### A reminder of steady-state Kalman filtering



- In Lesson 2.3.5, you learned that under common conditions, the Kalman-filter covariances converge to steady-state solutions as does the Kalman gain.
  - $\Box$  If we are willing to accept slightly suboptimal state estimates, we can implement a steady-state Kalman filter that uses constant  $L_{ss}$  instead of an optimal Kalman filter that uses time-varying  $L_k$ .
- The benefit of doing so is a huge reduction in computation.
  - $\Box$   $L_{ss}$  is computed once only, and the state-estimate-update equations are streamlined.
- To compute  $L_{ss}$ , we needed to solve a discrete algebraic Riccati equation (DARE), which is somewhat complicated, but is facilitated using dlqe in Octave.

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2.4.5: Steady-state  $\alpha$ - $\beta$ - $\gamma$  target-tracking filters

### A steady-state Kalman filter for target tracking



- For a steady-state solution to  $L_{ss}$  to exist, we must have constant A, B, C, D, and noise-covariance matrices.
- When these conditions are met, there is still no general closed-form solution to computing  $L_{ss}$ .
  - $\Box$  However, we can find a closed-form solution for the NCV model, producing an optimized  $\alpha-\beta$  filter.
  - $\Box$  We can also find a closed-form solution for the NCA model (to be introduced), producing an optimized  $\alpha-\beta-\gamma$  filter.
  - □ Since these models are commonly used for target-tracking applications, we consider steady-state forms of Kalman filters based on them in this lesson.
- In this lesson, you will learn how to derive the  $\alpha-\beta$  filter and will also learn the result for the  $\alpha-\beta-\gamma$  filter.

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# The generic $\alpha - \beta$ filter



■ An general  $\alpha$ - $\beta$  filter assumes an NCV model of the form:

$$p_{k+1} = p_k + (\Delta t)v_k + \text{noise}$$
  
 $v_{k+1} = v_k + \text{noise}.$ 

■ The update equations for the general  $\alpha$ - $\beta$  filter are of the form:

$$\hat{p}_{k+1} = \hat{p}_k + (\Delta t)\hat{v}_k + (\alpha)(z_k - \hat{p}_k) \hat{v}_{k+1} = \hat{v}_k + (\beta/\Delta t)(z_k - \hat{p}_k).$$

- There exists a volume of literature that proposes different ways to select  $\alpha$  and  $\beta$ .
- In this lesson, you will learn how to optimize the values of  $\alpha$  and  $\beta$  for this method using a closed-form steady-state Kalman filter.
- While the resulting KF is suboptimal because it is not using time-varying gains, the values of  $\alpha$  and  $\beta$  are optimal since the  $\alpha$ - $\beta$  filter assumes constant gains.

### Review of the NCV model in state-space form



■ To show how to derive the optimized  $\alpha$  and  $\beta$  values, recall that we can write the NCV model as:

$$x_{k+1} = \begin{bmatrix} 1 & \Delta t \\ 0 & 1 \end{bmatrix} x_k + \begin{bmatrix} (\Delta t)^2 / 2 \\ \Delta t \end{bmatrix} w_k'$$
  
$$z_k = \begin{bmatrix} 1 & 0 \end{bmatrix} x_k + v_k,$$

where  $w_k' \sim \mathcal{N}(0, \sigma_{\widetilde{w}}^2)$  and  $v_k \sim \mathcal{N}(0, \Sigma_{\widetilde{v}})$ , and where  $w_k'$  is a scalar.

■ We can rewrite the state equation as:

$$x_{k+1} = \begin{bmatrix} 1 & \Delta t \\ 0 & 1 \end{bmatrix} x_k + w_k,$$

where  $w_k \sim \mathcal{N}(0, \Sigma_{\widetilde{w}})$  and:

$$\Sigma_{\widetilde{w}} = \begin{bmatrix} (\Delta t)^2/2 \\ \Delta t \end{bmatrix} \mathbb{E} \left[ w_k'(w_k')^T \right] \left[ (\Delta t)^2/2 \ \Delta t \right] = \begin{bmatrix} (\Delta t)^4/4 & (\Delta t)^3/2 \\ (\Delta t)^3/2 & (\Delta t)^2 \end{bmatrix} \sigma_{\widetilde{w}}^2.$$

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### Solving for $L_{ss}$



■ We are now ready to solve for  $L_{ss}$ : we denote the components of the steady-state solution as:

$$L_{ss} = \left[ \begin{array}{c} L_1 \\ L_2 \end{array} \right] = \left[ \begin{array}{c} \alpha \\ \beta/\Delta t \end{array} \right], \quad \Sigma_{\tilde{x},ss}^- = \left[ \begin{array}{cc} \Sigma_{11}^- & \Sigma_{12}^- \\ \Sigma_{12}^- & \Sigma_{22}^- \end{array} \right], \quad \Sigma_{\tilde{x},ss}^+ = \left[ \begin{array}{cc} \Sigma_{11}^+ & \Sigma_{12}^+ \\ \Sigma_{12}^+ & \Sigma_{22}^+ \end{array} \right].$$

■ Using this notation, we can write the Kalman gain as:

$$\begin{split} L_{ss} &= \Sigma_{\tilde{x}, ss}^{-} C^{T} \left( C \Sigma_{\tilde{x}, ss}^{-} C^{T} + \Sigma_{\tilde{v}} \right)^{-1} \\ &= \begin{bmatrix} \Sigma_{11}^{-} & \Sigma_{12}^{-} \\ \Sigma_{12}^{-} & \Sigma_{22}^{-} \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \left( \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} \Sigma_{11}^{-} & \Sigma_{12}^{-} \\ \Sigma_{12}^{-} & \Sigma_{22}^{-} \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} + \Sigma_{\tilde{v}} \right)^{-1} \\ &= \frac{1}{\Sigma_{11}^{-} + \Sigma_{\tilde{v}}} \begin{bmatrix} \Sigma_{11}^{-} \\ \Sigma_{12}^{-} \end{bmatrix}. \end{split}$$

■ So, if we can solve for  $\Sigma_{\tilde{x}_{ss}}^-$ , we can find  $L_{ss}$  (and therefore  $\alpha$  and  $\beta$ ).

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# Solution #1 for $\Sigma_{ss}^+$



To find the steady-state covariance matrices, we first write (using the estimation-error covariance-matrix measurementupdate equation):

$$\begin{split} \Sigma_{\tilde{x},ss}^{+} &= (I - L_{ss}C)\Sigma_{\tilde{x},ss}^{-} \\ &= \left( \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} - \begin{bmatrix} L_{1} \\ L_{2} \end{bmatrix} \begin{bmatrix} 1 & 0 \end{bmatrix} \right) \begin{bmatrix} \Sigma_{11}^{-} & \Sigma_{12}^{-} \\ \Sigma_{12}^{-} & \Sigma_{22}^{-} \end{bmatrix} \\ &= \begin{bmatrix} (1 - L_{1})\Sigma_{11}^{-} & (1 - L_{1})\Sigma_{12}^{-} \\ (1 - L_{1})\Sigma_{12}^{-} & \Sigma_{22}^{-} - L_{2}\Sigma_{12}^{-} \end{bmatrix}. \end{split}$$

■ We call this "solution 1" for  $\Sigma_{\tilde{x}.ss}^+$ , which we will use in just a minute.

# Solution #2 for $\Sigma_{ss}^+$



■ We continue by finding another relationship for  $\Sigma_{\tilde{x}.xs}^+$ . Recall the prediction-error covariance-matrix time-update equation:

$$\Sigma_{\tilde{x},ss}^{-} = A \Sigma_{\tilde{x},ss}^{+} A^{T} + \Sigma_{\tilde{w}}.$$

■ We rewrite this equation in terms of  $\Sigma_{\tilde{x}}^+$  s:

$$\begin{split} \Sigma_{\tilde{x},ss}^{+} &= A^{-1} (\Sigma_{\tilde{x},ss}^{-} - \Sigma_{\tilde{w}}) A^{-T} \\ &= \begin{bmatrix} 1 & -\Delta t \\ 0 & 1 \end{bmatrix} \left( \begin{bmatrix} \Sigma_{11}^{-} & \Sigma_{12}^{-} \\ \Sigma_{12}^{-} & \Sigma_{22}^{-} \end{bmatrix} - \begin{bmatrix} (\Delta t)^{4}/4 & (\Delta t)^{3}/2 \\ (\Delta t)^{3}/2 & (\Delta t)^{2} \end{bmatrix} \sigma_{\tilde{w}}^{2} \right) \begin{bmatrix} 1 & 0 \\ -\Delta t & 1 \end{bmatrix}. \end{split}$$

- We can write this out and find all the terms for  $\Sigma_{\tilde{x},s}^+$
- We call this "solution 2" for  $\Sigma_{\tilde{r},ss}^+$ .

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### Final $\alpha - \beta$ filter solution



■ When we equate the terms of "solution 1" and "solution 2" for  $\Sigma_{\tilde{x},s}^+$  and do some algebra, we arrive at the solution:

$$L_{ss} = \begin{bmatrix} -\frac{1}{8} \left( \lambda^2 + 8\lambda - (\lambda + 4)\sqrt{\lambda^2 + 8\lambda} \right) \\ \frac{1}{4\Delta t} \left( \lambda^2 + 4\lambda - \lambda\sqrt{\lambda^2 + 8\lambda} \right) \end{bmatrix}$$
$$\Sigma_{\tilde{x},ss}^+ = \begin{bmatrix} L_1 \Sigma_{\tilde{v}} & L_2 \Sigma_{\tilde{v}} \\ L_2 \Sigma_{\tilde{v}} & \frac{L_2 \Sigma_{\tilde{v}}}{1 - L_1} \left( \frac{L_1}{\Delta t} - \frac{L_2}{2} \right) \end{bmatrix},$$

where:

$$\lambda = \frac{\sigma_{\widetilde{w}}(\Delta t)^2}{\sqrt{\Sigma_{\widetilde{v}}}}.$$

■ The term  $\lambda$  is the "target-maneuvering index" or "target-tracking index" and is proportional to the ratio of the motion uncertainty to the measurement uncertainty.

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# Compare Octave dlqe with $\alpha-\beta$ solution



```
% A generic sample period
sigmaW = 0.1; % Standard deviation of (scalar) process noise
SigmaV = 0.2; % Covariance of measurement noise
% Compute the DLQE solution
Ad = [1 dT; 0 1]; Cd = [1 0]; % Specify NCV matrices and compute process-noise
SigmaW = [(dT)^4/4 (dT)^3/2; (dT)^3/2 (dT)^2] * sigmaW^2; % vector covariance
[Lss1,SigmaMss1,SigmaPss1] = dlqe(Ad,eye(2),Cd,SigmaW,SigmaV);
% Compute the alpha-beta solution
lambda = sigmaW * (dT)^2 / sqrt(SigmaV);
Lss2 = [-(lambda^2+8*lambda-(lambda+4)*sqrt(lambda^2+8*lambda))/8; ...
       (lambda^2+4*lambda-lambda*sqrt(lambda^2+8*lambda))/(4*dT)];
SigmaMss2 = [Lss2(1)*SigmaV/(1-Lss2(1)), Lss2(2)*SigmaV/(1-Lss2(1)); Lss2(2)* ...
 SigmaV/(1-Lss2(1)), (Lss2(1)/dT + Lss2(2)/2)*Lss2(2)*SigmaV/(1-Lss2(1))];
SigmaPss2 = [Lss2(1)*SigmaV, Lss2(2)*SigmaV; Lss2(2)*SigmaV, ...
 Lss2(2)*SigmaV*(Lss2(1)/dT -Lss2(2)/2)/(1-Lss2(1))];
                      % They agree!
Lss1, Lss2
SigmaMss1, SigmaMss2 % These also agree!
SigmaPss1, SigmaPss2 % These agree as well!
```

### The NCA model used by the $\alpha - \beta - \gamma$ filter



■ The NCA model is also popular for tracking. For scalar  $w'_{k}$ :

$$x_{k+1} = \begin{bmatrix} 1 & \Delta t & (\Delta t)^2 / 2 \\ 0 & 1 & \Delta t \\ 0 & 0 & 1 \end{bmatrix} x_k + \begin{bmatrix} (\Delta t)^2 / 2 \\ \Delta t \\ 1 \end{bmatrix} w_k'$$
$$z_k = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} x_k + v_k.$$

■ We can rewrite the state equation for vector  $w_k \sim \mathcal{N}(0, \Sigma_{\tilde{w}})$  as:

$$x_{k+1} = \begin{bmatrix} 1 & \Delta t & (\Delta t)^{2}/2 \\ 0 & 1 & \Delta t \\ 0 & 0 & 1 \end{bmatrix} x_{k} + w_{k}, \text{ where}$$

$$\Sigma_{\widetilde{w}} = \begin{bmatrix} (\Delta t)^{4}/4 & (\Delta t)^{3}/2 & (\Delta t)^{2}/2 \\ (\Delta t)^{3}/2 & (\Delta t)^{2} & \Delta t \\ (\Delta t)^{2}/2 & \Delta t & 1 \end{bmatrix} \sigma_{\widetilde{w}}^{2}.$$

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### The $\alpha - \beta - \gamma$ solution framework



■ We denote the components of the steady-state solution as:

$$L_{ss} = \begin{bmatrix} \alpha \\ \beta/\Delta t \\ \gamma/(2(\Delta t)^2) \end{bmatrix}, \quad \Sigma_{\tilde{x},ss}^+ = \begin{bmatrix} \Sigma_{11}^+ & \Sigma_{12}^+ & \Sigma_{13}^+ \\ \Sigma_{12}^+ & \Sigma_{22}^+ & \Sigma_{23}^+ \\ \Sigma_{13}^+ & \Sigma_{23}^+ & \Sigma_{33}^+ \end{bmatrix}.$$

■ The solution is computed via some auxiliary variables ( $\lambda$  is the same as earlier):

$$b = \frac{\lambda}{2} - 3$$

$$c = \frac{\lambda}{2} + 3$$

$$p = c - \frac{b^2}{3}$$

$$q = \frac{2b^3}{27} - \frac{bc}{3} - 1$$

$$z = \left[ \frac{-q + \sqrt{q^2 + 4p^3/27}}{2} \right]^{1/3}$$

$$s = z - \frac{p}{3z} - \frac{b}{3}.$$

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# The $\alpha - \beta - \gamma$ filter solution



- The details for using these variables to find a solution are straightforward but tedious; I omit them here.
- For reference, the final answer is:

$$\alpha = 1 - s^{2}, \qquad \Sigma_{11}^{+} = \alpha \Sigma_{\tilde{v}}, \qquad \qquad \Sigma_{22}^{+} = \frac{8\alpha\beta + \gamma(\beta - 2\alpha - 4)}{8(\Delta t)^{2}(1 - \alpha)} \Sigma_{\tilde{v}}$$

$$\beta = 2(1 - s)^{2}, \qquad \Sigma_{12}^{+} = \beta \Sigma_{\tilde{v}}/\Delta t, \qquad \qquad \Sigma_{23}^{+} = \frac{\beta(2\beta - \gamma)}{4(\Delta t)^{3}(1 - \alpha)} \Sigma_{\tilde{v}}$$

$$\gamma = 2\lambda s, \qquad \qquad \Sigma_{13}^{+} = \gamma \Sigma_{\tilde{v}}/(2(\Delta t)^{2}), \qquad \Sigma_{33}^{+} = \frac{\gamma(2\beta - \gamma)}{4(\Delta t)^{4}(1 - \alpha)} \Sigma_{\tilde{v}}.$$

#### Summary



- Target-tracking applications often use simplified dynamic models such as NCV and NCA.
- Closed-form solutions exist for steady-state Kalman filters for these two models.
- They are optimized versions of  $\alpha$ - $\beta$  and  $\alpha$ - $\beta$ - $\gamma$  filters, respectively.
- In this lesson, you learned how the  $\alpha$ - $\beta$  filter is derived, and compared its solution to that computed by dlqe for the same model (they matched!)
- You also learned the solution to the  $\alpha$ - $\beta$ - $\gamma$  filter.
- The big advantage of knowing these solutions is that they can be computed using standard algebra, without requiring an algebraic-Riccati-equation solver.

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