

Maintaining Robust Stability and Performance through Sampling and Quantization

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Outline

- ① Robust Control Context
- ② Notations
- ③ Key Questions
- ④ Proposed Method Summary
- ⑤ Conclusions

Robust Control: μ -Synthesis [I]

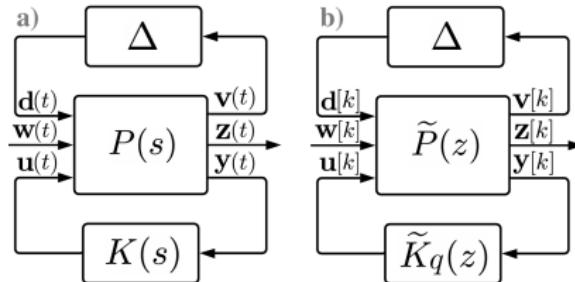
- Continuous-time augmented plant P and $P-K-\Delta$ connection;

$$P : \begin{pmatrix} \dot{\mathbf{x}}(t) \\ \mathbf{v}(t) \\ \mathbf{z}(t) \\ \mathbf{y}(t) \end{pmatrix} = \begin{pmatrix} A & B_d & B_w & B_u \\ C_v & D_{vd} & D_{vw} & D_{vu} \\ C_z & D_{zd} & D_{zw} & D_{zu} \\ C_y & D_{yd} & D_{yw} & D_{yu} \end{pmatrix} \begin{pmatrix} \mathbf{x}(t) \\ \mathbf{d}(t) \\ \mathbf{w}(t) \\ \mathbf{u}(t) \end{pmatrix}; \quad (1)$$

$$\Delta = \{\text{diag}(\Delta_1, \dots, \Delta_f, \delta_1 I_{n_1}, \dots, \delta_s I_{n_s})\}; \quad (2)$$

$$K = (A_K, B_K, C_K, D_K) \in \mathcal{G}. \quad (3)$$

- Sampled with fixed period** $\tau > 0 \Rightarrow$ discretized block equivalents: $\tilde{P}-\tilde{K}-\tilde{\Delta}$;
- Quantization with a resolution** $q > 0$ applied to \tilde{K} further symbolized by \tilde{K}_q .



Notations

- Nominal process model: $G_n = (A, B_u, C_y, D_{yu})$;
- Arbitrary uncertainty block $\Delta \in \Delta$, $\|\Delta\|_\infty \leq 1$;
- Uncertain plant $G_\Delta \equiv G$; nominal plant model becomes $G_n = G_\Delta|_{\Delta=0}$;
- Mapping \mathcal{T} between G_n and G :

$$\mathcal{T} : \mathcal{G} \times \mathcal{G} \rightarrow \mathcal{G}, \quad G = \mathcal{T}(G_n, U), \quad \Delta \in \Delta, \quad \|\Delta\|_\infty \leq 1. \quad (4)$$

- Mapping \mathcal{A} between G and P :

$$\mathcal{A} : \mathcal{G} \times \mathcal{G} \rightarrow \mathcal{G}, \quad P = \mathcal{A}(G, W). \quad (5)$$

Example

SISO nominal plant $G_n = G_{n_1} G_{n_2}$ with:

- multiplicative and additive uncertainties, $U = [U_1 \quad U_2]$, i.e.
 $G = G_{n_1} (1 + U_1 \Delta_1) (G_{n_2} + U_2 \Delta_2)$, $\Delta = \text{diag}([\Delta_1 \quad \Delta_2])$, $\|\Delta\|_\infty \leq 1$;
- augmented plant model using the closed-loop sensitivity weight $W = W_S$ as
 $P = [-W_S G \quad W_S]$.

Robust Control: μ -Synthesis [II]

Structured singular value (SSV):

$$\mu_{\Delta}(\text{LLFT}(P, K)) = \sup_{\omega \in \mathbb{R}_+} \frac{1}{\min_{\Delta \in \Delta} \{\bar{\sigma}(\Delta), \det(I - M_{\omega}\Delta) = 0\}}, \quad (6)$$

with $M_{\omega} = \text{LLFT}(P, K)(j\omega)$.

μ -synthesis problem:

$$K \text{ stabilizing s.t. } \mu_{\Delta}(\text{LLFT}(P, K)) < 1. \quad (7)$$

Assumed closed-loop properties imposed through $K(s)$:

- **robust stability (RS);**
- **robust performance (RP).**

Discretization

The numeric regulator \tilde{K}_q is obtained through:

- **sampling with a fixed period** $\tau \in \mathbb{R}_+$, followed by a discretization of K ;

$$\mathcal{D} : \mathcal{G} \times \mathbb{R}_+ \rightarrow \mathcal{G}_D, \quad \tilde{K} = \mathcal{D}\{K, \tau\}. \quad (8)$$

- **uniform quantization** of its coefficients with a fixed step $q \in \mathbb{R}_+$:

$$\mathcal{Q} : \mathcal{G}_D \times \mathbb{R}_+ \rightarrow \mathcal{G}_D, \quad \tilde{K}_q = \mathcal{Q}\{\tilde{K}, q\}. \quad (9)$$

Denote $\xi = (\tau, q) \in \mathbb{R}_+^2$ encompassed into a single two-dimensional variable.

Example

For $K = (A_K, B_K, C_K, D_K) \in \mathcal{G}$:

- discretization using the Tustin method, leading to $\tilde{K} = (A_d, B_d, C_d, D_d) \in \mathcal{G}_D$;
- rounding (midtread) quantizer:

$$\tilde{K}_q = \left(q \left\lfloor \frac{A_d}{q} \right\rfloor, q \left\lfloor \frac{B_d}{q} \right\rfloor, q \left\lfloor \frac{C_d}{q} \right\rfloor, q \left\lfloor \frac{D_d}{q} \right\rfloor \right) \in \mathcal{G}_D. \quad (10)$$

Key Questions

- ① How to guarantee that $\tilde{K}_q \in \mathcal{G}_D$ still maintains RS and RP?

- ② How to select such $\tilde{K}_q \in \mathcal{G}_D$ configurations?

Contributions

Lemma

Given the continuous models G_n , U , W , P and mappings \mathcal{T} , \mathcal{A} from (4) and (5), the discrete augmented plant counterpart \tilde{P} can be computed based on the discretization of its individual components \tilde{G}_n , \tilde{U} , \tilde{W} through $\mathcal{D}\{\cdot, \tau\}$, $\tau > 0$. Moreover, the block Δ is invariant from the continuous domain to the discrete domain. As such:

$$\tilde{G} = \mathcal{T}(\mathcal{D}\{G_n, \tau\}, \mathcal{D}\{U, \tau\}), \quad \Delta \in \Delta; \quad (11)$$

$$\tilde{P} = \mathcal{D}\{P, \tau\} = \mathcal{A}(\tilde{G}, \mathcal{D}\{W, \tau\}). \quad (12)$$

Main advantage: Reuse the models and weights from the continuous to discrete case.
Reformulate the μ -synthesis problem in terms of the equivalent discrete-time models:

$$\mu_{\Delta}(\text{LLFT}(\tilde{P}, \tilde{K}_q)) = \sup_{\omega \in \Omega_N} \frac{1}{\min_{\Delta \in \Delta} \{\bar{\sigma}(\Delta), \det(I - \tilde{M}_{\omega} \Delta) = 0\}} < 1 \text{ (mandatory constraint)},$$

with $\tilde{M}_{\omega} = \text{LLFT}(\tilde{P}, \tilde{K}_q)(e^{j\omega\tau})$ and domain $\Omega_N = [0, \omega_N]$, where $\omega_N = \pi/\tau$ is the Nyquist frequency for the period τ .

Optimization functionals

In addition to the mandatory SSV constraint, we are left with a separate degree of freedom to select the pairs $\xi = (\tau, q) \in \mathbb{R}_+^2$. Two dichotomic approaches are suggested:

- **Implementability functional:**

$$\min_{\xi \in \mathbb{R}_+^2} f_1(\xi) = -\xi_1 \xi_2. \quad (13)$$

- **Fidelity functional:**

$$\min_{\xi \in \mathbb{R}_+^2} f_2(\xi) = \mathcal{I}(\xi), \quad (14)$$

$$\mathcal{I}(\xi) = \int_{\Omega} \left| \bar{\sigma}(K) - \bar{\sigma}(\tilde{K}_q) \right| (1 + \|\nabla^2 K\|) d\omega. \quad (15)$$

Hands-On Summary

Preconditions:

- ① establish $G_n, \Delta, U \rightarrow \mathcal{T}, W \rightarrow \mathcal{A} \Rightarrow K$;
- ② define discretization methods $\mathcal{D}_p, \mathcal{D}_c$;
- ③ define quantization method \mathcal{Q} for \tilde{K}_q ;
- ④ discretize plant \tilde{P} (Lemma 1);
- ⑤ define cost functional $f : \mathbb{R}_+^2 \rightarrow \mathbb{R}_+$;
- ⑥ define nonlinear constrained optimization problem:

$$\min_{\xi \in \mathbb{R}_+^2} f(\xi) \text{ s.t. } \mu_\Delta(\text{LLFT}(\tilde{P}, \tilde{K}_q)) < 1,$$

\Leftrightarrow the interior-point function, $\rho > 0$:

$$F(\xi) = f(\xi) - \rho \ln \left(1 - \mu_\Delta(\text{LLFT}(\tilde{P}, \tilde{K}_q)) \right).$$

Algorithm 1: Optimal selection of sampling rate and quantization step for a continuous-time regulator K

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Input:  $K, G_n, U, W, \mathcal{T}, \mathcal{A}, \Delta, F \in \{F_1, F_2\}, \Omega, \rho$ 
        Discretization operators  $\mathcal{D}_p \{\cdot, \tau\}, \mathcal{D}_c \{\cdot, \tau\}$ .
        {For the plant and controller models}
        Solver algorithm  $\xi_{k+1} = \Sigma(F, \xi_k)$ .
Output: Optimum  $\xi^* = (\tau^*, q^*) \in \mathbb{R}_+^2$ .
1 Initialize  $\xi \leftarrow \xi_0 = (\tau_0, q_0) \in \mathbb{R}_+^2$ .
2 while stopping criterion is not satisfied do
3    $\tilde{G}_n \leftarrow \mathcal{D}_c \{G_n, \xi_1\}$ .
4    $\tilde{U} \leftarrow \mathcal{D}_c \{U, \xi_1\}; \tilde{W} \leftarrow \mathcal{D}_c \{W, \xi_1\}$ .
5    $\tilde{G} \leftarrow \mathcal{T} \{\tilde{G}_n, \tilde{U}\}; \tilde{P} \leftarrow \mathcal{A} \{\tilde{G}, \tilde{W}\}$  {Lemma 1}.
6    $\tilde{K}_q \leftarrow \mathcal{Q} \{\mathcal{D} \{K, \xi_1\}, \xi_2\}$  {Eq. (9), (10)}.
7   Compute  $\mu_\Delta \left( \text{LLFT} \left( \tilde{P}, \tilde{K}_q \right) \right)$  approximation.
8   Compute  $F(\xi)$  {Eq. (16) or (21)}.
9   Update  $\xi \leftarrow \Sigma(F, \xi)$ .
10  Verify stopping criterion.
11 end

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Numerical Example

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Conclusions and Future Works

Conclusions:

- mathematical tools to quantify the influence of (τ, q) on the transient response of the closed-loop system;
- **smaller (τ, q) pairs do not automatically lead to improved performance;**
- (τ^*, q^*) sensitivity analysis using the condition number.

Future Works:

- general-purpose functional $f(\xi)$ directly tied to hardware specifications (τ – operating frequency; q – data handling capability, i.e. memory and bandwidth);
- extensions on the modelling framework directly for discrete-time systems with uncertainties, i.e. $e^{\frac{\tau}{T}}$, $T > 0$.

Thank you for your attention!

