Improved 1-D FDTD Modeling of Parallel and Series RLC Loads in a Lossless Transmission Line

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This paper presents improved finite-difference time-domain (FDTD) update equations to model a parallel RLC load in series and a series RLC load in parallel with a simple, lossless transmission line. A one-dimensional (1-D) FDTD model of a simple, lossless transmission line was developed [1], and extended to model discrete circuit elements (e.g., capacitors, inductors, and resistors) [2]. Later, FDTD update equations were developed for parallel or series RLC loads placed in parallel or series with a transmission line [3], and for use with voltage sources and/or terminations [4].

Fig. 1a shows the circuit representation of an incremental section of a simple, lossless transmission line with a parallel RLC load in series. Applying Kirchoff's current law (KCL) to the top left and right nodes yields

$$i(z,t) - \frac{v_{PS}}{R} - i_{LPS} - C \frac{\partial v_{PS}}{\partial t} = 0$$
 (1)

where

$$i_{LPS}(z,t) = \frac{1}{L} \int_{t_0}^{t} v_{PS}(z,t) \, \partial t + i_{LPS}(z,t_0)$$
 (2)

and

$$i(z,t) - i(z + \Delta z, t) - c\Delta z \frac{\partial v(z + \Delta z, t)}{\partial t} = 0$$
(3)

Applying Kirchoff's voltage law (KVL) clockwise around the outside loop yields

$$v(z + \Delta z, t) - v(z, t) + v_{PS} + l\Delta z \frac{\partial i(z, t)}{\partial t} = 0$$
(4)

In these equations, c and l are the capacitance and inductance per-unit-length of the transmission line respectively.

For the FDTD method, the current and voltage are spatially discretized as shown in Fig. 2 for the incremental transmission line sections. Further, they are interleaved in time with voltage nodes placed at times $t = n\Delta t$, and current nodes placed at times $t = (n + 0.5)\Delta t$.

For clarity, the usual shorthand FDTD notation where
$$V^n(k) = v(z = k\Delta z, t = n\Delta t)$$
 and $I^{n+0.5}(k+0.5) = i(z = (k+0.5)\Delta z, t = (n+0.5)\Delta t)$ is used.

For the parallel RLC load in series, discretizing (1) at $t = (n - 0.5)\Delta t$, using the usual second-order accurate central-difference approximation to the time derivative, yields

$$I^{n-0.5}(k+0.5) - \frac{V_{PS}^{n-0.5}(k+0.5)}{R} - I_{LPS}^{n-0.5}(k+0.5) - C \left[\frac{V_{PS}^{n}(k+0.5) - V_{PS}^{n-1}(k+0.5)}{\Delta t} \right] = 0.$$

Letting $V_{PS}^{n-0.5}(k+0.5) \approx \left[V_{PS}^{n}(k+0.5) + V_{PS}^{n-1}(k+0.5)\right]/2$ and solving for the voltage v_{PS} across the parallel RLC load in series, yields the update equation

$$V_{PS}^{n}(k+0.5) = \frac{\left(\frac{C}{\Delta t} - \frac{1}{2R}\right)}{\left(\frac{C}{\Delta t} + \frac{1}{2R}\right)} V_{PS}^{n-1}(k+0.5) + \frac{1}{\left(\frac{C}{\Delta t} + \frac{1}{2R}\right)} \left[I^{n-0.5}(k+0.5) - I_{LPS}^{n-0.5}(k+0.5)\right] (5)$$

where the update equation for the inductor current i_{LPS} , found by discretizing (2), is

$$I_{LPS}^{n-0.5}(k+0.5) = I_{LPS}^{n-1.5}(k+0.5) + \frac{\Delta t}{L} V_{PS}^{n-1}(k+0.5)$$
 (6)

Discretizing (3) at $t = n\Delta t$ and (4) at $t = (n+0.5)\Delta t$, using the usual second-order accurate central-difference approximation to the time derivatives, yields the update equations for the current and voltage

$$I^{n+0.5}(k+0.5) = I^{n-0.5}(k+0.5) - \frac{1}{Z_C} \frac{v_p \Delta t}{\Delta z} \left[V^n(k+1) - V^n(k) \right] - \frac{1}{Z_C} \frac{v_p \Delta t}{\Delta z} V_{PS}^n(k+0.5)$$
 (7)

and

$$V^{n+1}(k+1) = V^{n}(k+1) - Z_{C} \frac{v_{p} \Delta t}{\Delta z} \left[I^{n+0.5}(k+1.5) - I^{n+0.5}(k+0.5) \right]$$
 (8)

where $v_p = 1/\sqrt{lc}$ is the phase velocity and $Z_c = \sqrt{l/c}$ is the characteristic impedance.

Fig. 1b shows the circuit representation of an incremental section of a simple, lossless transmission line with a series RLC load in parallel. Applying KVL counterclockwise around the right-hand loop yields

$$i_{SP}R + L\frac{\partial i_{SP}}{\partial t} + v_{CSP} - v(z + \Delta z, t) = 0$$
(9)

where

$$v_{CSP}(z + \Delta z, t) = \frac{1}{C} \int_{t_0}^{t} i_{SP}(z + \Delta z, t) \, \partial t + v_{CSP}(z + \Delta z, t_0)$$

$$\tag{10}$$

Applying KVL clockwise around the outer loop yields

$$v(z + \Delta z, t) - v(z, t) + l\Delta z \frac{\partial i(z, t)}{\partial t} = 0$$
(11)

Applying KCL to the top right node yields

$$i(z,t) - i(z + \Delta z, t) - c\Delta z \frac{\partial v(z + \Delta z, t)}{\partial t} - i_{SP} = 0$$
(12)

Discretizing (9)-(12) yields the update equations, placed in time order,

$$V_{CSP}^{n}(k+1) = V_{CSP}^{n-1}(k+1) + \frac{\Delta t}{C} I_{SP}^{n-0.5}(k+1)$$
(13)

$$I_{SP}^{n+0.5}(k+1) = \frac{\left(\frac{L}{\Delta t} - \frac{R}{2}\right)}{\left(\frac{L}{\Delta t} + \frac{R}{2}\right)} I_{SP}^{n-0.5}(k+1) + \frac{1}{\left(\frac{L}{\Delta t} + \frac{R}{2}\right)} \left[V^{n}(k+1) - V_{CSP}^{n}(k+1)\right]$$
(14)

$$I^{n+0.5}(k+0.5) = I^{n-0.5}(k+0.5) - \frac{1}{Z_C} \frac{v_p \Delta t}{\Delta z} \left[V^n(k+1) - V^n(k) \right]$$
 (15)

and

$$V^{n+1}(k+1) = V^{n}(k+1) - Z_{C} \frac{v_{p} \Delta t}{\Delta z} \left[I^{n+0.5}(k+1.5) - I^{n+0.5}(k+0.5) - I_{SP}^{n+0.5}(k+1) \right]$$
(16)

To demonstrate the validity of the improved FDTD update equations, the voltages reflected from each type of RLC load are shown in Figs. 3 and 4 along with those based on [3] and analytically derived results. For the examples, a long section of transmission line containing each type of RLC load is driven by a Gaussian voltage pulse $v_G(t) = \exp\left(-0.5\left((t-t_d)/\tau_p\right)^2\right)$ launched by a one-way voltage source injector and terminated at each end by an absorbing boundary condition [1]. For the simulations, $\Delta t/\tau_p = 0.125$ and $S = v_p \Delta t/\Delta z = 0.5$. As shown, there is excellent agreement between the analytic and improved FDTD results, an improvement in accuracy over prior FDTD updates [3], [4]. E.g., the percent error in Fig. 4 is 0.6% for the improved updates versus 2.2% using the previous updates at the first peak, and 0.25% versus 5.05% at the second.

References

- [1] J.G. Maloney, K.L. Shlager, and G.S. Smith, "A Simple FDTD Model for Transient Excitation of Antennas by Transmission Lines," *IEEE Trans. Antennas Propagation*, vol. 42, no. 2, pp. 289-292, Feb. 1994.
- [2] T. P. Montoya and G. S. Smith, "Modeling Transmission Line Circuit Elements in the FDTD Method," *Microwave and Optical Technology Letters*, vol. 21, no. 7, pp.105-114, April 20, 1999.
- [3] T. P. Montoya and W. R. Scott, Jr., "Modeling Parallel and Series RLC Loads in a 1-D FDTD Transmission Line," *USNC/URSI National Radio Science Meeting*, Salt Lake City, UT, p. 30, July 16-21, 2000.
- [4] T. P. Montoya, "Modeling 1-D FDTD Transmission Line Voltage Sources and Terminations with Parallel and Series RLC Loads," 2002 *IEEE APS International Symp.*, San Antonio, TX, vol. 4, pp. 242-245, June 16-21, 2002.

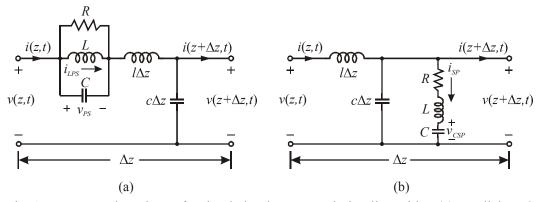


Fig. 1 Incremental sections of a simple lossless transmission line with a (a) parallel RLC load in series and (b) series RLC load in parallel.

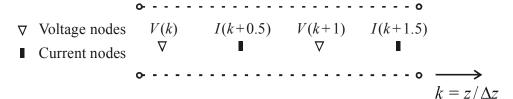


Fig. 2 FDTD model of 1-D transmission line

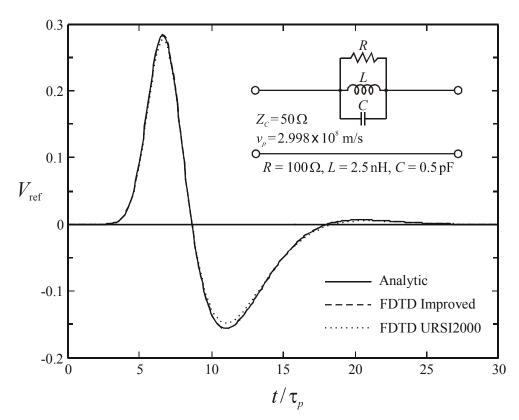


Fig. 3 Voltage reflected from parallel RLC load placed in series with a transmission line.

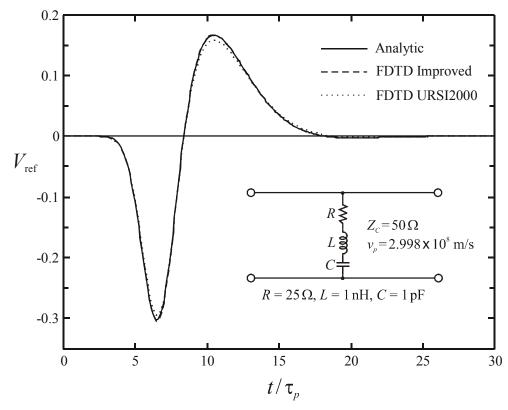


Fig. 4 Voltage reflected from series RLC load placed in parallel with a transmission line.