

# ElectroMagnetic radiation and Maxwell's equations

$$V = C = \frac{1}{\sqrt{\mu_0 \epsilon_0}} = 3.00 \times 10^8 \text{ m/s}$$

• this is the speed of light

In simple words this equation states that light travels at the speed it does because the electric and magnetic nature of empty space fixes it at  $C = 1/\sqrt{\mu_0 \epsilon_0}$ .

- light is an electromagnetic wave
- electromagnetic waves can travel in absence of vacuum

- These electromagnetic waves form a broad spectrum, as seen in the table below:

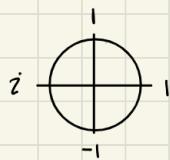


Region:	Wavelengths	Frequency
Radio	10 km to 10 cm	$10^5$ to $10^8$ Hz
Microwave	10 cm to 1 mm	$10^8$ to $10^{11}$ Hz
Infrared	1 mm to 1 μm	$10^{11}$ to $10^{14}$ Hz
Visible	350 nm to 700 nm	$4.5 \times 10^{14}$ to $7.5 \times 10^{14}$ Hz
Ultraviolet	100 nm to 10 nm	$10^{15}$ to $10^{16}$ Hz
X-rays	10 nm to 10 pm	$10^{16}$ to $10^{19}$ Hz
Gamma rays	10 pm to 1 pm	$10^{19}$ to $10^{20}$ Hz

Review and explanation on complex numbers

Since EM-waves are represented by a sine wave, it is often easier to represent them using complex exponential

$z = \sqrt{-1} \rightarrow z^n$  forms a cyclic group of order 4  
this cyclic group can only produce  $[z, -1, -z, 1]$  and the cycle repeats starting at  $z$



• and it just goes around it over and over

Mathematically,  $i^n$  forms a cyclic group of order 4 so:  $i^n = i^{n \bmod 4}$

Complex number:

$$z = a + bi$$

real part      imaginary part

- a complex number consists of a imaginary number and a real number

the complex conjugate of  $z$  denoted  $\bar{z}$ , is defined such that:

$$z = a + bi \longleftrightarrow \bar{z} = a - bi$$

• all we are doing here is Flipping the sign around

Notice the product:

$$\begin{aligned} z\bar{z} &= (a+bi)(a-bi) && \text{Apply Foil Method} \\ &\rightarrow z\bar{z} = a^2 + b^2 \end{aligned}$$

You get a purely real solution

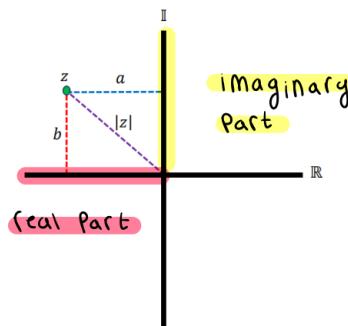
We define the absolute value of a complex number such:

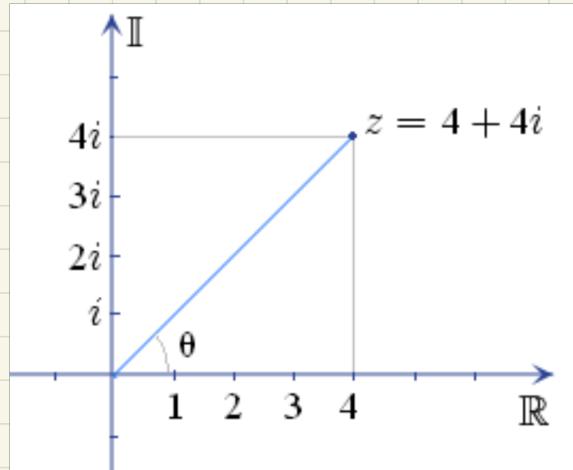
$$|z| = \sqrt{z\bar{z}} = \sqrt{a^2 + b^2}$$

• Notice this looks like the Pythagorean theorem

### The Complex Plane

- We can depict complex numbers graphically!
- Notice the absolute value of  $z$  is then distance from  $z$  to the origin.





Complex numbers in Polar Form:

- this gives us a equivalent way of writing a generic complex number

$$z = Re^{i\theta}$$

$$e^{i\theta} = \cos\theta + i\sin\theta \rightarrow \text{Euler's Formula}$$

$$\overline{Re^{i\theta}} = \overline{\underline{R}\underline{\cos\theta} + \underline{i}\underline{R}\sin\theta}$$

$$\begin{matrix} \downarrow \\ z \end{matrix} \quad \begin{matrix} \downarrow \\ a \end{matrix} \quad \begin{matrix} \downarrow \\ b \end{matrix}$$

so now  $\bar{z}$  will be:

$$\begin{aligned} Re^{-i\theta} &= R(\cos(-\theta) + i\sin(-\theta)) \\ (\text{rewrite}) \quad \rightarrow \bar{z} &= Re^{-i\theta} = R(\cos(\theta) - i\sin(\theta)) \end{aligned}$$

hence:

$$|z| = \sqrt{z\bar{z}} = \sqrt{Re^{i\theta}Re^{-i\theta}} = \sqrt{R^2} = R$$

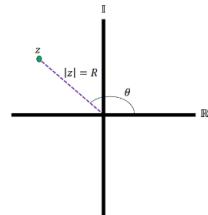
$$\begin{aligned} e^{ix} &= 1 + ix + \frac{x^2 i^2}{2!} + \frac{x^3 i^3}{3!} + \frac{x^4 i^4}{4!} + \frac{x^5 i^5}{5!} + \frac{x^6 i^6}{6!} + \frac{x^7 i^7}{7!} + \dots \\ &= 1 + ix - \frac{x^2}{2!} - i\frac{x^3}{3!} + \frac{x^4}{4!} + i\frac{x^5}{5!} - \frac{x^6}{6!} - i\frac{x^7}{7!} + \dots \\ &= 1 - \frac{x^2}{2!} + \frac{x^4}{4!} - \frac{x^6}{6!} + \dots + i\left(x - \frac{x^3}{3!} + \frac{x^5}{5!} - \frac{x^7}{7!} + \dots\right) \end{aligned}$$

$$\begin{aligned} 1 - \frac{x^2}{2!} + \frac{x^4}{4!} - \frac{x^6}{6!} + \dots &= \sum_{n=0}^{\infty} \frac{(-1)^n x^{2n}}{(2n)!} = \cos x \\ x - \frac{x^3}{3!} + \frac{x^5}{5!} - \frac{x^7}{7!} + \dots &= \sum_{n=0}^{\infty} \frac{(-1)^n x^{2n+1}}{(2n+1)!} = \sin x \end{aligned}$$

$$e^{ix} = \cos x + i \sin x$$

### Polar Form

- Notice polar form is analogous to using polar coordinates when graphing  $z$  in the complex plane.



## • EM-Plane Waves

For convenience, it is common to pick coordinates such that the wave is traveling along the positive z axis

$$V = \lambda f$$

$$V = \frac{\omega}{k}$$

$$\vec{A} = \bar{R} (\vec{x} \cdot \hat{z} - vt) = \bar{R} e^{i(kz-wt)}$$

$$k = \frac{2\pi}{\lambda}$$

$$\omega = \frac{2\pi}{T}$$

- represents a traveling wave

R = the amplitude of the wave

$e^{i(kz-wt)}$  = describes the oscillation of the wave in both space and time

- k = wave number
- z = position along the waves path (usually the direction wave is traveling)
- w = angular frequency
- t = time

- We choose to write our wave equation solution in this way for two reasons:

**Physics Reason:** In general, the velocity of a wave is a function of the frequency (a phenomenon called dispersion).

**Math Reason:** Any reasonably well-behaved function can be built by summing together complex exponentials (a process called Fourier decomposition).

- Wave equation for Electric Field

$$\vec{E}(\vec{x}, t) = \vec{E}_0 e^{i(kz-wt)}$$

- $\vec{E}_0$  = the amplitude of the electric field

- k = wave number

- z = position along the waves path (usually the direction wave is traveling)

- w = angular frequency

- t = time

- Wave equation For magnetic Field

$$\vec{H}(\vec{x}, t) = \vec{H}_0 e^{i(kz - \omega t)}$$

• in Many EM-Waves treatment especially in Free Space  $B$  and  $H$  are essentially interchangeable

- $H_0$  = amplitude
- $k$  = wave number
- $z$  = Position along the waves path (usually the direction wave is traveling)
- $\omega$  = angular frequency
- $t$  = time

hence the Source-Free Maxwell equations in linear media becomes

$$0 = i k \hat{z} \cdot \vec{E}$$

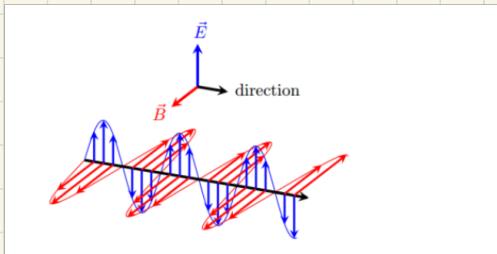
$$0 = i k \hat{z} \cdot \vec{H}$$

EM waves are transverse  
(meaning  $\vec{E}$  and  $\vec{H}$  have No component along propagation  $\rightarrow \vec{E}$  and  $\vec{H} \perp \hat{z}$ )

$$\vec{H} = \frac{1}{\mu V} \hat{z} \times \vec{E}$$

$$\vec{E} = -\frac{1}{\epsilon V} \hat{z} \times \vec{H}$$

$\vec{E} \perp \vec{H}$



Using Euler's relation our Plane wave solution can be thought as real valued Sinusoid

$$\vec{E} = \vec{E}_0 \cos(kz - \omega t), \quad \vec{H} = \vec{H}_0 \cos(kz - \omega t)$$

## Conclusions

due to  $\vec{E}$  being  $\perp \vec{H}$ , we can uniquely specify as an EM wave by specifying how  $\vec{E}$  behaves in 2 linearly independent directions

these independent directions are termed: Polarization

an EM wave that oscillates in only one such direction (and not the other) is said to be: Polarized

## Polarizers

Material that only transmits one polarization direction:

- Polymer chains absorb polarization aligned with their direction (allowing polarization  $\perp$  with their direction to pass)  $I_T = I_I / 2$

Passing light through 2 polarizers oriented  $90^\circ$  blocks all light

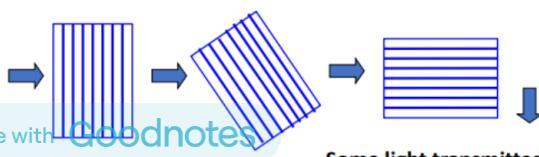
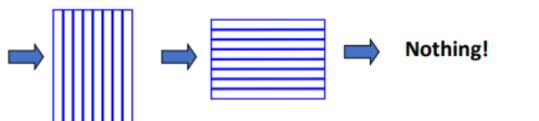
Rotated polarizers transmit varying intensity  $I_T = I_I \cos^2 \theta$

The equation is for transmitted light intensity after a polarizer

$$I_T = I_I \cos^2 \theta$$

↓  
intensity of transmitted light  
↓  
Intensity of incident light

• So:



## Impedance (Z)

and units of  $\Omega$  (ohms)

Impedance is a measure of how much a circuit resists the

flowing of alternating current (AC)

due to Euler's relation, we can model circuits with sinusoidal varying emfs using complex exponentials

$$I = |I| e^{i(\omega t + \phi_I)}$$

$$\Delta V = |\Delta V| e^{i(\omega t + \phi_V)}$$

$$Z = \frac{|\Delta V| e^{i(\phi_V - \phi_I)}}{|I|}$$

$Z_R = R$  • resistor

$Z_C = \frac{1}{i\omega C}$  • capacitor

$Z_L = i\omega L$  • inductor

## Equivalent impedance

### Parallel

$$\frac{1}{Z_{eq}} = \frac{1}{Z_1} + \frac{1}{Z_2} + \frac{1}{Z_3} \dots$$

### Series

$$Z_{eq} = Z_1 + Z_2 + \dots$$

## Inductance (L)

(L is in henries H)

If 2 current loops are placed next to each other, putting a current through one will induce a current in the other by Faraday's law

$$\mathcal{E} = -\frac{dI}{dt} \quad I = MI$$

$$\mathcal{E} = -M \frac{dI}{dt} \quad \begin{array}{l} \text{- induced emf} \\ \text{due to mutual} \\ \text{inductance} \end{array}$$

$$\mathcal{E} = -L \cdot \frac{dI}{dt} \quad \begin{array}{l} \text{- induced emf due} \\ \text{to inductors self} \\ \text{inductance} \end{array}$$

$$\Delta V_{total} = L_1 \cdot \frac{dI}{dt} + M_2 \frac{dI}{dt}$$

## Equivalent inductance

### Parallel

$$\frac{1}{L_{eq}} = \frac{L_1 + L_2 - 2M}{L_1 L_2 - M}$$

### Series

$$L_{eq} = L_1 + L_2 + 2M$$

## Important notes

$$M_1 = M_2 = M \quad (\text{For simplicity})$$

1. A *leaky* capacitor is one in which the charge can flow through the dielectric medium separating the two capacitor plates. Such a capacitor can be modeled by an ideal capacitor and a resistor placed in parallel with one another. Consider a  $50 \mu\text{F}$  leaky capacitor with an effective resistance of  $300000\Omega$ .

- If this leaky capacitor is being driven by an AC voltage source with a frequency  $f = 60 \text{ Hz}$  (i.e., the same as a typical US wall outlet), then what is the absolute value of the total impedance,  $Z$ ?
- If this leaky capacitor is brought to Europe (where the AC voltage sources operate at  $50 \text{ Hz}$ ), would the absolute value of the total impedance increase or decrease? Explain briefly.

a) You have a capacitor and a resistor in parallel when you have a situation like this you use this equation:

$$\frac{1}{Z} = \frac{1}{R} + i\omega C$$

If the resistor and capacitor are in series use this equation

$$Z = R - \frac{i}{\omega C}$$

$$\frac{1}{Z} = \frac{1}{R} + i\omega C \rightarrow \frac{1}{Z} = \frac{1 + i\omega CR}{R} \rightarrow Z(1 + i\omega CR) = R$$

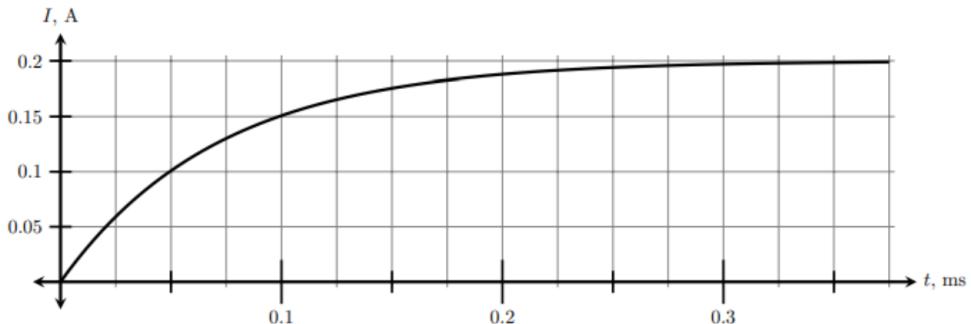
$$Z = \frac{R}{(1 + i\omega CR)} \cdot \frac{(1 - i\omega CR)}{(1 - i\omega CR)}$$

Never keep  $i$  in the denominator  
So you will need to rationalize it

$$Z = \frac{R}{1 + \omega^2 R^2 C^2} - i \frac{\omega R^2 C}{1 + \omega^2 R^2 C^2} \rightarrow Z = \sqrt{\left(\frac{R}{1 + \omega^2 R^2 C^2}\right)^2 + \left(\frac{\omega R^2 C}{1 + \omega^2 R^2 C^2}\right)^2}$$

\* square both side to  
get rid of  $i$  ( $i^2 = -1$ )

3. Two 0.25 mH inductors are wired together in series and then connected (in series) to a real 2.4 V battery with an internal resistance of  $1.8\Omega$ . The current as a function of time through the inductors is then plotted, as shown in the graph below.



- (a) (*Draws on Class #21*) Assuming inductors have identical internal resistances, what is the internal resistance,  $R$ , of one of the inductors?
- (b) What is the mutual inductance between the two 0.25 mH inductors?

a)

$$\begin{aligned} \mathcal{E} &= (1R + 1R + 1.8I) \\ \mathcal{E} &= 2IR + 1.8I \end{aligned}$$

↗

$$R = \frac{2.4 - 1.8(0.2)}{2(0.2)} = \boxed{5.1\Omega}$$

$$R = \frac{\mathcal{E} - 1.8I}{2I}$$

if they are all in Series you can apply a Kirchoff's rule around the circuit where you spin around the circuit and get the equation

b)

For series:

$$L = L_1 + L_2 + 2m$$

Now you isolate for  $m$  (mutual inductance) (remember  $L$  = self inductance and  $m$  = mutual inductance)

$$\gamma = \frac{t \cdot \alpha}{LN(\alpha)} = \frac{0.0005}{LN(2)} = 7.21 \times 10^{-4}$$

$$\gamma = \frac{L}{R} \rightarrow \gamma R = L \rightarrow (7.21 \times 10^{-4})(5.1) = 3.67 \times 10^{-3} \text{ H}$$
$$= 3.67 \text{ mH}$$

$$M = \frac{L - L_1 - L_2}{2} = \frac{3.67 - 0.25 - 0.25}{2} = \boxed{1.58 \text{ mH}}$$