

MicroBooNE

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ABSTRACT: This paper describes the design and construction of the MicroBooNE liquid argon time projection chamber and associated systems. Details of design specifications, assembly procedures, and acceptance tests are reported.

KEYWORDS: Time projection chambers; Noble-liquid detectors; Data analysis.

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¹ Date: June 21, 2016; Paper Version: 1.0

1. Introduction and Physics Motivation

2 Neutrino detection is entering a new era with a number of large precision neutrino detectors, specifically
3 Liquid Argon Time Projection Chambers (LArTPCs), now in design and construction both for
4 long and short baseline neutrino oscillation experiments. The **Micro Booster Neutrino Experiment**,
5 henceforth referred to as MicroBooNE, will be the first large (~ 100 tons) LArTPC to operate in the
6 United States and the second only after ICARUS to operate worldwide. LArTPC detectors com-
7 bine fine-grained tracking with total absorption calorimetry to provide excellent signal efficiency
8 and background rejection.

9 The MicroBooNE experiment combines physics goals of short-baseline oscillations and neu-
10 trino cross-section measurements with development goals to inform larger scale construction of
11 LArTPCs for the long-baseline program. MicroBooNE's principal physics goal is to address long
12 standing unsolved puzzles generated by short-baseline neutrino oscillation results from past exper-
13 iments. The LSND collaboration measured an oscillation rate of antimuon neutrinos to antielectron
14 neutrinos not compatible with the solar or atmospheric oscillation parameters [?]. The size of the
15 observed signal suggests a mass squared difference of neutrino masses of the order of about 1 eV,
16 which could only be accommodated with an additional neutrino mass eigenstate. Since the weak
17 interaction only couples to three neutrino eigenstates, an additional neutrino is called sterile, i.e.
18 not coupling to other fermions except via the oscillation mechanism.

19 Other hints come from a re-analysis of the expected reactor antineutrino flux that suggests a
20 deficit of reactor neutrinos [?], and the analysis of calibration data taken by GALLEX and SAGE
21 [?]. The MiniBooNE experiment [?] also investigated (anti)muon- to (anti)electron-neutrino os-
22 cillations and found a further significant deviation from the expectation in the number of electron
23 (anti)neutrino candidates at low energy. The excess of electromagnetic-like events can be attributed
24 to electrons or single photons since these are indistinguishable in the MiniBooNE Cerenkov imag-
25 ing detector.

26 While each of the above results taken alone does not reach the statistical significance to claim
27 a discovery, combining them together can provide indication for oscillations or other unexpected
28 phenomena. Definitive evidence for sterile neutrinos would constitute a revolutionary discovery,
29 with strong implications for particle physics as well as for cosmology. Proposals to address these
30 signals by employing reactor, accelerator, and radioactive source experiments are in the planning
31 stages or underway worldwide.

32 MicroBooNE will address this issue at Fermi National Accelerator Laboratory (Fermilab)
33 with the same beam and baseline as MiniBooNE but by employing the LArTPC technology to
34 efficiently separate electrons from photons. In addition to MicroBooNE's signature oscillation
35 analyses, a suite of precision cross-section measurements will be performed, critical both for fu-
36 ture LArTPC oscillation experiments and for what can be learned about neutrino interactions in
37 general. In the energy range of interest, multiple interaction processes (quasi-elastic, resonances,
38 deep inelastic scattering) are possible, and complicated nuclear effects in neutrino interactions on
39 argon result in a variety of final states. These can range from the emission of several nucleons
40 to more complex topologies with multiple pions, all in addition to the leading lepton in charged-
41 current events. The LArTPC technology employed by MicroBooNE is particularly well suited to
42 this purpose because of its excellent particle identification capability and calorimetric energy re-

1 construction down to very low detection thresholds. MicroBooNE will also be supernova live and
2 capable of detecting a galactic supernova, and prepare the analysis of the search for proton decays.
3 For the latter analysis the experiment will not have a competitive sensitivity due to too small target
4 mass. However, this will be an important scientific goal towards future measurements in larger
5 detectors.

6 On a longer time scale, a phased program is being pursued at Fermilab centered around the
7 existing Booster Neutrino Beam (BNB) and the Neutrinos from the Main Injector (NuMI) beam.
8 The MicroBooNE experiment is located 470 m from the BNB production target and 600 m from
9 the NuMI production target. MicroBooNE began operations in late 2015 for a \sim 3 year time period,
10 extending into the era of the full Short Baseline Neutrino (SBN) program beginning in 2018 with
11 the commencement of operations of the SBND [?] and ICARUS experiments, located at 100m and
12 600m respectively from the BNB production target. MicroBooNE will definitively address whether
13 or not the MiniBooNE low energy excess in neutrino mode is due to electrons or photons. SBND
14 will look for this low energy excess at the near location (100m) from the production target, and the
15 three detectors combined will cover the entire LSND allowed region in neutrino parameter space
16 to 5σ through ν_e appearance.

17 As mentioned above, the LArTPC technique will be used by MicroBooNE at a relatively
18 large scale, so far only surpassed by the ICARUS T600 detector. The potential of liquified noble
19 gases to be employed as the detection media in particle detectors with high spatial resolution was
20 recognized already in the 1960's (see, e.g., Ref. [1]). Large calorimeters for the measurement of
21 particle energy were then realized by using cryogenic noble liquids as active components (see, e.g.,
22 Ref. [2]). Furthermore, the LArTPC was proposed at CERN by C. Rubbia in 1977 [3] as a detector
23 capable of providing uniform and precise imaging of large volumes.

24 The original LArTPC idea was extensively developed by an intense ICARUS R&D program.
25 Various prototype detectors were realized and operated to develop the systems needed for larger
26 physics experiments such as purification methods, signal readout electronics, and cryogenics. The
27 largest ICARUS test device had a mass of about 3 tons of LAr [4, 5] and allowed for the collection
28 of large samples of cosmic-ray events. In parallel, a smaller volume detector (50 liters of LAr) was
29 exposed to the CERN WANF neutrino beam allowing the detection of neutrino interaction events
30 for the first time in a LArTPC [6].

31 The largest LArTPC detector built so far is the 600-ton ICARUS T600 detector, which was
32 also tested with cosmic rays at the surface [7]. Measurements from this activity showed that the
33 technique can successfully operate at large scales with a drift length as long as 1.5 m [8–12].
34 The T600 was then moved to LNGS, and subsequently operated in the CNGS neutrino beam for a
35 muon- to tau-neutrino oscillation search.

36 The success of the construction of the T600 motivated several follow up ideas aimed at reaching
37 even higher target masses (see, e.g., Ref. [13]). A monolithic magnetized large mass device was
38 proposed in [14], based on the scaling up of the ICARUS approach. The ICARUS experience led
39 then to the broader development of the LArTPC concept. A LArTPC design envisioning a single
40 cylindrical volume of 70 m diameter and 20 m height, called GLACIER, was discussed in [15, 16],
41 based on a double-phase readout. This approach eventually evolved to the LAGUNA [17] proposal
42 and to the LAGUNA LBNO [18, 19] conceptual design, aimed at a large detector placed along a
43 long-baseline beam from CERN to an European underground site. A test device is currently under

1 construction at CERN, in the framework of the WA105 R&D project [20]. In the United States,
2 an analogous multi-kton detector was proposed using conventional single-phase wire readout for
3 the LBNE project [21], aimed at a long-baseline neutrino beam sent from Fermilab to the Sanford
4 Laboratory in South Dakota.

5 In parallel, LArTPCs were also considered for short-baseline neutrino experiments, motivat-
6 ing the use of relatively smaller size devices. A first proposal was submitted for the T2K neutrino
7 beam [22, 23]. These ideas were backed by vigorous R&D programs still in progress in Bern [24 –
8 27], at ETH Zurich [28 – 31], as well as in the United States [32]. At Fermilab, the first LArTPC to
9 be successfully operated in the NuMI beam was the ArgoNeuT detector [33 – 37] that, despite the
10 relatively small mass of ~ 170 kg, has been able to perform a series of detailed studies on the inter-
11 action of medium-energy neutrinos[?]. Other critical development projects are currently underway
12 at Fermilab such as [38 – 42], all geared towards development of the technology and of analyzing
13 particle interactions in the detectors. The goal of the development in the near term is towards the
14 SBN program experiments, and in the farther term, following the recent recommendations of the
15 P5 panel [43], the LBNF project and the DUNE experiment which envisions an internationally co-
16 ordinated effort for the realization of an intense neutrino beam facility at Fermilab and a far large
17 mass neutrino detector in a long-baseline configuration (> 1000 km).

18 More information on the LArTPC technology can be found in existing reviews (see, e.g., [44]
19 and references therein).

20 **Organization of Document** This document describes the design, construction, and technical de-
21 tails of the MicroBooNE experiment. Section 2 gives a brief review of the LArTPC technique
22 and its implementation in MicroBooNE. Section 3 describes the design and construction of the
23 cryogenic and purification systems which are required for maintaining a stable volume of highly
24 purified liquid argon. The LArTPC detector described in section 4 is the centerpiece of the experi-
25 ment, providing fine-grained images of neutrino interactions. A light collection system, described
26 in section 5, provides precise timing information by viewing the detector volume and recording
27 signals from the scintillation light that is produced therein. Signals from the light collection system
28 and from the LArTPC are amplified, sampled, and recorded by a custom-designed electronic and
29 readout system, as described in section 6. Section 7 describes the auxiliary instrumentation that
30 monitor and control the detector and all of its associated systems, as well as provide an electrically
31 quiet environment for the experiment to operate. Finally, one of the main calibration sources for
32 the experiment is an ultraviolet laser system, described in section 8, that provides the capability to
33 map out the performance of the LArTPC detector.

34 **2. Experiment Overview**

35 The MicroBooNE detector is hosted at Fermilab in Batavia, Illinois and is located on-axis to the
36 Booster Neutrino Beam, which delivers a beam of predominantly muon-type neutrinos with ener-
37 gies peaking at 700 MeV. The characteristics of this beamline are well-understood from the work
38 of the MiniBooNE experiment [45], which operates directly downstream of the MicroBooNE lo-
39 cation. MicroBooNE also records neutrino interactions from an off-axis component of the NuMI
40 beamline [?]. Figure 1 shows the arrangement of MicroBooNE with respect to the BNB and NuMI

beamlines at Fermilab. The physics program of MicroBooNE will utilize both BNB and NuMI samples. MicroBooNE will also collect data that is out-of-time with either beam, which will be useful for developing analyses (e.g. proton decay searches) relevant for next-generation detectors. The primary detector technology employed by MicroBooNE is the LArTPC , whose operating principle is described in this section.

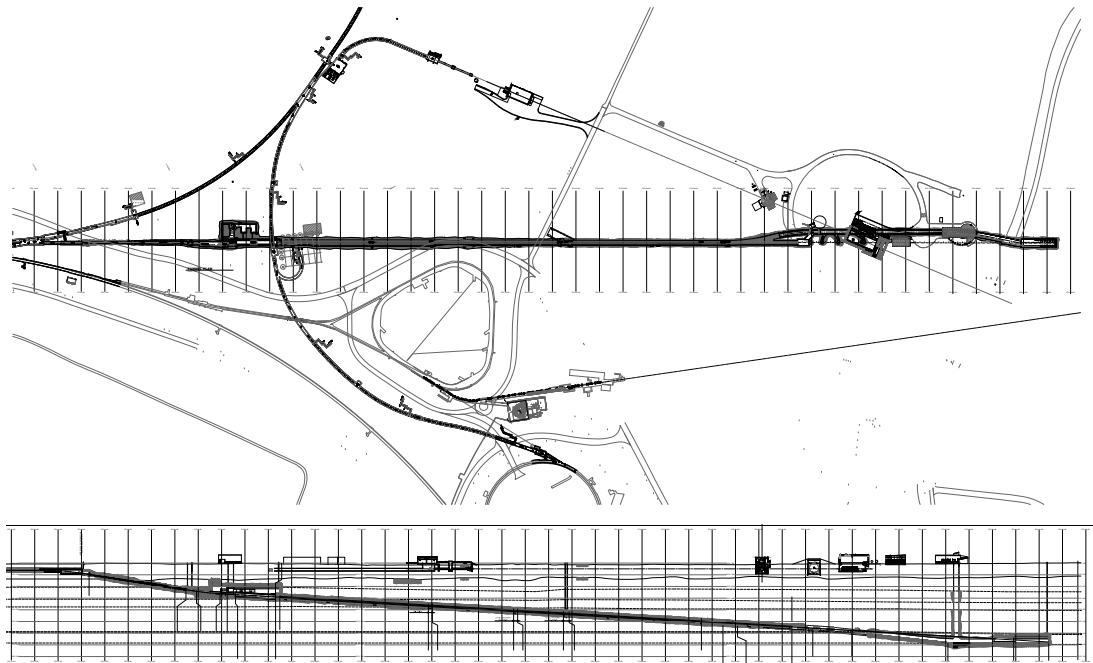


Figure 1: Plan (top) and cross-sectional (bottom) views of MicroBooNE. MicroBooNE is on-axis to the BNB, and off-axis to the NuMI beam.

2.1 The MicroBooNE LArTPC

Charged particles traversing a volume of highly-purified liquid argon leave trails of ionization electrons in their wake and also create prompt vacuum ultraviolet (VUV) scintillation photons. In a LArTPC , the liquid argon is highly purified and the ionization trails are transported practically undistorted over distances of the order of meters [46] under the influence of a uniform electric field in the detector volume, until they reach anode planes located along one side of the active volume. The uniform electric field is created by introducing voltage onto a cathode plane and gradually stepping that voltage down in magnitude across a field cage, which is formed from a series of equipotential rings surrounding the drift volume. The anode plane is arranged parallel to the cathode plane, and in the case of the MicroBooNE LArTPC, are three planes of sense wires composed of wires with a characteristic pitch, held at a predetermined bias voltage, that continuously sense the signals induced by the ionization electrons drifting towards them. The electrostatic potentials of the sequence of anode planes allow ionization electrons to pass undisturbed by the first planes before ultimately ending their trajectory on a wire in the last plane. The drifting ionization thus induces signals on the first planes (referred to as Induction planes) and directly contributes

1 to the signals in the final plane (referred to as the Collection plane). Figure 2 depicts a simplified
2 arrangement for a LArTPC.

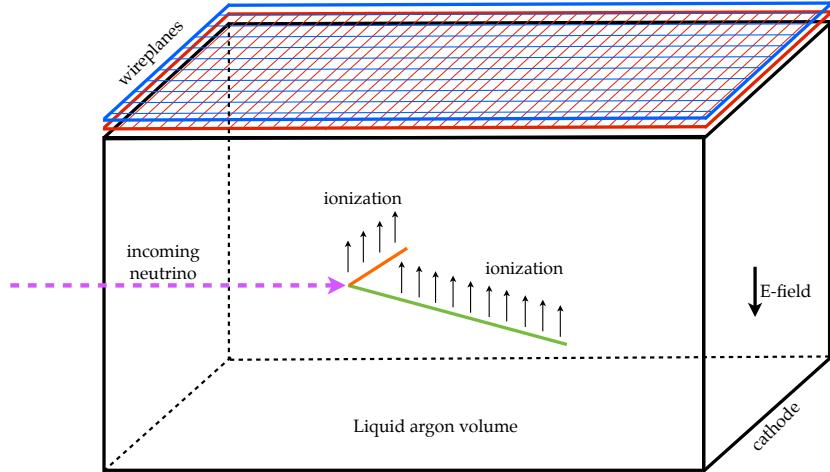


Figure 2: Operational principle of a LArTPC.

3 The charged particle trajectory is reconstructed using the known positions of the anode plane
4 wires and the recorded drift time of the ionization. The drift time is the difference between the
5 arrival times of ionization signals on the wires and the time the interaction took place in the detector
6 (t_0) which is provided by an accelerator clock synced to the beam (e.g. - BNB or NuMI) or from
7 a trigger provided by the light collection system. The characteristics of the waveforms observed
8 by each wire provide a measure of the energy deposition of the traversing particles near that wire,
9 which, when taken as a whole for each contained particle's trajectory, allow for determination of
10 momentum and particle identity.

11 The scintillation photons are detected by a light collection system that is immersed in the
12 liquid argon and faces into the detector volume. This system provides signals that can establish the
13 event t_0 and supplies trigger information to an electronic readout system. The precision timing of
14 a typical light collection device relative to that of the comparatively coarser beam structure means
15 that t_0 derived from the light collection system is more precise, and hence the knowledge of the
16 drift coordinate is more precise. The light collection system signals are vital in distinguishing
17 detector activity that is in-time with the beam (e.g. originating from beam interactions) from that
18 which is out-of-time (e.g. likely not originating from beam-related interactions), benefiting event
19 reconstruction. Spatial information transverse to the drift direction can also be inferred from the
20 light collection signals, further aiding in the reconstruction of the activity occurring inside the
21 detector.

22 The choice of liquid argon as both the neutrino target and detector medium in LArTPCs is well
23 motivated, though not without introducing unique considerations for experimental design. Liquid
24 argon as a target for neutrinos is attractive due to its density, allowing a more compact detector with
25 a substantial boost in event rate over a comparable detector using less dense media, such as water.
26 A tradeoff to this aspect is the short radiation length (14 cm) and that the complicated structure
27 of the argon atom, relative to simpler targets such as hydrogen or helium, does introduce nuclear

1 effects that must be considered during data analysis. The cryogenic temperatures at which the
2 noble elements are in the liquid phase also introduces the need for additional design considerations
3 to ensure stable and safe operations.

4 Table 1 lists some of the salient properties of liquid argon. The noble liquids are all char-
5 acterized by their excellent dielectric properties, able to maintain high voltages without suffering
6 electrical breakdown. The long drift lengths over which ionization electrons must travel in large
7 LArTPC detectors requires the presence of a uniform electric field over that distance. This is
8 achieved via the introduction of extremely high voltage ($\mathcal{O}(100\text{ kV})$ magnitude) onto the LArTPC cath-
9 ode; this voltage is suitably maintained due to the dielectric properties of liquid argon and by ensur-
10 ing the detector components have no sharp edges. The noble liquids are all very bright scintillators,
11 with wavelengths deep in the UV, and also produce copious amounts of ionization. Both the scin-
12 tillation and the ionization signals are necessary for a robust accounting of the activity occurring
13 inside the LArTPC. Finally, the desire to build LArTPCs on increasingly larger scales, such as
14 those necessary in neutrino experiments, is bolstered by the abundance (1% of atmosphere) and
15 low cost of argon.

Table 1: Liquid argon properties.

Property	Value
Boiling Point [K]	87.3
Density [g/cm ³]	1.4
Scintillation λ [nm]	128
Rad. Length [cm]	14

16 The successful implementation of the LArTPC technique depends critically on several factors.
17 The liquid argon must be purified of any electronegative contaminants, such as water or oxygen, to
18 accommodate the very long drift path of ionization through a MicroBooNE-sized LArTPC without
19 significant charge loss. The signals that the ionization electrons create on the anode wires are very
20 small, requiring low-noise electronics to discern signal pulses. The MicroBooNE collaboration
21 has designed and constructed an experiment that addresses all of these considerations, providing
22 critical technological development for the next generation of LArTPC experiments to build upon.

23 2.2 MicroBooNE LArTPC Implementation

24 MicroBooNE’s LArTPC active volume, which is defined as the volume immediately within the
25 confines of the LArTPC field cage, is a rectangular liquid argon volume with dimensions 2.3 m
26 vertical x 2.5 m horizontal x 10.4 m along the beam direction. This is the maximum volume that
27 can be used for physics analyses. The cathode (anode) defines the beam-left (beam-right) side of
28 the active volume. The end of the LArTPC that the beam first encounters is referred to as the
29 “upstream” end, while the opposite end is referred to as “downstream”. Anode plane-to-plane
30 spacing is 3 mm, and each plane has 3 mm wire pitch. The induction plane wires are oriented at
31 $\pm 60^\circ$ relative to vertical, and the collection plane has vertically oriented wires. Field cage loops
32 are employed to maintain uniformity of the electric field across the entire width of the detector,
33 and these loops also act to define the top, bottom, upstream, and downstream sides of the active
34 volume.

1 MicroBooNE uses a right-handed Cartesian coordinate system, with the origin defined to be
2 located on the upstream face of the LArTPC, centered halfway up the vertical height of the active
3 volume and horizontally centered on the anode plane closest to the cathode (the innermost anode
4 plane). In this system, x ranges from 0.0 m at the innermost anode plane to +2.5 m at the cathode,
5 y ranges from -1.15 m on the bottom of the active volume to +1.15 m at the top of the active
6 volume, and z ranges from 0.0 m at the upstream end of the active volume to +10.4 m at the
7 downstream end.

8 The light collection system, which is an array of photomultiplier tubes and scintillator pads,
9 is located directly behind the anode planes on beam-right, facing the detector volume through
10 the anode planes. The LArTPC and light collection system are immersed in liquid argon con-
11 tained within a single-walled cryostat with a 170 ton capacity. Electronics mounted directly on the
12 LArTPC amplify the signals on the wires, which are then passed out of the cryostat for further pro-
13 cessing and storage on disk. Table 2 lists the primary detector design parameters of MicroBooNE,
14 and figure 3 depicts the detector arrangement. Details of these design parameters and construction
15 of all detector subsystems will be provided in the subsequent sections.

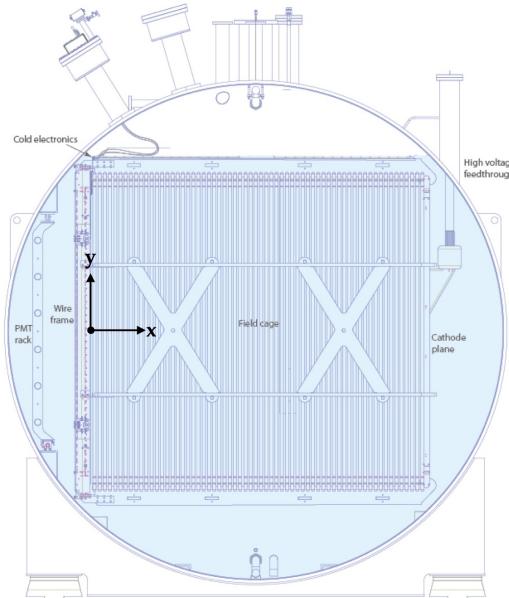


Figure 3: Diagram of MicroBooNE LArTPC. In this view, the beam would be directed out of the page (in the z direction).

16 3. Cryogenic System

17 The use of large quantities of highly-purified liquid argon as a detector medium in MicroBooNE
18 requires a sophisticated cryogenic infrastructure that can maintain stable operations for years at a
19 time. Not only must the purity of the liquid argon be maintained, but the pressure and temperature
20 gradients within the LArTPC active volume must be tightly controlled as the drift velocity of elec-
21 trons is dependent on these quantities. A customized cryogenic system that serves these purposes
22 has been built, and the requirements for this system are shown in table 3.

Table 2: Primary detector design parameters for MicroBooNE.

Parameter	Value
LArTPC Dimensions	2.325 m vertically 2.560 m horizontally 10.368 m longitudinally
LArTPC argon mass	90 tons
Total Number of Wires	8256
Induction0 Wires	2400
Induction1 Wires	2400
Collection Plane Wires	3456
Drift field	500 V/cm
Light collection	30 8" diameter PMTs 4 scintillator paddles
Cryostat liquid argon capacity	170 tons
Operating temperature	87 K
Operating pressure	2 psig

Table 3: Primary design requirements for MicroBooNE cryogenic and purification systems.

Parameter	Value	Motivation
Argon purity	<100 ppt O ₂	MIP identification at longest drift
Argon purity	<2 ppm N ₂	Scintillation light output
LAr Temperature gradient	<1° K	Drift-velocity uniformity
LAr recirculation rate	1 volume change/day	Maintain purity
Cryostat heat load	<15 W/m ²	Minimize convection currents and bubbles
Cryogenic capacity	10 kW	Capacity to deal with expected heat load
Cryostat maximum pressure	30 psig	Determines relief sizing

The MicroBooNE cryogenic system is represented in figure 4. The central component of the system is a cryostat that houses the complete LArTPC and light-collection detector systems. The cryostat is supported by three major subsystems: the argon purification system, the nitrogen refrigeration system, and the controls and monitoring system. These systems each represent the next generation of LArTPC cryogenic system after the Liquid Argon Purity Demonstrator (LAPD) [40] and make considerable use of the expertise gained during the design and implementation of that apparatus.

3.1 Cryostat Design Overview

Three major components make up the MicroBooNE cryostat: a stainless steel (type 304) vessel to contain the liquid argon and all the active detector elements, front and rear supports to carry the weight of the fully loaded cryostat, and foam insulation covering the cryostat outer surfaces. The foam insulation serves to reduce heat input from the ambient environment to a sufficiently low level to prevent large temperature gradients and boiling of the liquid argon. The cryostat and the



Figure 4: Three-dimensional renderings of the MicroBooNE cryogenic system installed at LArTF.

1 cryogenic systems are designed to achieve the high-purity liquid argon needed to allow ionization
 2 electrons to drift to the anode wires with low probability of capture, and the high degree of thermal
 3 homogeneity needed to avoid the introduction of non-constant drift velocities for the ionization
 4 electrons. Finally, the outer diameter of the vessel is designed to be the maximum standard size for
 5 over-the-road transport.

6 Ionization electrons must not be significantly attenuated, via attachment to electronegative
 7 contaminants in the liquid argon, as they drift up to 2.5 m across the active volume. This dictates
 8 that the argon be kept free of electronegative contaminants to the level of 100 parts-per-trillion
 9 (ppt). The cryostat is designed to minimize outgassing (desorption) and to avoid leakage and
 10 diffusion of air into the system. This requirement imposes strict quality assurance demands on all
 11 welds for penetrations into the cryostat and on cleaning and handling procedures for the finished
 12 vessel. Achieving the required level of purity is accomplished with a purification system, described
 13 in section 3.2, that removes electronegative contaminants from the argon during the initial fill and

1 those introduced over time by leaks and outgassing of system components.

2 The electron drift velocity ($v_d = 1600$ m/s at an electric field of 500 V/cm, with a liquid argon
3 temperature dependence $\Delta v_d/v_d = -0.019\Delta T$) must remain constant in magnitude and direction
4 throughout the active liquid argon volume to avoid distortion of the mapping of drift time into the
5 position along the drift (\hat{x}) direction. This requirement limits the allowable temperature variations
6 of the liquid argon to less than 0.1 K and the laminar and turbulent flow rate of liquid argon to
7 less than 1 m/s. These requirements limit fractional errors in velocity, and therefore in the drift-
8 coordinate determination, to be less than 0.1%. The constraints on constancy of drift velocity affect
9 the design by imposing limits on the acceptable heat flux through the insulation.

10 The cryostat is constructed to the latest American Society of Mechanical Engineers (ASME)
11 boiler code requirements [?] and features a single-walled construction, cylindrical shape, and
12 domed caps closing each end, as shown in figure 5. One end was removed for installation of
13 the active detectors, welded back in place upon completion of that task, and then recertified to
14 the ASME code requirements. Upon installation of the sealed cryostat in its final location at the
15 Liquid Argon Test Facility (LArTF), 16 inches of spray-on, closed cell, Polyurethane insulation
16 was applied to the exterior of the cryostat, as shown in figure 6. At LArTF, to avoid ground loops
17 that could interfere with the LArTPC signals, the cryostat vessel is grounded in only one place,
18 allowing it to act as a Faraday cage. This grounding scheme is explained further in section 7.1.

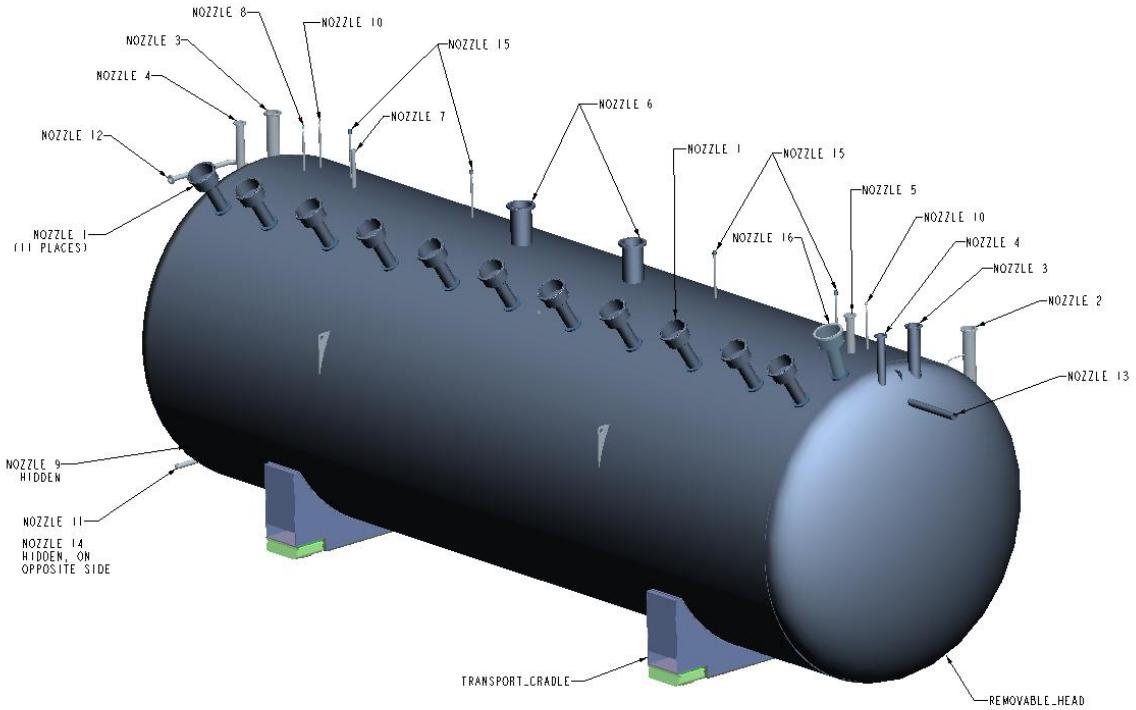


Figure 5: MicroBooNE cryostat with nozzle penetrations labeled.

19 The vessel surface has 34 nozzle penetrations for cryogenic and electrical services, detailed in
20 table 4. All nozzles are sealed with feedthroughs, flanges, or pipes that are suitable for operation at
21 the nominal pressure and temperature of the cryostat.

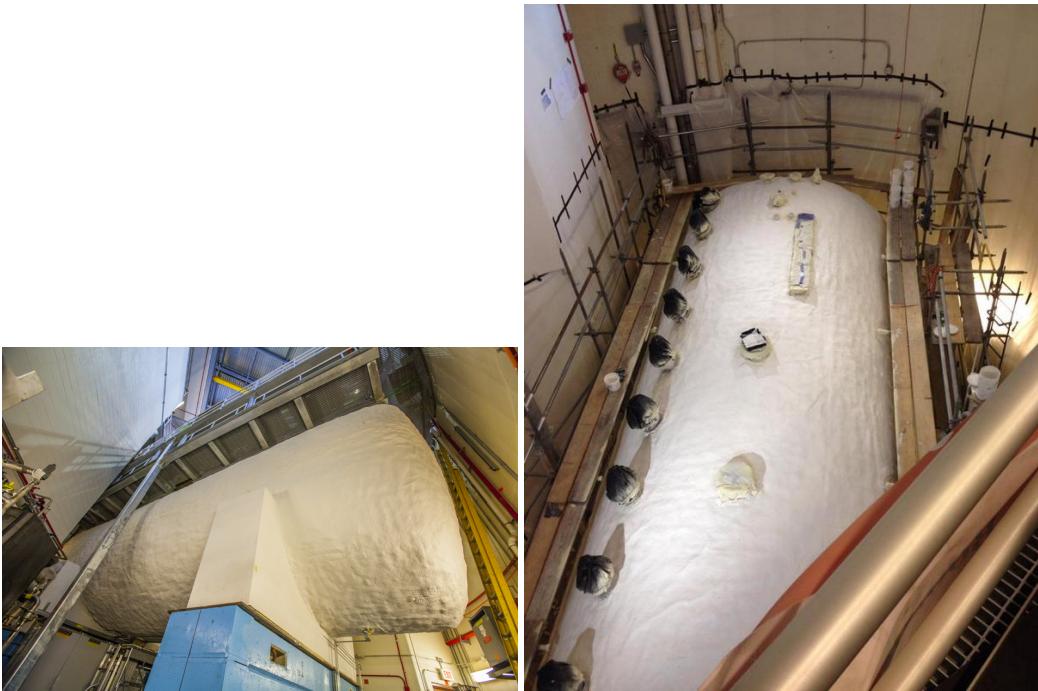


Figure 6: Photographs of the cryostat after application of exterior foam insulation.

3.2 Liquid Argon Purification Subsystem

The heart of the cryogenic system is the liquid argon purification subsystem. The primary requirement of this subsystem is to keep the level of electronegative contamination to below 100 ppt of oxygen-equivalent contaminants. This requirement was determined by the physics needs of the experiment, namely the need to keep the attenuation of ionization electrons to less than 20% over the longest drift distance in the LArTPC. In addition to the requirement on the electronegative contamination, the system must maintain the level of nitrogen contamination in the argon at less than 2 parts per million (ppm) [47] to keep the attenuation of the scintillation photons in the argon to a minimum.

The MicroBooNE argon purification subsystem consists of liquid argon pumps and filters that serve to circulate the argon and remove impurities that degrade the quality of the data collected by the active detectors. There are two pumps in the system arranged in parallel in order to allow for continuous recirculation while one pump is being serviced. Similarly, there are two sets of filters arranged in parallel in the system. Figure 7 schematically depicts the flow of liquid and gaseous argon in the MicroBooNE cryogenic system.

The recirculation pumps are Barber-Nichols [48] BNCP-32B-000 magnetically-driven partial-emission centrifugal pumps. Each pump isolates the liquid argon from the electric motor. The impeller, inducer, and driving section of the magnetic coupling each have their own bearings that are lubricated by the liquid argon at the impeller. The motor is controlled by a variable frequency drive (VFD) that allows adjustment of the pump speed to produce the desired head pressure and flow within the available power range of the motor.

Each filter skid contains two filters, each having identically-sized filtration beds of 77 liters.

Table 4: List of nozzle penetrations in MicroBooNE cryostat. CF=ConFlat flanges, RFWN=raised face weld neck flanges.

Nozzle ID	Function	Flange
N1A-N1K	LArTPC Signal Feedthrough	14" CF
N2	LArTPC HV Feedthrough	8" CF
N3A-N3B	Purity Monitor	8" CF
N4	Temperature Signals	6" CF
N5	Safety Vent	4" RFWN
N6A-N6B	Vacuum Pump-Out	10" RFWN
N7	Condensor	3"
N8	Top Instrument Port	3/4"
N9	Bottom Instrument Port	3/4"
N10A-N10B	Liquid Level Probe	3/4"
N11	Gas Circulation In	2"
N12	Gas Circulation Out	2"
N13	From LAr Filters	3"
N14	To LAr Pumps	2"
N15A-N15B	Laser Calibration	2-3/4" CF
N16	PMT Signal Feedthrough	14" CF
N17	Spare	6" CF
N18	Temperature Signals	6" CF
N19	Spare	6" CF

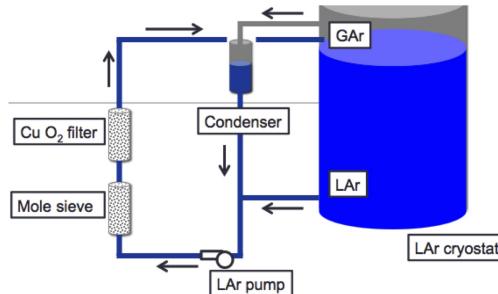


Figure 7: Flow diagram of argon in MicroBooNE, showing direction of liquid and gaseous argon in the cryogenic system. Gaseous argon from the cryostat is condensed and directed through the purification subsystem. Liquid argon drawn from the cryostat volume is directed into the purification subsystem. **NOTE: Update with better figure!**

- ¹ The first filter that the argon stream enters contains a 4A molecular sieve supplied by Sigma-Aldrich [49] that primarily removes water contamination but can also remove small amounts of nitrogen and oxygen. The second filter contains BASF CU-0226 S, a highly-dispersed copper oxide impregnated on a high-surface-area alumina, which removes oxygen [50] and, to a lesser extent, water. Thus, the oxygen filter is placed downstream of the molecular sieve to maximize oxygen fil-

1 tration. The oxygen-filtering media must be reduced to copper with the procedure described below
2 before it can remove oxygen from the liquid argon. The filters are insulated with vacuum jackets
3 and aluminum radiation shields. The metallic radiation shields were chosen because the filter re-
4 generation temperatures, described below, would damage traditional aluminized mylar insulation.
5 Pipe supplying the filter regeneration gas is insulated both inside the filter vacuum-insulation space
6 and outside the filter with Pyrogel XT which is an aerogel-based insulation [51] that can withstand
7 temperatures up to 650°C.

8 The filters are regenerated *in situ* using heated gas, by a procedure developed for LAPD. The
9 filters are regenerated using a flow of argon gas that is heated to 200°C, supplied by a commercial
10 500 liter liquid argon dewar. Once the argon gas reaches 200°C, a small flow of hydrogen is mixed
11 into the primary argon flow and exothermically combines with oxygen captured by the filter to
12 create water. Too much hydrogen mixed in with the primary argon flow would induce temperatures
13 that are sufficiently high to damage the copper-based filter media. The damage is induced by
14 sintering of the copper, which reduces the available filter surface area. Thus, precautions are taken
15 to maintain a hydrogen fraction below 2.5% of the heated gas mixture. During the heated gas
16 regeneration, five filter-bed temperature sensors monitor the filter-material temperature and the
17 water content of the regeneration exhaust gas is measured. To remove any remaining trace amounts
18 of water, the filters are then evacuated using turbomolecular vacuum pumps while they cool.

19 A particulate filter with an effective filtration of 10 microns, positioned between the cryostat
20 and the filter skids, prevents any debris in the piping from being introduced into the cryostat.
21 The particulate filter consists of a commercial stainless steel sintered-metal cylinder mounted in a
22 custom cryogenic housing and vacuum jacket. Filtration is accomplished by flowing liquid argon
23 to the interior, then outward through the walls, of the sintered-metal cylinder. Flanges on the argon
24 piping, along with flanges and edge-welded bellows on the vacuum jacket, allow removal of the
25 particulate filter.

26 The argon-purification piping is 2.54 cm diameter stainless steel that was pre-insulated by the
27 manufacturer with 10.2 cm of polyurethane foam. During the fabrication process, all piping was
28 washed with distilled water and detergent to remove oil and grease, then cleaned with ethanol.
29 All valves associated with the argon-purification piping utilize a metal seal with respect to am-
30 bient air, either through a bellows or a diaphragm, to prevent the diffusion of oxygen and water
31 contamination. The exhaust side of each relief valve is continuously purged with argon gas to
32 prevent diffusion of oxygen and water from ambient air across the o-ring seal. Where possible,
33 ConFlat flanges with copper seals are used on both cryogenic and room-temperature argon piping.
34 Pipe flanges in the system are sealed using spiral-wound graphite gaskets. Smaller connections are
35 made with VCR fittings with stainless steel gaskets.

36 **3.3 Nitrogen Refrigeration**

37 The cryostat and purification systems that contain the liquid argon are subject to heat load from
38 the environment, as well as from the active detectors that have electrical power enabled. To keep
39 these systems operating at a stable temperature and pressure, a liquid nitrogen refrigeration system
40 is present to provide the necessary cooling power. The liquid nitrogen system contains two con-
41 densers that are arranged in parallel. One of these is utilized for normal operations and one serves
42 as a backup on standby. Each condenser contains two liquid nitrogen coils, an inner and an outer,

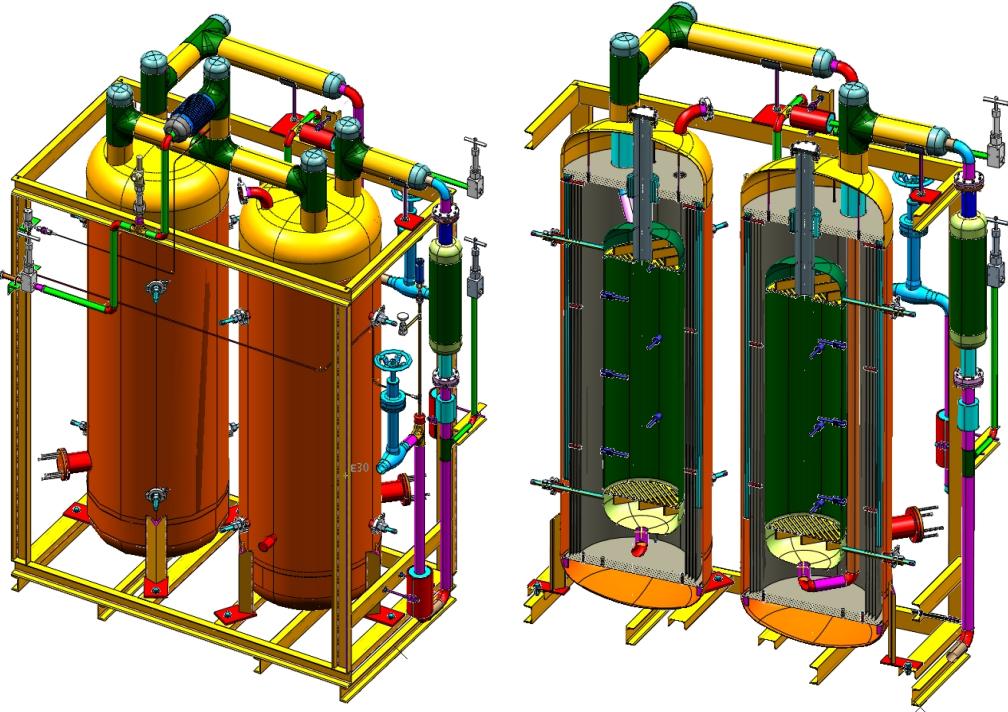


Figure 8: Three-dimensional rendering of a MicroBooNE filter skid. The left drawing shows the full skid, while the right drawing shows a cut-away of the vessels.

with the gas argon on the shell side, as shown in figure 9. Typically only one coil is actively running and the second can be manually activated during situations where the system heat load is higher than usual. Each condenser is sized to handle a heat load of approximately 9.5 kW. With the vessel full of liquid argon and no pump or liquid argon circulation running, the condenser uses 600-650 gallons of liquid nitrogen per day, which equates to a 3.9 kW system heat load. Once liquid argon begins circulating using the pumps, this usage rate will increase by 50-100%.

3.4 Controls and Purity Monitoring

MicroBooNE makes use of resistive thermal devices (RTDs) to measure temperatures throughout the experimental infrastructure. Twelve RTDs are located along the walls of the cryostat, and another ten RTDs are mounted inside screws attached to the structure of the LArTPC. Each of the filter vessels in the purification system contain nine RTDs. The RTDs within the filter vessels prevent overheating which could potentially occur during filter regeneration with heated argon-hydrogen gas.

Liquid argon contaminations ranging between 300 and 50 ppt oxygen equivalent can be measured using double-gridded ion chambers, henceforth referred to as purity monitors, immersed in liquid argon. The design of the purity monitors is based on the design presented in Ref. [52]. Reference [40] gives a thorough description of the purity monitors, the data-acquisition hardware and software used in LAPD. MicroBooNE uses the same type of purity monitors, and the same data-acquisition hardware and software.

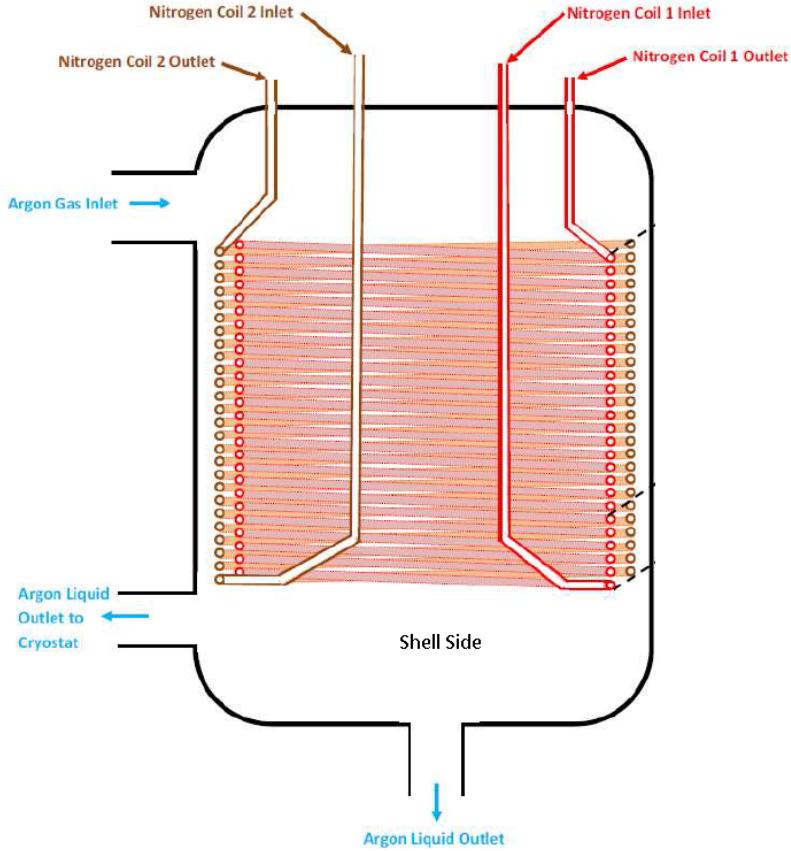


Figure 9: Diagram of a condenser.

1 A measure of the electronegative impurities is made with the purity monitor by examining the
 2 fraction of electrons generated at the purity monitor cathode that subsequently arrive at the purity
 3 monitor anode (Q_A/Q_C) after a drift time t . The ratio of (Q_A/Q_C) is related to electron lifetime, τ ,
 4 such that

$$Q_A/Q_C = e^{-t/\tau}. \quad (3.1)$$

5 Measurement of liquid argon purity in the MicroBooNE cryogenic system are provided by
 6 three purity monitors of varying lengths. One purity monitor with a drift distance of 50 cm sits
 7 in a vessel just downstream of the filters and is used to monitor filter effectiveness. Two purity
 8 monitors, one with a drift distance of 19 cm and the other with a drift distance of 50 cm, sit within
 9 the primary MicroBooNE vessel at each end of the LArTPC.

10 **4. Liquid Argon Time Projection Chamber**

11 The MicroBooNE LArTPC is the primary detector featured in the experiment, and it is capable of
 12 producing fine-grained images of the ionization that is liberated by charged particles traversing a

¹ volume of highly-purified liquid argon. This section describes the design and implementation of
² the LArTPC in the experiment.

³ The LArTPC is composed of three major structures: the cathode, the field cage, and the an-
⁴ ode. A negative voltage is introduced via a feedthrough passing through nozzle N2 on the cryostat
⁵ and applied at the cathode, which defines an equipotential surface. This cathode voltage is incre-
⁶ mentally stepped down in magnitude by means of a voltage divider chain spanning the field cage,
⁷ creating a region of uniform electric field within the LArTPC. Opposite the cathode, and oriented
⁸ parallel to it, are the anode wire planes: two induction planes (referred to as the “U” and “V”
⁹ planes) with wires oriented at $\pm 60^\circ$ from vertical, followed by one collection plane (referred to as
¹⁰ the “Y” plane) with vertically-oriented wires. The wires of the anode planes are the sensitive ele-
¹¹ ments that detect the ionization created by charged particles traveling through the LArTPC. Figure
¹² 10 depicts the assembled MicroBooNE LArTPC after insertion into the cryostat, showing details
¹³ of the cathode, field cage, and anode plane. Table 5 lists the main parameters of the MicroBooNE
¹⁴ LArTPC, which will be described in detail in this section.

Table 5: MicroBooNE LArTPC design parameters and nominal operating conditions.

Parameter	Value
LArTPC (active) dimensions ($h \times w \times l$)	2.325 m \times 2.560 m \times 10.368 m
LArTPC (active) mass	90 tons
# Anode planes	3
Anode planes spacing	3 mm
Wire pitch	3 mm
Wire type	SS, diam. 150 μm
Wire coating	2 μm Cu, 0.1 μm Ag
# wires (total)	8256
# Induction0 plane (U) wires	2400
# Induction1 plane (V) wires	2400
# Collection plane (Y) wires	3456
Wire orientation (w.r.t. vertical)	+60°,-60°,0° (U,V,Y)
Cathode voltage (nominal)	-128 kV
Bias voltages (U,V,Y)	-200 V, 0 V, +440 V
Drift-field	500 V/cm
U-V gap field	666.7 V/cm
V-Y gap field	1466.7 V/cm
Max. Drift Time, Cathode to U (at 500 V/cm)	1.6 ms
# Field-cage steps	64
Ring-to-ring voltage step	2.0 kV

¹⁵ **4.1 Cathode**

¹⁶ The cathode is assembled from 9 individual stainless steel sheets (Type 304, 0.09 inches thick) that
¹⁷ are fastened to a supporting frame by hex button head stainless steel screws. The outer edge of the

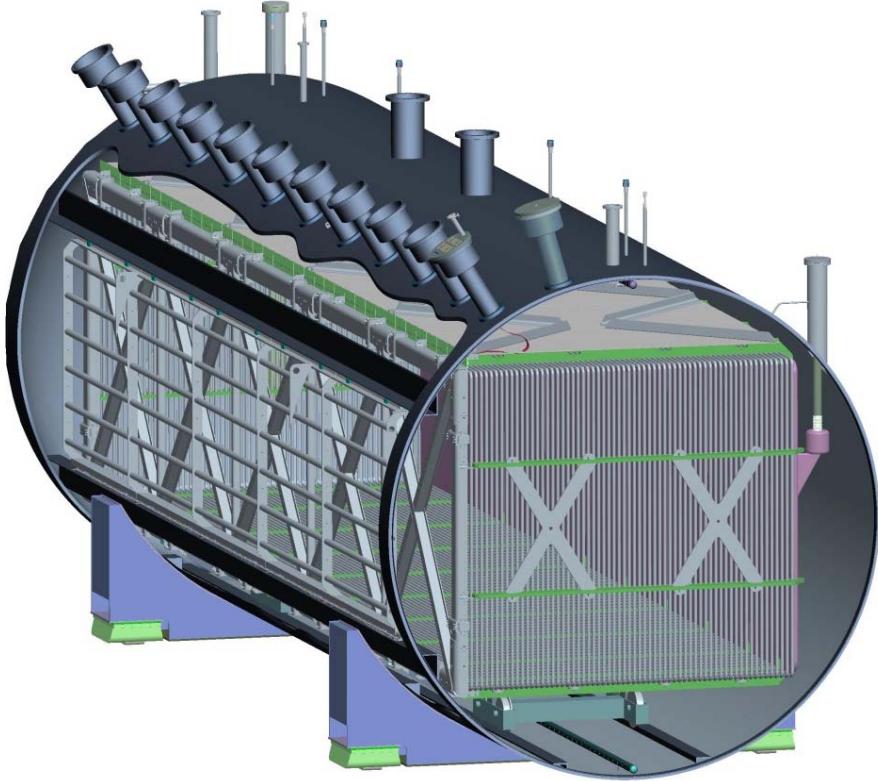


Figure 10: Schematic diagram of the MicroBooNE LArTPC , depicted as it is arranged inside the cryostat.

1 cathode frame consists of round stainless steel tubes of 50.8 mm outer diameter and 3.2 mm wall
 2 thickness. Within this outer edge, square tubes with 50.8 mm \times 50.8 mm cross-sectional area, and
 3 3.2 mm wall thickness, are fastened together with button head screws, forming a support structure
 4 upon which the cathode sheets are attached. The individual components of the support structure are
 5 further welded together to eliminate sharp features from this high-potential surface. The exterior
 6 frame and support structure of the cathode are shown in figure 11 and a view of the interior is shown
 7 in figure 12. The cathode plane sheets are shimmed according to survey data to make the cathode
 8 as flat and as parallel to the anode frame as possible. Flatness of the cathode is evaluated relative to
 9 a best fit plane of survey data (more than 10000 survey points recorded with a laser tracker). The
 10 largest deviations of the cathode from the best fit plane are +6.6mm and -6.5mm. Approximately
 11 55% of the measured survey points fall within +/-3mm of the best fit plane, and more than 90% of
 12 the points fall within ± 5 mm. Figure 13 shows the results of the survey, with deviations from flat
 13 represented as color-coded data extending away from the nominal plane of an ideal cathode.

14 4.2 Field Cage

15 The field cage encloses the volume between the cathode plane and the anode wire planes, and in
 16 doing so creates a region with a uniform electric field. The volume defined by the interior of the
 17 field cage, bounded by the anode and cathode planes, is referred to as the “active” volume. The field
 18 cage structure consists of 64 individual thin-walled stainless steel tubes (1.0 inch OD, 0.065 inch



Figure 11: Exterior view showing the cathode frame and structural supports to which cathode sheets are fastened.



Figure 12: Interior view of cathode plane as viewed from the upstream end of the LArTPC , showing cathode sheets.

- ¹ wall thickness), each shaped into a rectangular loop framing the perimeter of the active volume.
- ² These 64 loops are mounted parallel to the cathode and anode planes, as shown in figures 11 and 12,

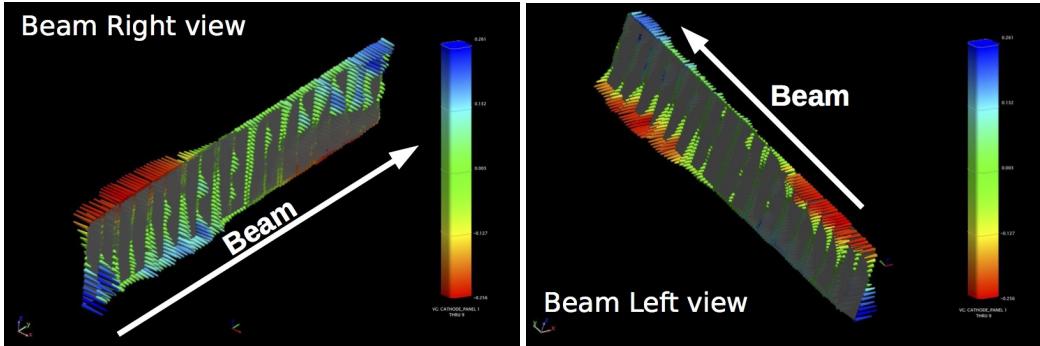


Figure 13: Survey results for interior (left) and exterior (right) of the cathode after shimming.

1 and are held in place by a G-10 rib support structure. Each field cage loop is electrically connected
 2 to its neighbors via a resistor divider chain (described in following section), causing each loop to
 3 operate at a different electrical potential which in turn maintains a uniform electric field between
 4 the cathode and anode planes. For a nominal -128 kV cathode voltage, the difference in potential
 5 between adjacent field cage loops is 2 kV, ramping down the total potential in equidistant steps from
 6 cathode to anode. The distance from center-to-center of adjacent field cage loops is 1.575 inches.
 7

8 Each field cage loop is assembled from 6.77 feet long vertical pipes on the upstream and
 9 downstream ends of the LArTPC , and on the top and bottom from two 17 feet long horizontal pipes
 10 connected by a stainless steel coupling in the center. Each tube has venting holes approximately
 11 every 6 inches so that they fill with liquid argon and allow air and gaseous argon to escape during
 cryostat cool down and filling.

12 The four corners of each field cage loop are curved with a radius of 2.0625 inch. Each corner
 13 is formed by three parts: two couplings and an elbow, shown in figure 14. The couplings make the
 14 connections between the pipes and the elbow. The thin-walled tubes and elbows slip-fit over the
 15 ends of the couplings with a 0.88-inch overlap. Each coupling has two 6-32 NC tapped holes and
 16 the connections to the adjoining pieces are made by hex head button screws and lock washers.

17 In order to avoid electrical breakdown between the inner cryostat surface and field cage parts
 18 at high potential on or near the cathode, the electric field strength is minimized at the corners and
 19 edges of the field cage. Loop 0 is the cathode frame, which is at the same potential as the cathode
 20 plane sheets attached to it. The frame is made from larger diameter piping of 2 inch OD, and it has a
 21 slightly smaller circumference than the other field cage loops, as shown in figure 15. The elbow of
 22 loop 1 has a specially designed shape in order to minimize the electric field potential, and the elbow
 23 of loop 2 has a softer radius of curvature than the standard elbows. For all three of these loops (loop
 24 0, 1, and 2), connections at corners and joints are made by welding instead of screws to avoid sharp
 25 edges that would result in higher electric fields and greater chance of electrical breakdown.

26 Another precaution to minimize the electrical field between the loops and the cryostat surface
 27 is the positioning of the coupling screws: for the first 20 loops, the screws are positioned on the
 28 sides facing the screws of the neighboring loops instead of facing inward to the LArTPC active vol-
 29 ume and outward toward the grounded cryostat surface. Hex head button screws and lock washers
 30 are also used here in order to minimize sharp metal edges.

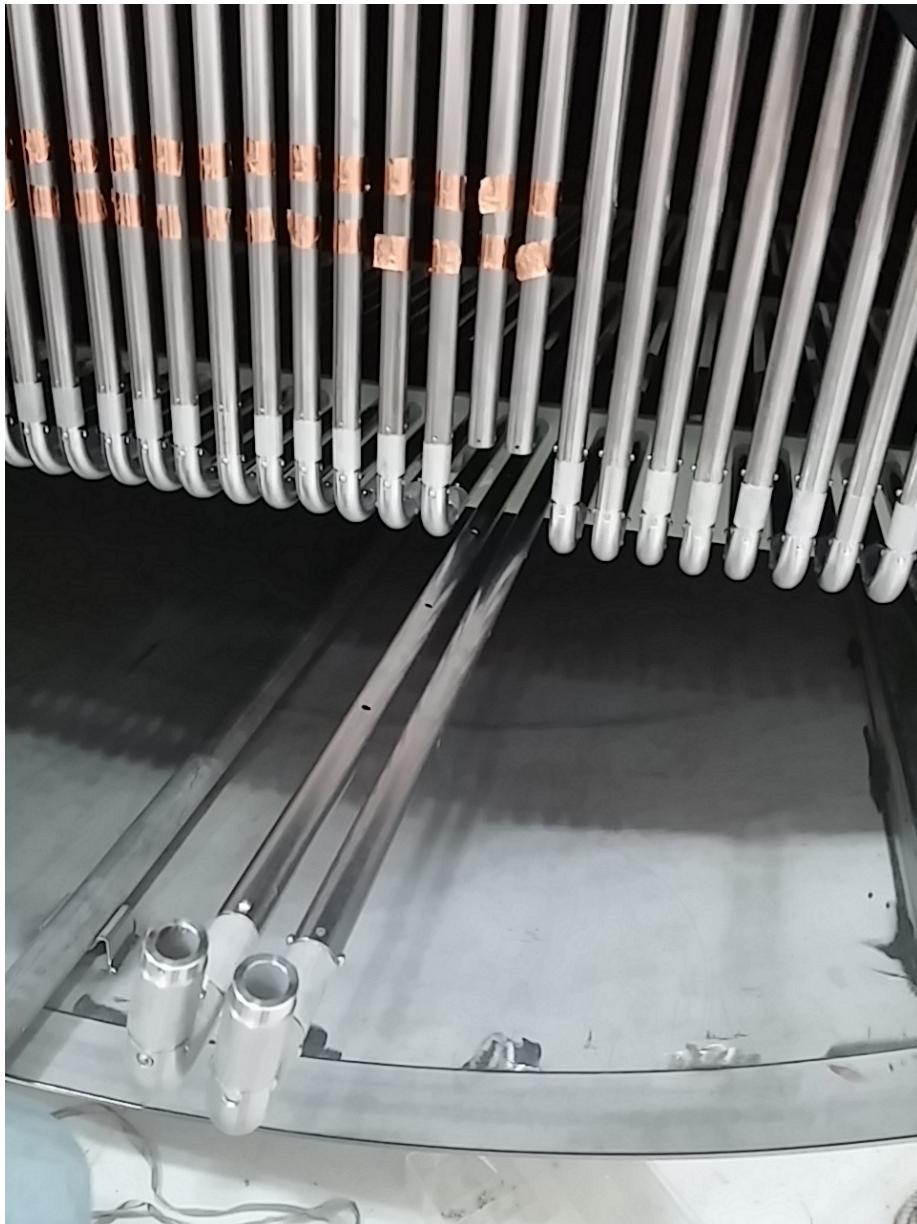


Figure 14: Field cage loops, with two partially disassembled to show elbow and couplings.

4.2.1 Resistor Divider Chain

A resistor divider chain installed across the field cage loops steps the voltage down in magnitude from the cathode plane to the anode wire plane in equidistant steps. 63 of the 64 loops between the cathode plane and the anode wire plane are electrically isolated. For a nominal value of -128 kV on the cathode, this results in a potential difference of 2 kV between each pair of loops. The value of the equivalent resistance between loops within the divider chain was chosen to be low enough such that the current flow through the divider circuit is much greater than the signal current flowing through the LArTPC . The signal current in this case is the total energy deposited by cosmic rays, and is estimated to be <50 nA. An equivalent resistance of 250 M Ω between each pair of field cage



Figure 15: Field cage loops closest to (and including) the cathode are modified to reduce sharp edges that would result in higher electric fields.

1 loops, corresponding to a current flow of $8 \mu\text{A}$, was chosen.

2 The voltage divider chain is mounted on the inside of the field cage at the upstream end of the
3 detector. The couplings at the top corner of each field cage loop have additional holes facing the
4 inside of the field cage, where the resistors are mounted. On the first 16 field cage loops, pairs of
5 Metallux HVR 969.23 499 M Ω resistors (rated to 23 W, 48 kV in air) are mounted electrically in
6 parallel to establish the beginning of the voltage divider chain. On the remaining loops, thick-film
7 Ohmite Slim-Mox 104E metal-oxide epoxy-coated resistors with a lower power and voltage rating
8 (1.5 W, 10 kV in air) are used. Extensive testing was done on these two types of resistors, as well

as others, all detailed in Ref. [?]. For redundancy, the desired $250 \text{ M}\Omega$ total resistance between each pair of field cage loops is composed of four individual $1 \text{ G}\Omega$ Slim-Mox 104E resistors placed electrically in parallel and attached to a printed circuit board. Printed circuit boards span across eight field cage gaps and therefore have eight $250 \text{ M}\Omega$ resistances in series, shown in figure 16. The electrical connection between the boards and each field cage loop is made by metal contact pads on the back side of the boards, held in electrical contact with the field cage tube by a button-head hex screw and lock washer.

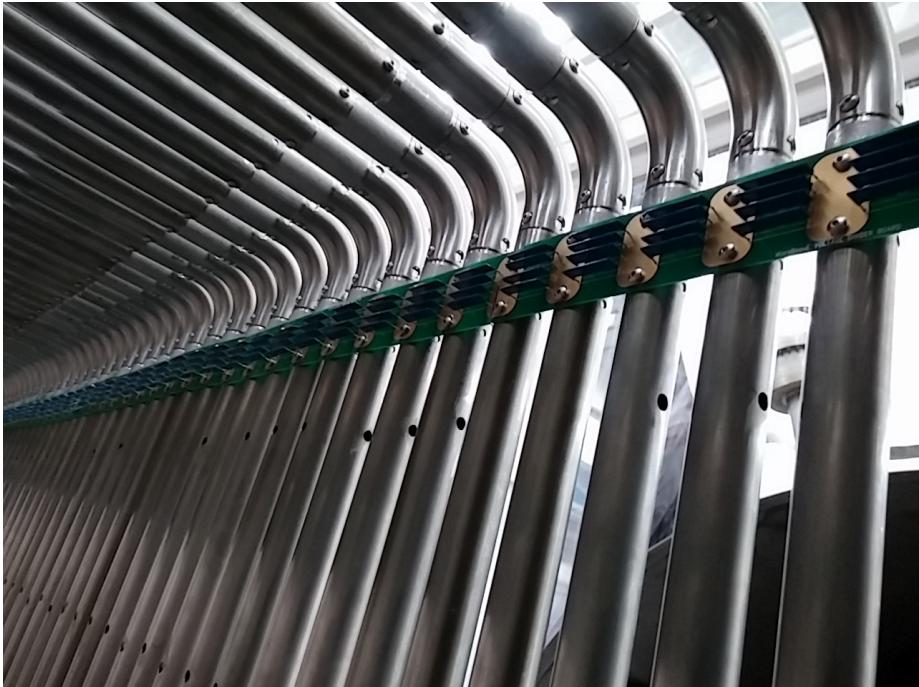


Figure 16: The Ohmite Slim-Mox 104E resistors arranged in parallel sets of four on printed circuit boards that span eight field cage tubes.

While the designed operating voltage difference across each resistor in the detector is 2 kV with a power flow of 4 mW , there is a slight possibility that these values could temporarily exceed the rating of the resistors in the case of discharge between the cathode plane or field cage loops and the cryostat wall, through the bulk liquid argon. Recent studies [?] have shown that the value of the minimum breakdown electrical field decreases with the increasing argon purity; for purities as high as that required in the MicroBooNE detector, breakdown has been observed at electric fields as low as 40 kV/cm .

The field cage behaves like a capacitance network. Based on measurements and simulations, the total energy stored inside the field cage when fully charged is estimated to be approximately 24 J . In the case of a discharge between the cryostat and the cathode or one of the field cage loops close to the cathode, simulations show that voltages of up to 80 kV can develop across the resistors over the timescale of a few seconds. The observed peak voltages in such discharge scenarios decrease the further the breakdown occurs from the cathode, such that discharges occurring between the cryostat and field cage loops 32 through 63 do not exceed the 10 kV rating of the resistors.

1 In order to protect the Ohmite Slim-Mox 104E resistors near the cathode from damage, two
2 strategies have been implemented. First, as previously mentioned, on the first 16 field cage loops
3 near the cathode the Ohmite Slim-Mox 104E resistors are replaced by $499 \text{ M}\Omega$ Metallux HVR
4 969.23 resistors that have a higher voltage rating of 48 kV. An extensive study of resistor breakdown
5 in liquid argon of several resistor brands and models [?] has revealed that the breakdown voltage
6 observed in liquid argon exceeds the rating in air substantially for all resistors. No breakdown
7 has been observed up to 32 kV for the Ohmite Slim-Mox 104E resistor. For the Metallux HVR
8 969.23 resistors, no breakdown has been observed up to the limit of the test apparatus of 130 kV.
9 Since the Metallux HVR 969.23 resistors are significantly larger physically than the loop-to-loop
10 distance, they are mounted diagonally between each pair of field cage loops. They are held by
11 copper brackets, which are attached to studs welded onto the field cage tubes, shown in figure 17.



Figure 17: The Metallux HVR 969.23 resistors mounted on the 16 field cage loops closest to the cathode.

12 The second type of protection installed for all resistors on field cage loops 1 through 32 is a
13 surge protection circuit. The chosen surge protection devices are designed to short the circuit in
14 the case of a voltage spike, which protects any other electrical components installed in parallel.
15 Below their clamping voltage, they exhibit a very high resistance and do not influence the circuit.
16 The behavior of Gas Discharge Tubes (GDTs) and varistors in liquid argon has been studied exten-
17 sively for application in the MicroBooNE field cage [?]. The surge protection device chosen is a
18 Panasonic ERZ-V14D182 varistor with a clamping voltage of 1700V. In order to obtain a very high
19 resistance in normal operation and a clamping voltage above the 2 kV in normal operation mode,
20 three of these devices are mounted in series across a block of four Slim-Mox 104E or two Metallux
21 969.23 resistors. These additional varistor boards make electrical contact with the field cage via

- ¹ brass mounting brackets that are fastened to the field cage with button head screws, as shown in
² figure 18.

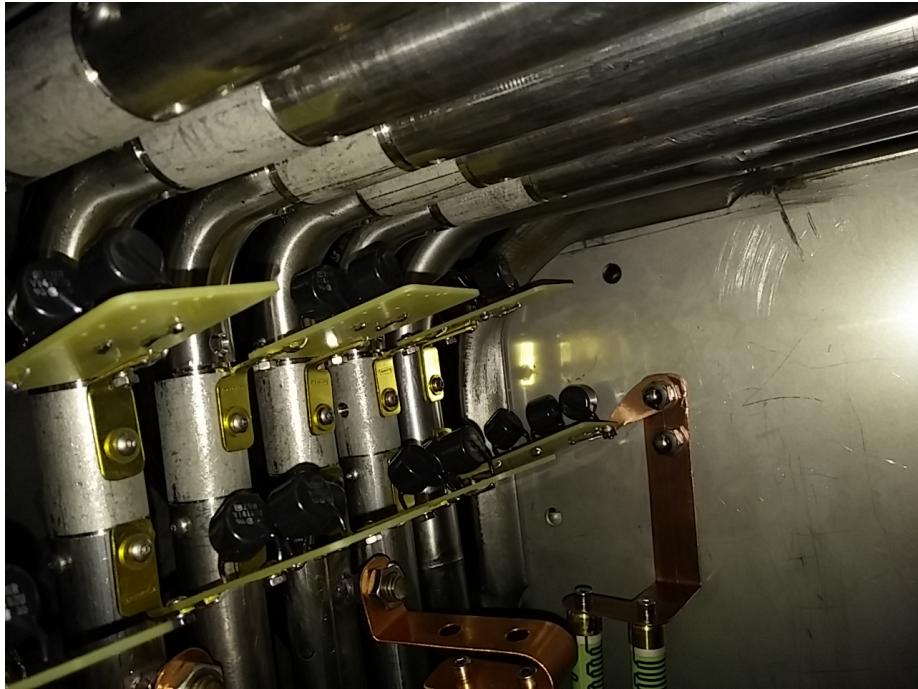


Figure 18: Surge-protecting varistors are installed in parallel with the voltage divider resistors for the first 32 field cage loops. Here, they are shown mounted on small boards in sets of 3, and attached to the field cage by means of 6-32 hex button head screws.

³ **4.3 Anode**

⁴ The anode frame holds the induction and collection plane sense wires at tension and provides
⁵ overall structural support for the beam-right side of the LArTPC. Individual sense wires for all
⁶ anode planes are held in place by wire carrier boards, which are printed circuit board assemblies
⁷ that locate the wires as well as provide the electrical connection to the electronic readout system of
⁸ the experiment.

⁹ **4.3.1 Mechanical structure**

¹⁰ The anode frame is comprised of a stainless steel C-channel hosting adjustable tensioning bars to
¹¹ which the wire carrier boards are attached. The C-channel and tensioning bar assembly is depicted
¹² for one corner of the anode frame in figure 19. Wire carrier boards attach to precision alignment
¹³ pins distributed along the length of the tensioning bars.

¹⁴ **4.3.2 Wire winding and quality assurance**

¹⁵ The three anode planes are constructed from wire carrier boards that have had individually-prepared
¹⁶ wires attached to them in groups of 16 (for the U- and V- angled planes) or 32 (for the vertical Y-
¹⁷ plane). Consistent quality in wire preparation was achieved by a semi-automated winding machine,

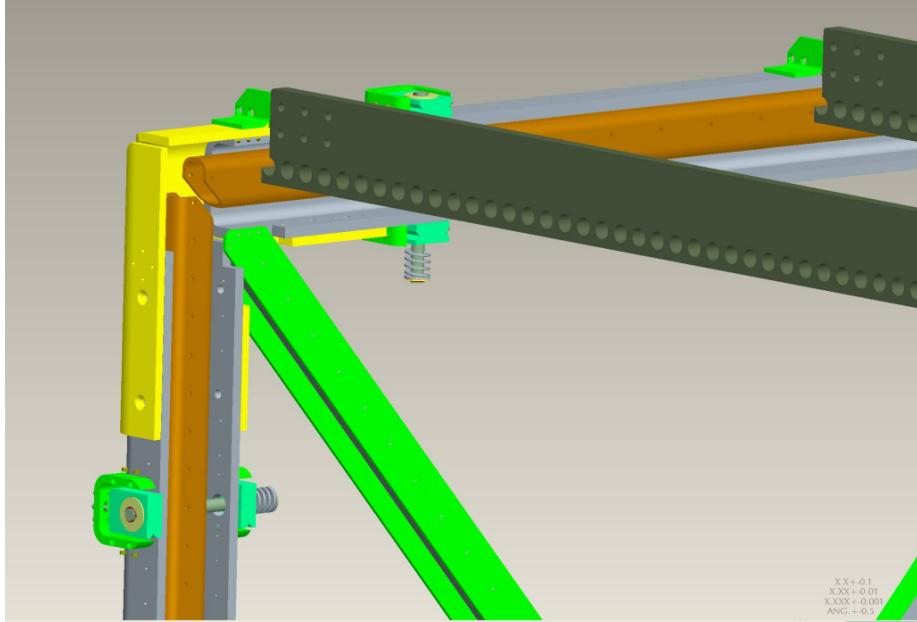


Figure 19: Rendering of the anode frame assembly. The C-channel is depicted in gray, and the adjustable tensioning bar assembly is shown in orange.

which terminated the ends of each wire via wrapping around 3 mm diameter brass ferrules as shown in figure 20. Each wire was tested for strength by being placed on a tensioning stand where a load of 2.5 kg (more than 3 times the nominal load of 0.7 kg) was applied for 10 minutes, ensuring that the wire preparation did not leave any weaknesses that could result in a breakage later on. Upon successful completion of the quality assurance testing for each wire, it was placed onto a wire carrier board, shown in figure 21

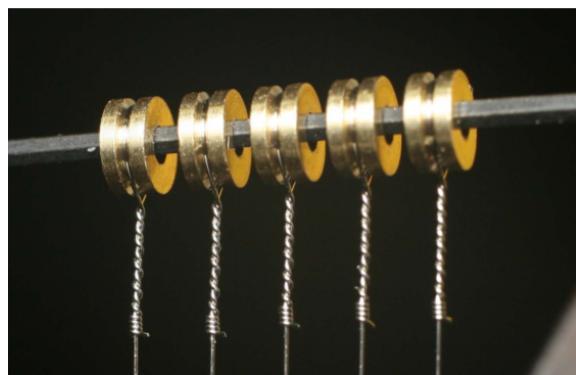


Figure 20: Picture of the wire termination on the brass ferrules. Each ferrule is 3 mm in diameter, and 1.5 mm thick.

When installed, the wires make contact with gold pins positioned on the wire carrier board, and these pins are connected to a trace that routes to the cold electronics described in section 6.1. After fully filling a wire carrier board with wires, a cover plate was installed and press-fit rivets were installed to hold the assembly together. The assembled wire carrier board was then placed

1 onto a tension stand, to reapply a 2.5 kg tension/wire to the whole board for 10 minutes. This is
2 to ensure that the wires were not weakened during the board assembly process. The tension stand
3 is depicted in figure 22. A comprehensive description of the MicroBooNE wire preparation and
4 associated quality assurance studies can be found in [?].



Figure 21: Picture of a collection plane wire carrier board that has been filled with wires, but has not yet had cover plate installed.

5 4.3.3 Wire installation and tension measurements

6 During detector construction the completed wire carrier assemblies were manually installed onto
7 the adjustable tensioning bars residing in the C-channel of the supporting anode frame. A team
8 of two people installed each assembly (consisting of wires and supporting carrier boards on either
9 end) onto the anode frame. The collection plane was the first installed, followed by the middle
10 induction plane, and then finally the inner induction plane. Once all three anode planes were
11 completely installed, the tensioning bars were adjusted and a survey was taken of the tension of all
12 anode wires. Tension was measured via measuring the resonant frequency of a laser beam reflected
13 off of a plucked wire and incident on a photodiode connected to a spectrum analyzer program [?].
14 The tensioning bars were adjusted iteratively until the surveyed tension of all wires was within a
15 range where no single wire was too taught or loose to create detector performance issues. Figure
16 shows the final surveyed tension of the wires within the LArTPC.

17 4.4 Parts Preparation

18 The majority of the parts that make up the LArTPC are either stainless steel or G-10. These two
19 material types, as well as any others used in the LArTPC, were tested in the Fermilab Materials
20 Test Stand (MTS) [41], whose purpose was to investigate the suitability of materials for use in
21 LArTPCs. The MTS confirmed that none of the materials used in the LArTPC assembly would



Figure 22: Picture of a collection plane wire carrier board on the tension stand.

1 contaminate the liquid argon. Before assembly, all LArTPC parts were cleaned according to the
2 procedures described in the following sections.

3 **4.4.1 Cleaning stainless steel**

4 The delivered stainless steel parts were often greasy due to machining, and those with holes or
5 interior cavities generally had a significant amount of trapped metal shavings due to the machining
6 processes. Many of the pieces also had markings from permanent ink pens, dirt smears, rust spots,
7 and/or dried oil from machining. These pieces were scrubbed with ScotchBrite 7447 general-
8 purpose hand pads before cleaning.

9 Parts that were small enough to fit in an ultrasonic bath were prepared according to the follow-
10 ing prescription. A pre-rinse with tap water was performed to remove particulate matter, followed
11 by deburring of sharp edges. The first ultrasonic wash was 15 minutes in heated distilled water
12 with a 3% solution of Citranox acid detergent [53]. After a first rinse in distilled water, a second

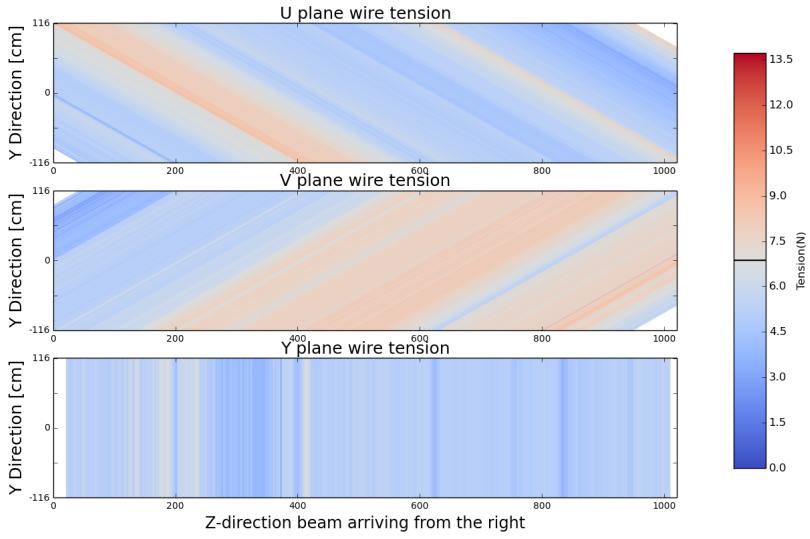


Figure 23: Final survey results for wire tension of the MicroBooNE LArTPC.

1 ultrasonic wash was performed, again for 15 minutes, but using heated distilled water with a weak
 2 solution of Simple Green detergent [54]. A second rinse in distilled water was performed, followed
 3 by a final rinse in a fresh bath of distilled water. The parts were then wiped dry with lint-free cloths,
 4 air dried completely, and wrapped in plastic film for storage.

5 Stainless steel parts which were too large to fit in the ultrasonic bath were prepared by a simpler
 6 prescription out of necessity. A pre-rinse with tap water was followed by deburring to remove sharp
 7 edges. Both the first and second washes were done with tap water and a weak solution of Simple
 8 Green detergent, scrubbing with brushes and lint-free sponges. Two tap water rinses were done,
 9 and the parts were then wiped dry with lint-free cloths. As a final additional step, each part was
 10 wiped with 200-proof ethyl alcohol, and then air dried completely. These parts were also wrapped
 11 in plastic film for storage.

12 **4.4.2 Cleaning G-10**

13 G-10 is known to absorb large quantities of water, which would outgas in the argon and could
 14 inhibit reaching the required argon purity in the detector. For this reason all G-10 parts were cleaned
 15 and then baked to remove moisture. The largest G-10 parts on the detector are beams that span the
 16 distance between the cathode and anode. These were washed in 500-gallon ultrasonic baths that are
 17 overseen by Fermilab Accelerator Division, typically used for cleaning large sections of accelerator
 18 beam pipes. An initial pre-wash was done with tap water to remove as much particulate matter as
 19 possible, since the machining process left a large amount of dust on the machined edges. Pieces
 20 were then placed in the ultrasonic bath with heated deionized water and a 2% solution of Elma-
 21 Clean 65 (EC 65) neutral cleanser [?]. Two ultrasonic bath rinses were performed, and the pieces
 22 were then sealed in plastic bags with clean dry nitrogen gas. In order to remove the absorbed water,
 23 the large G-10 parts were then transported to Fermilab Technical Division where they underwent
 24 an outgassing procedure to remove any remaining absorbed moisture. They were baked in a large

1 oven under vacuum until a plateau in the outgassing was reached, as reported by a monitor inside
2 the oven. Upon completion of the outgassing procedure, the parts were resealed in plastic bags for
3 storage.

4 **4.5 Assembly**

5 The full mechanical structure of the LArTPC is shown in figure 24, with the top image depicting
6 the cathode frame on the left, made semi-transparent to show the support structures on which the
7 cathode sheets are attached. The anode frame is on the right of this image, with I-beams configured
8 in a crossed pattern to maintain the shape and rigidity of the outer C-channel structure. Ribs of G-
9 10 connect the anode and cathode, electrically isolating them from each other while also providing
10 mounting holes to hold in place each of 64 field cage loops that define the active volume of the
11 LArTPC . The field cage loops are visible in the photograph in the bottom frame of figure 24.

12 Assembly was done inside of a clean tent, shown in figure 25, on a flat surface made up of
13 adjustable-height metal platforms that were installed on the assembly room floor. These platforms
14 were leveled to better than 0.5 mm before beginning assembly. The anode frame was the first part
15 of the detector to be assembled on this surface, shown in figure 26. It was temporarily placed
16 aside, and the cathode frame was assembled on the same set of platforms along with the G-10 ribs,
17 which stood vertically with the help of temporary unistrut support pieces. The combined cathode
18 and G-10 frame was then lifted and rotated to the proper orientation, with G-10 ribs extending
19 horizontally from the cathode to the anode, as shown in figure 27. Finally, the anode frame was
20 brought back over and attached to the G-10 ribs, and the stainless steel tubes that make up the field
21 cage loops were fed through the holes in the G-10 ribs to complete the mechanical structure of the
22 LArTPC .

23 **4.6 High Voltage System**

24 To create the drift electric field, an adjustable negative voltage (referred to as the "high-voltage"
25 or "HV") is supplied to the LArTPC cathode, generated outside of the cryostat by a Glassman
26 LX150N12 power supply. Before entering the cryostat, the output of the power supply is passed
27 through a current-limiting resistor series that serves as both a low-pass filter for the power supply
28 ripple, and a stored energy partition in case of a sudden reduction in voltage. The resistor series
29 is a set of eight $10\text{ M}\Omega$ resistors submerged in a transformer oil in an aluminum container. This
30 assembly was successfully tested to -200 kV. The capacitance of the cable is $\sim 50\text{ pF/m}$ making the
31 time constant of the component upstream of the LArTPC $\sim 35\text{ ms}$.

32 The potential is introduced into the cryostat by a custom-designed HV feedthrough. The
33 feedthrough is based on an ICARUS design [7]; an inner 1.0 in diameter stainless steel conductor
34 carrying the HV is surrounded by an insulating tube, 2.0 in outer diameter and 1.0 in inner
35 diameter, of ultra-high molecular weight polyethylene (UHMW PE), which is further encased in
36 an outer ground tube. A photograph and drawing of the production feedthrough are shown in
37 figure 28.

38 MicroBooNE's voltage target combined with its geometry of a cylindrical cryostat with a
39 single drift region necessitate that the HV feedthrough extend deeper into the cryostat than the
40 ICARUS feedthrough to avoid high field regions. Figure 29 shows the HV feedthrough extending

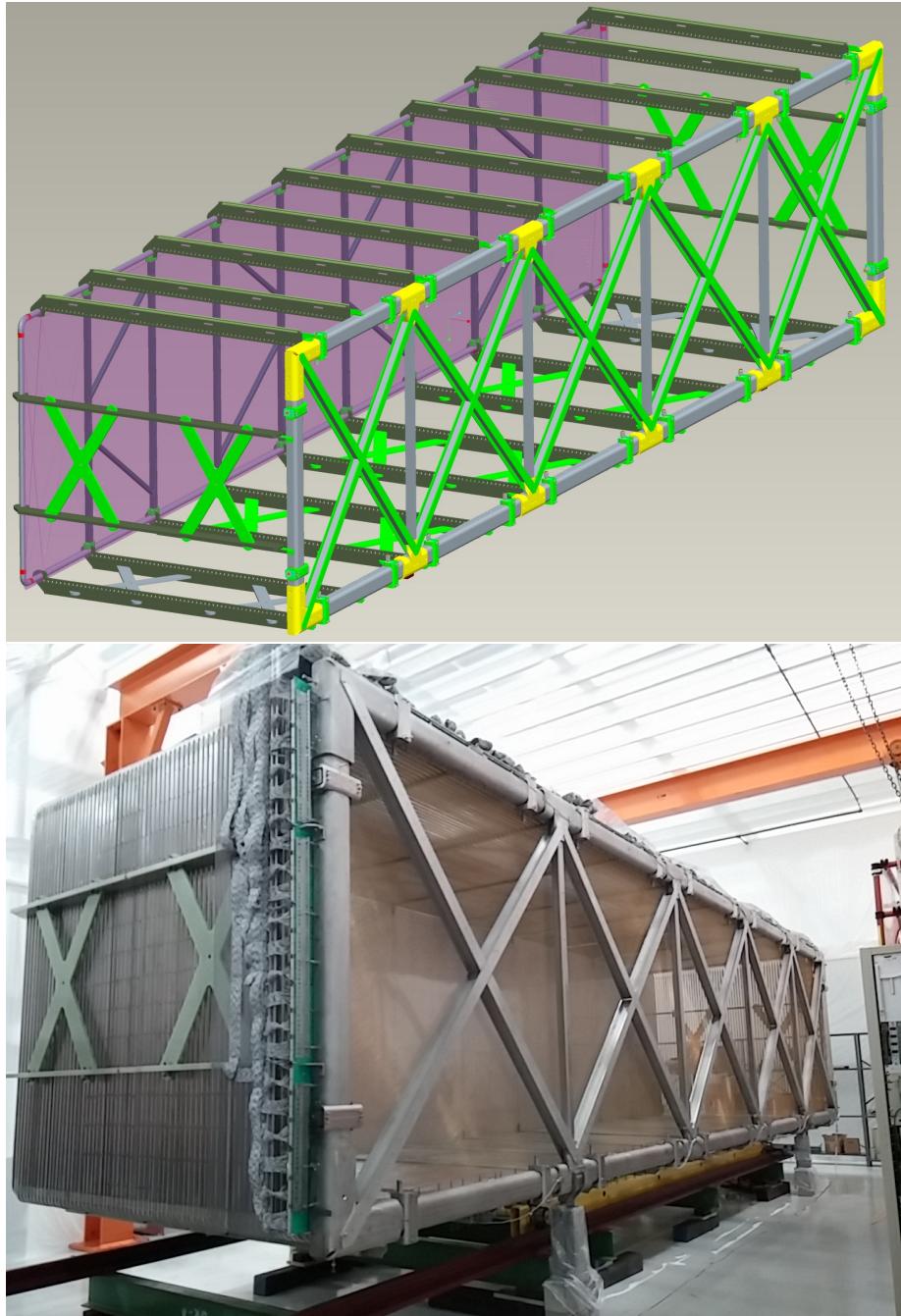


Figure 24: Top: Rendering of the full LArTPC frame assembly. Bottom: Assembled LArTPC after wire and electronics installation.

1 into its receptacle cup attached to the LArTPC cathode. The lower termination of the outer ground
 2 tube is a torus chosen to reduce the electric field between both the feedthrough and the cathode
 3 plane and along the feedthrough itself. The length of exposed UHMW PE is machined with grooves
 4 in an effort to improve performance. The electrical connection to the cathode is accomplished with
 5 a hemispherical spring-loaded tip attached to the inner conductor of the feedthrough. When the



Figure 25: Clean tent where MicroBooNE LArTPC assembly was conducted.



Figure 26: Anode frame in the process of being moved from the metal assembly platforms.

- ¹ feedthrough is installed on the cryostat, the spring-loaded tip makes contact with the receptacle
- ² cup attached to the cathode.



Figure 27: Cathode frame and G-10 ribs on metal assembly platforms.

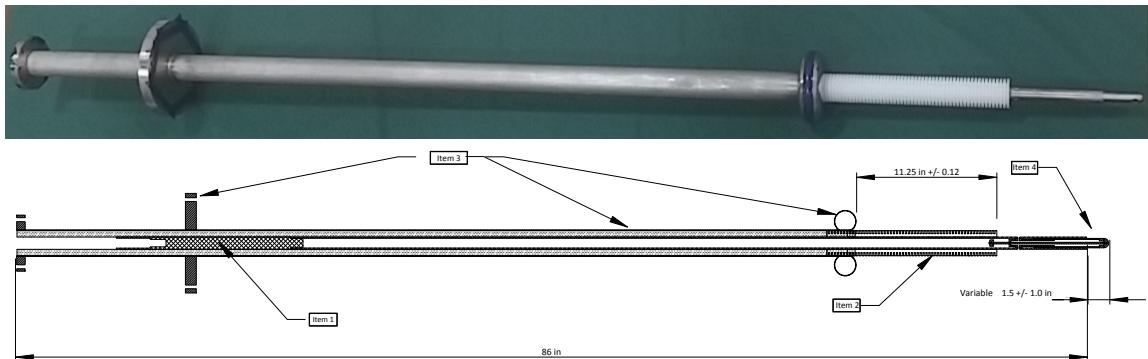


Figure 28: Photograph and drawing of the production HV feedthrough. The spherical probe tip is attached to the end of the inner conductor, on the right side of these figures.

5. Light Collection System

Liquid argon is a very bright scintillator, and collecting the light produced when charged particles travel through the LArTPC is an essential aspect of fully understanding the interactions taking place in the detector. This section describes the MicroBooNE light collection system.

5.1 Motivation and General Design

The installed system is designed to meet MicroBooNE's physics goals for light collection, which are twofold. First, for accelerator-induced events, the light collection system is required to trigger



Figure 29: The production HV feedthrough inserted into the cathode receptacle cup inside the cryostat.

1 on 40 MeV kinetic energy protons produced in neutrino interactions. Second, for non-beam events,
2 efficient observation of 5 to 10 MeV electrons from supernova neutrinos are required.

3 The light produced by neutrino interactions in MicroBooNE is an important input for both
4 event selection and reconstruction. One of the critical capabilities the light collection system pro-
5 vides is the ability to form a beam-event trigger when a pulse of light is observed in coincidence
6 with the beam spill. Because a vast majority of beam spills will not produce a neutrino interaction
7 in the detector, such a trigger will substantially reduce the data output rate. For non-beam physics
8 studies, the light system provides triggering and an event interaction time (t_0). For accelerator-
9 induced events, the start time of a beam spill is, in principle, sufficient. However, because of the
10 long window over which the ionization electrons of an event drift (about 1.6 ms maximum drift
11 time at 500 V/cm field), there are many accelerator-induced events in which a cosmic ray muon
12 crosses the detector during the drift time. Utilizing the more accurate t_0 from the light collec-
13 tion system and the position of hits in the photodetectors, one can better reject cosmic ray muon
14 tracks. The light can also be used to trigger and select specific types of cosmic-ray calibration
15 events (Michel electrons, straight-through muons, etc.) and non-beam events (supernova neutrinos,
16 cosmic background events to proton decay studies, etc.).

Table 6: Important properties of scintillation light in liquid argon that affect detection in MicroBooNE.

Property	Note	Ref.
Wavelength	128 nm	[?]
Singlet, Triplet state time constants	6 ns, 1.6 μ s	[?]
Photons/MeV for $E = 500$ V/cm	24000	[?]
Triplet lifetime quench due to N ₂ at 1 ppm	20%	[?]
Attenuation length due to N ₂ at 1 ppm	66 m	[47]
Rayleigh Scattering Length	\sim 90 cm	[?]
Transparency of the wireplanes at normal incidence	86%	-

1 Light produced in liquid argon arises from two processes: scintillation and Cherenkov radiation.
 2 Scintillation light is produced by the formation and eventual radiative decay of excited argon
 3 dimers (or eximers) and is emitted in an isotropic distribution. Liquid argon is an excellent scintil-
 4 lator: it produces a large amount of light per unit energy deposited (about 40,000 per MeV at no
 5 drift field) and is transparent to its own scintillation. The scintillation light has a prompt and slow
 6 component with decay times of about 6 ns and 1.6 μ s, respectively. The two lifetimes correspond
 7 to the two lowest-lying eximer states with the prompt component coming from the decay of a sin-
 8 glet state and the slow from the decay of a triplet state. The prompt to slow ratio is about 1:3 for
 9 minimum ionizing particles and varies with ionization density and particle type. Both components
 10 consist of photons with a wavelength of 128 nm.

11 There is a significant uncertainty on the expected triplet lifetime [?] and this may be further
 12 modified by quenching (non-radiative dissociation of excimers by impurities) [?]. Other factors
 13 that can affect the arrival of the light include Rayleigh scattering, absorption by impurities, and
 14 obstructions. For detailed discussion of the physics of scintillation light production and propagation
 15 in MicroBooNE, see Ref. [?]. Table 6 summarizes information about the scintillation light.

16 Because scintillation photons have a wavelength of 128 nm they are very difficult to detect
 17 using conventional photodetectors. Figure 30 summarizes the challenges involved in detection of
 18 the 128 nm scintillation light. In order to detect the scintillation light, with spectrum shown in
 19 red, the VUV photons must be shifted into the visible region. MicroBooNE employs tetraphenyl-
 20 butadiene (TPB). This organic fluor absorbs in the UV (green line) and emits in the visible with a
 21 peak at 425 ± 20 nm (green hatched region), the peak wavelength having a slight dependence on the
 22 micro-environment of the fluors. This is a favorable wavelength for detection by the photomulti-
 23 plier tubes (PMTs) employed by MicroBooNE. The efficiency for transmission through borosilicate
 24 PMT glass (black) and the quantum efficiency of the cryogenic tubes used in MicroBooNE (blue)
 25 are overlaid on the TPB spectrum.

26 The light collection system consists of primary and secondary sub-systems. The primary light
 27 collection system is made up of “optical units”, each one consisting of a PMT located behind a
 28 wavelength-shifting plate. In total, 32 optical units were installed, yielding 0.85% photocathode
 29 coverage. The secondary system consists of four light guide paddles. These paddles are introduced
 30 for R&D studies for future LArTPCs, and are placed near the primary optical units to allow a
 31 comparison of their performances. A flasher system, used for calibration, consists of optical fibers

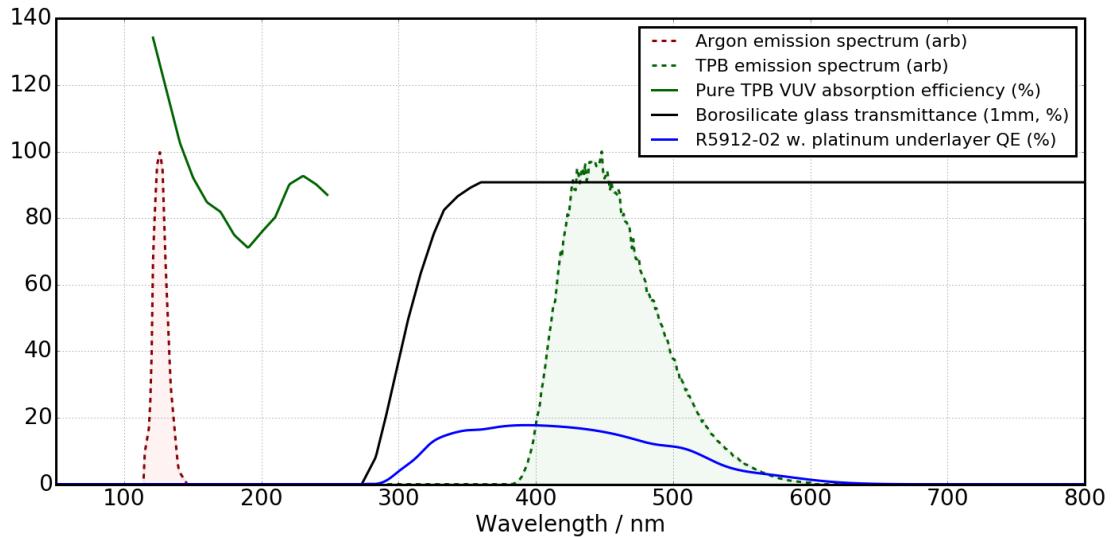


Figure 30: Scintillation light emission spectrum (red) and TPB re-emission spectrum (green), in arbitrary units. Superimposed are important efficiencies (see y axis): Dark green line – absorption of VUV light by TPB; Black line – transmission of borosilicate glass; Blue line – efficiency of a R5912-02mod cryogenic PMT [?]

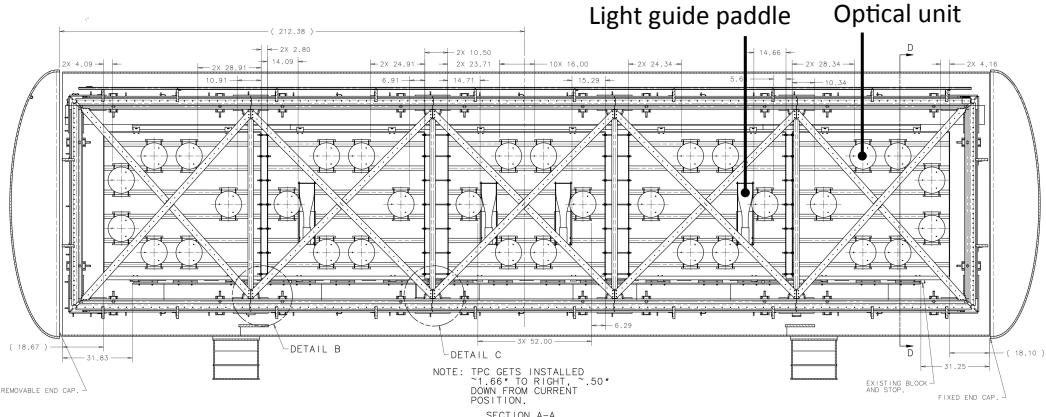


Figure 31: The MicroBooNE light collection system consists of a primary system of 32 optical units and a secondary optical system of four lightguide paddles [55]. These are mounted behind the anode wire planes such that the view is not obscured by structural cross bars of the LArTPC.

1 bringing visible light from an LED to each PMT face.

2 The light collection detectors are located in the y - z plane behind the anode planes of the
3 LArTPC, as shown in figure 31. The combined transparency of the three anode planes is 86%
4 for light at normal incidence. This transparency value assumes 100% of VUV photons impinging
5 on the wires ($150\ \mu\text{m}$ diameter and 3 mm pitch) are absorbed. The detectors were placed so as
6 not to be obscured by the LArTPC structural cross-bars, shown in figure 31. Locating the light
7 detectors behind the anode plane places them in a very weak electric field due to the +440 V bias of

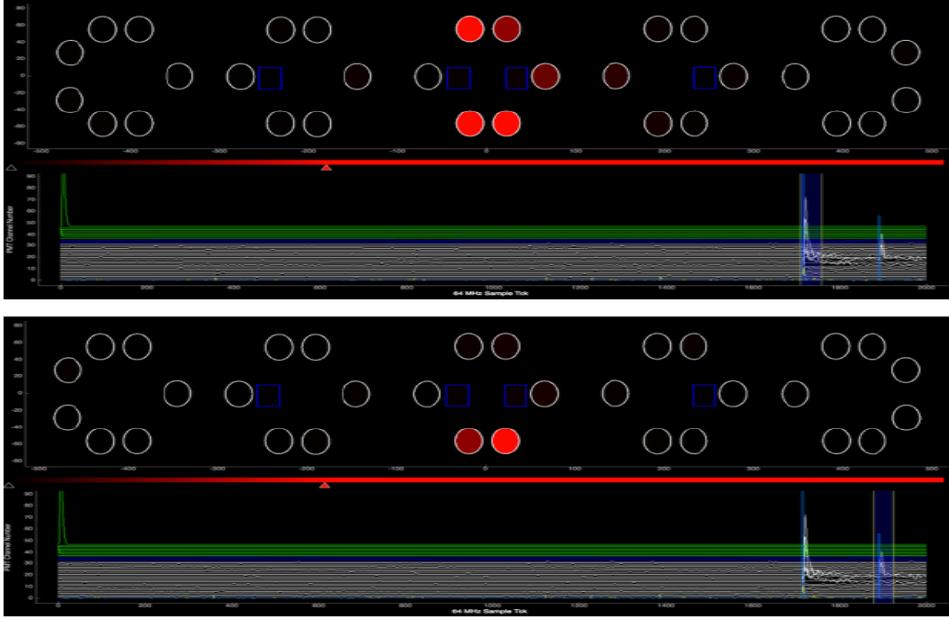


Figure 32: Two sequential event displays for the light collection system. The sequence is consistent with a muon that stops (top) and decays (bottom). The circles correspond to the optical units. Red circles indicate those units with hits. The waveforms versus time are shown below.

the collection plane. To test for an effect from weak electric fields, the response of a PMT placed between a +700 V mesh and ground, separated by 50 cm, was studied. The PMT zenith was 25 cm from the HV. No effects on the signal were observed. This was expected, as the photocathode is held at ground, effectively acting as a Faraday cage.

Throughout the design and construction of the light collection system, substantial R&D was performed. The reader should refer to Refs. [56, 57, ?, ?, 58, 47, 59, 55, 60–64] for detailed results of these studies. A useful overall review is available in Ref. [?].

Figure 32 shows the light observed in two sequential events, consistent with a muon entering the detector followed by a Michel electron from the decay. One can see that the light is relatively well localized. This allows the light to be correlated with specific tracks in the detector. This “flash-track matching” is used to identify and reconstruct the tracks that are in time with the beam spill—an important goal of the light collection system.

5.2 The Primary Light Collection System

Each of the 32 optical units of the primary light collection system consist of a cryogenic Hamamatsu 5912-02MOD PMT seated behind an acrylic plate coated with a TPB-rich layer and surrounded by a mu-metal shield. Figure 33 shows a diagram of one unit (left) and a photograph

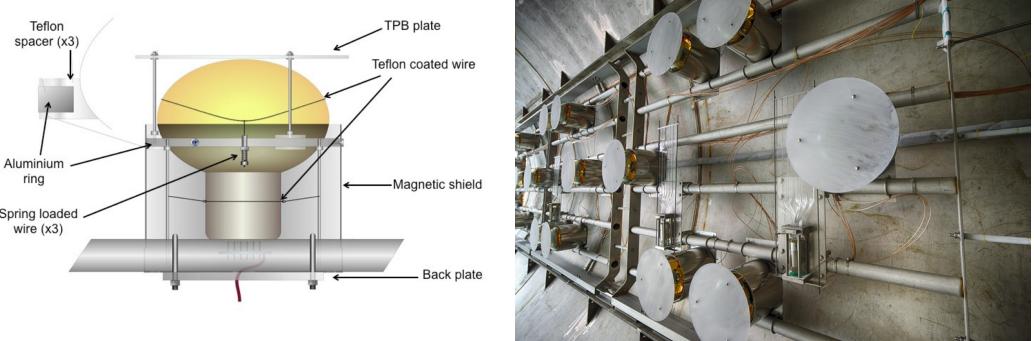


Figure 33: Left: diagram of the optical unit; Right: units mounted in MicroBooNE, immediately prior to LArTPC installation.

of installed units (right). Past experiments have directly coated PMTs with wavelength shifter [?]. However, the MicroBooNE design separates the PMT from the wavelength-shifting plate for simplicity of quality control and installation. This proved important, as R&D indicated that TPB is particularly vulnerable to environmental degradation (see section 5.2.3). In this section, description is provided for each component of the optical unit, as well as for the overall assembly.

5.2.1 Photomultiplier Tubes and Bases

Ref. [58] provides detailed information on the selection and testing of the 8" Hamamatsu R5912-02mod cryogenic PMTs employed in MicroBooNE. In this section a brief summary of the findings from this testing is presented.

The R5912-02mod employs a bi-alkali photocathode. Because the PMT is designed for cryogenic use, the R5912-02mod also features a thin platinum layer between the photocathode and the borosilicate glass envelope to preserve the conductance at low temperatures. While this allows the PMT to function below 150 K [?], absorption in the platinum reduces the efficiency of the PMT by 20%. Figure 34 provides quantum efficiency curves for the 8 inch photomultiplier tubes. The manufacturer's specifications do not include the effects of the platinum photocathode coating, but a wavelength dependent quantum efficiency was provided by Hamamatsu for 4 of the 32 installed PMTs. The mean and standard deviation of these curves is shown in figure 34.

The R5912-02mod is a 14-stage PMT. The high gain at room temperature (10^9 at ~ 1700 V), compensates for known reduced gain at 87 K in the liquid argon. The high gain also has the additional advantage of allowing operation at lower than nominal voltage which reduces heat-loss in the cables and the potential for high voltage breakdown at the feedthroughs.

The PMT base is designed such that the photocathode is grounded and the last dynode in the chain is held at large, positive voltage. Thus the PMT bulb, which is closest to the LArTPC anode plane, is at ground and does not disturb the electric field on the wires. The result is that the high voltage (HV) can be provided and the signal can be extracted from the PMT using a single cable, reducing the cable volume in the vapor region and removing a possible source of out-gassing water impurity.

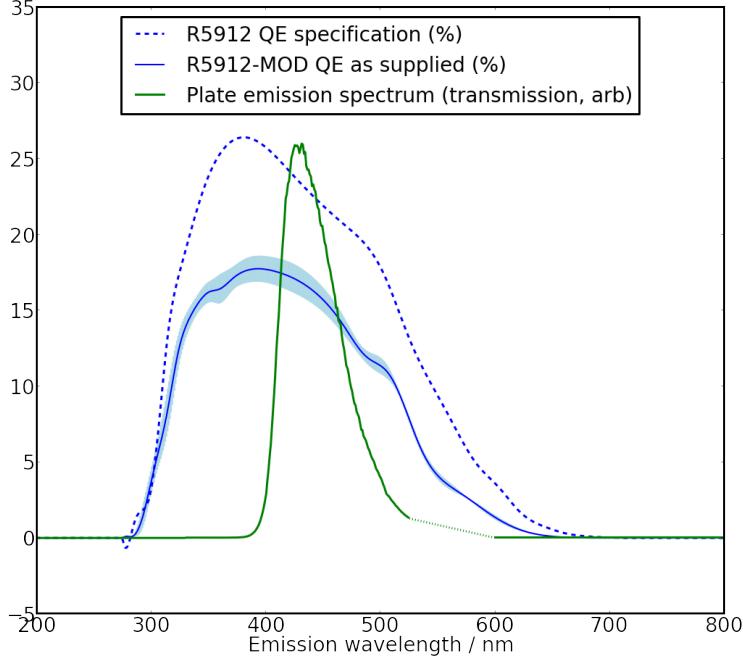


Figure 34: The specification for the non-platinum undercoated PMT from [65]. The blue band shows the mean and standard deviation of the four quantum efficiency curves provided by Hamamatsu for installed MicroBooNE PMTs. Also shown is the measured emission spectra of MicroBooNE wavelength-shifting coatings, discussed in section 5.2.2. [?]

1 The flat PC-board base was made of Rogers RO4000-series woven glass-reinforced laminate,
 2 which is the same material with the LArTPC cryogenic front end boards. The base is attached ~ 1
 3 cm from the bottom of the PMT and is the closest possible distance of safe approach to the PMT
 4 vacuum seal tip. A schematic and photos of the PMT base are provided in Ref. [58]. The passive
 5 components include only metal film resistors and C0G/NP0 capacitors, which have the minimum
 6 temperature coefficients, and the performance at cryogenic temperatures were tested. Because the
 7 supplied HV and return signal share a single cable, the signal must be split from the HV through
 8 an AC-coupling capacitor, as is discussed in section 5.2.5.

9 All installed PMTs were tested in a PMT test stand, both at room temperature and in liquid
 10 nitrogen which is at 77 K. The details are described in Ref. [58]. In brief, the test stand consisted of
 11 a light-tight 346 L, liquid nitrogen filled dewar into which up to four PMTs could be installed. The
 12 PMTs were immersed and maintained in the dark environment for up to three days before most
 13 measurements of the dark rate and gain were performed. A fiber brought in light from a pulsed
 14 blue LED, which was tested for linearity with bias voltage.

15 Among the important results from cryogenic testing were the following [58]:

- 16 • The PMTs could be ramped to voltage quickly in the cold environment, and after 30 minutes,
 17 the gains were found to be stable.
- 18 • If the room temperature PMTs were immersed in liquid nitrogen, dramatic changes in the
 19 gain were observed after initial turn-on. The PMT gain remained high in the first ~ 5 hours

1 after immersion, and then suddenly dropped by more than a factor of two, afterwards reaching
2 a stable value with a small drift.

- 3
- 4 The PMT response showed good linearity up to 100 photoelectrons (PE), which was the maximum attainable by the PMT test stand LED.

5

 - 6 The HV for each PMT was selected to produce a gain of 3×10^7 in liquid nitrogen, and was typically chosen to be ~ 1300 V.

7

 - 8 The dark current plateaus extended up to 1800 V in liquid nitrogen, and the dark current is higher in liquid nitrogen than at room temperature.

9

 - The PMT performance depended on rate of the pulsed LED.

10 No PMTs were rejected on the basis of the testing. However, there were three unexpected results
11 to note here.

12 The first unexpected behavior was that, at room temperature, most of the PMTs showed gains
13 that were 10 to 30% higher than manufacturer's specifications. As expected at cryogenic temperatures,
14 the gain is reduced by $\sim 10\%$ to 50%. To measure the PMT gain, the LED was set to produce
15 one to two PEs. The gain was found from the separation of the single PE peak from the pedestal,
16 where the single PE response is fitted using the procedure described in Ref. [66].

17 Second, it was found that the PMTs are noisier in the cryogenic environment. It would typically be expected that thermal emission is suppressed at cryogenic temperatures and one would expect a lower dark current for PMTs operating in this regime. However, the dark current measured in the liquid nitrogen is higher than at room temperature. Although the cause is unknown,
21 this phenomenon has previously been observed, see Ref. [67]. A proposed explanation is provided
22 in Ref. [68].

23 Third, an LED pulse-rate-dependent gain shift was found during testing, as shown on figure 9
24 of Ref. [58]. This behavior is described qualitatively in [69]. With 10 kHz LED pulsing, the gains
25 of cold and dark-adapted PMTs were shown to steadily increase, requiring nearly 24 hours from
26 turn-on to stabilize. The effect was not observable at 10 Hz. This is relevant to MicroBooNE
27 because the cosmic muon rate in the MicroBooNE detector is ~ 5 kHz. Therefore a similar effect
28 is expected in the MicroBooNE detector. Preliminary results from measurements of the PMTs
29 installed in the LAr-filled MicroBooNE cryostat shows the expected effect.

30 5.2.2 Wavelength-Shifting Plates

31 In order to be sensitive to 128 nm VUV liquid argon scintillation photons, the optical assemblies use
32 a wavelength-shifting coating to convert this VUV light to visible wavelengths that are detectable
33 by photomultiplier tubes. In MicroBooNE, the active ingredient of this coating is TPB, an organic
34 fluor which absorbs efficiently in the vacuum ultraviolet with an emission spectrum peaked around
35 425 nm [70] as shown in figure 30.

36 The optical unit PMTs observe light transmitted through TPB-coated, 12-inch diameter acrylic
37 plates (see figure 35). The coating consists of a 1:1 TPB-to-polystyrene ratio, with 1 g of each
38 dissolved in 50 ml of toluene. A small amount of ethyl alcohol is added as a surfactant. The coating
39 is applied to the acrylic plate in three layers by brush-coating. The solution dries in air at room

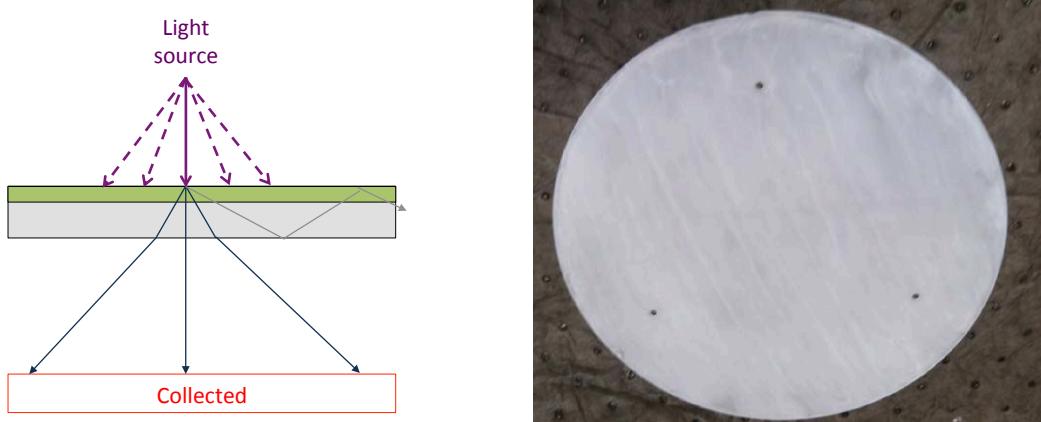


Figure 35: Left : Illustration of transmission mode, used by the optical units. Right: photograph of a coated plate.

temperature. This leads to a final layer which is oversaturated with TPB, and white crystals form on the surface as the coating dries. The presence of surface crystallization gives the MicroBooNE plates a white, opaque finish. Details of the process are described in Ref. [71].

The emission spectrum in figure 34 was measured using a Hitachi F-4500 fluorescence spectrophotometer with an incident beam of wavelength 270 nm, selected by a diffraction grating from a xenon lamp. A standard rhodamine dye calibration sample was used to correct for drift in the lamp and spectrometer. The measurement was made in transmission mode, and an artifact peak at a harmonic of the twice incident wavelength was observed between 525 and 600 nm. This region is omitted from the reported spectrum of figure 34.

As shown in figure 34, although the absolute quantum efficiency for the platinum-undercoated cryogenic PMT is lower than the non-cryogenic version, the wavelength dependence is similar, and overlap between the TPB emission spectrum and the sensitive wavelength range of the PMT remains high. Using the measured TPB emission spectrum for plate coatings and the PMT quantum efficiency curve provided, the spectrum-averaged PMT quantum efficiency is $15.3 \pm 0.8\%$ per visible photon incident on the photocathode.

The wavelength-shifting performance of the coating was measured at 128 nm relative to evaporative coatings of the type studied in [72]. Coating efficiencies were measured as a function of wavelength between 128 and 250 nm using a vacuum monochromator at room temperature. They were also measured in liquid argon using 128 nm scintillation light, relative to the same evaporatively coated plate. These data are shown in figure 36, and more information about these measurements can be found in [71].

The absolute efficiency of the MicroBooNE coatings can be obtained by multiplying the relative efficiencies of figure 36 by the measured absolute efficiencies from [72], and accounting for the temperature dependence in the wavelength-shifting efficiency of pure TPB reported in [73]. The expected efficiency of MicroBooNE coatings is found to be 0.98 ± 0.17 emitted visible photons per incident 128 nm photon for the plate coating.

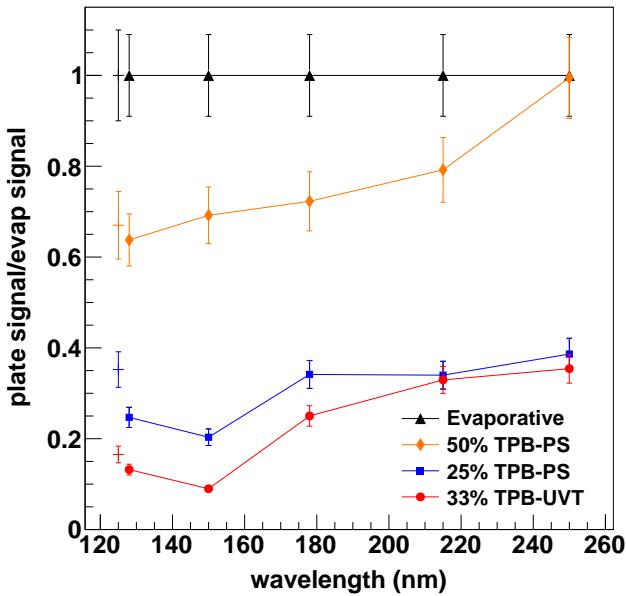


Figure 36: Measured efficiencies of various wavelength-shifting coatings, from [71]. In this plot, 50%TPB-PS is the MicroBooNE plate coating used in the optical units. The 33%TPB-UVT is the light guide coating for the secondary light collection system described below. Connected points were measured in a vacuum monochromator at room temperature, and non-connected points were measured in liquid argon with 128 nm scintillation light. All points are normalized to the performance of an evaporatively coated plate.

5.2.3 UV Light Protection for the Wavelength-Shifting Plates

TPB coatings have been shown to degrade under exposure to ultraviolet light [74] through a radical mediated photo-oxidation to the UV-blocker and photo-initiator benzophenone [75]. Several measures were taken to ensure that degradation was minimized during the construction of the experiment. The TPB powder and coated elements were stored in the dark at all times, with coated plates and light guides being kept wrapped in foil and stored in a dark container before installation. The detector construction area was covered with a UV blocking plastic [76], and test plates were placed at various positions in the clean tent to check for degradation from stray light. After several weeks of exposure, one test plate with a clear line of sight to the tent entrance demonstrated a few percent degradation, and all others showed no observable loss of efficiency. The open end of the MicroBooNE cryostat was shielded from light by a black curtain after installation, and the feedthroughs of the cryostat were blocked when not in use to prevent stray light from entering. The coated plates were the final component of the optical system to be installed into the detector to give the minimum possible light exposure during the detector construction process.

5.2.4 Cryogenic Mu Metal Shields

The trajectories of electrons within the 8" PMT can be deflected by the Earth's magnetic field. This effect can be reduced or removed by surrounding the PMT with mu metal, a metal of low magnetic

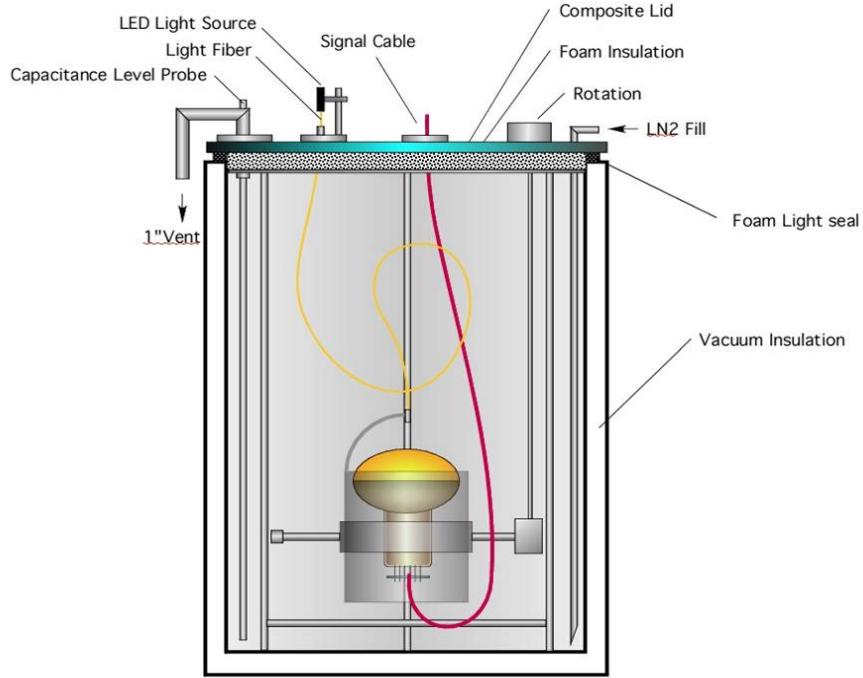


Figure 37: Schematic of the system used to study the mu-metal shields. The design allows rotation along all three axes.

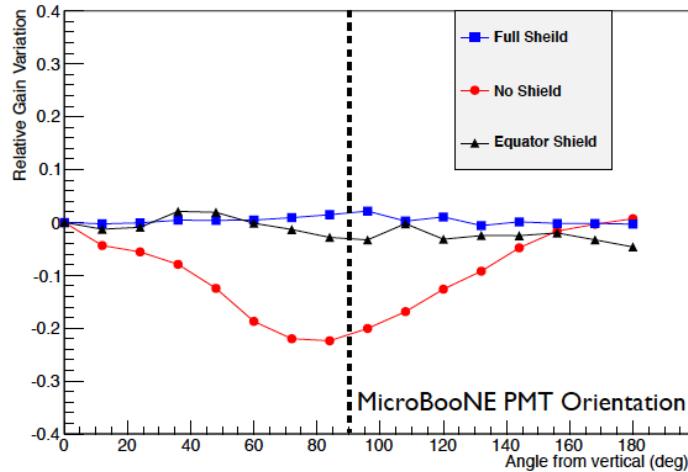


Figure 38: Red: Angular dependence of a PMT response with no shield; Black: for a shield that reaches the tube equator; Blue: for a shield that fully covers the tube to the zenith.

permeability. Commonly used mu-metal fails to provide shielding at cryogenic temperatures. Two types of cryogenic mu metal, Cryoperm 10 and A4K, both products of Amuneal, were identified that did provide shielding at cryogenic temperature.

The mu metal shields were tested in the apparatus shown in figure 37. The system allowed for the PMT to be positioned at an angle relative to the vertical axis, with the rotator set to 30 positions

1 from 0 to 348° . The set-up was on a dolly that allowed for rotation about the vertical axis. PMT
2 tests were performed in air and in liquid nitrogen. PMTs were dark adapted for 5 hours before
3 testing. A blue LED provided 1 to 2 single PEs of light through an optical fiber. The fiber was
4 fixed to the PMT mount such that the endpoint was stationary with respect to the tube as the system
5 was rotated.

6 The mean charge from the PMT as a function of angle was recorded. The error was primarily
7 systematic. The $\sim 10^6$ to LED pulses per data point give $< 1\%$ statistical error on the mean.
8 However, 24-hour studies of a single point showed $\sim 5\%$ variations in collected charge due to
9 gain drift. Results showed that A4K and Cryoperm function essentially identically, to within the
10 measurement error, and so MicroBooNE chose A4K, based on significant cost savings. A plot of
11 the relative PMT gain variation versus angle from vertical, figure 38, shows that adding the shield
12 significantly improves PMT performance. This plot shows the effect of the A4K shield with the
13 height aligned with the equator of the PMT (black) and aligned with the zenith of the bulb (blue),
14 compared to no shield (red) as a function of PMT azimuthal angle. Given the 5% systematic error,
15 the two shield positions are indistinguishable.

16 As a result of these tests, the A4K shields were designed to extend just past the equator of the
17 PMT. The shield has small holes in the backplate that allows the PMT cables to exit the shield.
18 MicroBooNE is the first LArTPC to use cryogenic mu metal shields in its light collection system.

19 **5.2.5 Implementation of the Primary System**

20 The light collection system is composed of optical units assembled as shown in the left-hand picture
21 of figure 39. The PMT is seated within the mu metal shield on three teflon pads attached to an
22 equatorial support ring. The neck of the tube slides inside a loose wire guide-loop that prevents
23 the PMT from tipping. The PMT is held within this assembly using teflon-encased wires that
24 extend across the bulb and connect to wire hooks attached to the equatorial ring with stainless
25 steel springs. Legs extending from the support at the equator are screwed into a backplate for final
26 mounting. Concern about differences in contraction of the materials led to this design which holds
27 the PMT in place, but with only moderate rigidity. The units were tested to ensure the PMTs would
28 not be displaced during installation and filling. Three posts extend upward from the equatorial
29 ring to hold the plate 3.0 cm above the apex of the bulb. The optical units slide into a cylindrical
30 mu-metal shield, which screws into the equatorial ring. The unit is then mounted on stainless steel
31 back-plates affixed to a support rack, as shown in the right-hand picture of figure 39.

32 The support rack consists of five stainless steel components, or modules, for ease of instal-
33 lation. Each module has vertical height 1.83 m and horizontal length 2.07 m, resulting in a total
34 horizontal length of 10.36 m. Unlubricated Thomson bearings fitted to the lower edge allow each
35 module to slide into the cryostat on rails mounted in the vessel. The system was designed to allow
36 the light detection system to slide into the vessel after the LArTPC was installed. However, in the
37 end, scheduling permitted installation of the system before the LArTPC installation. This had the
38 advantage of making installation and surveying easier, but the drawback that the system would be
39 exposed to UV light for a long period. Therefore, the units were installed with dummy clear acrylic
40 plates, and the TPB coated plates were installed only just before the LArTPC was moved in and
41 the detector could be easily protected from light. During optical unit installation, each rack module
42 was supported by a temporary mounting rail. The optical units were then mounted in positions

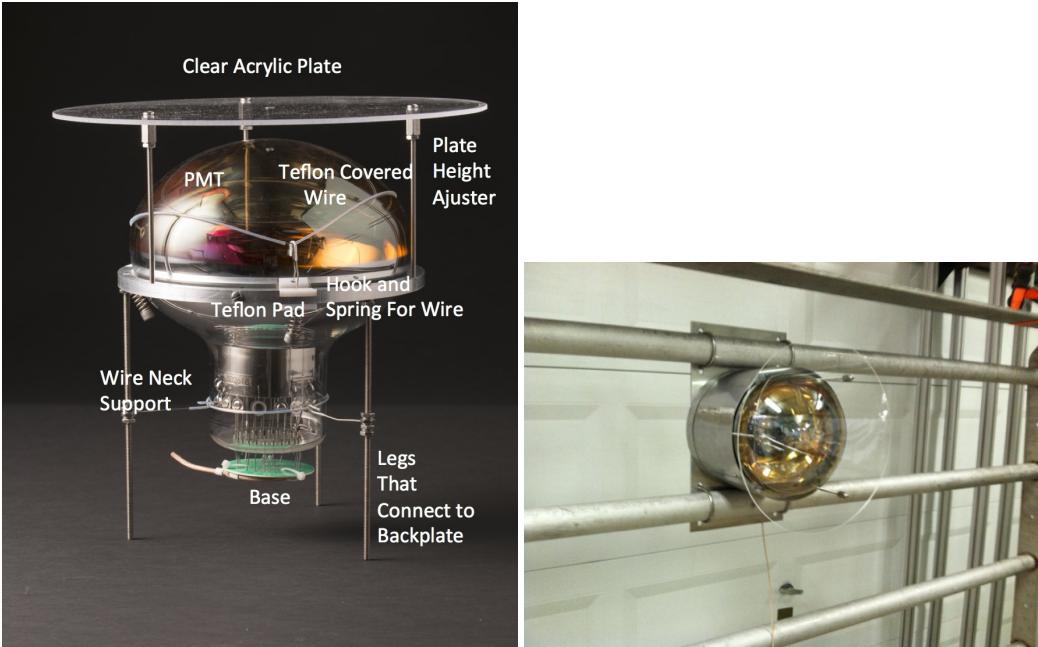


Figure 39: Left: The optical unit mount internal to the shield, with components labeled; Right: Unit mounted on rails. The clear plates were replaced with TPB-coated plates immediately before LArTPC installation, as discussed in the text [58, 55].

1 chosen to avoid obstruction by the LArTPC cross-bars, as shown in figure 31. As the units were
2 mounted and slid into the cryostat, the cables were loosely tied to the bars of the rack for support
3 and constraint.

4 The “splitter” circuit, located outside of the cryostat, is shown in figure 40. The splitter sepa-
5 rates the HV of the PMT from its output signal which is subsequently split into a high-gain (HG)
6 and a low-gain (LG) channel. The HG and LG channels respectively carry 18% and 1.8% of the
7 output signal. This allows a wide dynamic range for ADC readout of the PMT pulses. The ca-
8 pacitance was chosen to minimize reflections, since the bases are not back-terminated. The HV is
9 supplied to the splitter using BiRa Corporation, Model 4877PS modules.

10 The PMT cable system delivers HV and returns signals between the external splitter and the
11 optical unit in the cryostat. A single cable runs from an external connector, through the feedthrough
12 filled with Stycast 2850 epoxy [?], into the cryostat and to the PMT base. The RG316/U coaxial
13 cable has 50 Ohm impedance. Cables were terminated with Pasternack PE4498 SHV to accom-
14 modate that the cable carries HV to the tube as well as signals from the tube. The cable carries
15 an AC voltage rating of 1100 V; however tests showed the DC rating to be at least three times
16 higher and so suitable for this use. The cables were routed through feedthroughs consisting of a
17 pipe filled with solidified epoxy mounted on a conflat disk. On the warm side of the feedthrough,
18 the cables were terminated at a patch panel with SHV connectors. The SHV connector impedance
19 has a negligible effect on the 20-30 ns PMT signals. SHV cables connect the patch panels to the
20 splitters. The impedance of every channel was tested at the feedthrough patch panel for a stable
21 and correct value for the base resistance, which was $4.04 \pm 0.02 \text{ M}\Omega$ for the 8-inch PMTs.

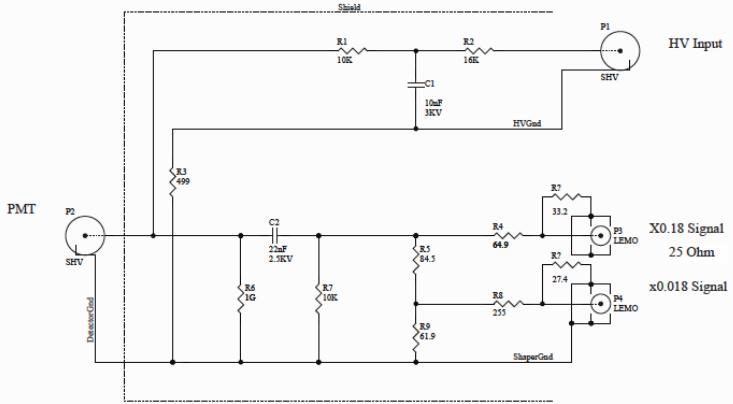


Figure 40: The "splitter" circuit. The circuit connects the HV source to the PMT. It also provides a pathway for signal pulses from the PMT to reach the readout electronics via an AC-coupling capacitor (C_2). The signal is split into two copies, one provided with an attenuation factor of 0.18 and another at 0.018. Both signal sources are recorded by the readout electronics in order to provide two dynamic ranges.

5.3 The Bo Vertical Slice Tests

The Bo cryostat at Fermilab is a 250-liter vacuum-insulated vessel with an inner diameter 22 inches and a depth of 40 inches used for R&D studies, and, relevant to this paper, a vertical slice test of the MicroBooNE optical units. The system is described in detail in Ref. [?]. The cryostat can be filled with purified LAr with parts per billion level contamination of oxygen and water and parts per million level of nitrogen. The light collection system was tested with light from visible (420 nm) and UV (250 nm) LEDs piped in via fiber, as well as scintillation light from ^{210}Po alpha sources and cosmic rays.

A vertical slice test (VST) of the MicroBooNE optical system was performed in the Bo Cryostat. The slice consisted of two 8 inch PMTs with base electronics, mu-metal shield, TPB plates, cable feed throughs, splitters, the HV power supply and the interlock system. Tests were performed without and with the mu-metal shield. The test made use of the DAQ components described in section 6, including the shaper, FEM, trigger card, control card, and server. The MicroBooNE trace impurity monitors were also used.

The VST informed the final design of many components, as well as producing results relevant to understanding the running conditions and performance expectations. For example, during studies of the response of the slice to 128 nm scintillation light from the alpha source, valuable information was gathered on the single photo-electron dark rate and cosmic ray rate that could be scaled to the MicroBooNE detector expectations. As a second example, these runs allowed characterization of the pulse shape nonlinearities of the optical units, as seen in figure 41. These were shown to be significant at ~ 300 photo-electron in pulse amplitude. Full amplitude saturation occurred at ~ 670 PE. Thus, it is concluded that for pulses of more than 300 PE, pulse shape cannot be described by linear superposition of single photoelectron pulses.

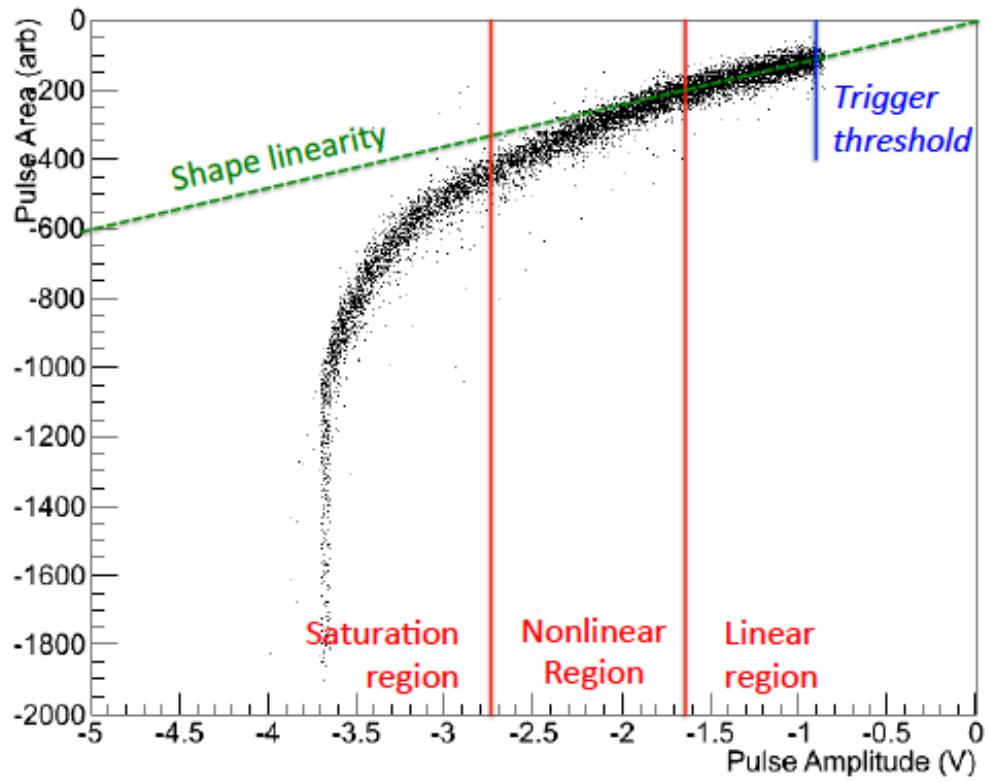


Figure 41: As shown by the VST, linearity of cosmic-ray induced PMT pulses is maintained up to amplitudes of around 1.7 V (300 PE), and amplitude saturation occurs at 3.7 V (670 PE) [?]

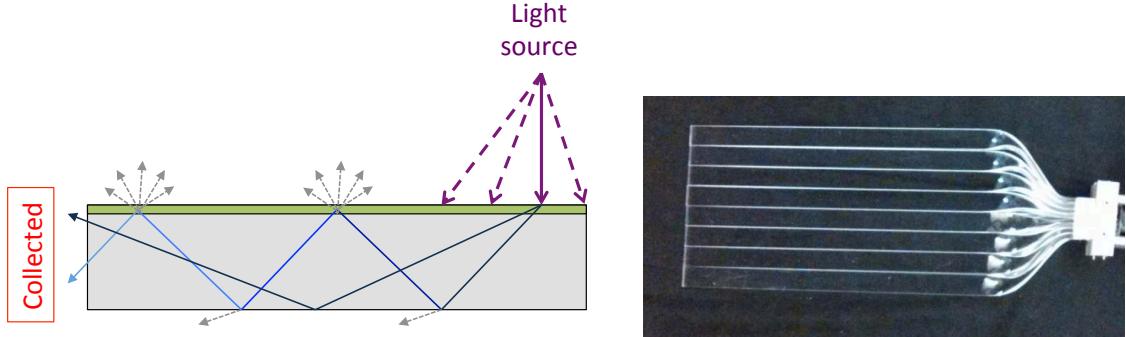


Figure 42: Left : Illustration of guiding mode, used by the paddles. Right: photograph of a coated paddle.

5.4 Acrylic Light Guides as a Secondary System for R&D

- ¹ A secondary light collection system consisting of four lightguide paddles was also installed. This
- ² concept has several advantages for future large detectors such as DUNE. First, the collection area

1 per channel is larger than the optical units, providing more coverage for the same number of elec-
2 tronics channels, cables, and feedthroughs. Second, the detectors have a flat profile so they can
3 be slid between chambers in a multi-LArTPC detector, minimizing space requirements of the light
4 collection system. In the case of MicroBooNE, the design gradually guides light in bent acrylic
5 bars to a PMT. This design was an early alternative to a perfectly flat design that guides the lights
6 to SiPMs. Running this system will provide long-term information on performance of lightguide
7 based systems. It also enhances the MicroBooNE dynamic range, since the lightguide detectors
8 saturate at a much higher light level than the optical units.

9 In the case of the lightguides, the 128 nm light is absorbed and shifted by a clear wavelength-
10 shifting coating, and the re-emitted light is guided to a 2-inch Hamamatsu R7725-MOD PMT, as
11 illustrated in figure 42, left. The installed paddles consist of six bars. A photograph of one coated
12 paddle with eight bars is shown in figure 42, right. The active length of each bar is 20 inches. This
13 system was added for R&D purposes and made use of 8 spare channels available of HV, cables,
14 feedthroughs, and electronics. The impedance of the 2 inch PMTs, tested at the feedthrough, was
15 $4.89 \pm 0.01 \text{ M}\Omega$. As shown in figure 31, each paddle is installed next to an optical unit for direct
16 comparison of performance.

17 The coating requirements for plate assemblies and light guides are different, and so the com-
18 position and coating methods for each were separately optimized. In the case of the light guide
19 coatings, the figure of merit is the light emitted in guided mode. Guided mode light is the light
20 that is detected at one of the ends of a test sample, which is orthogonal to the illuminated face of
21 the sample. In addition to the wavelength-shifting efficiency of the active layer, the detected light
22 yield is affected by the reabsorptive and scattering losses in the coating as visible light propagates
23 along the bar. The light guide assemblies have a TPB coating of 33% TPB to 67% UVT acrylic
24 by mass, also with ethanol surfactant. The coating is applied as a single layer and the TPB re-
25 mains suspended in the acrylic matrix as the coating dries, leading to a smooth, visibly transparent
26 surface. The performance and attenuation behavior of similar light guides to those installed in Mi-
27 croBooNE were studied experimentally in [77]. The reported non-exponential attenuation suggests
28 that surface losses dominate over bulk losses as the attenuation mechanism, and that the fractional
29 loss per reflection within the light guide is of order 2-3% [75].

30 In the light guide coatings, the TPB is suspended in an acrylic matrix which leads to a slight
31 broadening of the emission spectrum compared to the spectrum from the plates. This is an ex-
32 pected effect—TPB fluorescence has been shown to have dependence upon its microenvironment
33 [78, 79, 73, 80, 81], and Ref. [73] demonstrated spectral broadening in the presence of a polystyrene
34 substrate. Using the monochromator described for the plate spectrum studies, the light guide coat-
35 ing spectrum was measured in guided mode. A 10 cm section of light guide, with the incident
36 beam perpendicular to the TPB coated surface, was used. Based on reference [73] it is expected
37 that the emission spectrum for the light guide coating, with TPB embedded in the substrate, will
38 not change significantly as it cools to 87 K. The expected efficiency for the light guide coating is
39 found to be 0.25 ± 0.05 emitted visible photons per incident 128 nm photon.

40 **5.5 Flasher System**

41 The flasher system for the optical units and the light guides is described in Ref. [62]. This system
42 was developed to check the timing of the installed optical units, exercise the optical units dur-

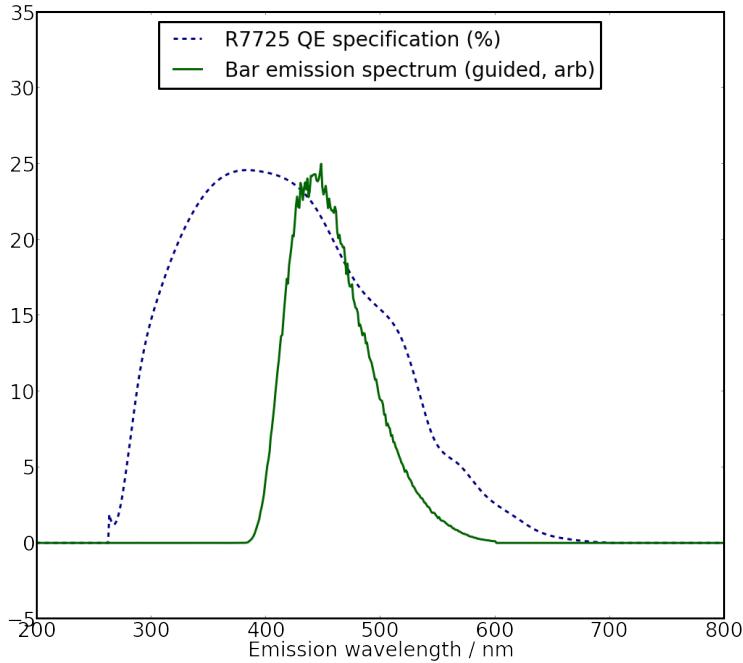


Figure 43: Measured emission spectra of the light guide coating in guided mode, and the R7725 quantum efficiency. Only the quantum efficiency of the non-undercoated PMT model is shown, from [82].

ing construction and commissioning, and to calibrate them once the detector is operational. The reference provides engineering drawings and details. The system is briefly summarized here.

A control board pulses an array of 400 nm LEDs, each of which is coupled through an optical feedthrough to 10 m optical fibers within the cryostat. The custom feedthrough/patch panel design encases the fibers in Arathane CW 5620 blue with HY 5610 hardener. The internal fibers are Molex FVP polyimide fibers with diameter of 600 mm, cladding of 30 μm , and an additional buffer layer of 25 μm . Each PMT has an individual calibration fiber. Each fiber is routed along the rack and attached to the PMTs by a fiber holder constructed of an aluminum standoff with a nylon-tipped set screw at a distance of about 2 inches from the PMT glass.

Figure 44 shows the results of flasher tests on all 32 optical units and four PMTs on the paddles. Time is along the x axis and PMT channel number is along the y axis. The white region separating the optical unit PMTs from the paddle PMTs represents unused channels. The colored bars indicate charge detected in the PMTs during flashing. One can see that all tubes respond properly.

5.6 Testing Coupling of PMT Signals to the Anode Wires

The ICARUS detector has observed an unexpected cross-talk between their PMTs and collection wire plane [?]. The Argontube test stand at the University of Bern also observed this effect [?]. Therefore, while the LArTPC was under final testing, before being rolled into the cryostat, an experiment was devised in order to determine the implication of this cross-talk for MicroBooNE. One of the production 8" Hamamatsu R5912-02mod PMTs was placed 5" from the LArTPC collection wire plane. This PMT was encased in a dark box with an optical fiber delivering light from a LED

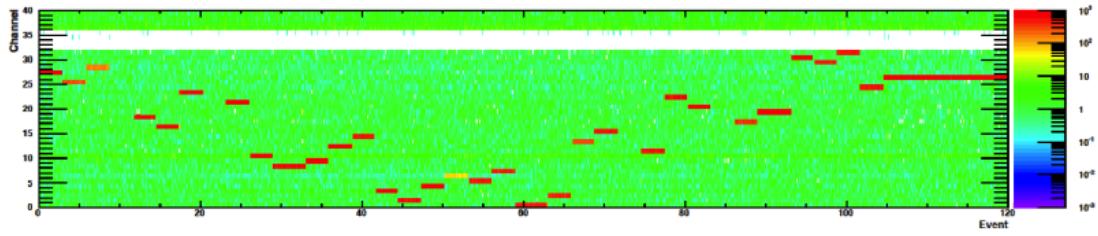


Figure 44: Charge detected in each PMT due to flasher tests, shown as a function of time. Each PMT is flashed for a short period. The white band indicates unused channels.

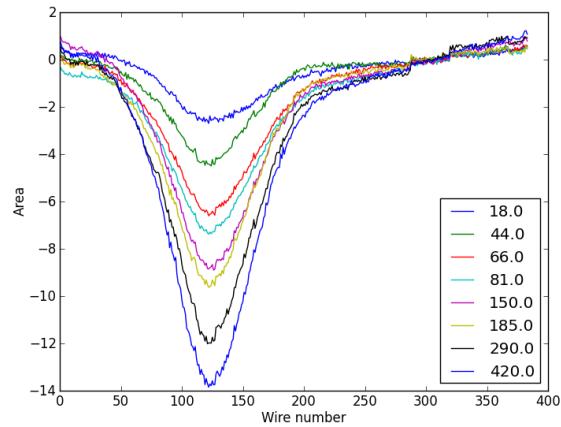


Figure 45: Area of signal pulses recorded on collection plane (in arb. units) as a function of wire number (arb. offset). The legend indicates the calibrated photo-electron count. The signal pulses are averaged over 50,000 repetitions. The plots are pinned together at wire 300, as the distribution baseline fluctuated over the course of the experiment due to intermittent noise.

1 flasher. The collection wire plane was read out using a DAQ test-stand that reflects the final DAQ
2 design.

3 A clear signal was observed on the collection plane when the LED fired. The magnitude of the
4 signal was characterized as a function of photo-electron count (figure 45) and separation between
5 the PMT and wire-plane. The signal induced on the collection plane was estimated to be no more
6 than ~ 10 ADC counts for a cosmic ray under normal operating conditions. This was reduced
7 to ~ 2 ADC counts when the PMT was encased in the μ -metal shields. This was deemed an
8 acceptably small amount that further shielding was not required.

9 The effect appears to saturate with photo-electron count, and is reduced when shorting the
10 resistors between the anode and last dynode. A later test with Argontube showed that the signal
11 was drastically amplified when using capacitors not able to withstand cryogenic temperatures in the
12 PMT base [?]. This suggests that the effect is electrostatic in nature, and probably due to capacitive
13 coupling between the PMT and wire-plane. This is also suggested by the similarity between the
14 signals observed and those produced by anode-coupled readout of a light collection system [64].

5.7 Initial Performance of the MicroBooNE Light Collection System

The system was first powered on after the cryostat had been filled with liquid argon. All the PMTs in both the primary and secondary system were found to be operational. Figure 46 shows the waveforms for all the PMT channels around the time that a large pulse, potentially from a cosmic ray muon, is observed. After initial checks of the system's health, the single photoelectron response of each PMT was set such that a one photo-electron pulse has an amplitude of 20 ADC counts as seen by the PMT readout system. The flasher LED system, described previously, is used to set this response. Figure 47 shows a candidate single photoelectron pulse following arrival of a TTL logic pulse driving the LED flasher system. The LED light level is set so that the majority (about 80%) of waveforms see no response in the region where pulses from the LED are expected to occur. This ensures that for windows with pulses, the pulses are of single photoelectrons. Figure 48 shows an example of the area vs. maximum amplitude of such pulses seen by a PMT during the flasher runs.

6. Electronics and Readout Systems

The analog signals that develop on a LArTPC during its operation must be amplified, digitized, and written to disk for use in analysis. Custom low-noise electronics that are capable of operating in the liquid argon environment have been developed for this purpose in MicroBooNE. The data from these LArTPC electronic channels, as well as from the PMTs, is sent to a readout system that digitizes and organizes the information before passing it along to a Data Acquisition (DAQ) system that stores it on disk. The stages of signal processing are illustrated in figure 49. The following subsections describe the LArTPC cold electronics, the LArTPC and PMT readout electronics systems, and the DAQ system in more detail. Details of triggering capabilities available in MicroBooNE are also provided in this section.

6.1 Cryogenic Low-Noise Electronics

To obtain optimum detector performance, MicroBooNE uses cryogenic low-noise front-end electronics for readout of the LArTPC. To reduce electronic noise, the interconnection length between the LArTPC wires and preamplifier should be as short as possible thus minimizing the total capacitance seen at the preamplifier input. To accomplish this, the analog front-end ASICs, which include a preamplifier, shaper, and signal driver are located inside the cryostat in addition to the wire bias voltage distribution system, decoupling capacitors, and calibration networks. The front-end ASIC and associated circuits are implemented on a cold mother board which is directly attached to wire carrier boards on the LArTPC itself. Cold cables are used to transmit output signals from cold motherboards to warm interface electronics installed on the top of signal feed-through flanges.

6.1.1 CMOS ASIC

The analog front end ASIC is designed in 180 nm CMOS technology, which integrates both the preamplifier and shaper on a single chip. Each chip has 16 channels to read out signals from 16 wires. Each channel also has a charge injection capacitor for precision calibration. In MicroBooNE, the shaper has 4 programmable gain settings (4.7, 7.8, 14 and 25 mV/fC) and 4 programmable peaking time settings (0.5, 1.0, 2.0 and 3.0 μ s) that provide increased flexibility to the readout

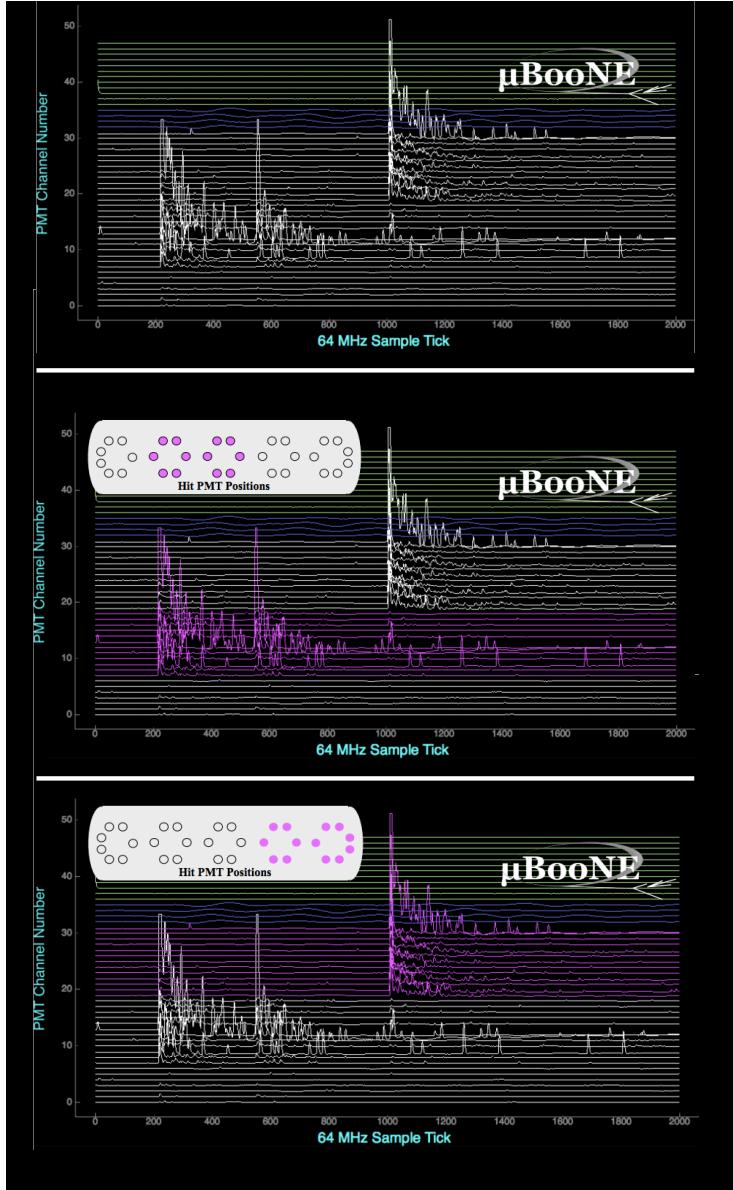


Figure 46: Top: Example waveforms from all 32 PMTs of the primary light collection system over a 31.25 microsecond readout window. The waveforms from the PMTs are in white. The blue waveforms are from the secondary lightguide PMTs, which are off in this picture. Waveforms from channels reserved for logic inputs are shown in green. In this image, the PMTs see two successive flashes of light at different parts of the detector. Middle and Bottom: Magenta highlights the early and late pulse, with the insets showing the PMTs which fired. Each pulse is likely from two cosmic ray muons traveling through the detector.

1 system. The ASIC also has programmable baseline settings (200 or 900 mV) to accommodate
 2 different detection wire configurations: either collection or induction plane. It has a selectable
 3 AC/DC coupling mode with a $100 \mu\text{s}$ time constant for the AC coupling mode which can be used
 4 to reduce low frequency noise. The ASIC also has built-in band-gap reference and temperature

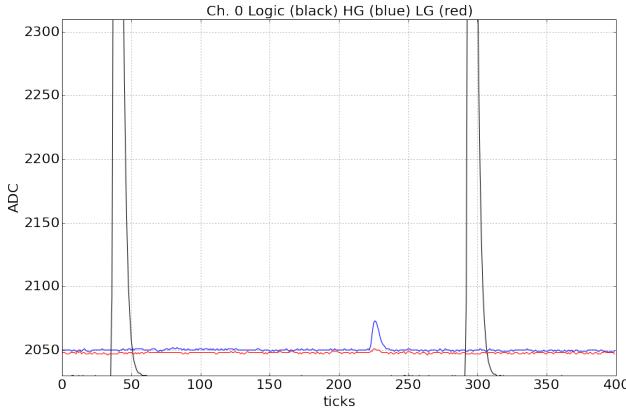


Figure 47: Example waveform captured by the PMT readout electronics during single photoelectron calibrations. The black waveform is of logic pulses that mark time at which an LED in the flasher system is driven. After some delay coming from PMT cable lengths and the flasher system, a candidate single photoelectron pulses is seen.

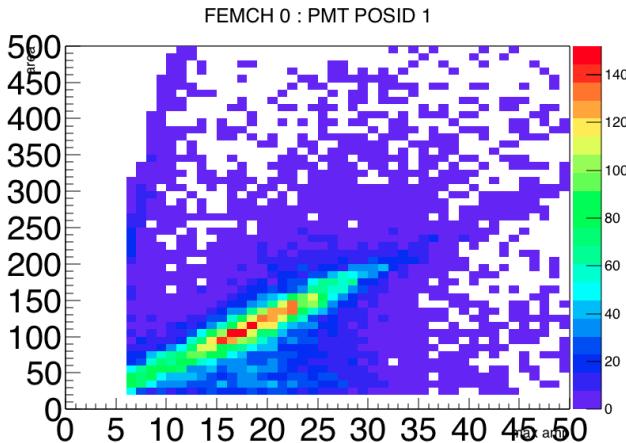


Figure 48: Distribution of the maximum amplitude and charge of pulses collected during an LED flasher calibration run. The central distribution of events is due to single photoelectron pulses.

¹ sensors to facilitate biasing and monitoring.

² The CMOS ASICs consume only \sim 6mW/channel in its default configuration. The front end
³ ASICs of the entire detector generate \sim 50 W of heat load that is easily handled by the cryogenics
⁴ system. Design guidelines that constrain the electric field and the current density to address the
⁵ lifetime of CMOS devices operated at cryogenic temperatures have been applied to every single
⁶ transistor (total \sim 15,000 transistors) in the ASIC design. A picture of the layout of the CMOS
⁷ ASIC is shown in figure 50. The test results [?] agree well with simulations and indicate that the
⁸ analog and the digital circuits (including the digital interface) operate as expected in the cryogenic
⁹ environment.

¹⁰ The MicroBooNE LArTPC has 8,256 readout channels and a total of 516 CMOS ASICs are
¹¹ required to fully instrument the detector. The production testing of the CMOS ASICs required
¹² two steps: both a warm and cold test. The warm test was performed with a dedicated test board

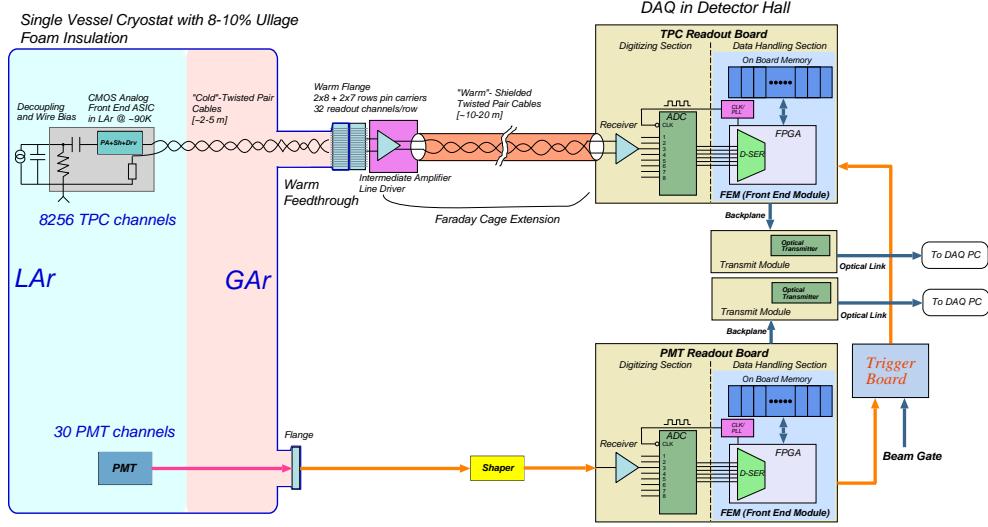


Figure 49: MicroBooNE LArTPC and PMT signal processing and readout stages.

1 housed in a Faraday box containing a socket to house the ASIC for ease in chip exchange. All
 2 programmable parameters (gain, peaking time, baseline, AC/DC coupling etc.) were exercised
 3 with the warm test setup for careful screening at room temperature. The yield of the warm testing
 4 of the ASICs was $\sim 89\%$. ASICs must have passed the warm test before going through cold testing.
 5 The cold test was performed with a dedicated test board containing 6 sockets to facilitate testing
 6 of multiple chips in liquid nitrogen at the same time. A total of 201 ASICs went through cold
 7 testing with a yield of $\sim 97\%$. Based on this high yield, it was decided not to continue the cold
 8 screening test on the rest of the production chips, as they were tested cold after being installed on
 9 the motherboard. After enough (~ 600) ASICs passed the production screening test, they were sent
 10 to an assembly house to equip the cold motherboard.

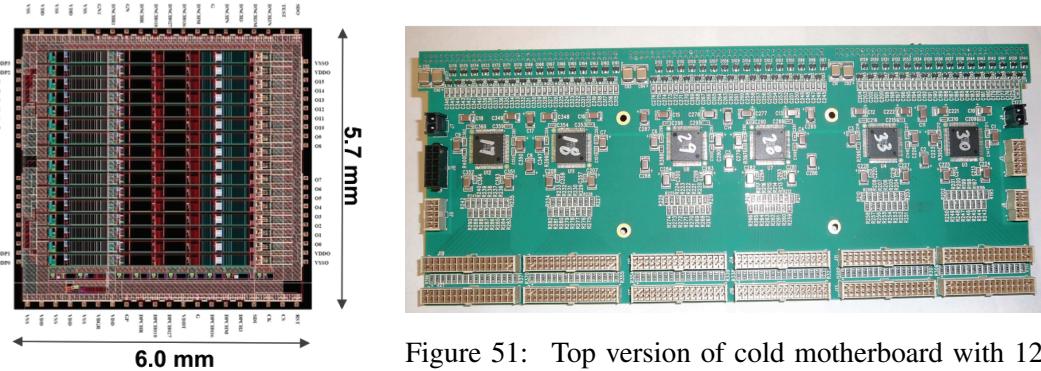


Figure 50: Layout of the CMOS analog front end ASIC

Figure 51: Top version of cold motherboard with 12 ASIC chips including 6 chips are mounted on top layer and 6 chips on bottom layer

6.1.2 Cold Motherboards

A cold motherboard was designed to house the MicroBooNE CMOS ASICs. In this capacity, the motherboard provides signal interconnections both between the detector wires and preamplifier inputs as well as between the driver outputs and cold cables to the signal feed-through. The cold motherboard design provides sufficient protection of the ASICs against electrostatic discharge during installation. It also provides a calibration network and bias voltage distribution for the wire planes. Specifically, a calibration signal enters the cryostat via a feed-through and reaches the preamplifiers through the motherboard. Each preamplifier channel in the ASIC has a built-in switch to individually cycle the calibration injection. The bias voltage reaches the LArTPC wires via a two-fold redundant path on the motherboard that allows the detector to operate normally even if one bias voltage channel fails. Use of Rogers 4000 series as the base material of the cold motherboard avoids the potential risk of contamination of the liquid argon. The Rogers material has a similar temperature expansion coefficient as the surface mounted components which enhances the reliability of the electronics assembly when operated at cryogenic temperatures. The different positions of the wire attachments along the top and sides of the LArTPC requires 2 types of cold motherboard. The top version of the motherboard has 192 readout channels that includes 96 Y channels, 48 U channels, and 48 V channels. The side version of the motherboard has 96 readout channels that are either U or V channels. A picture of the top version of a cold motherboard with 12 mounted ASIC chips is shown in figure 51.

The MicroBooNE LArTPC required a total of 36 top version motherboards and 14 side version motherboards to instrument the full detector. A test stand was built for testing of the front end electronics. This test stand included a full readout chain from the cold motherboard, cold cable, signal feed-through, warm interface electronics, warm cable and Receiver ADC board to a DAQ board based on a Xilinx ML605 FPGA evaluation board which sends acquired data to a PC over a Gigabit Ethernet. The production test of each motherboard also involved both a warm and cold test. Both tests used the same test stand, except the motherboard was placed in a Faraday box for the warm test and submerged in a liquid nitrogen dewar for the cold test. Noise, gain, peaking time, and linearity parameters were measured in both warm and cold to screen the motherboard. Motherboards had to pass both tests before being installed on the detector.

6.1.3 Cold Cables

Cold cables transmit the detector signals from the cold motherboard to an intermediate amplifier on top of the signal feed-through and distribute power to the CMOS ASICs. The cold cable is a custom-built 32-pair twisted pair flat ribbon cable with Teflon FEP insulation and 100Ω ($\pm 10\%$) impedance, using AWG 26 stranded wire with silver-plated copper. Custom designed shells with jack screws used in the cable assembly ensured proper alignment of the insertion on the signal feed-through pin carriers. The twisted pair cable was ordered from the manufacturer and sent to a company to be woven into flat ribbon cables. Cold cables of two different formats were assembled by an assembly house: signal cables and service cables. Signal cables are used to transmit amplified detector signals while the service cable is used to transmit calibration pulses and slow control/monitoring signals. Signal cables were produced in three different lengths: 80 inch, 100 inch and 180 inch to accommodate the different lengths between the cold motherboards

1 and the signal feedthroughs, while the service cables were produced in two different lengths: 100
2 inch and 180 inch. All of the cold cable assemblies were tested with a cable tester in the assembly
3 house before they were shipped out. In addition, \sim 10% of the cold cables were tested in the test
4 stand at BNL to confirm the quality of the cable assembly.

5 **6.1.4 Electronic Calibration**

6 The MicroBooNE cold electronics include a precision charge calibration system. Through the cold
7 cable and calibration network on the motherboard, a calibration signal enters the cryostat via a
8 feed-through and reaches the preamplifiers. A built-in switch in the ASIC makes it possible to
9 power cycle the calibration injection for every channel individually. The electronics calibration is
10 based on charge injection through known capacitances (180 fF) in the ASIC. This system enables
11 gain (charge sensitivity) calibration, verification of sense wire integrity and noise measurements.
12 The built-in electronics calibration capability is an important tool in testing and characterizing the
13 overall performance of the detector readout system. It was extensively used in the cold electronics
14 production testing and the electronics checkout during installation, commissioning, and data taking.

15 **6.1.5 Performance Tests**

16 The development of the analog front end ASICs was initiated using 180 nm CMOS technology
17 and 300 K models, though the performance parameters are extracted at 77 K. CMOS was found
18 to function at cryogenic temperatures with increased gain and lower noise. The noise, gain, and
19 pulse shaping were found to be as expected in evaluation tests of the ASICs. Extensive testing of
20 the ASICs mounted on the motherboards was performed; these tests were done in liquid nitrogen
21 rather than at room temperature, since noise levels and characteristics of the ASIC performance in
22 liquid nitrogen are similar to the performance in liquid argon. Thus, cold tests were performed on
23 all production cold motherboards fully populated with 12 chips. A total of \sim 2,200 chip-immersions
24 were accumulated in liquid nitrogen without any failures due to thermal contraction or expansion.

25 The test results show the noise of the front end readout electronics system decreasing uni-
26 formly for all 768 channels from \sim 1,200e $^-$ at 293 K to less than 600e $^-$ at 77 K with 150 pF
27 detector (sense wire) capacitance. A plot of noise versus temperature of 12 ASICs for a total of
28 192 channels is shown in figure 52. The response of the front end electronics exhibits excellent uni-
29 formity at cryogenic temperatures. As shown in figure 53, the gain variation of a cold motherboard
30 with 12 ASICs is only \sim 7% peak-to-peak across 192 channels. The spread of the gain variation is
31 only \sim 1% of the gain setting.

32 **6.2 Warm Electronic Amplification**

33 Signals from the cold electronics are carried over the cold cables to dedicated feedthroughs mounted
34 on the cryostat. The cold cables are connected to pin-carriers located on 14-inch CF signal feedthrough
35 flanges that are mounted on nozzles N1A-N1K of the cryostat (see figure 5). The signal feedthrough
36 design must accommodate 100% hermeticity and high signal density. A design based on the AT-
37 LAS pin carrier style was developed for this purpose. Two 8-row pin carriers and two 7-row pin
38 carriers are welded onto the 14-inch CF flange, as shown in figure 54, and create a vacuum-tight
39 seal. Nine of the 11 signal feedthroughs receive signals from the three LArTPC anode planes (384

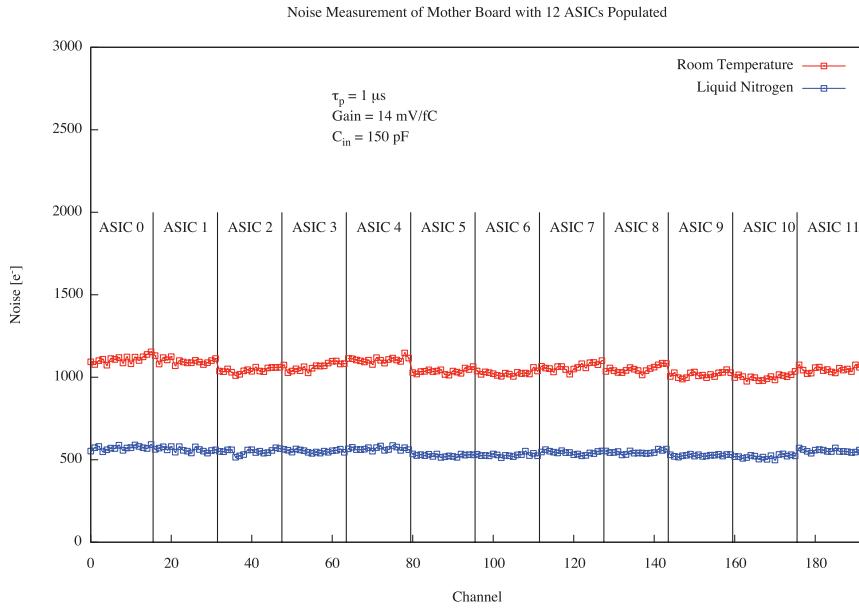


Figure 52: Plot of noise vs. temperature of 12 ASICS, total 192 channels. Noise is $\sim 1,200e^-$ at 293K, and $\sim 550e^-$ at 77 K with 150 pF C_d

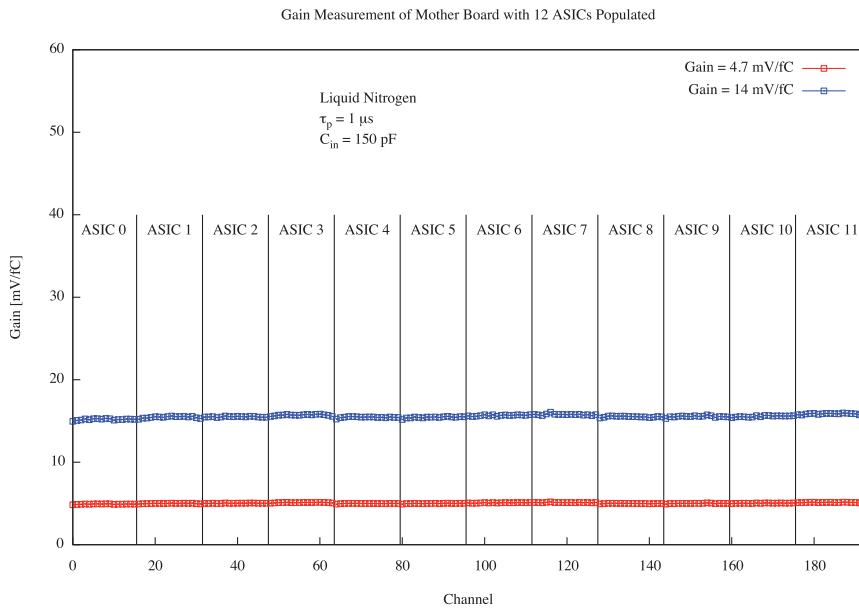


Figure 53: Plot of gain uniformity of 12 ASICS, total 192 channels, at 77 K with two different gain settings

- ¹ Y-plane, 192 U-plane, 192 V-plane), while the remaining two on the extreme ends of the cryostat
- ² only receive signals from one of the angled induction planes (672 U-plane on one feedthrough, 672

¹ V-plane on the other).

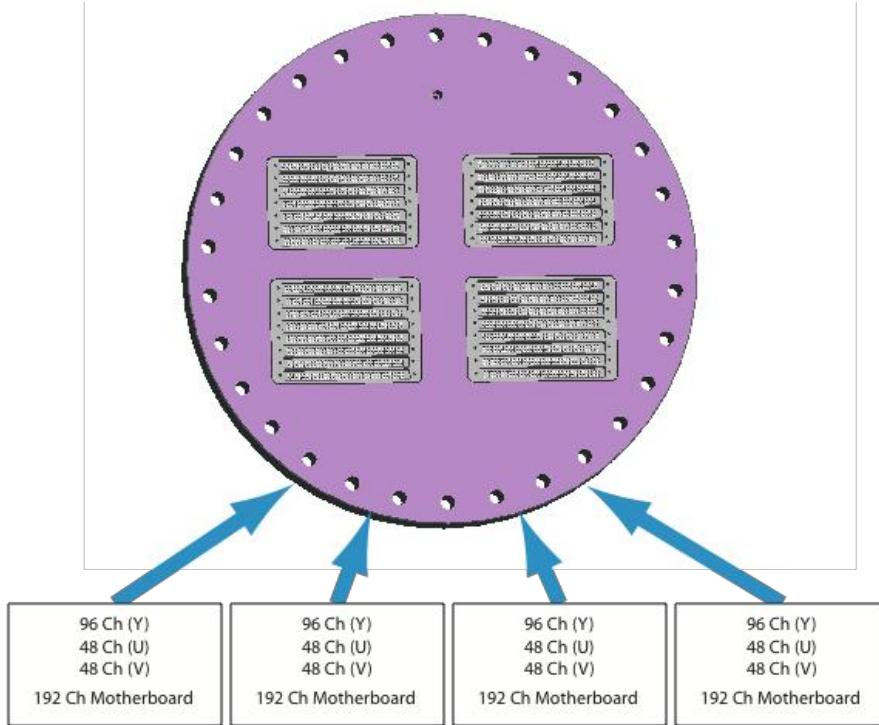


Figure 54: Signal feedthrough flange consisting of a 14-inch CF flange with two 8-row and two 7-row pin carriers welded in place.

² A Faraday cage is mounted on the external, warm, side of the signal feedthroughs to provide
³ shielding for the intermediate amplifiers located inside. The bias voltage feedthrough, which sup-
⁴ plies anode plane bias voltages into the cryostat, is built onto a small 2.75-inch CF flange welded
⁵ onto the signal feedthrough flange. A filter board mounted on the bias voltage flange filters noise
⁶ and ensures a good ground connection. Figure 55 shows details of the signal feedthrough assembly
⁷ with electronics boards, bias voltage feedthrough, and Faraday cage.

⁸ The intermediate amplifiers provide \sim 12 dB gain to the LArTPC signals to make them suit-
⁹ able for transmission over a 20 m long cable to the readout electronics (see section 6.3). Each
¹⁰ intermediate amplifier has 32 channels installed on the signal feedthrough flange and housed inside
¹¹ the Faraday cage to provide noise isolation. Figure 56 shows a picture of a prototype intermediate
¹² amplifier plugged on the signal feedthrough pin carrier. The intermediate amplifier uses a 68-pin
¹³ SCSI-3 connector to drive the 32 channels of signal differentially for better noise immunity. The
¹⁴ layout and connector position have been carefully designed to ensure the card can be plugged on
¹⁵ the pin carrier in either direction. This efficiently utilizes the limited available space on the top of
¹⁶ the feedthrough, which also makes the design of the Faraday cage easier.

¹⁷ In addition to the intermediate amplifiers, there are two service boards mounted on the top of
¹⁸ each signal feedthrough. The service board provides regulated low voltage, control and monitoring
¹⁹ signals to the analog front end ASICs. It also provides pulse injection to the preamplifiers for

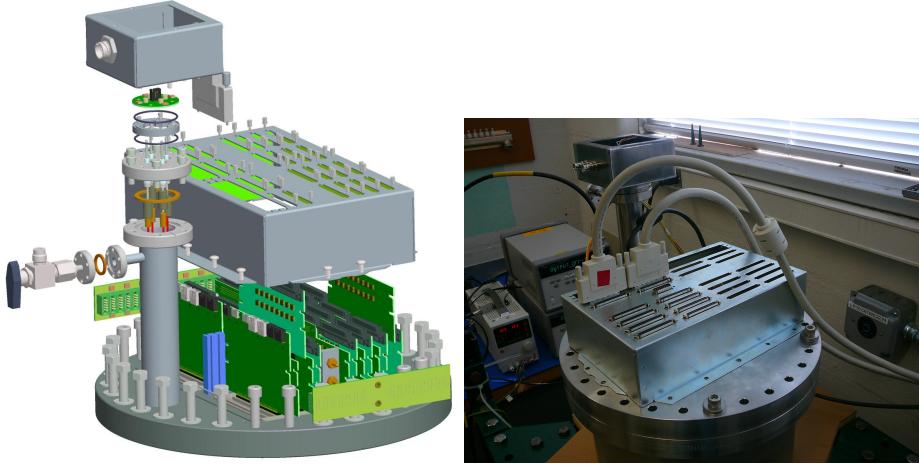


Figure 55: Left: Diagram of the signal feedthrough assembly, which includes intermediate amplifiers, Faraday cage, bias-voltage feedthrough and filtering circuit. Right: Photograph of one of the feedthrough assembly, partially constructed and being tested.

precision calibration. The control, monitoring, and calibration signals are provided to the front end electronics with two-fold redundancy. Should one set of signals become defective the detector can still operate normally with the redundant set. Each service board plugs onto a 64-pin carrier row. Figure 56 shows a picture of a prototype service board.



Figure 56: Left: Photograph of one of the intermediate amplifier boards plugged into a pin carrier during testing. Right: Photograph of a prototype service board.

6.3 LArTPC Readout Electronics

The MicroBooNE readout electronics system consists of two subsystems: the LArTPC and PMT readout electronics. The LArTPC readout electronics are responsible for the readout, digitization, and processing of the induction and collection wire signals after amplification. The PMT readout electronics are responsible for the amplification, shaping, digitization, and handling of PMT signals, as can be used to provide a trigger signal for the readout and data acquisition systems. While

1 the LArTPC and PMT readout systems share the same back-end design that organizes and packages
2 the data for delivery to the DAQ system, they employ different analog front-end and digitization
3 designs, which are described in this and the following subsection.

4 The LArTPC readout electronics are responsible for processing the signals from the 8,256
5 wires in MicroBooNE after pre-amplification and shaping in the cold electronics (section 6.1).
6 The pre-amplified and shaped analog signals from the cold electronics are transmitted to the warm
7 electronics outside the cryostat, as described in section 6.2, and then passed to custom-designed
8 LArTPC readout modules distributed evenly over nine readout crates. The readout modules digitize
9 the analog signals and then process and prepare them for shipping to designated DAQ machines
10 (one per readout crate) (section 6.6).

11 The LArTPC readout crates communicate with the DAQ machines via three duplex 3.125
12 Gbits/sec optical links that connect to a crate controller module and data transmitter (XMIT) mod-
13 ule on the crate end, and to three PCI Express boards on the DAQ machine end. The controller
14 is responsible for configuration, trigger and run control command distribution as well as the slow
15 monitoring of each readout crate. The controller occupies one of the optical links while the XMIT
16 is responsible for sending two separate streams of readout data to the DAQ machines via the two
17 other optical links. The first XMIT stream contains losslessly compressed LArTPC data associ-
18 ated with event triggers received by the LArTPC readout crates, such as the BNB trigger, and
19 is referred to as the “NU” data stream. The second stream is a continuous LArTPC data stream
20 which is compressed with some data loss. The continuous data stream is used for beam-unrelated
21 physics analyses, such as the study of potential supernova neutrino events, and is referred to as
22 the “SN” data stream. The compression schemes used in the NU and SN streams are described in
23 section 6.3.3.

24 All readout crates are synchronized to a common 16 MHz clock. The clock sync is provided
25 by a clock fanout board which shares the same ground as all readout crates, and is sent via coaxial
26 cable to a distribution board which is mounted on each crate backplane. The readout frame size
27 is set to 1.6 ms, which is equivalent to the time it takes for charge produced on the far end of the
28 LArTPC to drift to the wire planes at the design cathode voltage of -128 kV.

29 **6.3.1 LArTPC Data Digitization**

30 The amplified and shaped analog LArTPC signals are differentially received and digitized in the
31 first section of the readout modules. Each ADC module holds 8 AD9222 octal-channel 12-bit
32 ADCs. Each ADC module handles signals from 64 wires. The wire signals are grouped in two
33 sets of 32 consecutive wire channels: either 32 induction wires plus 32 collection wires or two sets
34 of 32 induction wires. The induction channel sequence alternates wires between the two induction
35 planes. The ADC module digitizes the signals continuously at 16MHz. Each channel has a config-
36 urable baseline, which is either set low (450 ADC counts) for collection channels or at the middle
37 of the dynamic range (2055 ADC counts) for induction channels, thus ensuring that both the collec-
38 tion plane unipolar differential signals and the induction plane bipolar differential signals can make
39 use of the full ADC analog input range. The requirement to observe a MIP produced at the far end
40 of the LArTPC in the induction plane determines the lower end of the dynamic range, while the
41 requirement to observe a highly-ionizing stopping proton at the close end of the LArTPC without
42 saturation sets the upper end. The digitized outputs from the ADC board are passed directly to a

1 Front End Module (FEM) in the second section of the LArTPC readout module. The FEM houses
2 an FPGA for data processing, data reduction, and preparation for readout by the DAQ system as
3 described in the following section.

4 **6.3.2 LArTPC Data Handling**

5 The FEM board consists of a 14-layer printed circuit board which is mechanically integrated with
6 the ADC board as illustrated in figure 57. The choice of a smaller board allows for short trace
7 lengths which is beneficial for high speed signals. The full assembly comes together as a standard
8 VME 9U card in height, with a 280 mm depth. Differential outputs from the ADCs connect to the
9 FPGA through HM-Zd connectors that have individual ground shielding on each differential pair.

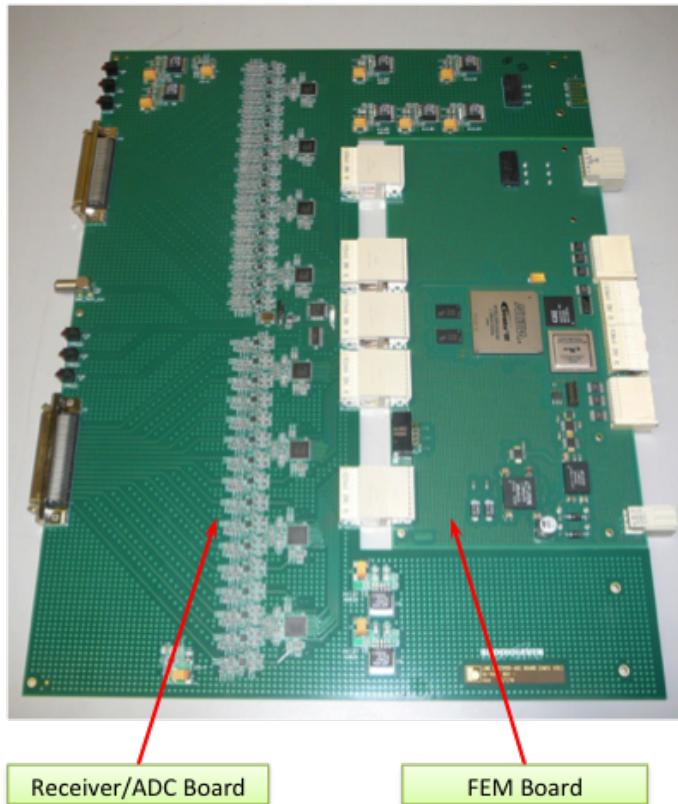


Figure 57: LArTPC ADC+FEM board.

10 The digitized data stream moves from the ADCs to a Stratix III Altera FPGA, which reduces
11 the sampling rate of the ADC from 16 MHz to 2 MHz. The 2 MHz sampling rate is optimized by
12 taking into consideration the expected pulse shape provided by the convolution of the cold electron-
13 ics, the expected LArTPC field responses, and the $O(1\mu s)$ diffusion effects which govern charge
14 drift within the liquid argon. The FPGA stores the data from all 64 wires per board sequentially in
15 time in a $1M \times 36$ bit 128 MHz SRAM, grouping two ADC words together in each 36 bit memory
16 word. This requires a data storage rate of $(64/2) \times 2$ MHz = 64 MHz. The SRAM chip size and
17 memory access speed allow for continuous readout of the LArTPC data. Since data reduction and
18 compaction algorithms rely on the sequential time information of a given wire, the data readout out

- 1 from this SRAM takes place in wire order in alternate clock cycles, again at the rate of 64 MHz.
 2 This read in/out sequence is illustrated in figure 58.

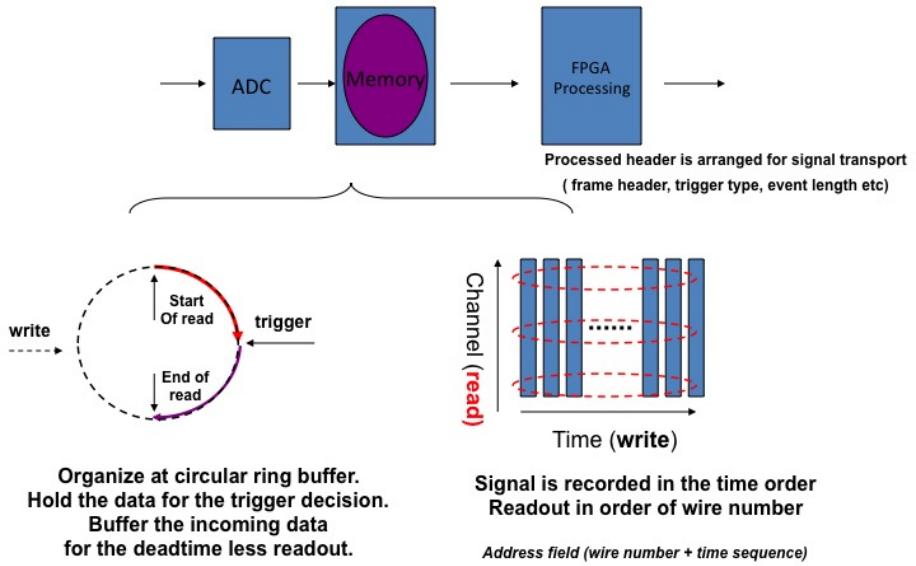


Figure 58: LArTPC readout sequence in the TPC FEM.

3 Separate DRAM multi-event buffers on the FEM store the NU and SN data streams. The data
 4 divert into the NU readout stream when a trigger is issued and received (as shown in figure 59)
 5 which signals, for example, an accelerator neutrino-induced event. When an event trigger is re-
 6 ceived, 4.8 ms worth of data, relevant to that event, are packeted per channel and sent to the DAQ
 7 through the NU data stream. The 4.8 ms readout size is governed by the maximum drift time and
 8 spans three or four frames. In order to reduce the amount of data being transmitted, the FPGA
 9 trims the three or four frames to span the exact 4.8 ms required, 1.6 ms before the trigger plus 3.2
 10 ms after the trigger. In parallel, the data is continually sent out through the SN data stream, frame
 11 by frame. The compression and data reduction algorithms applied to each of the two streams are
 12 described in the following section.

13 After processing by the FPGA, the data passes to the crate backplane dataway on connectors
 14 shown in figure 57. A token-passing scheme is utilized to transfer data from each FEM board to the
 15 data transmitter module (XMIT) in a controlled way, whereby each FEM, in the order of closest to
 16 furthest from the XMIT module, receives a token, transmits its data to the XMIT, and passes the
 17 token on to the next FEM in the sequence. For the NU stream, each FEM sends all data associated
 18 with a particular trigger number; while for the SN stream, each FEM sends all data associated with
 19 a particular frame number. This data transfer is relayed via the otherwise passive crate backplane,
 20 and is limited to 512 MB/s. In the XMIT module, the data is buffered temporarily and sent to the
 21 DAQ machine through the two streams, SN and NU, which proceed effectively in parallel.

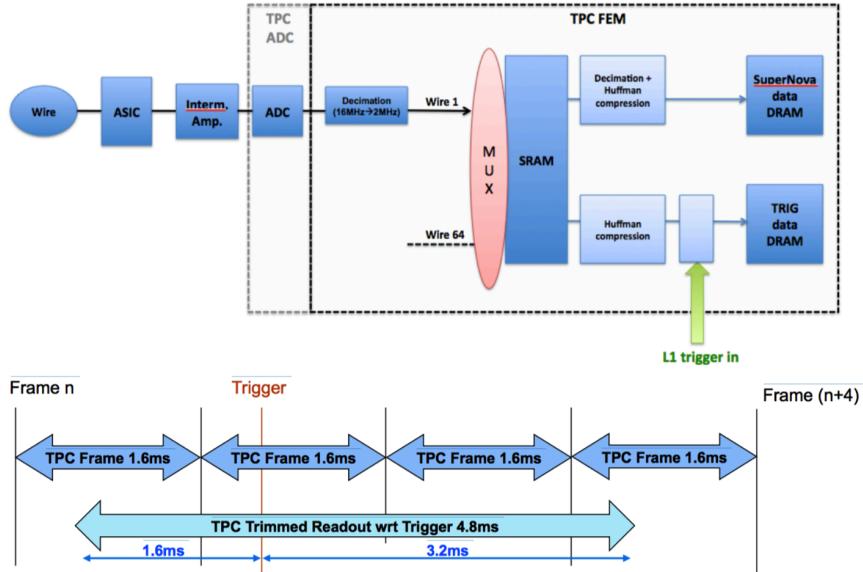


Figure 59: LArTPC trigger readout.

6.3.3 LArTPC Compression Schemes

In the case of the NU data stream, a lossless Huffman coding scheme implemented in the FEM FPGA compresses the data by approximately a factor of five. Further reduction in the overall rate is achieved by exploiting a PMT trigger in coincidence with the BNB trigger, as described in section 6.5. Huffman coding provides for lossless data compression by taking advantage of the slow variation of the waveform TPC data in any given channel. In particular, this compression scheme relies on the fact that successive data samples on any given wire vary relatively slowly in time. As such, when noise levels are low, any two adjacent data samples either coincide or differ by 1 ADC count. The most frequent values for the difference in ADCs between successive data samples are assigned pre-specified bit patterns with the lowest number of bits possible. Those bit patterns are encoded in the 16-bit data words that would otherwise be used for a single 12-bit ADC sample value. As such, data reduction of up to a factor of 14 is theoretically possible¹. In practice, the data reduction is sensitive to noise levels and LArTPC activity, and is also dependent on the gain setting. The compression factor achieved by MicroBooNE is shown in figure 60.

Because of the low trigger rate², lossless Huffman coding compression proves sufficient for the NU data stream. However, for the continuous SN stream, further compression becomes necessary, resulting in unavoidable data loss. A method called “dynamic decimation” (DD) handles this case. The DD scheme relies on recognizing regions of interest (ROI) in the data stream that contain waveforms corresponding to drift ionization charges. Portions of the data stream not containing ROI contribute to pedestal determination, and ROI are identified as deviations from the continually-updated pedestal, buffered, and read out to disk. At the time of this writing, the MicroBooNE SN stream compression scheme is being finalized and the SN readout stream will be commissioned in

¹The uppermost two bits are always reserved for header information, in each 16-bit word.

²The BNB trigger dictates an upper bound on the trigger rate of 15 Hz.

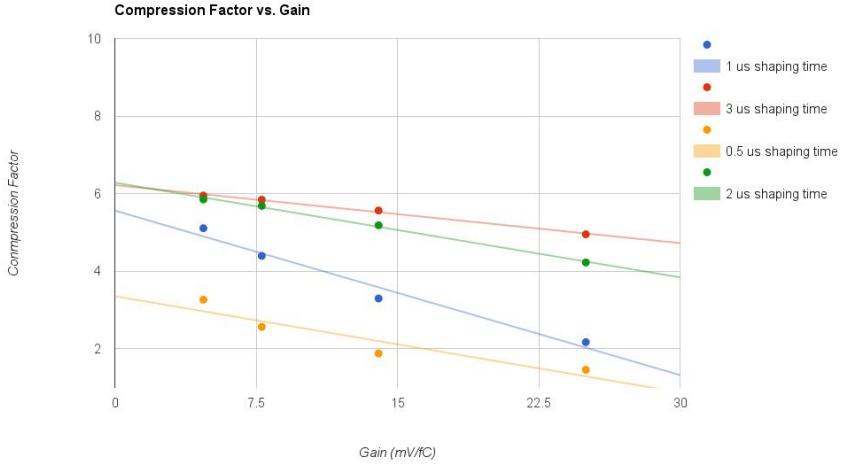


Figure 60: Compression factors achieved on ADC data with Huffman compression.

¹ September 2016.

² 6.4 PMT Readout Electronics

³ The PMT readout electronics are responsible for processing signals from the 32 PMTs described
⁴ in section 5 and identifying light signatures coincident with the BNB and NuMI beam spills. The
⁵ coincidences generate PMT triggers that can be later mixed with other triggers in the Trigger Board
⁶ (TB). Signals from the 4 light paddle guides installed in the LArTPC are also recorded by the PMT
⁷ readout electronics, but these signals do not participate in the PMT trigger generation.

⁸ The stages of signal processing are illustrated in figure 61. First, each PMT signal (with the
⁹ exception of the light paddle guide signals) is split into two different gains, as described in sec-
¹⁰ tion 5.2.5, with the HG channel carrying 18% of the PMT signal and the LG channel carrying
¹¹ 1.8% of the PMT signal. Each gain is split once again into HG1 and HG2, and LG1 and LG2, in
¹² order to allow different processing of beam-related and beam-unrelated PMT signals. All $32 \times 2 \times 2$
¹³ plus 4 signals are pre-amplified and shaped in 16-channel pre-amp/shaper boards (section 6.4.1).
¹⁴ Four PMT readout modules receive the analog shaped signals differentially and digitize them (sec-
¹⁵ tion 6.4.2) at 64 MHz. The PMT readout modules then process the signals in order to prepare them
¹⁶ for shipping to a designated DAQ machine and to form a possible PMT trigger (section 6.4.3).

¹⁷ Each one of the 4 PMT ADC+FEM readout boards used in the PMT readout system handles
¹⁸ one of the following:

- ¹⁹ • Readout of and PMT trigger generation using the HG1 PMT signals associated with neutrino
²⁰ beam events. The paddle signals are also readout by this board.
- ²¹ • Readout of and PMT trigger generation using the HG2 PMT signals that are out of beam
²² time (i.e. cosmic rays and other cosmogenic backgrounds)
- ²³ • Readout of the LG1 PMT signals associated with neutrino beam events

- 1 ● Readout of the LG2 PMT signals that are out of beam time

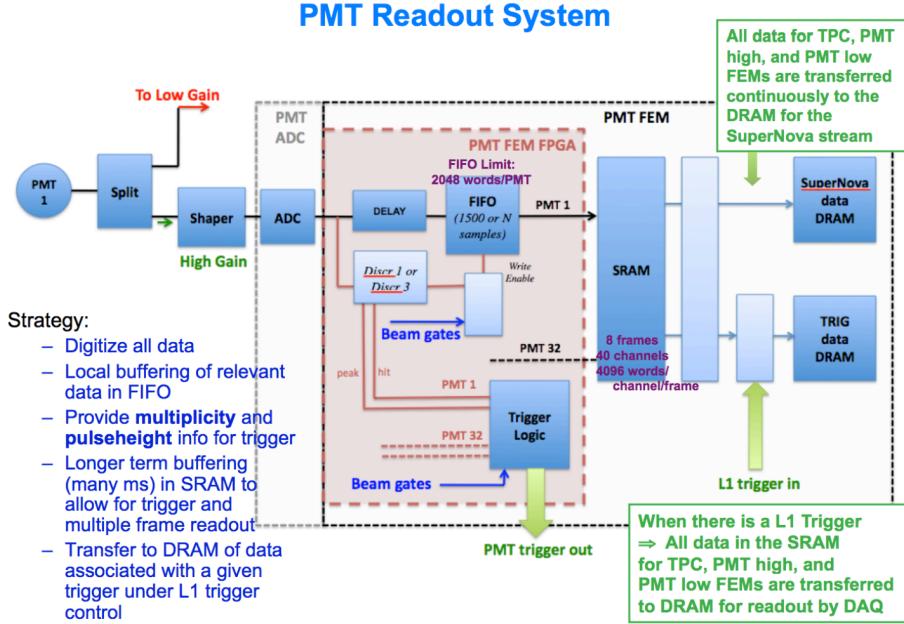


Figure 61: Signal processing in PMT FEM.

2 After signal processing, the data is sent to a designated DAQ machine via a transmitter (XMIT)
 3 module in the same way as is done for LArTPC data. Two data streams are provided: a NU data
 4 stream associated with event triggers and a SN data stream which a continuous version of the NU
 5 stream readout.

6 6.4.1 Signal Amplification and Shaping

7 The preamp/shaper boards read raw PMT signals from the PMT HV/signal splitters and shape them
 8 into unipolar signals with a 60 ns rise time. The shaped signals are sent to the LArTPC readout
 9 boards differentially via short front-panel cables, in order to minimize noise, where they are digi-
 10 tized at 64 MHz. The 60 ns peaking time allows digitization of two or three samples on the rising
 11 edge. This in turn enables an accurate determination of the event start time (t_0) needed to determine
 12 the x coordinates of ionization signals along the drift direction. An accurate time measurement also
 13 helps reject other tracks, such as cosmic rays, that cross the detector during the drift time.

14 6.4.2 PMT Data Digitization

15 The ADC (Texas Instruments, ADS5272) module part of the PMT readout board (figure 62) is
 16 responsible for digitization of up to 48 differentially-driven input signals. The differential signals
 17 are digitized at 64 MHz. The 64 MHz clock used by the PMT readout is generated starting from
 18 the 16 MHz clock that is common to all readout crates (LArTPC and PMT).

19 In addition to HG and LG waveforms from the PMTs, each readout board also receives, digi-
 20 tizes, and processes beam gate signal markers which arrive $4\mu s$ before the BNB $1.6\mu s$ and NuMI
 21 $10\mu s$ beam gates. These gates are used to (a) specially mark regions of interest where PMT data

TEMPORARY PLACEHOLDER

Figure 62: The PMT readout board digitizes 48 input signals. **NOTE: Missing Image!**.

- ¹ are read out continuously with no compression and (b) look for coincident PMT light signatures
- ² for trigger generation.

³ 6.4.3 Data Handling and PMT Trigger Generation

⁴ PMT information is recorded in the NU data stream for four 1.6 ms frames associated with an event
⁵ trigger: the frame containing the (asynchronous) trigger, the frame preceding the trigger frame,
⁶ and two frames following the trigger frame. To avoid the inordinate amount of data that would
⁷ be generated at a 64 MHz sampling rate, the FEM applies a zero-suppression immediately after
⁸ digitization, retaining only samples above a given threshold as well as enough information before
⁹ and after this useful data to establish a local baseline value; this collection of information is referred
¹⁰ to as a PMT readout ROI. An exception is formed for beam-related or other likewise-triggered data
¹¹ where, for example, the $4\mu\text{s}$ -early BNB and NuMI beam gates mentioned in section 6.4.2 instruct
¹² readout of 1500 consecutive samples ($23.4\mu\text{s}$) surrounding and including the beam gates regardless
¹³ of signal activity.

¹⁴ Two different discriminators are used: one that is active inside the beam gate(s), and one
¹⁵ that is also active outside the beam-gate-surrounding $23.4\mu\text{s}$. The latter discriminator governs the
¹⁶ readout activity due to cosmic rays and other non-beam related activity. The first discriminator
¹⁷ enables PMT channels with pulse heights above a configurable threshold (e.g. corresponding to 1
¹⁸ photoelectron) to participate in trigger multiplicity and pulse height sum conditions, as described
¹⁹ in the following section. The thresholds for those two discriminators are set to different levels and
²⁰ configured with different dead times for the HG and LG signals.

1 **6.5 Level-1 Trigger Generation**

2 The TB, which physically resides in the PMT readout crate, issues a “Level-1” trigger in order to
3 flag frames that must be treated differently. In the case of the LArTPC readout, the TB flags the 4
4 frames that must be trimmed and readout through the NU data stream and, in the case of the PMT
5 readout, it flags the 4 frames that must be readout in full through the NU data stream.

6 The inputs to the TB include a BNB trigger input (maximum rate of 15 Hz), a NuMI trigger
7 input (1.25 Hz), a Fake Beam trigger input (configurable), a PMT trigger input, and two calibra-
8 tion trigger inputs, provided by the laser calibration system and the cosmic ray muon telescope,
9 respectively. The TB also has the ability to receive, via the crate controller, DAQ-issued calibration
10 triggers, which are used explicitly for cold electronics and PMT calibration. The various input
11 triggers can be independently pre-scaled, masked, and mixed together (OR or AND) to generate an
12 event trigger.

13 The FPGA firmware in the PMT FEM can generate two different types of PMT triggers based
14 on the PMT signals: a cosmic PMT trigger and a beam gate PMT trigger. Beam gate PMT triggers
15 are configured in the same way for the BNB, NuMI, and Fake Beam. The nominal criteria for these
16 triggers are (1) PMT multiplicity ≥ 1 and (2) summed PMT pulse-height ≥ 2 photoelectrons (p.e.)
17 summed over all 32 HG1 PMT channels. Both criteria must be met during any 100 ns time interval
18 coincident with the beam spill duration ($1.6\mu s$ in the case of the BNB and Fake Beam gates and
19 $10\mu s$ in the case of the NuMI gate), and only channels enabled by the beam gate discriminator can
20 participate in the active pulse-height and multiplicity sums. The criteria for a cosmic PMT trigger
21 are (1) PMT multiplicity ≥ 1 and (2) summed PMT pulse-height ≥ 40 p.e. summed over any one of
22 28 preset groups of 5 HG2 PMT channels that are grouped based on their spatial correlation. Again,
23 only channels enabled by the cosmic discriminators can participate in the trigger generation.

24 In addition, a software-based algorithm has been written to mimic the capabilities of the beam
25 gate PMT trigger performed in the FPGA and provide more flexibility in trigger criteria settings.
26 Details of this higher-level software trigger are described in section 6.7.

27 Figure 63 diagrams the PMT readout and trigger logic. Activation or masking of each of the
28 trigger inputs and outputs is DAQ-controlled. The trigger condition and explicit PMT trigger type,
29 if applicable, is available for every event in the NU data stream at both the event-building stage and
30 offline; this information is read out via a dedicated optical data stream, directly from the TB. The
31 trigger number and trigger time are also propagated and available in the NU data streams arriving
32 independently at the assembler DAQ machine from each LArTPC and PMT crate, and can therefore
33 be used to correctly associate data from the same event.

34 The readout control sequence is illustrated in figure 64. When a trigger is generated by the
35 TB it is passed to a fan-out module on a single cable and from there it is distributed to all crate
36 controllers (LArTPC and PMT). Through the crate backplane, the trigger gets propagated to each
37 FEM. An FEM that receives a trigger temporarily inhibits the SN stream with its associated deci-
38 mation and initiates the loss-less readout scheme to direct the data to the appropriate readout path.
39 SN readout resumes once the XMIT is done sending all NU data associated with an event to the
40 DAQ.

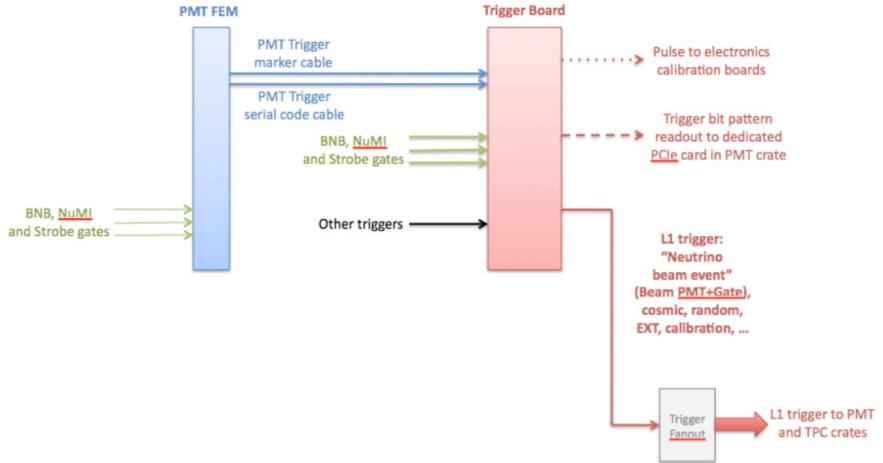


Figure 63: PMT readout and trigger logic.

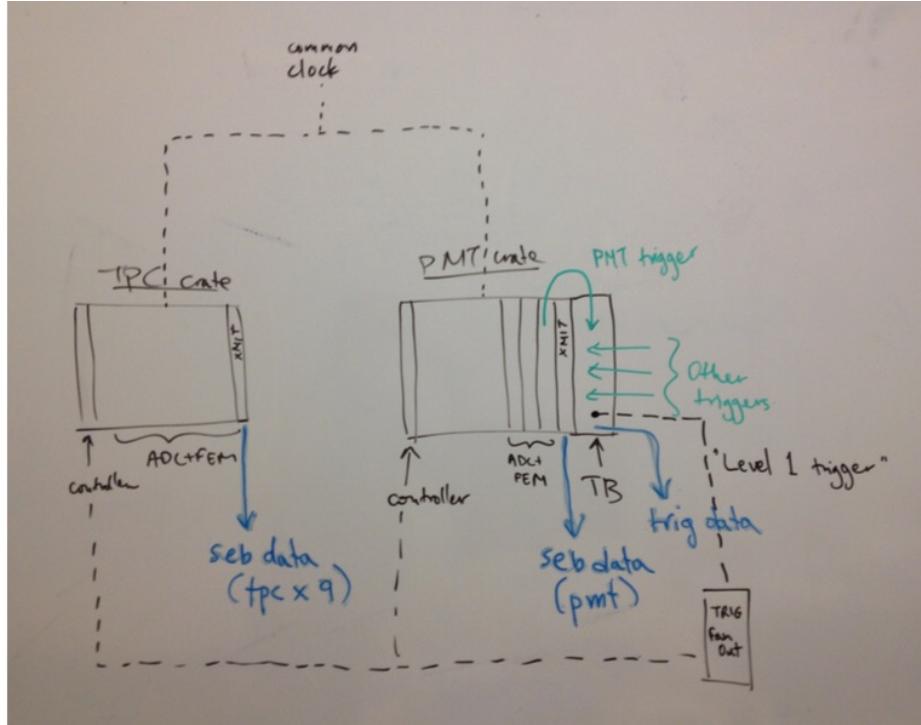


Figure 64: Readout control sequence. The MicroBooNE Level-1 trigger is a hardware trigger which consists of the OR between a BNB, NuMI, and EXT (strobe) trigger. Once received, the Level-1 trigger is propagated to all readout crates and instructs PMT and TPC data readout into ten dedicated sub-event buffer DAQ machines. The data across different DAQ machines is correlated at event building stage by the trigger number and corresponding trigger frame and sample numbers recorded in each data stream, per event. All readout crates are synchronized and correlated to the same, 16 MHz clock. **NOTE: Placeholder figure!**

1 **6.6 DAQ Design**

2 The MicroBooNE DAQ system acquires data from the readout electronics, writes data to local disk
3 before transferring it to long-term storage, configures and controls the readout electronics during
4 data-taking periods, and monitors the data flow and detector conditions. These tasks are performed
5 on a network of commodity servers running both custom and open-source software.

6 The data from each crate of the backend electronics is sent to a dedicated server (called the
7 sub-event buffer, or SEB) via an optical fiber, arriving in a card on the SEB’s PCIe bus. A real-time
8 application places these data in an internal buffer, collects all segments belonging to an event, and
9 creates a sub-event fragment that may be routed to a specified destination. For the NU stream,
10 in which the data arrives with every trigger, these fragments are sent to a single event-building
11 machine (EVB) over an internal network. Full events are checked for consistency and written to
12 local disk on the EVB before being sent offline for further processing. A high-level software trigger,
13 described in section 6.7, is applied to the data to determine whether events should be written locally
14 or ignored. For the SN stream the data remains on the SEB where it is written to disk and only sent
15 for offline analysis on explicit requests.

16 Data writing to either triggered or SN streams is limited by the RAID6 disk write speeds which
17 are roughly 300 MB/sec. This is much less than the network bandwidth bottleneck, which is 10
18 Gbps. The 300 MB/sec disk write speed therefore sets the maximum aggregate rate at which all
19 SEB fragments can ship data to the EVB without loss of data. With Huffman compression, which
20 gives a data reduction of approximately a factor of five (figure 60) and the PMT trigger, which
21 reduces the data rates by another factor of > 70, this is more than sufficient for MicroBooNE’s
22 maximum 15 Hz beam spill rate. MicroBooNE expects a total triggered write rate of around 12
23 MB/sec. The SN stream circular buffers, which will be aggressively (non-losslessly) compressed
24 beyond what the triggered stream experiences, will fill each server’s 14 TB in on the order of one
25 day, which is ample time to respond to a Super Nova Early Warning System (SNEWS) alert [?].

26 After data is written to disk, it is then copied to another server on the internal DAQ network,
27 where the raw data is further compressed, shipped, and queued to be stored on in tape and disk
28 cache using the Fermilab central data management system known as SAM. Offline applications
29 then begin processing the raw data, converting the binary data format into a LArSoft ROOT-based
30 format which can be used as input for reconstruction algorithms. A separate process collects beam
31 data and, during binary to LArSoft conversion, inserts that data into the built events. A duplicate
32 copy of the data is also stored offsite at Pacific Northwest National Laboratory (PNNL). This
33 collection of approximately 15 “projects” and the database which holds and monitors the state of
34 the data flow is known as the Python/Postgres for MicroBooNE Scripting system (PUBS), and
35 is patterned after a similar database state machine that the Double Chooz experiment used for
36 data management. As PUBS pushes the data through this process, the progress of each project
37 is monitored and viewable via GUI. PUBS can also monitor the state of the SN stream data, held
38 locally on the SEBs. A separate offline PUBS instance controls the processing of the data, including
39 applying newly calculated calibration constants as part of data quality management. Figure 65
40 schematically depicts the flow of data throughout the MicroBooNE DAQ system.

41 Additional software components handle the management of the main DAQ processes moving
42 the data. A run control application issues configuration and state-progressing commands to the

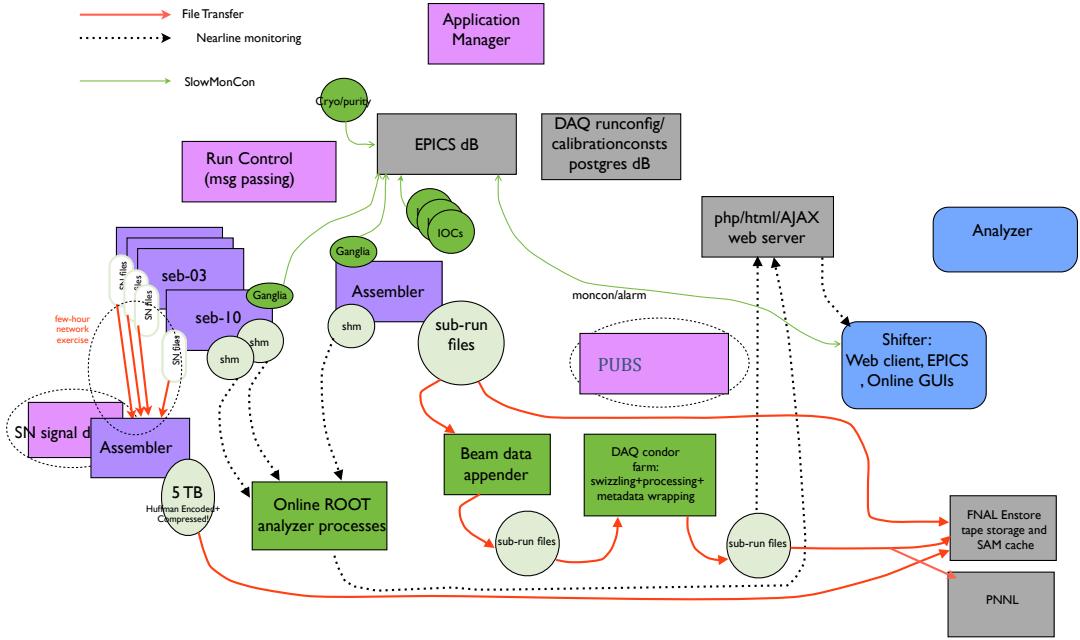


Figure 65: Data flow for MicroBooNE DAQ from raw to processed.

1 SEBs and EVB. Configuration states are stored in a dedicated run configuration database, which
 2 allows for the setting and preserving of configuration information for the DAQ, readout, and addi-
 3 tional components. This database not only allows for creating the large (~ 200 parameters) intricate
 4 DAQ run configuration files, but also enforces certain conditions which must hold for consistency.
 5 For example, the configuration that initiates the ASICS charge-injection calibration also dials and
 6 captures the settings on the external pulser that drives the calibration signal and assures that ASICS
 7 gains and peaking time parameters are enforced and recorded.

8 Another important aspect of the system is monitoring the health of the DAQ. Monitoring of
 9 DAQ components is accomplished through Ganglia which monitors basic system states (such as
 10 CPU, memory, and network usage) as well as allows use of custom metrics to monitor the data flow
 11 and status of the readout electronics [?]. These metrics are sampled and collected by the EPICS
 12 slow monitoring and control processes, which archives desired quantities and provides alarms when
 13 pre-defined thresholds are exceeded [83]. Some examples of Ganglia metrics that are monitored
 14 and alarmed in EPICS are the rates of growth of the SEB data buffers, the fragment rates leaving
 15 each SEB, and fragment arrival rates at the EVB. Figure 66 shows examples of Ganglia metrics.

16 Additional online monitoring exists to check data quality in more detail, through both pro-
 17 grammatical checks and visual checks including a real-time event display. The online monitoring
 18 takes snapshots into shared memory segments on the SEBs and the EVB, and thus provides the
 19 desired low latency checks of newly-arriving data. It continually walks through these ≈ 150 MB

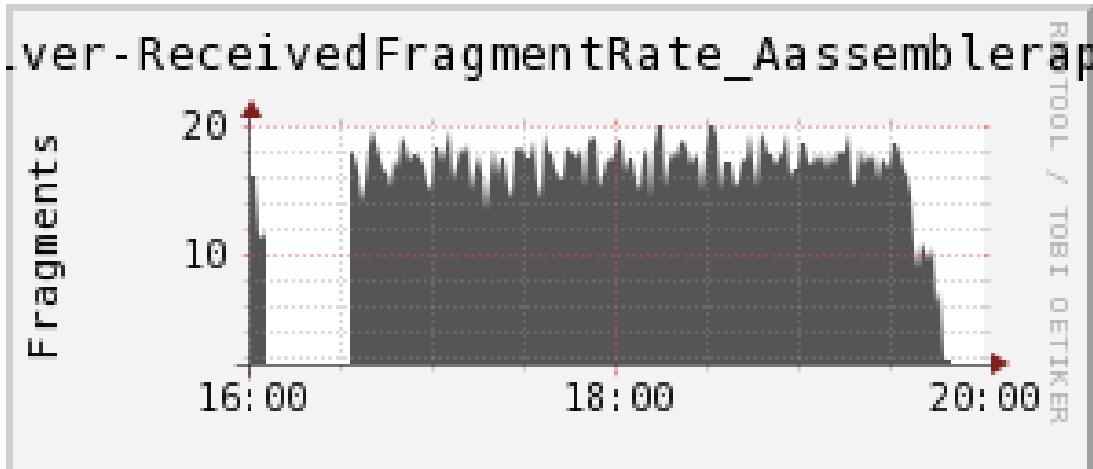


Figure 66: Ganglia metrics showing data flow on the EVB machine during a 2 Hz test run. (ul) the event disk write rate (ur) The number of assembled queued-up events waiting to be written (ll) The total EVB received data rate, and the (lr) received fragment rate.

1 snap-shotted events and outputs histograms of occupancies and rates which are saved in ROOT files
 2 [?]. The histograms are then displayed in a web-based monitoring system that is easily accessible
 3 by the shift crew. Channels are aggregated in a variety of formats, including the order in which
 4 they appear in crates or across the wires and PMTs themselves. In this way, potential problems
 5 across connectors or crates, for example, may be more readily identified. Noisy, quiet, and unre-
 6 sponsive channels are easily marked and displayed to the shift crew. Figure 67 shows an example
 7 of available online monitoring information.

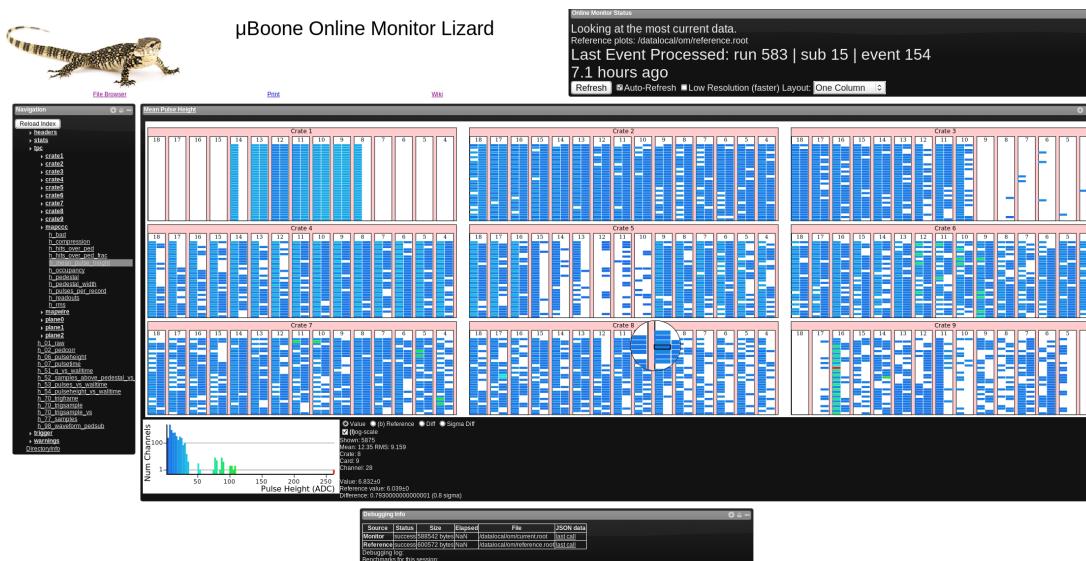


Figure 67: The online monitoring GUI (Lizard). The pedestal-subtracted ADC values for all 8256 wires in one minimum bias event are shown. Each box is a channel. Not all crates are fully populated with electronics.

6.7 High-level software trigger

After events are collected on the DAQ event-builder server, a suite of trigger algorithms are applied to the data. Currently, these algorithms mimic the PMT readout electronics' FPGA-based beam gate trigger algorithms, described in section 6.5. A search over the PMT digitized waveforms from the beam-spill period is performed, and if there is a significant amount of light in coincidence with the expected arrival time of the neutrinos, the event is marked and saved. This selection is performed on data that passes the level-1 trigger, which typically includes data from the BNB and NuMI beams, and randomly selected off-beam data from an "external" trigger. A fraction of the data from each of these level-1 trigger input streams is also retained via a random prescale, which provides a selection of data that has not been biased by the trigger. The high-level trigger algorithms take approximately 10 ms to return a result, a latency that is well-below the event-taking rate and so does not impact data-taking performance. The average pass rate for data-events in the PMT beam gate trigger algorithm is roughly 5%.

7. Infrastructure and Monitoring Systems

MicroBooNE is housed at LArTF, which is located on-axis to the BNB. Complete knowledge of the electrical and cryogenic systems housed within LArTF is necessary to maintain acceptable operating conditions for the experiment. Continuous monitoring of the beam being delivered to LArTF is also necessary for subsequent physics analyses. This section describes the details of infrastructure within LArTF, as well as monitoring of the experiment and beam conditions.

7.1 Electronics Infrastructure at LArTF

This section describes the electronics infrastructure of the experiment, which is essential for proper operation of the electronics that control the functioning of and extract data from the LArTPC, PMTs, and other detector subsystems. Figure 68 shows a diagram of this system at LArTF, depicting racks located on a platform directly above the cryostat that house electronics for: LArTPC control and readout, light collection system, drift high voltage, purity monitor, calibration laser, trigger, and cryogenic control systems. Additional server racks containing the data acquisition, beam timing, and external network electronics are located in a separate computer room above and adjacent to the above-cryostat platform. The distribution of power, data, and network connections to and from all of these racks are also presented in figure 68, and is described in more detail in the following sections, as are electronics safety systems and interlocks.

7.1.1 AC Power Distribution and Grounding for Low-Noise LArTF Data-taking

As with any large detector operating with a high dynamic range, prevention of electromagnetic interference and its attendant effects on MicroBooNE data is an essential aspect of detector design. MicroBooNE's strategy for producing a low-noise environment for the LArTPC and associated readout electronics can be largely summarized in a few key points. AC power distribution-related items will be described here, while cabling, connections, and shielding will be described in a following section.

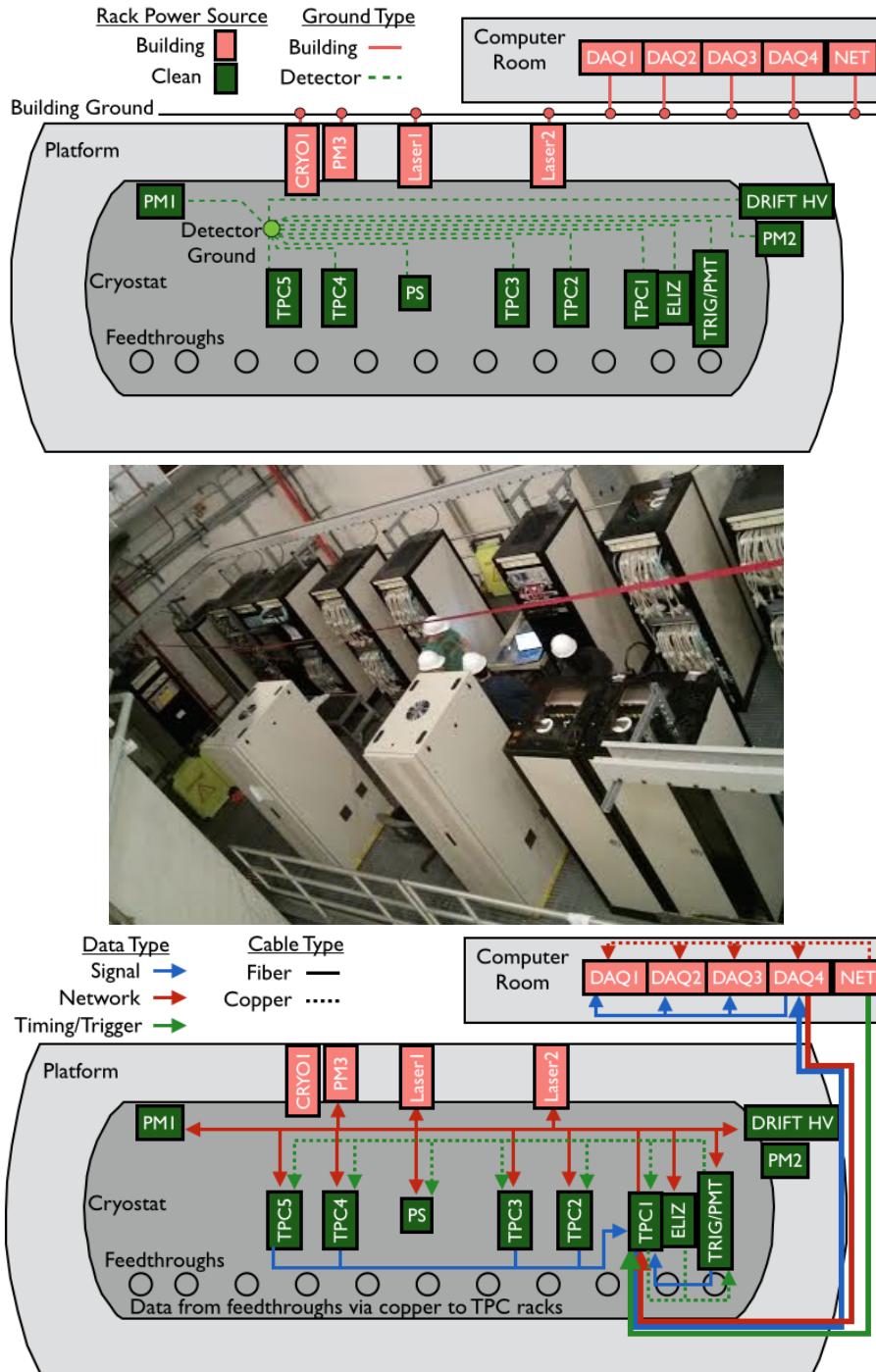


Figure 68: Top: Diagram illustrating location of deployed electronics racks and separation of detector and building grounds and clean and building power for differing racks. Middle: Photograph of installed electronics racks on the LArTF platform. Bottom: A diagram illustrating the general scheme of signal, network, and timing signal cabling in the LArTF computer room and platform.

- ‘Clean power,’ or AC power electrically isolated from AC power for the rest of LArTF (‘building power’), is supplied to all sensitive electronics via an isolation transformer.
- Highly sensitive electronics are housed inside the Faraday cage provided by the detector cryostat or inside Faraday cages directly grounded to the cryostat.
- The detector cryostat is grounded to a ‘detector ground,’ which is physically and electrically isolated from the ground provided to all other LArTF power circuits, or ‘building ground.’
- No direct electrical connections are present between detector ground and building ground. This is accomplished through the use of insulating platform and cryostat saddle materials, insulating cable trays and cables, and by inserting insulating ‘breaks’ (i.e. fiber data links or insulating cryo pipe sections) when connections between sensitive and potentially noisy detector components are necessary.
- Indirect pickup on clean signals through capacitive coupling to adjacent noise sources is minimized through use of detector-grounded shielding and electrically-insulating cable trays.
- Ground loops on detector ground are avoided wherever possible by connecting all electronics racks directly to the cryostat and by minimizing direct electrical connections between racks.
- Direct or capacitive couplings between building and detector ground are constantly monitored during installation and operation with a custom-designed impedance monitor.

A line drawing describing the production of clean power and clean ground are shown in figure 69. Two 200-Amp clean power circuits produced at isolation transformers are used to power all sensitive racks, which are indicated in figure 68. All racks containing LArTPC readout electronics are placed on one circuit, while all other sensitive equipment is placed on the alternate circuit. On the platform, all racks utilize clean power with the exception of the calibration laser and in-line purity monitor racks, which either contain noise-producing elements or support building-grounded components. All racks in the LArTF computer room utilize building power.

208-3phase power is distributed to each individual electronics rack. For racks with significant power requirements or a large number of components, this power is delivered to a Fermilab-designed ‘AC switch box,’ which distributes power to an Eaton Power Distribution Unit (PDU) only upon receiving an interlock signal from a smoke detection system in each rack, which will be described in more detail below. Rack components then receive power from one of the three phases on this PDU. For racks with fewer requirements, power is supplied to components directly from an interlocked simplified AC switch box or SurgeX SX-1120-RT PDU. Racks with sensitive electronics are grounded to the cryostat via copper sheeting running throughout insulated cable trays above the cryostat. Sensitive components within each rack are connected to a tin-plated copper grounding bar electrically connected to the rack bottom and running the height of the rack. Mechanical attachments to the rack provide grounding for less sensitive rack components. As mentioned before, any unintentional direct connection between building and detector ground can be quickly identified by the impedance monitor located on the LArTF platform.

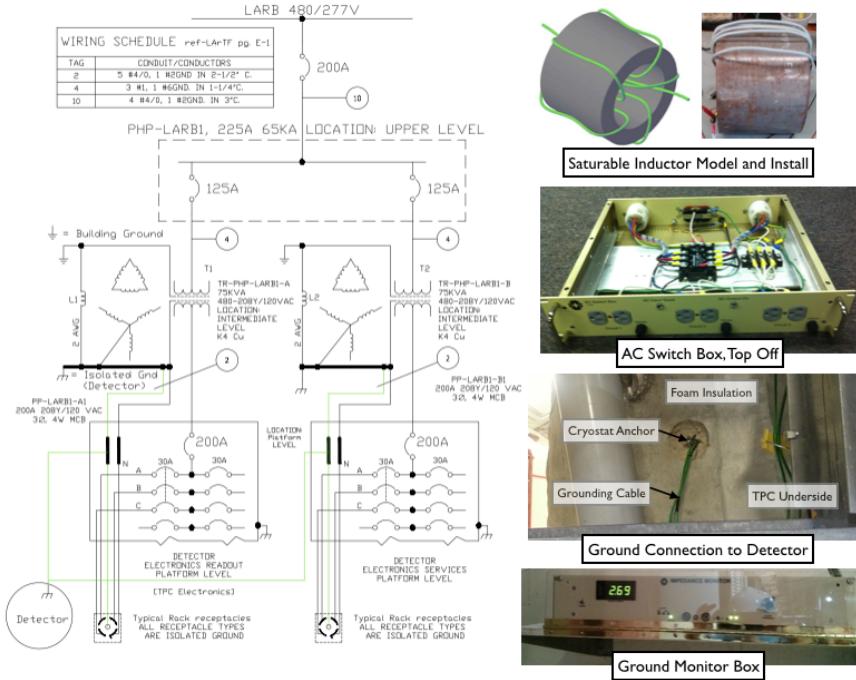


Figure 69: Line drawing of clean power generation and distribution and connections to detector ground, along with pictures of the installed saturable inductor (top), AC switch box (top middle), detector ground strap and connection (bottom middle), and impedance monitor (bottom).

7.1.2 DC Power Distribution to the MicroBooNE Detector

DC power is provided to the LArTPC and readout electronics by power supplies in clean-powered, detector-grounded racks for a variety of purposes:

- Holding the LArTPC cathode plane at voltage to produce the desired ionization electron drift speed
- Holding the two ungrounded anode planes at the proper constant voltage to ensure the planes are transparent to drifting electrons
- Operating the light collection system PMTs
- Powering the cold electronics located inside the LArTPC
- Powering the warm electronics located in the LArTPC readout electronics racks
- Powering auxiliary systems, such as LArTPC temperature sensors and purity monitors

Table 7 summarizes the required voltages or currents for each of these purposes as well as the power supply make and model utilized in each case. Power supplies are located in the relevant subsystem's electronics rack, with power and grounding connections as dictated by that rack.

DC supply power consumption is minimal in most cases, with the exception of the warm LArTPC electronics PL-508 supplies, which can draw in the vicinity of XXX W during normal

Table 7: Overview of MicroBooNE DC power distribution. Delivered voltages or currents are listed, along with power supply makes and models and whether each supply utilizes clean or building power.

Sub-System	Supplied Value	Supply Made/Model	Clean Power?
Drift High Voltage	120 kV	Glassman LX150N12	Y
Anode Plane Voltage	1 kV / 8 mA	Wiener MPOD ISEG_EHS 82 10X_805F	Y
PMT High Voltage	2 kV / 3 mA	BIRA T4	Y
Cold Electronics Power	8 V / 10 A	MPOD MPV 800I	Y
Warm LArTPC/PMT Electronics	+/-5V MDH 2-8V/25A +12V MEH 8-15V/92A +3.3V MEH 2-7V/115A	Wiener PL508	Y
Auxiliary Systems Power	Various	Various	Laser: N Inline PM: N Cryostat PM: Y

¹ operation. Care was taken to distribute the AC power load by limiting the number of high-draw PC supplies per electronics rack.

³ 7.1.3 Network, Timing, and Data Distribution for Low-Noise LArTF Data-taking

⁴ Network, timing, and data connections must be made between the detector, building-ground, and ⁵ detector-ground racks to properly read out MicroBooNE data. However, as described above, these ⁶ connections must be made while maintaining strict detector-building electrical isolation. Deployed ⁷ interconnections meeting both of these requirements are displayed in figure 68.

⁸ Timing and LArTF-external network signals are brought into LArTF via electronics in the ⁹ computer room, where all racks are building grounded and powered. These signals are distributed ¹⁰ and processed in the computer room via copper cable, while network and processed timing signals ¹¹ to be sent to the platform are converted onto fiber cables and aggregated into a central fiber ¹² termination box. A fiber trunk line then delivers these signals to the platform, where another fiber ¹³ termination box on a detector-ground rack is used to fan out these signals. Network connections are ¹⁴ fanned out via fiber to a network switch in each platform rack, while timing signals are re-converted ¹⁵ to copper and further processed for use by the trigger system on a different detector-grounded rack. ¹⁶ All rack-to-rack cables are run in insulating cable trays beneath the platform.

¹⁷ PMT and LArTPC data are transferred from each detector feedthrough to readout crates in ¹⁸ detector-ground racks via insulated copper cable whose shield is tied to detector ground. Digitized ¹⁹ crate output is then sent to the aforementioned platform fiber termination box, where these signals ²⁰ are sent via fiber trunk line to the computer room. In the computer room, these fibers are then ²¹ fanned out to the appropriate DAQ computer. Readout crate and cold electronics control commands ²² are transmitted in the opposite direction utilizing a similar scheme, with crate controls delivered ²³ directly via fiber, and cold electronics commands delivered via ranger fiber to a copper fanout in a ²⁴ detector-ground rack.

1 Clock and trigger signals must also be sent from a central trigger rack to all detector-ground
2 LArTPC/PMT readout racks. These signals are transmitted via copper connections, and represent
3 the only source of ground loops on detector ground. To further reduce the possible impact of
4 induced noise in these and all copper cables mentioned above, all insulating cable trays beneath
5 the platform are lined with copper sheeting grounded to the detector. As an additional precaution,
6 all LArTPC signal copper cables are run in separate cable trays from power and auxiliary cabling
7 beneath the platform as well as inside every rack.

8 All cables between all detector components have been uniquely labelled with serial number,
9 source, and destination to allow for ease of replacement and reconnection. Ample fiber and copper
10 spares for every major cable type are also installed along with the production cables to allow for
11 quick replacement of any failed cable.

12 **7.1.4 Interlocks and Safety Systems**

13 All electronics racks contain smoke-sensing and temperature-monitoring systems, which, when
14 interlocked with AC and DC power transmission in each rack, constitute a rack protection system
15 (RPS) designed to meet Fermilab safety requirements and reduce the risk of fire and related damage
16 in LArTF and to individual rack components.

17 The RPS principally consists of a smoke sensor connected to a Fermilab-designed rack pro-
18 tection box. This box produces and outputs a 12 V interlock signal when the rack protection box
19 is on and receiving a 'no-smoke' signal from the smoke sensor. This 12 V signal can be sent to the
20 AC distribution box located in each rack, as described above, to allow AC transmission to all rack
21 components only if the RPS is on and not detecting smoke. A similar 12 V 'RPS Status' signal
22 is also produced by the RPS box for input into the MicroBooNE slow control box, which will be
23 described in following sections. Alternate contacts are available on the rear of the RPS box for
24 coupling the status of additional subsystems, such as the DAQ and calibration laser uninterruptible
25 power supply (UPS), to smoke sensor or rack power status.

26 Temperature sensors deployed in two or three locations in each electronics rack sample air
27 temperature within each rack. Temperatures at each sensor are read out and recorded in the slow-
28 control database by the slow-control monitoring box. In addition, the box also produces a 5 V
29 interlock signal if all sampled temperatures are within pre-programmed thresholds. In electronics
30 racks distributing PMT- or LArTPC-related DC power these temperature interlock signals are input
31 into each relevant power supply, allowing DC power distribution only when this interlock signal is
32 present, for safety purposes.

33 Additional hardware interlocks ensure the non-simultaneous operation of particular subsys-
34 tems. In particular, the PMT system is disabled when cryogenic system liquid level sensors detect
35 a level below that of the highest PMT bases, or when the UV laser system is active. The former
36 requirement is enforced with a dry-contact hardware interlock, while the latter is enforced with a
37 software interlock in the MicroBooNE online software.

38 **7.1.5 Performance Measurements**

39 The proper operation of each production electronics rack's AC and DC distribution and RPS sys-
40 tems has been tested prior to installation at LArTF. Furthermore, test stands exercising function-
41 ality of DAQ, PMT and LArTPC electronics, trigger, and drift HV systems have successfully in-

1 incorporated and tested various aspects of these same AC and DC distribution and RPS systems.
2 Impedances between detector and building grounds were recorded throughout the installation of
3 the rack infrastructure at LArTF using the impedance monitor located on the LArTF platform.

4 **7.2 Slow monitoring and control system**

5 MicroBooNE uses the Experimental Physics and Industrial Control System (EPICS) [83] for con-
6 trolling and monitoring most devices and conditions important to the experiment. These include
7 power supply controls, temperatures, fan speeds, rack protection interlock status, and various en-
8 vironmental conditions. The DAQ, cryogenics systems, and beam data collection systems operate
9 independently of the EPICS slow monitoring, but export data which are imported into EPICS for
10 archiving and status displays. Applications from the Control System Studio software collection
11 [84] are used for providing displays, alarm notifications, and data archiving.

12 An EPICS system consists of any number of server programs implementing the EPICS Chan-
13 nel Access (CA) protocol [85] to provide client programs access to any number of process vari-
14 ables, where each process variable represents a quantity being controlled (an output) or measured
15 (an input). The EPICS base distribution provides a standard type of channel access server called an
16 Input/Output Controller (IOC), which can be extended to support specific hardware as desired.

17 Most power supplies are controllable over the network through the NetSNMP protocol [86].
18 Several EPICS driver modules are available for SNMP, and MicroBooNE utilizes one written at
19 NSCL [87]. An IOC with this SNMP module runs on a central computer and contacts the power
20 supplies over a private network for monitoring and control. The photomultiplier power supplies are
21 reused from the D0 experiment and have custom IOCs running in their own controllers [?]. The
22 main high voltage power supply has only a simple RS-232 serial interface; control and monitoring
23 for it is provided by a nearby computer running an IOC with the EPICS asynDriver [88] and
24 StreamDevice [89] modules.

25 MicroBooNE has a number of racks in various positions above the detector and in an adjacent
26 server room. Each is equipped with a rack-protection system and multiple digital temperature sen-
27 sors, and most contain one or two fan packs, each containing 6 fans. To monitor and control these
28 devices, each rack has an 1U rack-mount enclosure containing an ARM-based single-board com-
29 puter (SBC) running Linux. An off-the-shelf GESBC-9G20 from Glomation Inc. [90] is utilized
30 for the SBC. A custom interface board [91] connects the SBC to front panel LEDs, temperature
31 probes, fan packs, and rack-protection-status input. The temperature sensors are DS1621 chips,
32 controlled and read out over an I2C bus by the SBC's I2C controller. The DS1621 also has a ther-
33 mostat output with programmable trip and reset temperatures, which are connected via the interface
34 board to outputs that can be used to interlock devices in the racks, such as power supplies. The fans
35 provide pulse-per-rotation outputs, which are monitored by a 12-channel tachometer implemented
36 via a PIC16F887 microcontroller, and also read out by the I2C bus. An EPICS IOC runs in each
37 SBC, with custom device drivers for reading all status information and controlling the heartbeat
38 LED and temperature sensor trip and reset points.

39 Data are imported into EPICS channels from a number of external sources. The primary rea-
40 son for duplicating these data in EPICS is to integrate displays and warnings into one system for
41 the experiment operators, and to provide integrated archiving for sampled data in the archived
42 database. An IOC running on a central computer provides “soft” process-variables channels for

1 these data. The data acquisition system provides many metrics describing its operation via the
2 Ganglia system[92, 93], which makes the data available in an XML format easily read by a Python
3 script, which in turn writes to EPICS using the PyEPICS module [94]. The hardware and system
4 status of the DAQ computers is monitored through the industry standard Intelligent Platform Man-
5 agement Interface (IPMI); rather than writing a script to import data from IPMI directly into EPICS,
6 a IPMI-to-Ganglia interface provided by the FreeIPMI’s “ipmi-sensors” package [95] is used, al-
7 lowing data to be imported via the same mechanism used for the DAQ metrics. Separate Python
8 scripts periodically retrieve data about outside weather conditions from various sources, cryogen-
9 ics system data from a file retrieved non-intrusively from the IFIX cryogenics control system, and
10 beam data from Fermilab’s Intensity Frontier Beam Database (IFDB) [96].

11 **7.3 Beam Monitoring**

12 The primary source of neutrinos for the MicroBooNE experiment is the BNB. The primary beam-
13 line is lined with instrumentation including toroids which indicate beam intensity, “multiwires”
14 showing beam profile in the horizontal and vertical planes, and beam position monitors measuring
15 the mean beam position. Data from these monitors are stored on a spill-by-spill basis in the IFDB.
16 Many of MicroBooNE’s physics analyses require that beam data are recorded for each spill and
17 matched to detector events.

18 Primary beam monitoring in MicroBooNE is done using a “dashboard” interface to IFDB.
19 By using the IFBD instead of the accelerator control system, the experiment can also verify that
20 data are being acquired by the IFBD. The dashboard is accessible over the network using a web
21 browser. The final monitoring step includes a post-data-merge check, ensuring that beam data are
22 successfully matched with detector data for all beam spills. This is done once the detector DAQ
23 binary data file is closed.

24 The dashboard presents a graphic representation of the data, allowing for easy error identifica-
25 tion, as shown in figure 70. The experiment monitors: two toroids which indicate beam intensity;
26 three multiwires, each of which shows beam profile in each plane; and beam position monitors
27 along the beamline which show the vertical and the horizontal position. Parameters pertaining to
28 the target and horn, such as cooling air temperature and horn current, can also be monitored. The
29 dashboard allows the experiment to easily add additional devices if experience demonstrates the
30 need for their monitoring.

31 Data are monitored in near real-time. A reasonable history is also kept so that changes are
32 easily identified. The accelerator control system provides detailed diagnostics tools to experts and
33 can be used in case of any problems.

34 **8. UV Laser System**

35 The knowledge of the electric field inside the drift volume of a LArTPC is a necessary aspect
36 for performing subsequent event reconstruction. Distortions of particle tracks due to field non-
37 uniformities affect the accuracy of the particle momentum reconstruction based on multiple scat-
38 tering. Deviations from a uniform drift field may arise mainly due to accumulation of positive argon
39 ions in the drift volume. These ions are produced by ionizing particles from neutrino interactions,
40 as well as by cosmic rays. While electrons produced by ionizing particles are quickly (within few

Booster Neutrino Beamline Status Display (e,1d events) - Historical data for 2015-07-01 11:15:03

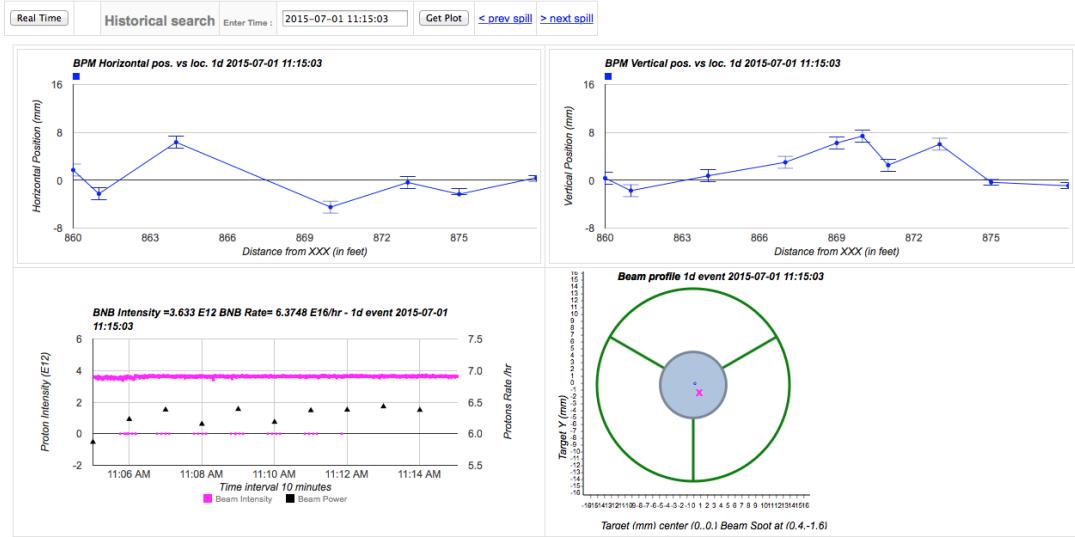


Figure 70: The BNB dashboard, showing graphical representation of beam instrumentation data, is used to monitor the beam. The top box shows the timestamp of the beam spill and indicates if data is stale by changing the color. The two top plots show the primary proton beam position along the BNB. The bottom left plot shows the recent beam spill intensity and rate. The bottom right plot shows the beam as projected onto the Beryllium target (the grey circle in the middle with radius of 0.5cm). The dashboard is accessible via web page providing both real time updates and the review of past data. The page can be easily extended to monitor additional beam devices.

1 milliseconds) swept towards the readout system, ions have significantly lower mobility. Their drift
 2 velocity in the MicroBooNE detector at nominal drift field is of the order of 0.8 cm/s. The rate of
 3 cosmic muons in the LArTPC volume is estimated to be 11,000 muons/s within the active volume
 4 (assuming a cosmic rate of 200 muons/m²/s through a horizontal plane at the earth's surface and
 5 63 muons/m²/s through a vertical plane), traversing a combined length of 1.9×10^4 m through the
 6 liquid argon. Assuming that cosmic muons are minimum-ionizing (2.1 MeV/cm) and produce 23.6
 7 eV per ion pair, positive ion charge is produced at a rate of 2.8×10^{-8} C/s in the MicroBooNE TPC.
 8 These ions are continuously neutralized at the cathode. The resulting charge distribution in equi-
 9 librium is shown in figure 71. Such accumulated space charge leads to noticeable distortion of the
 10 drift field and, consequently, to deviations of reconstructed track coordinates by up to 10cm (see
 11 figure 72). The ion drift velocity is comparable to local argon flow velocities, produced by global
 12 argon recirculation flow and thermal convection. Therefore the distribution of positive space charge
 13 inside the drift volume may be not only nonuniform (figure 73), but also non-stationary.

14 A nonuniform drift field in the LArTPC leads to bending of initially straight tracks of high-
 15 momentum ionizing particles. In principle, a set of events from such particles allows for the re-
 16 construction of the field in any small region of the LArTPC drift volume, using the systematic
 17 apparent curvature of tracks at different angles passing through that region. In practice, the rate
 18 of such events from cosmic muons is too low to acquire sufficient statistics in reasonable time. A

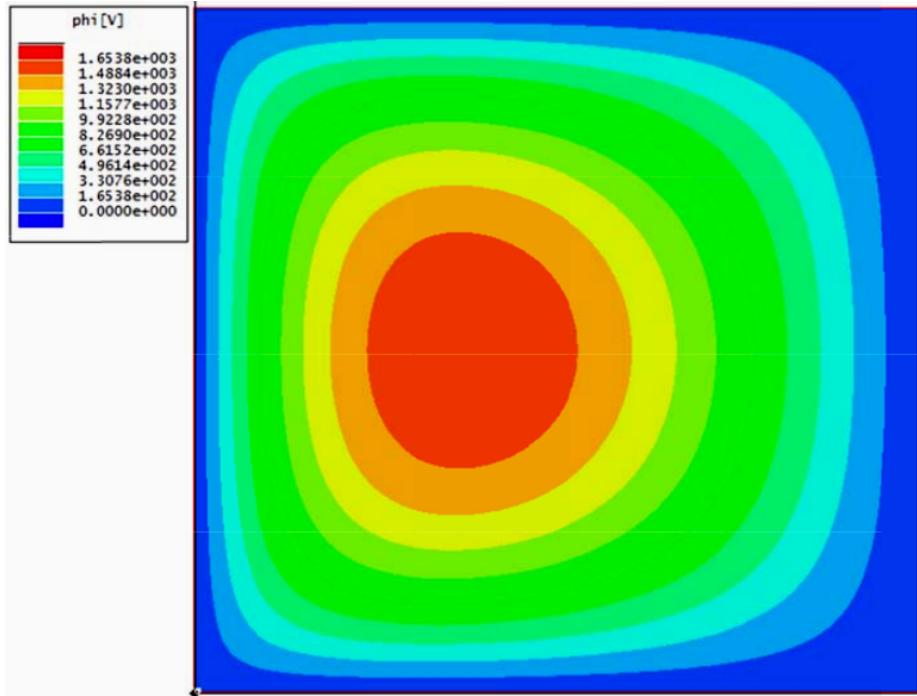


Figure 71: Distorting potential distribution due to positive space charge in equilibrium in the MicroBooNE detector.

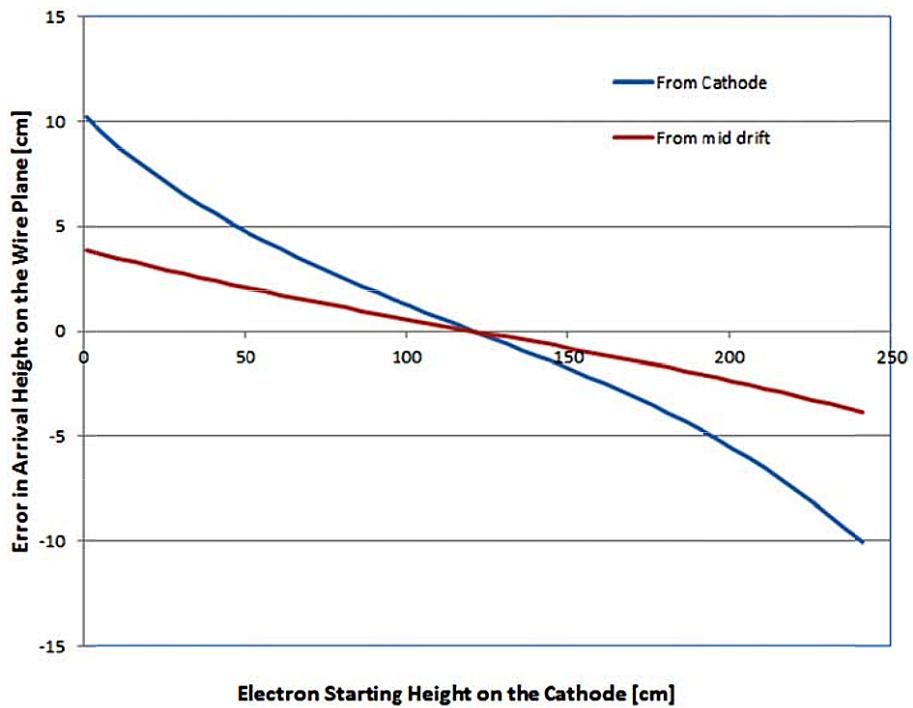


Figure 72: Deviation of a crossing track from its true coordinates due to positive ion space charge.

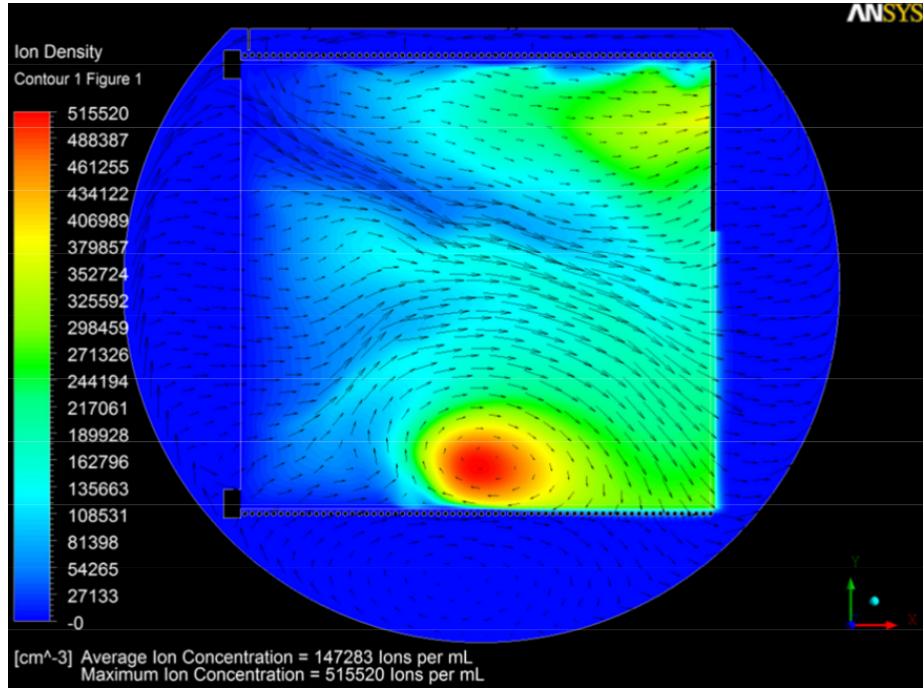


Figure 73: Distribution of the positive space charge in presence of argon circulation.

method to generate straight ionization tracks at a defined location in liquid argon is described in [97]. A collimated photon beam from a pulsed UV laser with $\lambda=266$ nm can ionize liquid argon via multi-photon absorption. The resulting ionization track is straight, characterized by low electron density, therefore featuring little charge recombination loss, unlike cosmic muon tracks. Laser tracks are also free from δ -electrons, which complicate track reconstruction in the case of muons. The method was successfully exploited in the Argontube long drift LArTPC [25 – 27] to derive the non-uniformity of the electric field along its 5 m long drift volume (see [98]). The MicroBooNE LArTPC requires a set of such tracks in order to cover the whole sensitive volume to reconstruct field distortions. Tracks are generated one at a time by steering a pulsed laser beam with the use of a custom-designed opto-mechanical feedthrough (see [99]). The pulse rate of the laser generator is 10 Hz. This allows production of a minimum required set of 100 tracks within one minute (taking into account steering time). Details of this solution are described in the following sections.

8.1 UV Laser Calibration

In order to unambiguously reconstruct a drift field vector at any point within the detector fiducial volume, a minimum of two ionization tracks are required to cross in the region of interest. The total number of crossing points is determined by the required reconstruction granularity. In MicroBooNE the initial scenario is to acquire 100 tracks from each direction, producing a reconstructed 3-D map with voxels that are approximately $10 \times 10 \times 50$ cm 3 in volume. This map provides a rough picture of the space charge distribution. Depending on the results of this measurement, areas of interest can be studied in more detail. Repetitive study of small volumes may reveal dynamics in the space charge distribution due to turbulent circulation, and should further inform an optimized scenario for a standard UV laser calibration procedure.

1 An algorithm of drift field calibration utilizes an input array of detector events with one straight
 2 ionization track in each element of the array. The result of the calibration is a coordinate correction
 3 map, to be applied to each track, which converts apparently curved track images back to the true
 4 coordinate system where they are straight. The algorithm is iterative with an optimizable iteration
 5 step and required accuracy. An example of simulated reconstruction in 2-D space is depicted in
 6 figure 74, showing that the magnitude of the field distortions is reduced from 10 cm down to several
 7 millimeters in 99% of the detector volume.

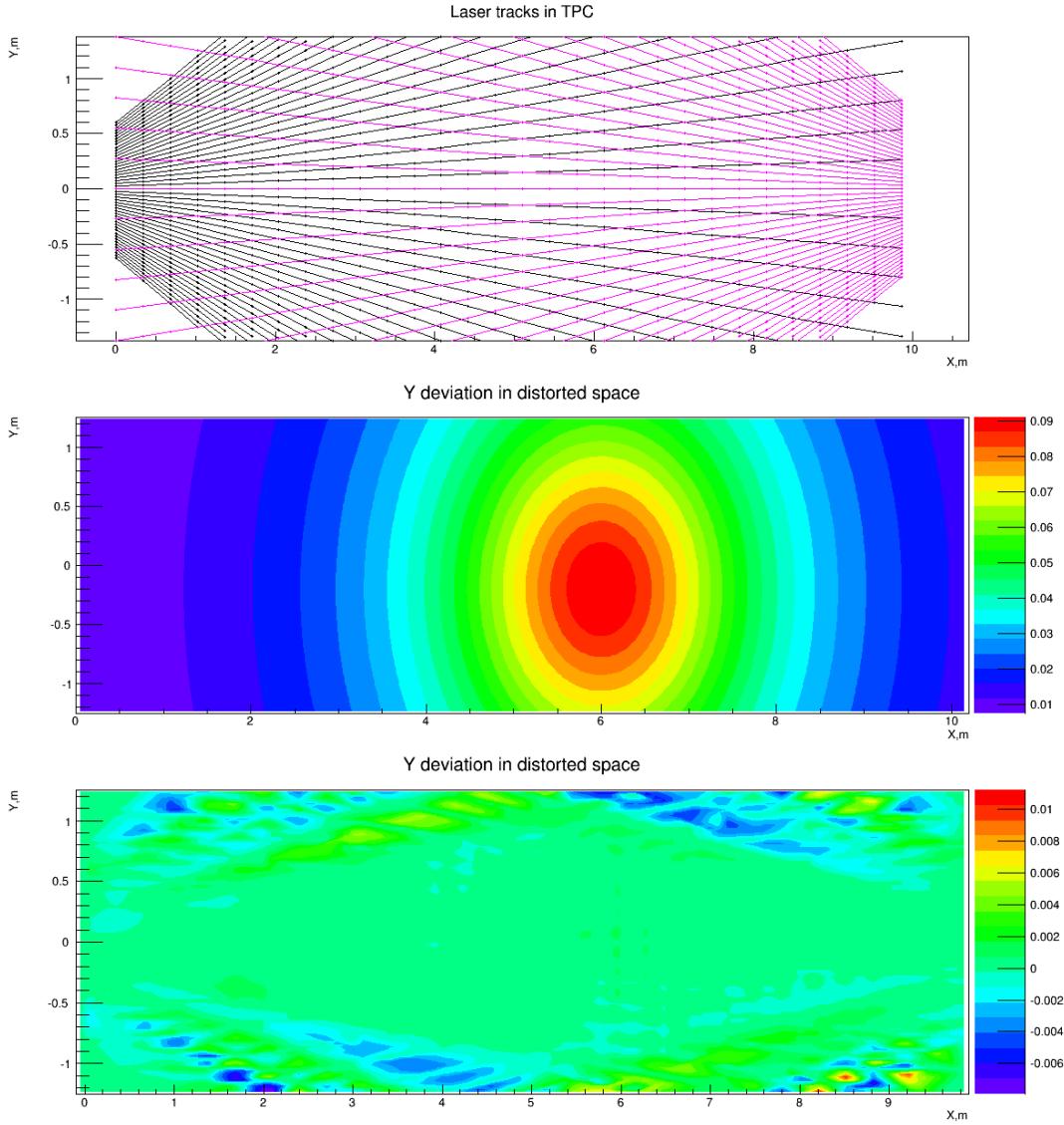


Figure 74: Top: Simulated true laser beam trajectories in the detector; Middle: map of Y coordinate of track deviation under influence of non-uniform electric field; Bottom: residual map of reconstructed track Y-coordinates versus true ones.

8 **8.2 Laser Source and Optics**

9 A Nd:YAG laser emitting light at a wavelength of 1024 nm is used as the primary light source
1 [?]. Inside the laser head, nonlinear crystals are installed in the beam line for frequency doubling
2 and summing, resulting in a wavelength of 266 nm needed for ionization of liquid argon. For this
3 wavelength, the Nd:YAG laser is specified to produce an output energy of 60 mJ for each 4 to 6 ns
4 long pulse and a horizontal polarization. The maximal repetition rate is 10 Hz; the beam has a
5 divergence of 0.5 mrad.

6 An optical table, as seen in figure 75, was developed to introduce the necessary parts in a
7 stable and compact environment. With regard to the operation of the optical table on MicroBooNE,
8 the parts were chosen to be accessible remotely where necessary. The emitted laser beam contains
9 not only ultraviolet light but also all other harmonics generated in the crystal and the primary light
10 of 1024 nm. Dichroic Mirrors optimized to reflect only wavelengths in the UV region are used
11 to filter higher wavelengths out. To absorb the transmitted wavelength behind the mirrors, glass-
12 ceramic plates are installed. The beam leaving the laser head is reflected by the first 45°-mirror
13 into an attenuator. For optical adjustment and verification of the non-visible UV-beam, a green
14 alignment laser is placed behind this mirror and adjusted such that its path is coincident with the
15 UV-laser beam. In the attenuator (Altechna Wattpilot) a turnable $\lambda/2$ -plate enables rotation of the
16 orientation of the laser beam polarization. Behind the attenuator two parallel plates are installed
17 such that the angle of the incident beam matches the Brewster angle of the reflector. Modulating
18 the polarization of the beam adjusts the intensity of the reflected beam. After the attenuator an
19 aperture is put in the optical path of the beam to control the beam diameter. The last part in the
20 beam line is a remotely-controllable mirror mount (Zaber T-OMG), which directs the beam to the
21 laser feedthrough on the cryostat. A photodiode (Thorlabs DET10A/M), which is sensitive in the
22 ultraviolet region, detects the scattered light when a laser pulse is fired. Its signal is then used as a
23 trigger for data taking. Both the UV-laser head and the optical table are mounted on a 15 mm thick
24 aluminum plate.

25 **8.3 Steering System**

26 One of the main challenges of the laser calibration system is the introduction of a steerable laser
27 beam into the detector. Earlier, an evacuated quartz-glass [100] was utilized to introduce a laser
28 beam into liquid argon, however this beam had a fixed path through the detector. For the purpose of
29 scanning the full detector a fully steerable mirror in liquid argon is necessary. In the MicroBooNE
30 detector, this is achieved by mounting a mirror on a horizontally-rotatable support structure. A
31 rack and pinion construction, where the mirror is mounted on the frontside of a half gear (pin-
32 ion), provides the necessary freedom for the vertical movement (see figure 76 right). To steer
33 the horizontal movement from outside the cryostat, a commercial differentially-pumped rotational
34 feedthrough is deployed (see figure 76 left). The rack and pinion construction is attached to a linear
35 feedthrough. Both feedthroughs are motorized to allow for remote control and automation of the
36 mirror movement. The mirror support structure was fabricated out of polyamide-imide (Duratron
37 T4301 PAI), which has a very low outgassing rate and low thermal expansion coefficient, and is
38 certified for operation at 87 K. To minimize the probability of discharges due to the close loca-
39 tion of the feedthrough to the field cage structure in MicroBooNE, no conductive parts were used

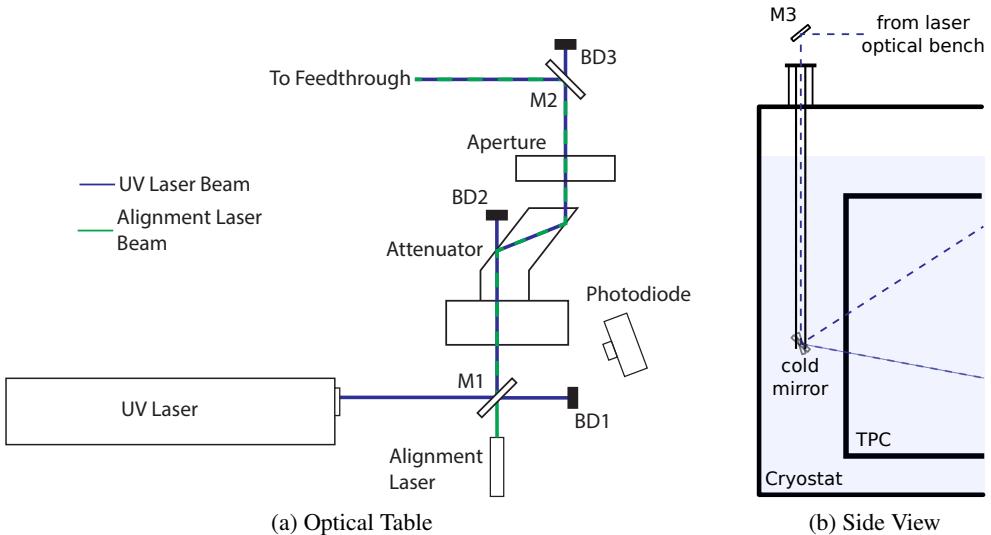


Figure 75: (a) A schematic drawing (not to scale) of the components used for laser beam configuration. An alignment laser (visible light) is introduced along the UV-laser path at the first dichroic mirror (M1), such that the paths overlap. In the attenuator the UV-laser beam intensity can be adjusted to the desired level, and the diameter of the beam is controlled by an aperture. A motorized mirror (M2) deflects the light into the direction of the feedthrough. Beam dumps (BD) are installed behind all mirrors to absorb the non-reflected laser light. (b) Side view of the cryostat indicating the mirror support structure with respect to the LArTPC.

40 in the support structure. The support structure has a total length of 2.5 m in MicroBooNE. Both
 41 feedthroughs are equipped with high precision position encoders from Heidenhain. The accuracy
 1 of the encoders is chosen such that a position accuracy of 2 mm for the laser beam spot over 10 m
 2 distance is achieved. An external interface box controls the encoders and records a position reading
 3 upon receiving a trigger signal from the photodiode. The DAQ computer accesses the position in-
 4 formation over an ethernet connection. The same computer is also used for steering the two motors
 5 via a motor driver system (over a RS232 interface).

6 **8.4 Performance Tests and Initial Operation**

7 A full performance test of the laser calibration system identical to the one installed in the Micro-
 8 BooNE LArTPC was performed prior to the final installation. Apart from the general proof-of-
 9 principle of the laser calibration system, several operationally relevant parameters were identified.
 10 These include scanning speed, positioning accuracy, positioning limits, optimal laser beam inten-
 11 sity, beam diameter, and the minimal achievable field distortion which can be resolved. The test
 12 system consists of a LArTPC equipped with 64 readout channels and an active area of about 400
 13 cm^2 , with a drift distance of 40 cm (see [101] for further details).

14 Several tests of the motorized feedthrough were performed under warm conditions before cold
 15 tests were conducted. One crucial parameter for the quality of the electric field calibration is the
 16 resolution at which laser tracks can be aimed in the detector. For the rotational axis this angle
 17 is directly measured on a circular scale. For the vertical movement the linear displacement of

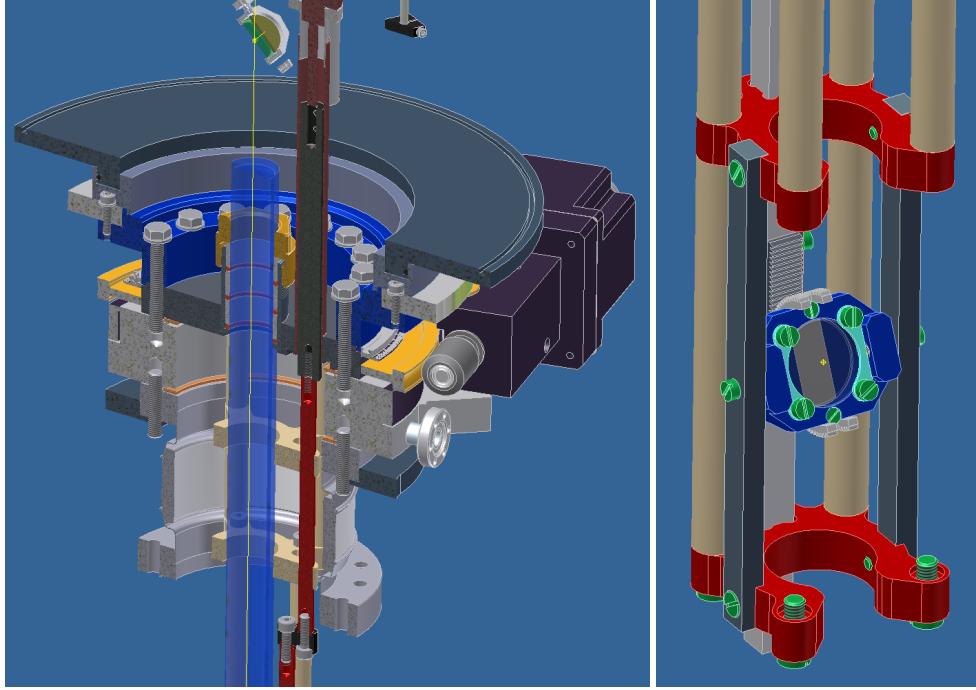


Figure 76: Left: CAD cutaway drawing of the feedthrough construction is shown. The yellow line indicates the laser path. Right, CAD drawing of the cold mirror including the support structure.

the bellow is translated into a rotation inside the cryostat, as can be seen in the CAD drawing in figure 76. This construction introduces uncertainties to the measurement position and backlash. The backlash can be compensated by always approaching positions from the same direction. For the translation of the linear movement ΔL into a rotation $\Delta\phi$ the translation ratio s according to $\Delta L = s \cdot \Delta\phi$ was measured with a laser alignment device³. The obtained ratio is $s = 0.3499 \pm 0.0002 \text{ mm/}^\circ$. The dominant uncertainty in the vertical position measurement is the accuracy of the encoder $\sigma_{\text{linear}} = \pm 1 \mu\text{m}$, which translates into a vertical rotation measurement accuracy $\sigma_{\text{vertical}} = 10.29''$.

Horizontal movement limitations arise from the construction of the feedthrough system, namely the warm mirror support structure. This limit has its origins in the way of mounting the laser table relative to the feedthrough on the MicroBooNE cryostat. Vertically the mirror can be rotated more than 45° relative to the horizon in both directions. In an upward looking configuration, no limitations arise which would affect the coverage of the detector with the beam. A limit arises when the mirror faces the opposite downward direction, when properly aligned onto the centre of the mirror the laser diameter and the size of the mirror limits the achievable coverage. However slight misalignment will affect this limit, since the beam will not be in the optimal spot anymore. In warm tests an maximal downward angle of the beam of 52.5° with respect to the horizon was achieved. During the cold tests the horizontal movement speed was set to $2.6 \text{ }^\circ/\text{s}$ and the vertical speed to $1 \text{ }^\circ/\text{s}$, horizontally an angle of 81° was covered and 22° vertically, respectively. Tests in warm of the fully expanded setup showed vibrations if to high speed was chosen. The vibrations

³Bosch GPL3.

19 are expected to be damped with a more stable fixation on the detector and the immersion of the
20 setup in liquid argon.

1 Modulation of the beam energy with respect to the vertical alignment of the cold mirror was
2 found to be crucial for obtaining sufficient ionization in the detector. Investigations showed that
3 the reflectivity of the selected dielectric mirrors, which were optimized for 45° in air, are very
4 sensitive to the angle of incidence in liquid argon. Therefore during a calibration run, the beam
5 energy has to be controlled. The emitted UV-laser beam has a diameter of 6 mm and will spatially
6 diverge during propagation. A beam with this diameter will produce an ionization signal larger
7 than the wire spacing, which will limit the capabilities of the full system. With the aperture a small
8 as possible diameter of the laser was selected to enter the detector. Measurements of the diameter
9 were performed with thermal paper (used for thermal printing) on which the selected beam spot
10 burns in. The minimal achieved diameter was 1 mm.

11 **9. Conclusion**

12 The MicroBooNE detector is the culmination of several years of development and construction. It
13 is the largest LArTPC ever constructed in the U.S., and represents a major technological advance
14 that future experiments will build upon.

15 **Acknowledgments**

16 We gratefully acknowledge support from the National Science Foundation, the United States De-
17 partment of Energy, and the EU funding gods.

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