible
  - Issue to be aware of:
    - Requirements for the baseband filters more stringent
    - Since the local oscillator frequency is exactly at the same freq. as the RX signal freq.  $\rightarrow$  can cause major DC offsets & drive the receiver front-end into non-linear region

## Example: Wireless LAN 802.11b & Bluetooth



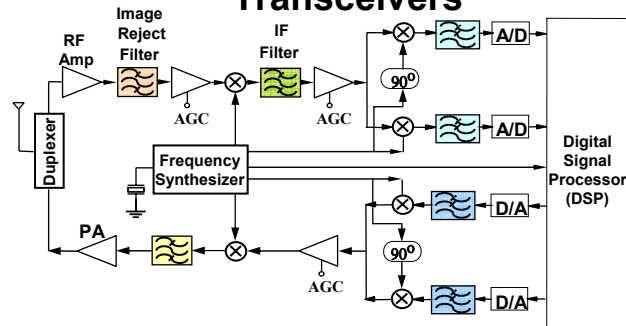
Ref: H. Darabi, et al, "A Dual Mode 802.11b/Bluetooth Radio in 0.35um CMOS," *IEEE ISSCC*, 2003 pp. 86-87.



- EECS 247 Lecture 27:

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## Analog Filters in Super-Heterodyne Wireless Transceivers



### Filters

RF Filter  
IF Filter  
Base-band Filters

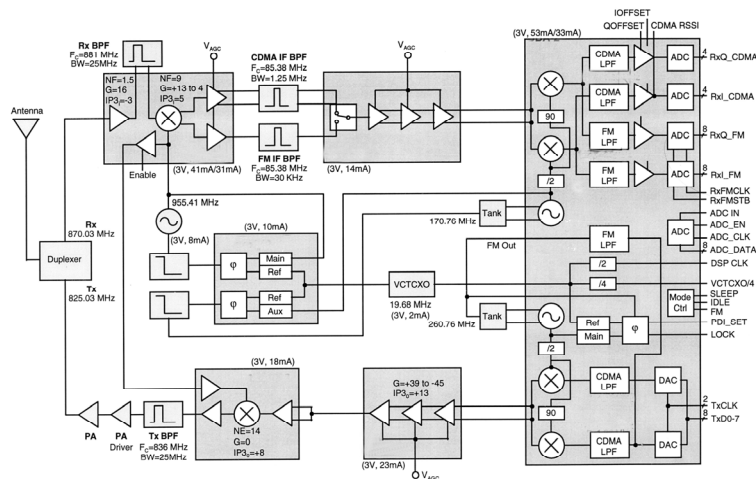
### Function

Image Rejection  
Channel selection  
Channel Selection  
& Anti-aliasing for ADC

### Type

Ceramic or LC  
SAW  
Integrated Cont.-Time  
or S.C.

## Example: Dual Mode CDMA (IS95)& Analog Cellular Phone



## Example: Dual Mode CDMA (IS95)& Analog Cellular Phone

- Baseband analog circuitry includes:
  - CDMA
    - 4bit flash type ADC clock rate 10MHz
    - 8bit segmented TX DAC clock rate 10MHz (shared with FM)
    - 7<sup>th</sup> order elliptic RX lowpass filter corner freq. 650kHz
    - 3<sup>rd</sup> order chebyshev TX lowpass filter corner freq. 650kHz
  - FM (analog)
    - 8bit successive approximation ADCs clock rate 360kHz
    - 5<sup>th</sup> order chebyshev RX lowpass filter corner frequency 14kHz
    - 3<sup>rd</sup> order butterworth TX lowpass filter corner frequency 27kHz

## Summary

- Examples of systems utilizing challenging analog to digital interface circuitry- in the area of wireline & wireless systems discussed
- Analog circuits still remain the interface  
→ connecting the digital world to the real world!