

UNIVERSITY OF BERN

BACHELOR THESIS

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# Indoor positioning using Raspberry Pi with UWB

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## Declaration of Authorship

I, Mischa WENGER, declare that this thesis titled, “Indoor positioning using Raspberry Pi with UWB” and the work presented in it are my own. I confirm that:

- This work was done wholly or mainly while in candidature for a research degree at this University.
- Where any part of this thesis has previously been submitted for a degree or any other qualification at this University or any other institution, this has been clearly stated.
- Where I have consulted the published work of others, this is always clearly attributed.
- Where I have quoted from the work of others, the source is always given. With the exception of such quotations, this thesis is entirely my own work.
- I have acknowledged all main sources of help.
- Where the thesis is based on work done by myself jointly with others, I have made clear exactly what was done by others and what I have contributed myself.

Signed:

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Date:

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*“Thanks to my solid academic training, today I can write hundreds of words on virtually any topic without possessing a shred of information, which is how I got a good job in journalism.”*

Dave Barry



UNIVERSITY OF BERN

Faculty Name  
Institute of Computer Science

Bachelor of Science in Computer Science

**Indoor positioning using Raspberry Pi with UWB**

by Mischa WENGER

## *Abstract*

The Thesis Abstract is written here (and usually kept to just this page). The page is kept centered vertically so can expand into the blank space above the title too...

## *Acknowledgements*

The acknowledgments and the people to thank go here, don't forget to include your project advisor...





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# List of Abbreviations

|                |   |
|----------------|---|
| <b>ACK</b>     | <b>ACKnowledgement</b>                            |
| <b>AN</b>      | <b>Anchor Node</b>                                |
| <b>API</b>     | <b>Application Programming Interface</b>          |
| <b>CDS</b>     | <b>Communication Distributed System</b>           |
| <b>GHz</b>     | <b>GigaHertz</b>                                  |
| <b>GPIO</b>    | <b>General Purpose Input Output</b>               |
| <b>GPS</b>     | <b>Global Positioning System</b>                  |
| <b>HMM</b>     | <b>Hidden Markov Model</b>                        |
| <b>IMU</b>     | <b>Inertial Measurement Units</b>                 |
| <b>IoT</b>     | <b>Internet of Things</b>                         |
| <b>kbps</b>    | <b>kilo bit per second</b>                        |
| <b>Mbps</b>    | <b>Mega bit per second</b>                        |
| <b>MCL</b>     | <b>Monte Carlo Localization</b>                   |
| <b>MF</b>      | <b>Magnetic Field</b>                             |
| <b>MHz</b>     | <b>MegaHertz</b>                                  |
| <b>M2M</b>     | <b>Machine 2(to) Machine</b>                      |
| <b>OS</b>      | <b>Operating System</b>                           |
| <b>PRF</b>     | <b>Pulse Repetition Frequency</b>                 |
| <b>RTLS</b>    | <b>Real Time Locating System</b>                  |
| <b>RTT</b>     | <b>Round Trip Time</b>                            |
| <b>RSSI</b>    | <b>Received Signal Strength Indication</b>        |
| <b>SDS-TWR</b> | <b>Symmetrical Double Sided - Two Way Ranging</b> |
| <b>TAG</b>     | <b>TArGet</b>                                     |
| <b>TDOA</b>    | <b>Time Difference Of Arrival</b>                 |
| <b>ToF</b>     | <b>Time of Flight</b>                             |
| <b>TWR</b>     | <b>Two Way Ranging</b>                            |
| <b>UDP</b>     | <b>User Datagram Protocol</b>                     |
| <b>UWB</b>     | <b>Ultra WideBand</b>                             |





# Physical Constants

Speed of Light  $c_0 = 2.997\,924\,58 \times 10^8 \text{ m s}^{-1}$  (exact)



# List of Symbols

|          |                   |                        |
|----------|-------------------|------------------------|
| $a$      | distance          | m                      |
| $P$      | power             | W (J s <sup>-1</sup> ) |
| $\omega$ | angular frequency | rad                    |



*For/Dedicated to/To my...*



# Chapter 1

## Introduction

### 1.1 Motivation

In the last twenty years, the number of mobile devices in use has tremendously increased. In the first quarter of 2018 more than 380 Million smartphones have been sold worldwide [6]. However, in the past few years, not only smartphones have been sold, but also a new market of mobile gadgets and connected devices, summed up as Internet of Things, has evolved. In 2017, more than 20 Billion devices were connected to the internet. Forecasts predict 30 Billion devices in 2020 and already more than 70 Billion in 2025. [14]

This increase in mobile computing has also increased the demand of accurate real-time positioning systems, which led to an active research mainly in indoor positioning system technologies, as there are established solutions for outdoor positioning.

#### 1.1.1 Indoor difficulties vs Outdoor

For outdoor applications, primarily the Global Positioning System (GPS) is in use. For indoor application in the other hand, GPS has limitations that make it almost useless. Due to the environmental conditions indoors, with heavy walls armoured with steel and other distractions, additional signal loss is encountered which makes it hard to detect and decode GPS signals. [7] In addition, higher buildings in the neighborhood can reflect transmitted signals, which leads to false position estimations. As GPS is mainly applied as 2D positioning system, it will not provide 3D indoor information such as the current floor level. For this purposes we are forced to use alternative technologies that provide even higher accuracy indoors than GPS would achieve outdoors. There are many different approaches to do indoor positioning, which made it an attractive and active research field.

#### 1.1.2 Important Applications

There are various possible use cases for devices that track their indoor position. These use cases can be grouped into two groups. In the one hand applications for pedestrians with a smartphone and in the other hand real machine to machine (M2M) applications.

Some examples for Smartphones:

**Location of person in need** For emergency services every second counts to get to the position of persons in need. An accurate positioning system that indicates additional information such as the floor level could save lives.

**Security Guards** Real time tracking of security guards on their patrol. A security system can check autonomous if all security guards are on the right tracks.

**Museum guidance** Tourists visiting a museum could easily be guided through the museum with customized location based information.

Examples for Machine to machine (M2M):

**Logistic** An autonomous storage system can find articles in a big storehouse according to the exact position of the carrier vehicle. Numerous vehicles can be in use at the same time.

**Cleaning** An autonomous cleaning machine keeps track of its position, such that the floor can efficiently be cleaned.

**Indoor post roboter** An autonomous roboter can collect letters in the building and bring them to the internal post office.

## 1.2 Idea

For an object in space, there are several basic ideas to keep track of its current location. We can define a starting position and keep track of every move the device registers. E.g. every visitor in the museum starts at the entrance and will then walk through the building. Alternatively the object can be tracked by defining at least three triangulation points and periodically measure the distance from these points to the device. There are various ways to measure this distance, some with higher and some with lower accuracy.

### 1.2.1 Ranging Positioning System with different Inputs

Our idea was to not only use one of the mentioned approaches, but to combine them to in one alorithm. We would use a range positoning system combined with motion detection of the device and even integrate environmental restrictions, given by floor topologies like walls. By combining different methods we hope to compensate measurement errors and thus minimize the overall errors.

## 1.3 Contributions

In this thesis we present a real-time indoor positioning system on Raspberry Pi based on a particle filter implementation in smartphones, developed in previous works of the University of Bern. [1] We adapted the inputs of the particle filter to range-based localization using ultra wideband (UWB) instead of Wi-Fi and added motions measured by inertial measurement units (IMU) of the target. We expound results of our experiments, where we tested different variants of our implementation and other algorithms in a real test scenario and compared the accuracy of the estimated position.

Our main contributions are:

- We implemented a real-time localization system on raspberry pi using UWB and IMU sensors.
- We created an extensive test scenario, where we placed several anchor nodes in a real building and collected data on complex indoor trajectories.
- We compared the results of our implementation to the results of an UWB based localization system provided by Uniset Company.



## 1.4 Overview

Our work compounds of five remaining chapters: Section 2 provides the theoretical background and related work. Chapter 3 presents the theoretical system design and chapter 4 more specifically explains our system implementation and the test bed. The evaluation of our experiments can be found in section 5. Finally the sixth part concludes the work, where our findings are summarized.



## Chapter 2

# Theoretical Background and Related Work

In this section we explain the different types of range measuring in range-based localization systems. For comparison reasons, we shortly introduce variants of received signal strength indication (RSSI), which is the mostly used indoor localization technique. We then briefly explain two slightly less common methods, which were used in our system - two way ranging (TWR) and time difference of arrival (TDOA). We also include some background theory about our implementation and the particle filter.

These are the main parts of this section:

First a short overview of range based localization with the main principles of RSSI, TWR and TDOA as well as the concept of triangulation/trilateration and the weighting process. Second we present background information about ultra wideband (UWB) and finally information about the particle filter is given.

### 2.1 Range based localization

Range based localization systems are depending on an infrastructure in the area of the localization:

- **Target Node (TAG)** which is the device that is localized.
- **Anchor Nodes (AN)** that are placed on carefully chosen points in the building, to encounter the best coverage of the whole area.



FIGURE 2.1: A simple ranging process.

As shown in figure 2.1, a simplified localization system work as follows: Either the TAG, the anchor or both of them collect data used for localization. The data can be a signal strength, a round trip time or IMU measurement. In a processing unit - on the TAG or on a separate server - the data is processed and converted into a distance. This is repeated for every anchor node. The last step contains trilateration of the position using the ranges of every AN to the TAG.

In this abridged scenario, some difficulties are left out. Full indoor localization systems are more complex, as they use ingenious algorithms to improve the accuracy of the estimated ranges or improve the system by adding weighting to deal with incorrect range measures.

### 2.1.1 Received signal strength indication

As already mentioned, many indoor positioning algorithms use received signal strength indication to calculate distances to the anchor nodes. Mainly because RSSI can be applied to almost every type of transmitted signal thus RSSI uses the universal theory of free-space path loss. The following formula describes the relation between the received signal strength and the distance to the transmitter. [4].

$$P_r = P_t(\lambda/4\pi r)^2$$

$P_r$  - received signal strength

$P_t$  - transmitted signal strength

$\lambda$  - wavelength

$r$  - radius (distance from transmitter to receiver)

However, this formula is restricted to free-space. There are several different kind of distractions that can affect the accuracy of the measurements in an indoor environment. For example the following occurrences:

- Multi-path propagation
- Reflections
- Diffraction
- Doppler effect
- etc.

Signal strength can often be obtained from the transmitter hardware as a discrete number. The higher the number, the stronger the signal. This discretization reflects another source of errors.

### 2.1.2 Round trip time: Two way ranging, time difference of arrival

Gathering round trip times (RTT) is a second method to get distance estimations. For accurate RTT results the hardware of transmitter and receiver, as well as the operating firmware are very important. For the presented two RTT-measuring communication techniques, the key characteristics are either a quick responding time or extremely well synchronized TAG and AN.

Operating in two way ranging (TWR) mode, the TAG sends a message to the ANs and registers the exact time of sending. As soon as the message arrives at its destination, the firmware of the AN instantly captures another timestamp. In an acknowledgement (ACK) message, the timestamp of reception and a timestamp of sending the response is transmitted. When this message arrives at the TAG, again a timestamp is registered. With the equation below, the time of flight can now be evaluated [13]. To achieve higher accuracy, the communication can be extended with a final message containing the timestamps of the requester. This is called symmetrical



FIGURE 2.2: Illustration of TWR and SDS-TWR communication

double-sided two-way ranging (SDS-TWR) [16] and is indicated as optional in Figure 2.2.

Time of flight for TWR:

$$ToF = [(T_{RA} - T_{SP}) - (T_{SA} - T_{RP})]/2$$

Time of flight for SDS-TWR:

$$ToF = [(T_{RA} - T_{SP}) - (T_{SA} - T_{RP}) + (T_{RF} - T_{SA}) - (T_{SF} - T_{RA})]/4$$

$ToF$  - Time of flight

$T_{SP}$  - sending of poll timestamp

$T_{RP}$  - reception of poll timestamp

$T_{SA}$  - sending of ACK timestamp

$T_{RA}$  - reception of ACK timestamp

$T_{SF}$  - sending of final timestamp

$T_{RF}$  - reception of final timestamp

For simplicity reasons normally ranging systems are designed such that the TAG gets the final message of the TWR communication, as the TWR is done with several ANs. In this case the requested information - the distances to every AN - is already on one device and can be further processed. However, the TWR has not necessarily to be initialized by the TAG - the receiver and the sender can easily be exchanged. When for example the computational power of the TAG is limited or the application runs on a separate server, we can imagine some benefits to trigger the TWR in the ANs and forward the collected data directly to the server.

While TWR does not need further synchronization between the devices, time difference of arrival (TDOA) requires a very precise synchronization of the anchor nodes. This is normally done by specifying a master node per three to five anchors. For bigger scenarios often multiple dedicated masters will send clock synchronizations every once in a while, such that every AN gets at least one sync package. It occurs as well that an anchor holds two differently synchronized times. To evaluate the time of flight, a TAG in range will broadcast a blink message. Every AN that receives this blink, will capture a timestamp of the time of arrival (or when holding more than one synctime, capture multiple timestamps). These timestamps are forwarded to the server together with a synch ID and a blink transmitter identity. When a server received at least three timestamps with the same synch ID for a target device, it can perform the position and therefore the distance estimation. A huge benefit of TDOA is the fact, that the tag only needs to send one blink message per timeinterval and will not have to communicate with every AN separately, as in TWR. For TWR the overhead grows enourmous with every anchor and every tag that is added. For TDOA hundrets of TAGs can be tracked, whith only proportional overhead growth and much lower energy consumption for the TAG.

Number of messages sent in one iteration:

TWR:  $3 * n_t * n_{an}$

TDOA:  $n_t$

Where  $n_t$  is the number of TAGs and  $n_{an}$  is the number of anchor nodes. [12]

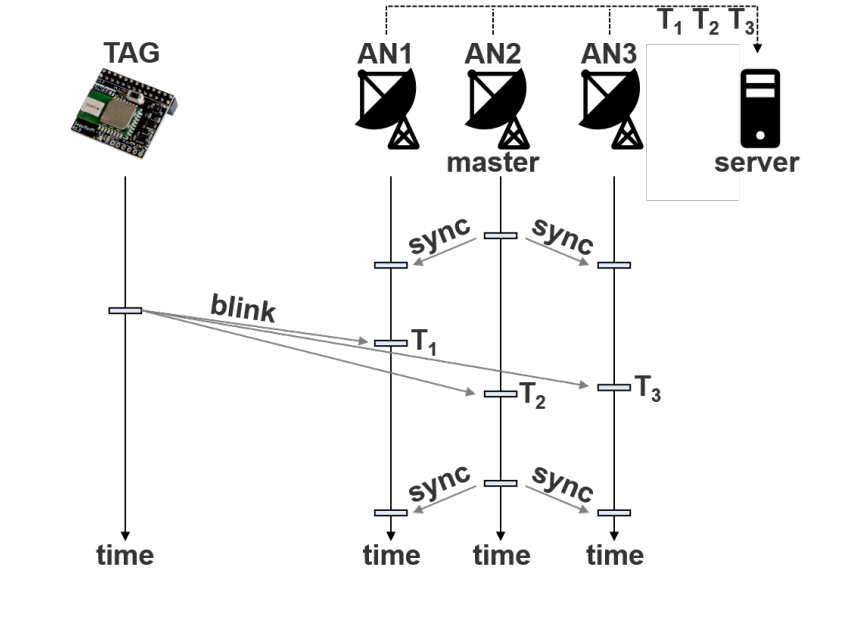


FIGURE 2.3: Typical Setup for Time Difference of Arrival communication

For both methods the same calculations to convert the ToF to the related distance can be applied. We assume that radio signals travel almost with speed of light, so we just multiply the ToF with speed of light ( $c_0$ ).

$$distance = ToF * c_0$$

ToF - Time of flight

$c_0$  - Speed of Light =  $2.997\,924\,58 \times 10^8 \text{ m s}^{-1}$  (exact)

### 2.1.3 Triangulation and Trilateration

Triangulation and trilateration use the mathematical concepts of triangles to find unknown lengths. Triangulation was already mentioned by the greek mathematician Thales, who used this concept for finding out the height of ancient egypt pyramids. [8] It was also used for cartography purposes, where angles between fixed points were measured and heights and distances could be calculated. Although trilateration and triangulation use the same mathematical triangle concept, they have one defined difference: We call it triangulation, when angles to anchor positions are measured, otherwise - when distances to anchors are measured - it's called trilateration. As it was easier to measure angles than distances in the past, triangulation was more often used. With modern electronic devices, it is more common to determine distances, rather than angles.

Figure 2.4 shows how trilateration is used for positioning. With the known distance to every AN, a circle with radius of this distance can be drawn around every AN. These circles do only have one common intersection point, that is where the TAG lies. However, this is a theoretical and idealized scenario, where every range can be determined accurately. In real applications, the ranges are not exactly calculated, what leads to the fact that we will not only get a single point for the calculated position, but several points, especially when we use more than three ANs.

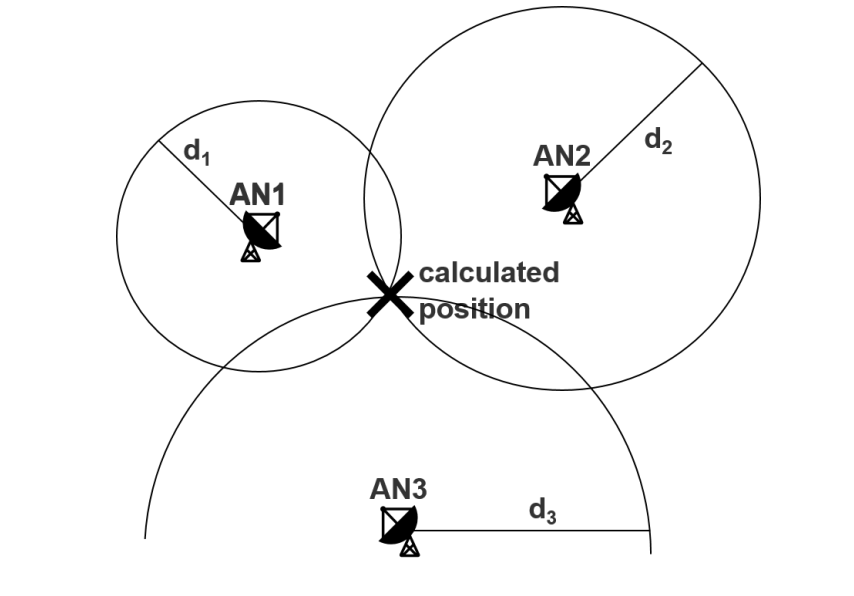


FIGURE 2.4: Graphical illustration of the trilateration concept.

## 2.2 UWB Theory

Ultra wide-band (UWB) is a radio technology in use for military communication, positioning and collecting sensor data. Unlike other communication technologies, UWB occupies a wide area of frequencies instead of just covering a small frequency spectrum. As showed in figure 2.5, UWB spans over a spectrum of more than 500 megahertz (MHz) that lies within the range of 3.1 gigahertz (GHz) and 13.6

GHz. UWB operates with less energy compared to other communication like Wi-

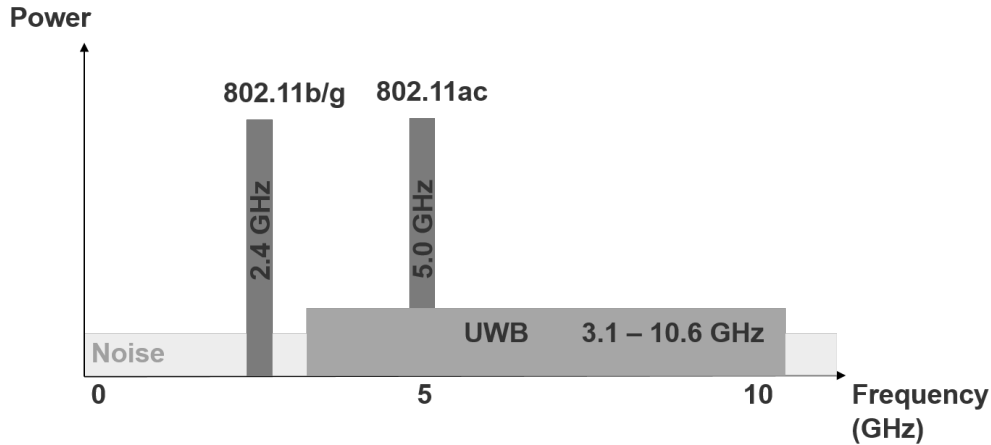


FIGURE 2.5: Comparison of Wi-Fi (802.11) and UWB frequencies.

Fi. However, the main difference between UWB and conventional radio transmissions is the underlying modulation technique. UWB transmits data by generating short radio energy at specific times instead of varying frequency and phase of sinus waves. In addition to the pulse position, the pulses can carry information either by their polarity, their amplitude or by using orthogonal pulses. A single pulse is kept as short as possible, such that more than 100 Million, sometimes even continuous streams with more than 1 Billion pulses per second are generated. As single pulses can be registered and identified by the receiver, UWB devices are able to determine very exact ToFs such that distance estimations can be done to high resolution. Using the large frequency spectrum, there are even methods to overcome multipath propagation, when at least some frequencies have direct line-of-sight trajectories.

## 2.3 Particle Filter

Sed ullamcorper quam eu nisl interdum at interdum enim egestas. Aliquam placerat justo sed lectus lobortis ut porta nisl porttitor. Vestibulum mi dolor, lacinia molestie gravida at, tempus vitae ligula. Donec eget quam sapien, in viverra eros. Donec pellentesque justo a massa fringilla non vestibulum metus vestibulum. Vestibulum in orci quis felis tempor lacinia. Vivamus ornare ultrices facilisis. Ut hendrerit volutpat vulputate. Morbi condimentum venenatis augue, id porta ipsum vulputate in. Curabitur luctus tempus justo. Vestibulum risus lectus, adipiscing nec condimentum quis, condimentum nec nisl. Aliquam dictum sagittis velit sed iaculis. Morbi tristique augue sit amet nulla pulvinar id facilisis ligula mollis. Nam elit libero, tincidunt ut aliquam at, molestie in quam. Aenean rhoncus vehicula hendrerit.



## Chapter 3

# System design

The given background theory of section 2 was applied for our system design. In the following we present our theoretical system and which aspects of the theory led to system design decisions. We provide a short overview in the first part, following a part where our theoretical setup is introduced. Finally we add our thoughts about the computations that are used.

### 3.1 Overview

As already mentioned in the introduction, our system combines different types of data input to achieve the best position estimation. This leads to a slightly more complicated ranging process than indicated on the simplified illustration in figure 2.1. We wanted to include not only measurements from one device, but from several different data sources: TAG and AN collect data and send it to the server, where the data is fed into the particle filter. The particle filter spreads particles according to TAG movement indication, restricted by the topology read from the floorplan and calculates the position likelihood according to range measurements, the movement vector and a rather complex zone indication. In this phase, trilateration is already done implicitly. With respect to the likelihood, a normalized weight is assigned to each particle. The position estimation corresponds to the weighted sum of all particle positions. An overview of the whole process is illustrated on figure 3.1.

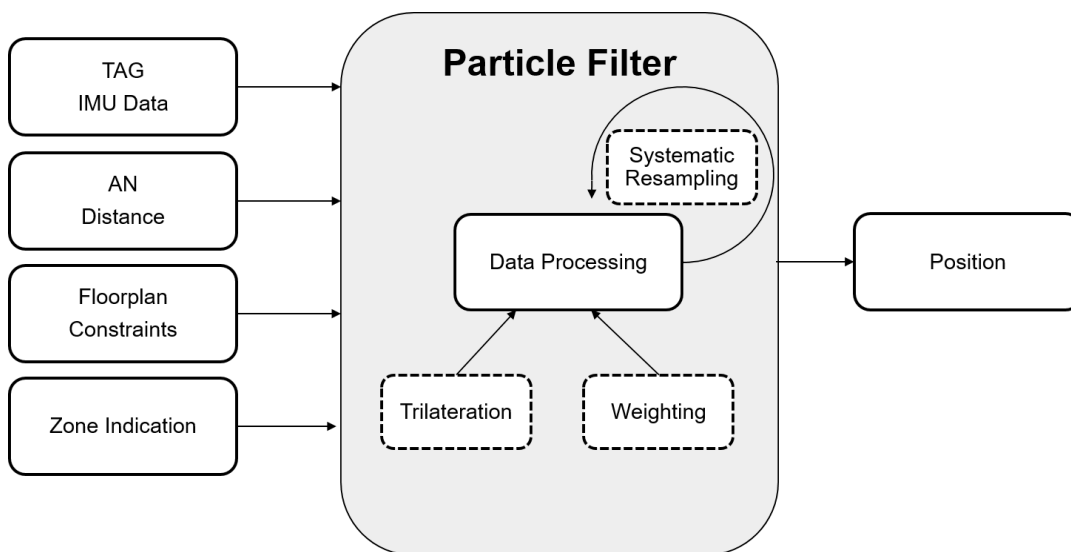


FIGURE 3.1: Overview of the theoretical components in our system.

### 3.2 Setup

The theoretical design on our system is shown in figure 3.2, it works as follows: At least three ANs, equipped with UWB technology, are placed in an indoor environment of one floor. Our algorithm runs on a separate server, where a zoneplan and floorplan of the floor are available. A single TAG is located somewhere on the floor, it is as well equipped with UWB technology, moreover it measures acceleration and magnetic energy with the onboard IMU. The TAG continuously collects data from the IMU and contemporarily waits for a request from the server. The server periodically requests data via UWB from the TAG. As soon as a request is noticed at the TAG, it performs a range estimation to every AN and sends this data together with the continuously collected IMU data to the server. For every estimation period, the server has all the data needed to feed the particle filter.

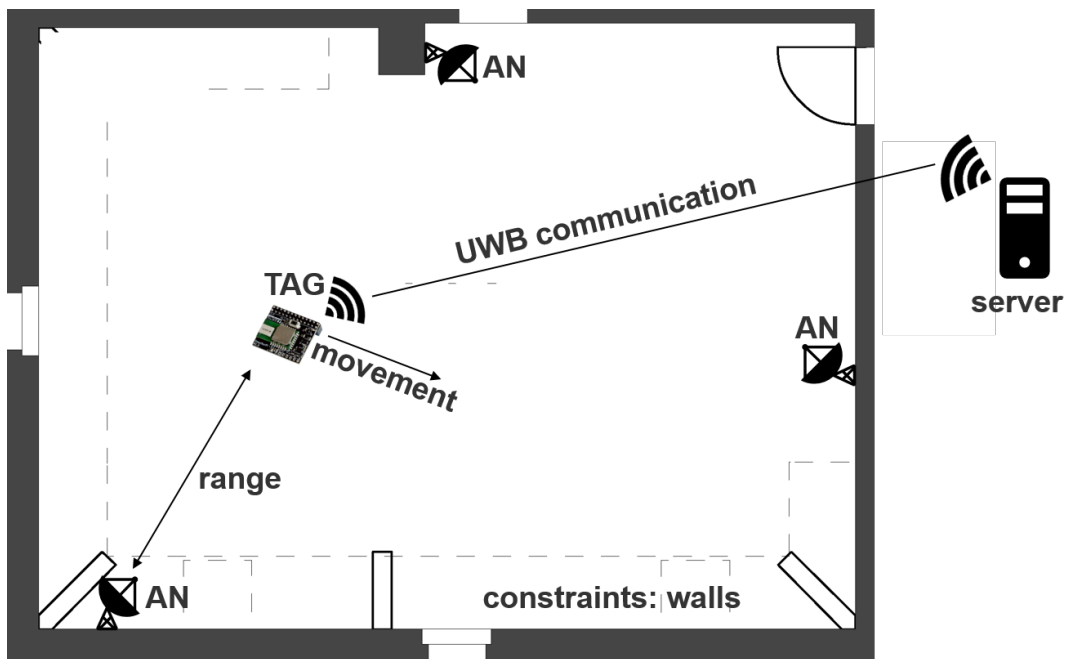


FIGURE 3.2: Overview of the theoretical system design.

### 3.3 Computations on the server

Most computations are done on the separate server, as we would like to minimize the required computational power and energy consumption of the devices. As already mentioned, the server receives IMU data and range estimations from the nodes. This data, as well as the floorplan constraints and the zone indication, flow into the particle filter. These operations are done on the server:

- Spread the particles and validate new positions with floormap constraints
- Evaluate UWB ranges, IMU measures and zone indication to assign likelihood
- Calculate weight function and systematically resample (and reposition) particles with low weights
- Sum up weighted positions

In the first part, every particle is moved from its current to a new position. To validate the new position, we check if the direct trajectory between the old and the new location intersects with an impediment. If so, a new position is generated, as the last was not reachable. The second item covers the likelihood of the new position. For every AN, the measured distance is compared to the distance from the new position to the AN. The less these two distances differ, the higher is the assigned likelihood. This is also in effect for the IMU motion, as the difference between the old and the new position is compared to the measured IMU motion to evaluate the likelihood. **@Mischa: ADD PART OF ZONE INDICATION** In the weighting step, every particle gets weighted by the likelihoods of the previous computations. The weights of all particles are normalized for further processing. The last part is simply a weighted sum of the particles locations.



## Chapter 4

# Implementation and experiment setup

In this section we explain in detail, what hardware we used in our implementation and with which technology the UWB communication and ranging was done. Finally we briefly explain the core of our work, the mathematical theory of the particle filter.

### 4.1 Hardware

We decided to use Raspberry Pis for all mobile devices such as the TAG and the ANs. As the Raspberry Pis were not equipped with UWB technology, we extended them with a SEQUITUR Pi board from UNISSET company running InGPS Lite firmware. In the following two subsections we present you these two hardware components.

#### 4.1.1 Raspberry Pi

A Raspberry Pi is a single-board computer not much bigger than a credit card. Raspberry Pis are mainly designed for educational purposes as an alternative to expensive notebooks or desk computers. Hence the focus lies also on easy-to-use and plug-and-play experiences. Raspberry Pis are useful for versatile types of projects, as they provide common state of the art hardware - like HDMI, USB and wireless LAN - direct on board and as they are extendable with selected components.

We used Raspberry Pi Model B [9], these were the most relevant specifications for our work:

- Quad Core 1.2GHz Broadcom BCM2837 64bit CPU
- 1GB RAM
- 100 Base Ethernet
- 40-pin extended general purpose input output (GPIO)
- Micro SD port for loading your operating system and storing data

#### 4.1.2 SEQUITUR Pi board with InGPS Lite

On the 40-pin extended GPIO, we connected the SEQUITUR Pi board from UNISSET Company. UNISSET is a company located in Italy that focuses on research, development and manufacturing of innovative sensors in two major application areas [15]:

- Access control security systems, enhancing the reliability of intrusion detection

- Indoor and outdoor tracking. Sequitur is a precise real time locating system (RTLS) for tracking any object in 2D or 3D with centimeter accuracy.

This hardware seemed perfect for our ambitions, as it provides a state-of-the-art UWB communication and ranging. Moreover the SEQUITUR Pi board of the TAG has IMU sensors like 3D-accelerometer and 3D-magnetometer on board. Together with the hardware, UNISSET delivers a firmware running on Raspberry Pis operating system (OS) to establish a connection via user datagram protocol (UDP). This firmware allows to communicate with the sensors in order to retrieve IMU sensor data, but also to get direct access to the range between two nodes. It is explained in detail throughout the next section.

## 4.2 UWB Communication and Ranging

The radio module of SEQUITUR Pi board is used on the one hand to transmit data - in order to obviate the need for additional communication hardware - and on the other hand to evaluate the ToF. As UNISSET is a commercial company, they do not provide full information of the underlying transmission techniques. Nonetheless in the two following subsections, the known parts are mentioned.

### 4.2.1 Transmission

SEQUITUR InGPS Lite enables single-hop wireless communication with the UWB interface between neighboring nodes of the same network. The radio module supports six different user-selectable frequency bands between 3.5 GHz and 6.5 GHz. There are six different operation modes to change the spectral occupation, listed in the table below:

| Channel Number | Central Frequency [MHz] | Bandwidth [MHz] |
|----------------|-------------------------|-----------------|
| 1              | 3494.4                  | 500             |
| 2              | 3993.6                  | 500             |
| 3              | 4492.8                  | 500             |
| 4              | 3993.6                  | 1300            |
| 5              | 6489.6                  | 500             |
| 7              | 6489.6                  | 1100            |

The data rate can be changed to three preset values of 110 kilobit per second (kbps), 850 kbps and 6.8 megabit per second (Mbps). All nodes have to operate in the same radiomode to communicate correctly. A lower data rate allows larger operating distances between the nodes. The default pulse repetition frequency (PRF) is assumed to 64 MHz for all the channels. The underlying modulation techniques are not indicated in the specifications. [11] [10]

### 4.2.2 Ranging with TWR

UNISSET company offers two different packages for positioning. InGPS Lite, which is the standard software and InGPS Pro, which is the advanced package. For our implementation InGPS Lite was sufficient, as the main difference of the two packages are the number of TAGs and anchors supported. With InGPS Lite only one TAG and a maximum of 10 anchors are supported as with InGPS Pro numerous TAGs and anchors are possible. InGPS Lite operates only in TWR mode other than InGPS

Pro, where a second mode with TDOA range estimation is available. The range estimation of two nodes is triggered by the application programming interface (API) command *CLIENT\_GET\_RANGE* (50). In our application we sent the command to the anchor in order to minimize the communication of the TAG. The flow of actions related to this API is a even more simplified version of the message exchange indicated in Figure 2.2. In our case, the request message performed by the client starts the TWR conversation via UWB between AN and TAG. The AN sends only one ranging request to the TAG, which immediately responds. By observing the difference between the time instants related to the transmission of the request packet and the reception of the response packet, the AN will directly determine the RTT and thus the range. Finally an answer message with the range is reported from the AN to the client and no messages are reported from the TAG to the client.

### 4.3 Zone Indication: Discriminative Ensemble Learning with Hidden Markov Models

As the zone indication we used is a rather complex and computationally demanding process, we will explain it to higher detail than the other inputs of the particle filter. The zone indication fuses Wi-Fi, UWB, magnetic field (MF) and room transition information in an enhanced learning model. A set of independent individual machine learning methods are combined in an ensemble learning model. The used approach integrates Hidden Markov Model (HMM) with discriminative learning techniques as presented in an earlier work of CDS. [2]

#### 4.3.1 HMM-Discriminative Ensemble Learning Method

In this approach different machine learning algorithms are combined to improve the zone prediction result. Where a zone can be defined as any subarea in the area of interest (e.g. a room). In the concept of Markov localization [3], the state of the system is estimated by the state transitions. For localization, the states of the model correspond to defined zones. Therefore, the HMM is specified by the following components, as in [2]:

- A set of  $n$  states  $Z = \{z_1, z_2, \dots, z_n\}$ , with  $z_i$  as the identifier value of the zone  $i$ . Resulting the discrete random variable  $x_t \in Z$  representing the hidden state at time  $t$ .
- A quadratic matrix  $A$  holding the transition probabilities,

$$A = \begin{pmatrix} a_{1,1} & a_{1,2} & \dots & a_{1,n} \\ a_{2,1} & a_{2,2} & \dots & a_{2,n} \\ \vdots & \vdots & \ddots & \vdots \\ a_{n,1} & a_{n,2} & \dots & a_{n,n} \end{pmatrix},$$

where  $a_{ij}$  represents the probability of moving from zone  $z_i$  to zone  $z_j$ .

- A set of observations  $O$ , defined as:

$$O = \{(o_1, o_2, \dots, o_m)_1, \dots, (o_1, o_2, \dots, o_m)_r\}$$

where  $o_i$  stands for the zone prediction result of the  $i$ -th individual machine learning algorithm. This leads to  $O$  being a set of  $rP_m$  permutations (repetitions allowed), with  $r$  being the number of zones and  $m$  the number of different machine learning algorithms. The machine learning methods have to be conditionally independent for  $y_t \in O$  as the random variable of the observations in time  $t$ .

- A matrix  $B$  holding the emission probabilities of observation likelihoods:

$$B = \begin{pmatrix} b_{1,1} & b_{1,2} & \dots & b_{1,r} \\ b_{2,1} & b_{2,2} & \dots & b_{2,r} \\ \vdots & \vdots & \ddots & \vdots \\ b_{n,1} & b_{n,2} & \dots & b_{n,r} \end{pmatrix},$$

where  $b_{ij}$  represents the probability of  $(o_1, o_2, \dots, o_m)_j$  being the observation generated in zone  $z_i$ .

- An initial probability distribution  $\pi = \pi_1, \pi_2, \dots, \pi_n$  over the  $n$  zones.

How the specified components interact in the HMM can be seen in figure 4.1.

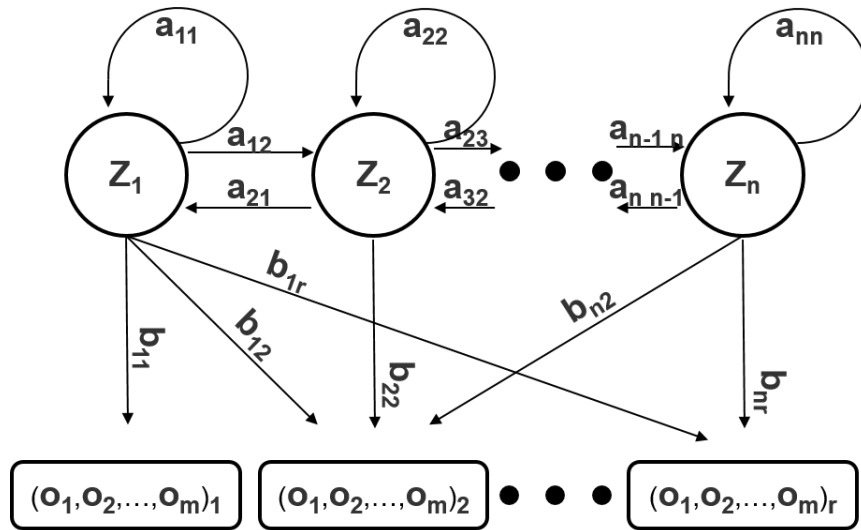


FIGURE 4.1: Application of the different probabilistic parameters in the hidden markov model.

The learning of each individual machine learning method only relies on the fingerprint of Wi-Fi, UWB and magnetic field, that's why prediction errors can still occur. However we can integrate the zone transition information (e.g. some areas are only reachable through other areas) in the HMM to improve the prediction. To determine the sequence of variables that is underlying source of some sequence of observation in a model with hidden variables, a decoding task is necessary. With the HMM model  $\lambda = \{\pi, A, B\}$  and the given observation sequence  $y_{t-i}, \dots, y_{t-1}, y_t$ , the Viterbi algorithm [5] is used to estimate the hidden states sequence  $x_{t-i}, \dots, x_{t-1}, x_t$ .

### 4.3.2 Transition and Emission Probabilities in the HMM

In advance, all zones are exactly defined. Then, the transition probabilities in matrix  $A$  express the likelihood of moving from a beforehand defined zone to another. The



connections between zones, retrieved from the floorplan, predetermine these probabilities. Therefore the values in the  $n \times n$ -matrix  $A = (a_{ij})$  are defined as follows:

$$a_{ij} = P(x_{t+1} = z_j | x_t = z_i)$$

where  $a_{ij}$  represents the transition likelihood from zone  $z_i$  to zone  $z_j$ . Thus,  $\forall i \in \{1, \dots, n\} : \sum_{j=1}^n a_{ij} = 1$ .

The emission probability is the likelihood of observing a particular set of observations  $y_j$  at zone  $z_i$ . Therefore the values of the  $n \times r$ -matrix  $B = (b_{ij})$  can be written as:

$$b_{ij} = P(y_j | z_i), \forall y_j \in O \wedge z_i \in Z,$$

where  $y_j = (o_1, o_2, \dots, o_m)_j$  and  $o_i$  is the zone prediction result from the  $i$ -th machine learning algorithm.

## 4.4 Particle Filter

We used the a particle filter approach to solve the localization problem. This method, also known as Monte Carlo Localization (MCL), is often used for indoor positioning. It combines various noisy measurements to estimate the system state and minimize errors. To introduce the particle filter, we explain in a first paragraph which inputs we fused into the particle filter. The following paragraphs define the different phases in our mathematical model whereas we discuss the different variations of our system in the last subsection.

### 4.4.1 Inputs

As stated above, various measurements are taken into account in our particle filter. The most important inputs are the range estimations between the TAG and ANs, the motion vector measured by the IMU of the TAG and the restrictions given by the floormap. We will refer to  $Zd_t$  as the range observation vector at time  $t$ , which is described as  $Zd_t = [d_t^j], j = 1 \dots N$ , where  $N$  is the number of ANs. Every distance measurement  $d_t^j$  itself is consisting of various errors, statistically it can be described as:

$$d_t^j = \hat{d}_t^j + d_{be,t}^j + \epsilon_{dj,t},$$

where  $\hat{d}_t^j$  is the actual distance to node  $j$ ,  $d_{be,t}^j$  is an environmental bias due to local conditions (obstacles) and  $\epsilon_{dj,t}$  is a measured random error.

The motion vector  $Mv_t = [\theta_t, \ell_t]$  is modeled by the heading direction  $\theta$  and the movement length  $\ell$ . Both of  $\theta$  and  $\ell$  are calculated by the IMU readings, where again several noises occur such that the heading direction is statistically described as:

$$\theta_t = \hat{\theta}_t + \theta_{bs,t} + \theta_{be,t} + \epsilon_{\theta,t},$$

with  $\hat{\theta}_t$  as the actual heading orientation,  $\theta_{bs,t}$  as a sensor bias introduced by uncalibrated sensor readings,  $\theta_{be,t}$  as an environmental angular bias due to magnetic field disturbances and  $\epsilon_{\theta,t}$  as a measured random error. In our implementation we calculated the heading direction with the formula  $\theta = \text{atan}(\frac{mag_x}{mag_y})$ , where  $mag_x$  and  $mag_y$  are low pass filtered magnetometer readings in x and y direction. The update frequency of the sensor was higher than the update frequency of the particle filter, hence we calculated an average of these different measurements, hereafter appearing as  $\theta_t$  for the average during time period from  $t-1$  to  $t$ .

Whereas the heading direction is directly calculated from IMU data, for the stride length we took the previous system state into account. This was necessary, because measured errors propagate over time, so it is almost impossible to use relative quantities (e.g. acceleration) to calculate absolute quantities (e.g. distance) over a longer period of time. That means that the movement length can be defined as:

$$\ell_t = \hat{\ell}_{t-1} + (\hat{a}_t + a_{bs,t} + \epsilon_{a,t}) * \Delta t^2 + \epsilon_{\ell,t},$$

where  $\hat{\ell}_{t-1}$  is the actual movement length of time period t-1,  $\hat{a}_t$  is the actual middle acceleration during time period t,  $a_{bs,t}$  is an other sensor bias due to uncalibrated sensor readings and  $\epsilon_{a,t}$  as well as  $\epsilon_{\ell,t}$  are measured random errors in acceleration and distance respectively. To minimize these errors, we approximated the observed movement length as follows:

$$\ell_t = v_{t-1} * \Delta t + \frac{\sum_{k=0}^N(a_k)}{N} * \Delta t^2$$

where  $v_{t-1}$  is the velocity of the estimated position change in the last system state update,  $a_k$  is the kth acceleration measurement and N the number of discrete acceleration measurements during the time period  $\Delta t$ , which corresponds to the time passed between t-1 and t. As the acceleration sensor - depending on pitch and roll of the device - had a huge non-zero mean noise, we decided to not use the accelerometer data directly, but to use the change in acceleration. To gather the change in acceleration we fed the sensor data into two low pass filters with different parameters, one with a high adaption and one with a low adaption, and only took their difference into account. This corrects a part of the long term sensor bias.

Although there were many sources of errors, for the likelihood calculation in our work we nevertheless used the actual obtained values  $d_t^i$ ,  $\theta_t$  and  $\ell_t$  since the errors are handled in the likelihood model by fusing different data sources and anchor node distances. However, to compensate the bias and error for the particle spreading, we assumed the heading direction  $\theta$  and the stride length  $\ell$  as random normal variables whose values are obtained from  $\mathcal{N}(\theta_t, \sigma_\theta^2)$  and  $\mathcal{N}(\ell_t, \sigma_\ell^2)$ .

#### 4.4.2 Prediction phase

Each particle has a state vector that is defined as follows:

$$X_t = [x_t, y_t, x_{t-1}, y_{t-1}]$$

where  $(x_t, y_t)$  corresponds to the Cartesian coordinates of the particle at time t and  $(x_{t-1}, y_{t-1})$  at time t-1 respectively. In the prediction phase each particle is updated depending of the current movement vector  $Mv_t = [\theta_t, \ell_t]$ . The coordinates of the particle are updated with the following pattern:

$$[x_t, y_t] = [x_{t-1} + \ell_t * \cos(\theta_t), \quad y_{t-1} + \ell_t * \sin(\theta_t)]$$

As mentioned in the last subsection, with  $\ell_t$  and  $\theta_t$  as random normal variables. In the remainder of this work we will also refer to the motion in Cartesian coordinates as  $M_{x,t} = \ell_t * \cos(\theta_t)$  for the motion in x-direction and  $M_{y,t} = \ell_t * \sin(\theta_t)$  for the motion in y-direction. Floorplan restrictions are applied in this phase, whereas movements through walls are not permitted, they lead to another prediction iteration for that particle.

### 4.4.3 Observation phase

In the observation phase an associated weight  $w_t^i$  is recalculated for every particle, since the weight does not anymore correspond to the current position. The weight is updated corresponding to the likelihood of the range observations conditioned on each particle  $p(Zd_t|X_t^i)$  at time  $t$ , respectively the likelihood of the motion observation conditioned on each particle  $p(Mv_t|X_t^i)$  at time  $t$ . Then, the probability is determined as:

$$p(Zd_t|X_t^i) = p(d_t^i|X_t^i)$$

and

$$p(Mv_t|X_t^i) = p(M_{x,t}|X_t^i) * p(M_{y,t}|X_t^i)$$

In order to avoid confusion between different likelihoods used in this work, hereafter we refer to  $p(d_t|X_t^i)$  as the ranging likelihood,  $p(M_t|X_t^i)$  for the motion likelihood and  $p(Z_t|X_t^i)$  as the overall likelihood.

The associated weight  $w_t^i$  of each particle is given by ranging as well as by motion information. A particle at the current position  $(x_t, y_t)$  with low probability to observe  $d_t^i$  in its position will be assigned a small weight. Additionally a particle that moved in x-direction by  $x_t^i - x_{t-1}^i$  with low probability to observe the movement  $M_{x,t}$  will also be assigned a small weight. Analogue for the movement in y-direction. That leads to the fact that particles with large weights will have a stronger effect to the determination of the state of the system. We assume that all these likelihoods - the ranges to each AN as well as the movement in direction x, y - are statistically independent from each other. Therefore, the overall likelihood is defined as:

$$p(Z_t|X_t^i) = \prod_{j=1}^N p(\hat{d}_{j,t}|X_t^i) * p(\hat{M}_{x,t}|X_t^i) * p(\hat{M}_{y,t}|X_t^i)$$

where  $\hat{d}_{j,t}$  is the measured distance to the AN  $j$  at time  $t$  and  $\hat{M}_{x,t}$  is the measured motion in x-direction in timeinterval  $t$ , respectively  $\hat{M}_{y,t}$  in y-direction.

The individual likelihood for the range observation can then be expressed as:

$$p(\hat{d}_{j,t}|X_t^i) = \frac{1}{\sqrt{2\pi\sigma_j^2}} * \exp\left(\frac{-[\sqrt{(x_t^i - x_j)^2 + (y_t^i - y_j)^2} - \hat{d}_{j,t}]^2}{2\sigma_j^2}\right)$$

where  $(x_j, y_j)$  are the known coordinates of the  $j$ th AN. Whereas the individual likelihood of the motion observation in x-direction (analogue for y-direction) is expressed as follows:

$$p(\hat{M}_{x,t}|X_t^i) = \frac{1}{2\pi\sigma_{Mx}^2} * \exp\left(\frac{-[(x_t^i - x_{t-1}^i) - \hat{M}_{x,t}]^2}{2\sigma_{Mx}^2}\right)$$

### 4.4.4 Resampling phase

The resampling phase is an essential component of our particle filter implementation, although it is a computationally expensive step. In the resampling, particles with low assigned weights are repositioned at identical positions as particles with high associated weights. This means, that after the repositioning of the prediction phase and the weight calculation in the observation phase, a resampling in systematic manner is done. This resampling relies on the overall likelihood  $p(Z_t|X_t^i)$ , which

means that every kind of likelihood is taken into account for this step. After repositioning the particles with low weights (and updating their weight), all weights are normalized to obtain in the next step the weighted center of all particles, which corresponds to the estimated position.

#### **4.4.5 Variants**

Variants of the system (one without spreading particles with movementvector and one without likelihood)

### **4.5 Experiment Setup**

#### **4.5.1 Likelihood and weighting**

Here goes the text.

#### **4.5.2 Ranging with TWR**

Here goes the text.

## Appendix A

# Frequently Asked Questions

### A.1 How do I change the colors of links?

The color of links can be changed to your liking using:

```
\hypersetup{urlcolor=red}, or  
\hypersetup{citecolor=green}, or  
\hypersetup{allcolor=blue}.
```

If you want to completely hide the links, you can use:

```
\hypersetup{allcolors=.}, or even better:  
\hypersetup{hidelinks}.
```

If you want to have obvious links in the PDF but not the printed text, use:

```
\hypersetup{colorlinks=false}.
```



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