

Introduction to PIC Programming

Baseline Architecture and Assembly Language

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Lesson 10: Analog-to-Digital Conversion

We saw in the [last lesson](#) how a comparator can be used to respond to an analog signal being above or below a specific threshold. In other cases, the value of the input is important and you need to measure, or *digitise* it, so that your code can process a digital representation of the signal's value.

This lesson explains how to use the analog-to-digital converter (ADC), available on a number of baseline PICs, to read analog inputs, converting them to digital values you can operate on.

To display these values, we'll make use of the 7-segment displays used in [lesson 8](#).

In summary, this lesson covers:

- Using an ADC module to read analog inputs
- Hexadecimal output on 7-segment displays

Analog-to-Digital Converter

The analog-to-digital converter (ADC) on the 16F506 allows you to measure analog input voltages to a resolution of 8 bits. An input of 0 V (or VSS, if VSS is not at 0 V) will read as 0, while an input of VDD corresponds to the full-scale reading of 255.

Three analog input pins are available: AN0, AN1 and AN2. But, since there is only one ADC module, only one input can be read (or *converted*) at once.

The analog-to-digital converter is controlled by the ADCON0 register:

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
ADCON0	ANS1	ANS0	ADCS1	ADCS0	CHS1	CHS0	GO/ $\overline{\text{DONE}}$	ADON

Before a pin can be selected as an input channel for the ADC, it must first be configured as an analog input, using the ANS<1:0> bits:

ANS<1:0>	Pins configured as analog inputs
00	none
01	AN2 only
10	AN0 and AN2
11	AN0, AN1 and AN2

Note that pins cannot be independently configured as analog inputs.

If only one analog input is needed, it has to be AN2. If any analog inputs are configured, AN2 must be one of them.

If only two analog inputs are needed, they must be AN0 and AN2.

By default, following power-on, **ANS<1:0>** is set to '11', configuring AN0, AN1 and AN2 as analog inputs.

This is the default behaviour for all PICs; all pins that can be configured as analog inputs will be configured as analog inputs at power-on, and you must explicitly disable the analog configuration on a pin if you wish to use it for digital I/O. This is because, if a pin is configured as a digital input, it will draw excessive current if the input voltage is not at a digital "high" or "low" level, i.e. somewhere in-between. Thus, the safe, low-current option is to default to analog inputs and to leave it up to your program to only enable digital inputs on those pins known to be digital.

All of the analog inputs can be disabled (enabling digital I/O on the RB0, RB1 and RB2 pins) by clearing **ADCON0**, which clears **ANS<1:0>** to '00'.

The **ADON** bit turns the ADC module on or off: '1' to turn it on, '0' to turn it off. The ADC module is turned on (**ADON** = 1) by default, at power-on.

Note: To minimise power consumption, the ADC module should be turned off before entering sleep mode.

Hence, clearing **ADCON0** will also clear **ADON** to '0', disabling the ADC module, conserving power.

However, disabling the ADC module is not enough to disable the analog inputs; the **ANS<1:0>** bits must be used to configure analog pins for digital I/O, regardless of the value of **ADON**.

The analog-to-digital conversion process is driven by a clock, which is derived from either the processor clock (**FOSC**) or the internal RC oscillator (**INTOSC**). For accurate conversions, the ADC clock rate must be selected such that the ADC conversion clock period, **TAD**, be between 500 ns and 50 µs.

The ADC conversion clock is selected by the **ADCS<1:0>** bits:

ADCS<1:0>	ADC conversion clock
00	FOSC/16
01	FOSC/8
10	FOSC/4
11	INTOSC/4

Note that, if the internal RC oscillator is being used as the processor clock, the **INTOSC/4** and **FOSC/4** options are the same.

But whether you are using a high-speed 20 MHz crystal, a low-power 32 kHz watch crystal, or a low-speed external RC oscillator, the **INTOSC/4** ADC clock option (**ADCS<1:0>** = '11') will always work, giving accurate conversions.

INTOSC/4 is always a safe choice.

Each analog-to-digital conversion requires 13 **TAD** periods to complete.

If you are using the **INTOSC/4** ADC clock option, and the internal RC oscillator is running at 4 MHz, **INTOSC/4** = 1 MHz and **TAD** = 1 µs. Each conversion will then take a total of 13 µs.

If the internal RC oscillator is running at 8 MHz, **TAD** = 500 ns (the shortest period allowed, making this the fastest conversion rate possible), each conversion will take 13 × 500 ns = 6.5 µs.

Having turned on the ADC, selected the ADC conversion clock, and configured the analog input pins, the next step is to select an input (or *channel*) to be converted.

CHS<1:0> selects the ADC channel:

CHS<1:0>	ADC channel
00	analog input AN0
01	analog input AN1
10	analog input AN2
11	0.6V internal voltage reference

Note that, in addition to the three analog input pins, AN0 to AN2, the 0.6V internal voltage reference can be selected as an ADC input channel.

Why measure the 0.6V absolute reference voltage, if it never changes?

That's the point – it never changes (except for some drift with temperature). But if you're not using a regulated power supply (e.g. running direct from batteries), VDD will vary as the power supply changes. Since the full scale range of the ADC is VSS to VDD, your analog measurements are a fraction of VDD and will vary as VDD varies. By regularly measuring (*sampling*) the 0.6 V absolute reference, it is possible to achieve greater measurement accuracy by correcting for changes in VDD. It is also possible to use the 0.6 V reference to indirectly measure VDD, and hence battery voltage, as we'll see in a later example.

Having set up the ADC and selected an input channel to be sampled, the final step is to begin the conversion, by setting the GO/ DONE bit to '1'.

Your code then needs to wait until the `GO/DONE` bit has been cleared to '0', which indicates that the conversion is complete. You can then read the conversion result from the `ADRES` register.

You should copy the result from **ADRES** before beginning the next conversion, so that it isn't overwritten during the conversion process¹.

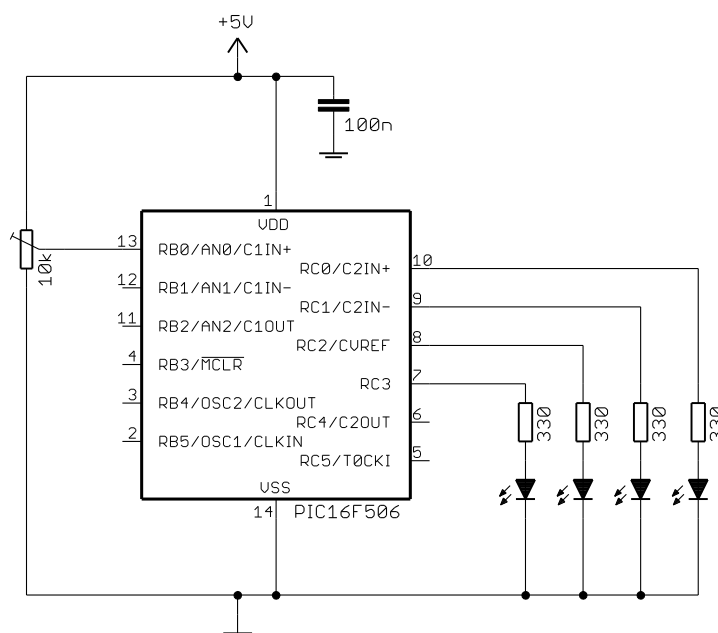
Also for best results, the source impedance of the input being sampled should be no more than 10 k Ω .

Example 1: Binary Output

As a simple demonstration of how to use the ADC, we can use a potentiometer to provide a variable voltage to an analog input, and four LEDs to show a 4-bit binary representation of that value, using the circuit shown on the right.

To implement it using the [Gooligum baseline training board](#), place a shunt across pins 1 and 2 ('POT') of JP24, connecting the 10 kΩ pot (RP2) to AN0, and shunts in JP16-19, enabling the LEDs on RC0-3.

If you are using Microchip's Low Pin Count Demo Board, the onboard pot and LEDs are already connected to AN0 and RC0 – RC3. You only need to ensure that jumpers JP1-5 are closed.



¹ The result actually remains in **ADRES** for the first four TAD periods after the conversion begins. This is the sampling period, and for best results the input signal should not be changing rapidly during this period.

To make the display meaningful (i.e. a binary representation of the input voltage, corresponding to sixteen input levels), the top four bits of the ADC result (in ADRES) should be copied to the four LEDs.

The bottom four bits of the ADC result are thrown away; they are not significant.

To use RC0 and RC1 as digital outputs, we need to disable the C2IN+ and C2IN- inputs, which can be done by disabling comparator 2:

```
clrf    CM2CON0          ; disable comparator 2 -> RC0, RC1 digital
```

This also disables C2OUT, making RC4 available for digital I/O, even though it isn't used in this example.

To use RC2 as a digital output, the CVREF output has to be disabled. By default, on power-up, the voltage reference module is disabled, including the CVREF output. But it doesn't hurt to explicitly disable it as part of your initialisation code:

```
clrf    VRCON            ; disable CVref -> RC2 usable
```

AN0 has to be configured as an analog input. It's not possible to configure AN0 as an analog input without AN2, so for the minimal number of analog inputs, set ANS<1:0> = '10' (AN0 and AN2 analog).

It makes sense to choose INTOSC/4 as the conversion clock (ADCS<1:0> = '11'), as a safe default, although in fact any of the ADC clock settings will work when the processor clock is 4 MHz or 8 MHz.

AN0 has to be selected as the ADC input channel: CHS<1:0> = '00'.

And of course the ADC module has to be enabled: ADON = '1'.

So we have:

```
movlw   b'10110001'      ; configure ADC:
                ; 10-----    AN0, AN2 analog (ANS = 10)
                ; --11----    clock = INTOSC/4 (ADCS = 11)
                ; ----00--    select channel AN0 (CHS = 00)
                ; -----1    turn ADC on (ADON = 1)
movwf   ADCON0           ; -> AN0 ready for sampling
```

Alternatively the 'movlw' could be written as:

```
movlw   b'10'<<ANS0|b'11'<<ADCS0|b'00'<<CHS0|1<<ADON
```

But that's unwieldy, and harder to understand.

Having configured the LED outputs (PORTC) and the ADC, the main loop is quite straightforward:

```
main_loop
    ; sample analog input
    bsf    ADCON0,GO      ; start conversion
w_adc   btfsc ADCON0,NOT_DONE ; wait until done
        goto    w_adc

    ; display result on 4 x LEDs
    swapf  ADRES,w        ; copy high nybble of result
    movwf  LEDS           ; to low nybble of output port (LEDS)

    ; repeat forever
    goto   main_loop
```

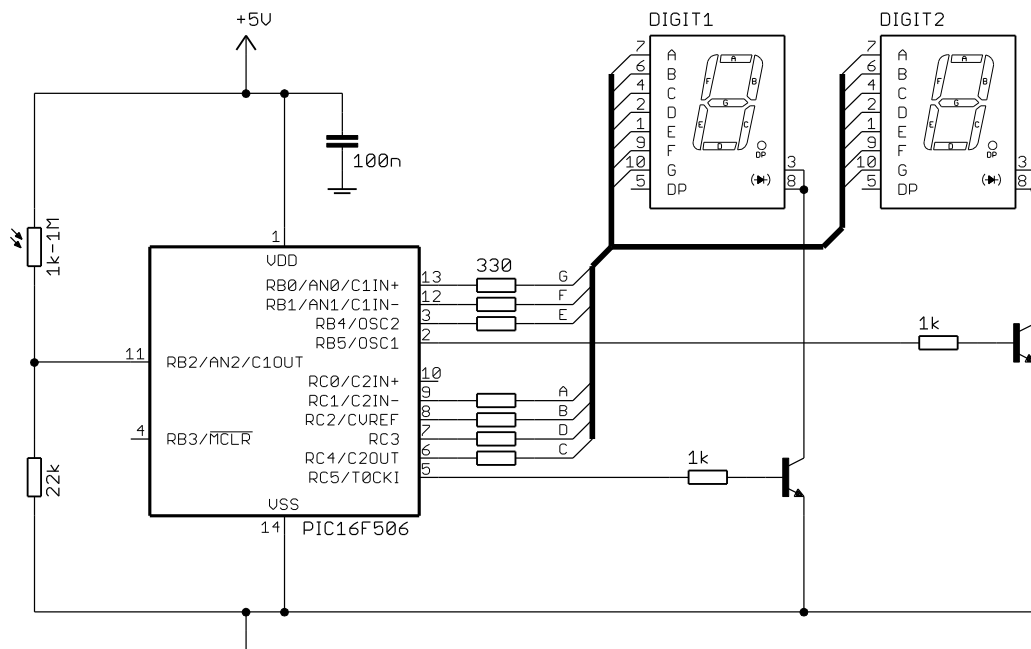
You'll see that two symbols are used for the $\overline{GO/DONE}$ bit, depending on the context: when setting the bit to start the conversion, it is referred to as "GO", but when using it as a flag to check whether the conversion is complete, it is referred to as "NOT_DONE".

Using the appropriate symbol for the context makes the intent of the code clearer, even though both symbols refer the same bit.

Finally, note the use of the 'swapf' instruction. The output bits we need to copy are in the high nybble of ADRES, while the output LEDs (RC0 – RC3) form the low nybble of PORTC, making 'swapf' a neat solution; much shorter than using four right-shifts.

Example 2: Hexadecimal Output

A binary LED display, as in example 1, is not a very useful form of output. To create a more human-readable output, we can modify the multi-digit 7-segment LED circuit from [lesson 8](#) by dropping one digit, and adding a photocell and resistor to supply a voltage that increases with light level (as we saw in [lesson 9](#)), as shown below:



To implement this circuit using the [Gooligum baseline training board](#), place shunts:

- across every position (all six of them) of jumper block JP4, connecting segments A-D, F and G to pins RB0-1 and RC1-4
- in position 1 ('RA/RB4') of JP5, connecting segment E to pin RB4
- across pins 2 and 3 ('RC5') of JP6, connecting digit 1 to the transistor controlled by RC5
- in jumpers JP8 and JP9, connecting pins RC5 and RB5 to their respective transistors
- in position 1 ('AN2') of JP25, connecting photocell PH2 to AN2.

All other shunts should be removed.

If you are using Microchip's Low Pin Count Demo Board, you will need to supply your own display modules, resistors, transistors and photocell, and connect them to the PIC via the 14-pin header on that board, as described in lessons [8](#) and [9](#).

To display a hexadecimal value representing the light level, we can adapt the multiplexed 7-segment display code from [lesson 8](#).

First, to drive the displays using RC1-RC5 and RB0, RB1, RB4 and RB5, we need to disable the comparators, comparator outputs, and voltage reference output:

```
; configure ports
clrw                ; configure PORTB and PORTC as all outputs
tris    PORTB
tris    PORTC
clrf    CM1CON0      ; disable comparator 1 -> RB0, RB1 digital
clrf    CM2CON0      ; disable comparator 2 -> RC0, RC1 digital
clrf    VRCON        ; disable CVref -> RC2 usable
```

To use RB0 and RB1 for digital I/O, it is also necessary to deselect AN0 and AN1 as analog inputs, configuring only AN2 as an analog input with ANS = 01.

Since we are using AN2 as an analog input, we need to select it as the active ADC input with CHS = 10.

So, to configure and select only AN2 as an analog input, we initialise the ADC using:

```
movlw    b'01111001'    ; configure ADC:
                ; 01-----    AN2 (only) analog (ANS = 01)
                ; --11-----    clock = INTOSC/4 (ADCS = 11)
                ; ----10--    select channel AN2 (CHS = 10)
                ; -----1    turn ADC on (ADON = 1)
movwf    ADCON0          ; -> AN2 ready for sampling
```

As we did in [lesson 8](#), the timer is used to provide a ~2 ms tick to drive the display multiplexing:

```
; configure timer
movlw    b'11010111'    ; configure Timer0:
                ; --0-----    timer mode (T0CS = 0) -> RC5 usable
                ; ----0---    prescaler assigned to Timer0 (PSA = 0)
                ; -----111    prescale = 256 (PS = 111)
option    ; -> increment every 256 us
                ; (TMR0<2> cycles every 2.048 ms)
```

This assumes a 4 MHz clock, not 8 MHz, so the configuration directive needs to include ‘_IOSCFS_OFF’:

```
__CONFIG    _MCLRE_ON & _CP_OFF & _WDT_OFF & _IOSCFS_OFF & _IntRC_OSC_RB4EN
```

Note also that, by clearing the T0CS bit, placing the timer in timer mode, the T0CKI input (see [lesson 5](#)) is disabled, making RC5 (which, on the PIC16F506, shares its pin with T0CKI) available as a digital output. So, even if we weren’t using Timer0 in this example, we’d still have to clear T0CS, to make it possible to use RC5 as an output.

The lookup tables have to be extended to include 7-segment representations of the letters ‘A’ to ‘F’, but the lookup code remains the same as in [lesson 8](#).

Since each digit is displayed for 2 ms, and the analog to digital conversion only takes around 13 µs, the ADC read can be completed well within the time spent waiting to begin displaying the next digit (~1 ms), without affecting the display multiplexing.

The main loop, then, simply consists of reading the analog input and displaying each digit of the result, then repeating that quickly enough for the display to appear to be continuous:

```
main_loop
    ; sample input
    bsf      ADCON0,GO          ; start conversion
w_adc      btfsc    ADCON0,NOT_DONE ; wait until conversion complete
           goto     w_adc

           ; display high nybble for 2.048 ms
w10_hi     btfss    TMR0,2          ; wait for TMR0<2> to go high
           goto     w10_hi
           swapf    ADRES,w          ; get "tens" digit
           andlw    0x0F             ; from high nybble of ADC result
           [code to display "tens" digit then wait for TMR<2> to go low goes here]

           ; display ones for 2.048 ms
w1_hi      btfss    TMR0,2          ; wait for TMR0<2> to go high
           goto     w1_hi
           movf     ADRES,w          ; get ones digit
           andlw    0x0F             ; from low nybble of ADC result
           [code to display "ones" digit then wait for TMR<2> to go low goes here]

           ; repeat forever
           goto     main_loop
```

Complete program

Here is the complete “hexadecimal light meter”, so that you can see where and how the various program fragments fit in:

```
;*****
;
;   Description:      Lesson 10, example 2
;
;   Displays ADC output in hexadecimal on 7-segment LED displays
;
;   Continuously samples analog input,
;   displaying result as 2 x hex digits on multiplexed 7-seg displays
;
;*****
;
;   Pin assignments:
;       AN2          = voltage to be measured (e.g. pot or LDR)
;       RB0-1,RB4,RC1-4 = 7-segment display bus (common cathode)
;       RC5          = "tens" digit enable (active high)
;       RB5          = ones digit enable
;
;*****

list      p=16F506
#include   <p16F506.inc>

radix     dec

;***** CONFIGURATION
           ; ext reset, no code protect, no watchdog, 4 MHz int clock
__CONFIG  _MCLRE_ON & _CP_OFF & _WDT_OFF & _IOSCFS_OFF & _IntRC_OSC_RB4EN
```

```

; pin assignments
#define TENS      PORTC,5      ; "tens" digit enable
#define ONES      PORTB,5      ; ones digit enable

;***** VARIABLE DEFINITIONS
        UDATA_SHR
temp     res 1                  ; used by set7seg routine (temp digit store)

;***** RC CALIBRATION
RCCAL    CODE    0x3FF          ; processor reset vector
        res 1                  ; holds internal RC cal value, as a movlw k

;***** RESET VECTOR *****
RESET    CODE    0x000          ; effective reset vector
        movwf    OSCCAL        ; apply internal RC factory calibration
        pagesel  start
        goto     start         ; jump to main code

;***** Subroutine vectors
set7seg  pagesel  set7seg_R
        goto     set7seg_R

;***** MAIN PROGRAM *****
MAIN     CODE

;***** Initialisation
start
        ; configure ports
        clrw                      ; configure PORTB and PORTC as all outputs
        tris     PORTB
        tris     PORTC
        clrf     CM1CON0          ; disable comparator 1 -> RB0, RB1 digital
        clrf     CM2CON0          ; disable comparator 2 -> RC0, RC1 digital
        clrf     VRCON            ; disable CVref -> RC2 usable

        ; configure ADC
        movlw    b'01111001'      ; configure ADC:
        ; 01-----      AN2 (only) analog (ANS = 01)
        ; --11-----      clock = INTOSC/4 (ADCS = 11)
        ; ----10--      select channel AN2 (CHS = 10)
        ; -----1      turn ADC on (ADON = 1)
        movwf    ADCON0          ; -> AN2 ready for sampling

        ; configure timer
        movlw    b'11010111'      ; configure Timer0:
        ; --0-----      timer mode (T0CS = 0) -> RC5 usable
        ; ----0---      prescaler assigned to Timer0 (PSA = 0)
        ; -----111      prescale = 256 (PS = 111)
        option                    ; -> increment every 256 us
                                ; (TMR0<2> cycles every 2.04 8ms)

;***** Main loop
main_loop
        ; sample input
        bsf      ADCON0,GO        ; start conversion
w_adc    btfsc    ADCON0,NOT_DONE ; wait until conversion complete
        goto     w_adc

```



```

        ; display high nybble for 2.048 ms
w10_hi  btfss    TMR0,2          ; wait for TMR0<2> to go high
        goto    w10_hi
        swapf    ADRES,w        ; get "tens" digit
        andlw    0x0F          ; from high nybble of ADC result
        pagesel  set7seg
        call     set7seg        ; then output it
        pagesel  $
        bsf      TENS          ; enable "tens" display
w10_lo  btfsc    TMR0,2          ; wait for TMR0<2> to go low
        goto    w10_lo

        ; display ones for 2.048 ms
w1_hi   btfss    TMR0,2          ; wait for TMR0<2> to go high
        goto    w1_hi
        movf     ADRES,w        ; get ones digit
        andlw    0x0F          ; from low nybble of ADC result
        pagesel  set7seg
        call     set7seg        ; then output it
        pagesel  $
        bsf      ONES          ; enable ones display
w1_lo   btfsc    TMR0,2          ; wait for TMR0<2> to go low
        goto    w1_lo

        ; repeat forever
        goto    main_loop

;***** LOOKUP TABLES *****
TABLES  CODE      0x200          ; locate at beginning of a page

; pattern table for 7 segment display on port B
;   RB4 = E, RB1:0 = FG
get7sB  addwf     PCL,f
        retlw     b'010010'      ; 0
        retlw     b'000000'      ; 1
        retlw     b'010001'      ; 2
        retlw     b'000001'      ; 3
        retlw     b'000011'      ; 4
        retlw     b'000011'      ; 5
        retlw     b'010011'      ; 6
        retlw     b'000000'      ; 7
        retlw     b'010011'      ; 8
        retlw     b'000011'      ; 9
        retlw     b'010011'      ; A
        retlw     b'010011'      ; b
        retlw     b'010010'      ; C
        retlw     b'010001'      ; d
        retlw     b'010011'      ; E
        retlw     b'010011'      ; F

; pattern table for 7 segment display on port C
;   RC4:1 = CDBA
get7sC  addwf     PCL,f
        retlw     b'011110'      ; 0
        retlw     b'010100'      ; 1
        retlw     b'001110'      ; 2
        retlw     b'011110'      ; 3
        retlw     b'010100'      ; 4
        retlw     b'011010'      ; 5
        retlw     b'011010'      ; 6

```

```

    retlw    b'010110'      ; 7
    retlw    b'011110'      ; 8
    retlw    b'011110'      ; 9
    retlw    b'010110'      ; A
    retlw    b'011000'      ; b
    retlw    b'001010'      ; C
    retlw    b'011100'      ; d
    retlw    b'001010'      ; E
    retlw    b'000010'      ; F

; Display digit passed in W on 7-segment display
set7seg_R
    ; disable displays
    clrf     PORTB           ; clear all digit enable lines on PORTB
    clrf     PORTC           ; and PORTC

    ; output digit pattern
    movwf    temp           ; save digit
    call     get7sB          ; lookup pattern for port B
    movwf    PORTB          ; then output it
    movf     temp,w          ; get digit
    call     get7sC          ; then repeat for port C
    movwf    PORTC
    retlw    0

END

```

Of course, most people are more comfortable with a decimal output, perhaps 0-99, instead of hexadecimal.

And you'll find, if you build this as a light meter, using an LDR (CdS photocell), that although the output is quite stable when lit by daylight, the least significant digit jitters badly when the LDR is lit by incandescent and, in particular, fluorescent lighting. This is because these lights flicker at 50 or 60 Hz (depending on where you live); too quickly for your eyes to detect, but not too fast for this light meter to react to, since it is sampling and updating the display 244 times per second.

So some obvious improvements to the design would be to scale and display the output as 0-99 in decimal, and to smooth or filter high-frequency noise, such as that caused by fluorescent lighting.

We'll make those improvements in [lesson 11](#). But first we'll look at one last example.

Example 3: Measuring Supply Voltage

As mentioned above, the 0.6 V absolute voltage reference can be sampled by the ADC, and this provides a way to infer the supply voltage (actually $V_{DD} - V_{SS}$, but to keep this simple we'll assume $V_{SS} = 0$ V).

Assuming that $V_{DD} = 5.0$ V and $V_{SS} = 0$ V, the 0.6 V reference should read as:

$$0.6 \text{ V} \div 5.0 \text{ V} \times 255 = 30$$

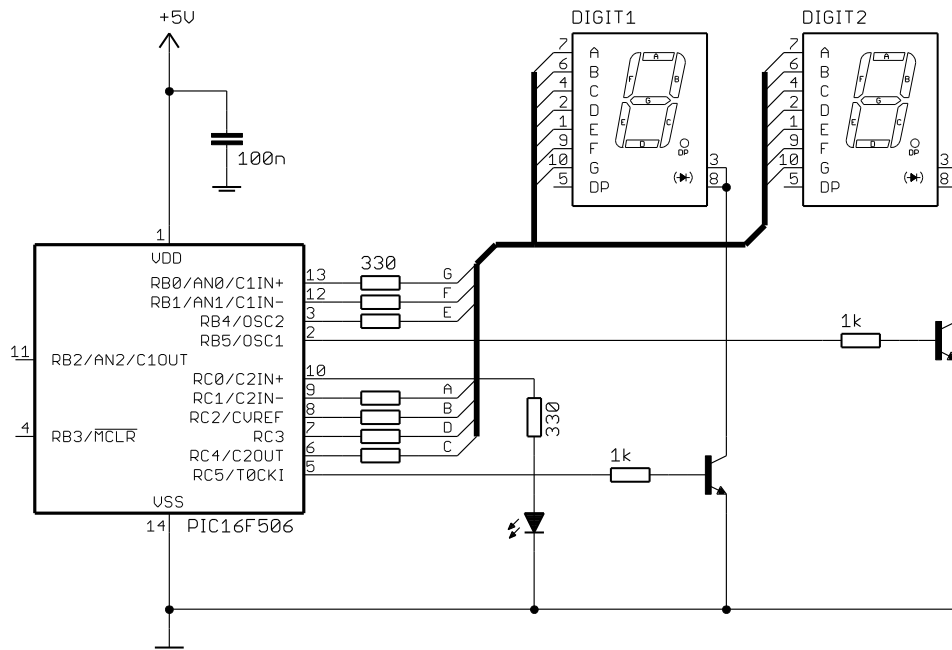
Now if V_{DD} was to fall to, say, 3.5 V, the 0.6 V reference will read as:

$$0.6 \text{ V} \div 3.5 \text{ V} \times 255 = 43$$

As V_{DD} falls, the 0.6 V reference will give a larger ADC result, since it remains constant as V_{DD} decreases.

So to check for the power supply falling too low, the value returned by sampling the 0.6 V reference can be compared with a threshold. For example, a value above 43 indicates that $V_{DD} < 3.5$ V, and perhaps a warning should be displayed, or the device shut down before power falls too low.

To illustrate this, we can use adapt the circuit and program from example 2, displaying the ADC reading corresponding to the 0.6 V reference as two hex digits, and light a “low voltage warning” LED attached to RC0 (as shown below) if VDD falls below some threshold.



If you are using the [Gooligum baseline training board](#), you should set it up as in the last example, but remove the shunt from JP25 (disconnecting the photocell from AN2) and close JP16 (connecting the LED on RC0).

To implement this low voltage warning, the code from example 2 can be used with very little modification.

To make the code easier to maintain, we can define the voltage threshold as a constant:

```
constant MINVDD=3500           ; Minimum Vdd (in mV)
constant VRMAX=255*600/MINVDD  ; Threshold for 0.6 V ref measurement
```

Note that, because MPASM only supports integer expressions, “MINVDD” has to be expressed in millivolts instead of volts (so that fractions of a volt can be specified).

The initialisation code remains the same, except that the ADC configuration is changed to disable all the analog inputs and selecting the internal 0.6 V reference as the ADC input channel:

```
movlw    b'00111101'      ; configure ADC:
; 00-----                no analog inputs (ANS = 00) -> RB0-2 digital
; --11----                clock = INTOSC/4 (ADCS = 11)
; ----11--                select 0.6 V reference (CHS = 11)
; -----1                turn ADC on (ADON = 1)
movwf    ADCON0            ; -> 0.6 V reference ready for sampling
```

After sampling the 0.6 V input, we can test for VDD being too low by comparing the conversion result (ADRES) with the threshold (VRMAX):

```

; sample 0.6 V reference
bsf      ADCON0,GO          ; start conversion
w_adc    btfsc    ADCON0,NOT_DONE ; wait until conversion complete
          goto     w_adc

```

```

; test for low Vdd (measured 0.6 V > threshold)
movlw    VRMAX
subwf    ADRES,w          ; if ADRES > VRMAX
btfsc    STATUS,C
bsf      WARN             ; turn on warning LED

; display high nybble for 2.048 ms
[wait for TMR0<2> high then display "tens" digit]

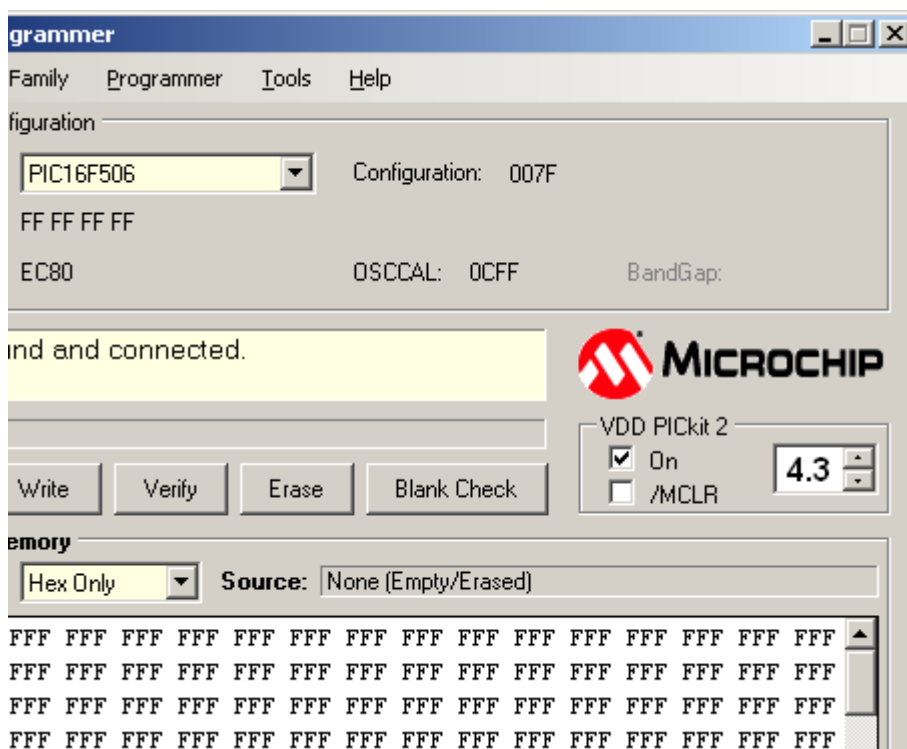
```

There's a slight problem with this approach: when a digit is displayed, the LED on RC0 will be extinguished, since the lookup table for PORTC always returns a '0' for bit 0. To avoid that problem, the 'set7seg_R' routine would need to be modified to include logical masking operations so that RC0 is not overwritten. But it's not really a significant problem; the LED will remain lit for ~1 ms, while the "display high nybble" routine waits for TMR0<2> to go high, out of a total multiplex cycle time of ~4 ms. That is, when the LED is 'lit', it will actually be on for ~25% of the time, and that's enough to make it visible.

To test this application, you need to be able to vary VDD.

If you are using a PICKit 2 or PICKit 3 to power your circuit, you can use the standalone PICKit 2 or PICKit 3 programming application (each downloadable from www.microchip.com) to vary VDD, while the circuit is powered. But first, you should exit MPLAB, so that you don't have two applications trying to control the PICKit 2 at once.

Although the PICKit 2 Programmer application is shown below, the PICKit 3 version looks almost identical, and the method for varying target power (VDD) is the same for both.



In the programmer application, select the Baseline device family, then the PIC16F506 device and then click 'On', as illustrated.

Your circuit should now be powered on, and, assuming the supply voltage is 5.0 V, the display should show '1E' (hexadecimal for 30), or something close to that.

You can now start to decrease VDD, by clicking on the down arrow next to the voltage display, 0.1 V at a time.

When you get to 3.5 V, the display should read '2b' (hex for 43) – but note that the PICKit 2 does not deliver as accurate a voltage as the PICKit 3, so if you are using a PICKit 2, you may see different values at "3.5 V".

When the voltage is low enough for the display to read '2b', the warning LED should light.

Conclusion

We've seen that it's relatively simple to setup and use the ADC on the PIC16F506, whether for the usual purpose of reading an analog quantity, or even to infer the PIC's supply voltage.

However, as mentioned earlier, the light meter project would be more useful if the output was converted to a range of 0-99 and displayed in decimal, and if the results were filtered to smooth out short term fluctuations.

The [next lesson](#) will complete our overview of baseline PIC assembler, by demonstrating how to perform some simple arithmetic operations, including moving averages and working with arrays, to implement these suggested improvements.