

EEN1047

CONTROL SYSTEMS ANALYSIS

Dr Brendan Hayes & Dr Mingming Liu

School of Electronic Engineering
Dublin City University

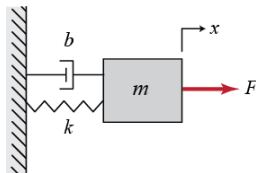
Semester 1 2024-2025

Section 3

Linear Systems Review

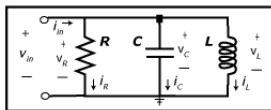
3.1 Physical Systems Behaviour

- Physical systems can be modelled using differential equations
 - Spring-mass-damper system:



$$m \cdot \frac{d^2x(t)}{dt^2} + b \cdot \frac{dx(t)}{dt} + k \cdot x(t) = F(t)$$

- Electrical circuits:



$$\frac{v_L(t)}{R} + C \cdot \frac{dv_L(t)}{dt} + \frac{1}{L} \cdot \int_0^t v_L(t) \cdot dt = i(t)$$

3.2 State Space Models

- State space models can be formed from these original differential equation models.
- **State space** models are a set of input, output and state variables related by first-order differential equations.
- State-space models can be used for models that are not linear, where initial conditions are not zero, and/or there are multiple inputs/outputs.
- For continuous, LTI systems, the **general state space description** is:

$$\begin{aligned}\dot{\mathbf{x}}(t) &= \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t) \\ \mathbf{y}(t) &= \mathbf{C}\mathbf{x}(t) + \mathbf{D}\mathbf{u}(t)\end{aligned}\tag{1}$$

- The **eigenvalues** of the system matrix \mathbf{A} indicate important performance behaviour of the system.
- The characteristic equation can be found using:

$$\lambda(s) = |\lambda\mathcal{I} - \mathbf{A}| = 0$$

3.3 Laplace Transform

- The time response can be difficult to obtain from differential equation models.
- The Laplace transform allows for the substitution of algebraic equations for these differential equations.
- It does this by transforming the time-domain model into a frequency-domain model.
- **Definition:**

$$F(s) = \mathcal{L}\{f(t)\} = \int_0^{\infty} e^{-st} f(t) dt, \quad s = \sigma + j\omega, \quad \sigma, \omega \in \mathbb{R}$$

- Example for spring–mass–damper:

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- Example for spring-mass-damper:

Laplace Transform Example

$$F(s) = m \left(s^2 X(s) - sx(0^-) - \frac{dx(0^-)}{dt} \right) + b \left(sX(s) - x(0^-) \right) + kX(s)$$

Assuming zero initial conditions:

$$F(s) = ms^2 X(s) + bsX(s) + kX(s)$$

3.4 Transfer Function Models

- **Definition:** a **Transfer Function** is defined as the ratio of the Laplace transforms of the output variable to the input variable, assuming *all initial conditions are zero*.
- Can only be defined for linear, time-invariant (**LTI**) systems.

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- Given:

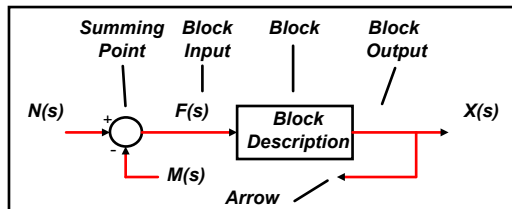
$$G(s) = \frac{Y(s)}{U(s)} = \frac{Num(s)}{Den(s)},$$

the denominator of the TF is called the **characteristic polynomial**, $Den(s)$.

- The **characteristic equation** is defined by letting the characteristic polynomial equal to zero, $Den(s) = 0$.
- The roots of this characteristic equation are called **poles**
 - These poles occur as real numbers or complex conjugate pairs, they are values for s .
- The roots of $Num(s) = 0$ are called the **zeros** of the TF.

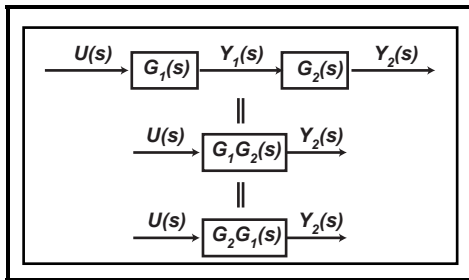
3.5 Block Diagrams

- Large systems may consist of many components, each of which may be represented by a transfer function.
- **Block diagrams** are used as diagrammatical means of representing the relationships between the inputs and outputs of different systems.
- They consist of uni-directional blocks of interest.



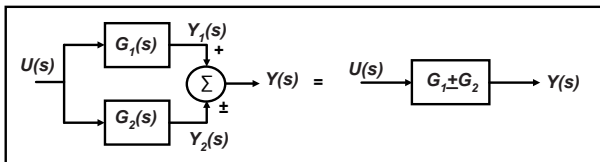
3.6 Blocks in Series

- Complicated block diagrams can be simplified or reduced by combining blocks together using block diagram algebra.
- There are 3 basic operations, depending on the connection between the blocks.
- Blocks that are in **series (cascade connection)** are multiplied together - if individual systems are linear, then multiplication is commutative.

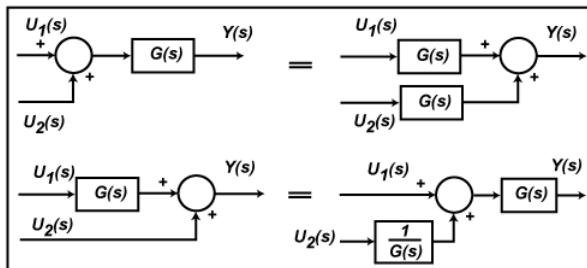


3.7 Blocks and Summers

- Blocks that are in **parallel** are added together - if individual systems are linear, then addition is commutative and distributive

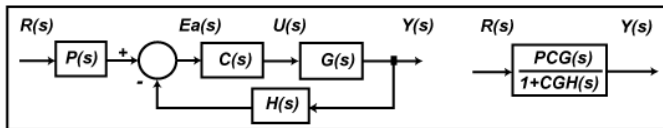


- Summation points can be moved in a diagram



3.8 Block Diagram Feedback

- A **feedback connection** occurs when the output of one block is fed back to the input of an earlier block.
- Loops in block diagrams are common and can be eliminated/simplified.
- Example: **the canonical block diagram**



- This is the most common block diagram for systems and control.
- The simplified representation represents the complete feedback system; it is the **Closed Loop Transfer Function, CLTF**.
- The path from the summer and the output is called the **Forward Path**
- The path from the output to the summer is called the **Feedback Path**

3.9 Block Diagrams CLTF

- From the diagram above:

Derivation

$$Y(s) = C(s)G(s).E_a(s)$$

$$E_a(s) = P(s).R(s) - H(s).Y(s)$$

$$Y(s) = CG(P.R(s) - H.Y(s))$$

$$(1 + CGH).Y(s) = PCG.R(s)$$

$$\text{CLTF: } T(s) = \frac{Y(s)}{R(s)} = P \cdot \frac{CG}{1 + CGH}$$

- $H(s)$ is the feedback transfer function; if $H(s) = 1$ the system is a **unity feedback system**.
- $P(s)$ is a pre-filter and is usually assumed to be unity.
- A unity feedback systems has CLTF:

$$T(s) = \frac{Y(s)}{R(s)} = \frac{CG}{1 + CG}$$

3.10 MATLAB for Polynomials

- Polynomials (e.g. denominator of TF) are entered as matrices in MATLAB.
 - Enter polynomial coefficients into a matrix; coefficients are placed in the matrix starting with the coefficient of the highest power of s & ending with the coefficient of s^0 term.
 - **N.B.** - remember to put the coefficient of the highest power of s (typically 1) into the matrix for the polynomial in MATLAB.
 - There is always 1 term more than the order of the polynomial in the MATLAB matrix. MATLAB assumes that the last entry in the polynomial matrix is the coefficient of s^0 and interprets the other coefficients accordingly.

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Example

polyCoeffs = [6 7 -8 9]

deng = [3 2 5]

numg = [1 0 5]

denh = [1 0 -1 0 0]

$$6x^3 + 7x^2 - 8x - 9$$

$$3s^2 + 2s + 5$$

$$s^2 + 5$$

$$s^4 - s^2$$

3.11 MATLAB for Manipulating Polynomials

- The **conv** function multiplies two polynomials.
- The **roots** function finds the roots of a polynomial.
- The **poly** function finds a polynomial from the roots.

Example

```
polynom1 = [1 2 3];  
polynom2 = [3 4 5];  
polyTotal = conv(polynom1 , polynom2 ); % [3 10 22 22 15]  
  
rootsPolyTotal = roots(polyTotal);  
% -1 +/- j1.41, -0.667 +/- j1.11  
  
polyNew = poly([-1-0.5j -1+0.5j -0.58 -100]);  
% [1 102.58 260.41 241.725 72.5]  
polyNew2 = poly([-1-j*0.5; -1+j*0.5; -0.58; -100]); % same result
```


3.12 MATLAB for LTI Models

- The **tf** function creates a Transfer Function model.
- The **zpk** function creates a Zero-Pole model.
- The **ss** function creates a State-Space model.

Example

```
numPoly = [4 5];  
denPoly = conv([1 3],[1 2]);  
sysGtf = tf(numPoly,denPoly)  
%      4 s + 5  
%      -----  
%      s^2 + 5 s + 6  
sysGzpk = zpk([-1.25],[-3 -2],4);  
%      4 (s+1.25)  
%      -----  
%      (s+3) (s+2)  
aG = [-2 0.866;0 -3];bG = [0;2];cG = [-1.732 2];dG=0;  
sysGss = ss(aG,bG,cG,dG);  
sysGcheck = tf(sysGss);           % can convert between models
```

3.13 MATLAB for Model Data

- The attributes of the LTI objects can be accessed directly.
- The **tfdata**, **ssdata**, **zpkdata** functions also extract this data.
- The poles and zeros the LTI objects can be accessed directly with **pole**, **zero**.

Example

```
numSysG = sysGtf.num{1};  
numSysG = cell2mat(sysGtf.num)  
% [0 4 5]  
gainG = sysGzpk.k; % 4  
sysGcMatrix = sysGss.c % [-1.7320    2.0000]  
  
[numGout,denGout] = tfdata(sysGtf,'v');  
[aGout,bGout,cGout,dGout] = ssdata(sysGss);  
[zGout,pGout,kGout] = zpkdata(sysGzpk,'v');  
% or [zGout,pGout,kGout] = zpkdata(sysGtf,'v');  
  
zeroGtf = zero(sysGtf);  
poleGtf = pole(sysGtf);
```

3.14 MATLAB for Polynomials

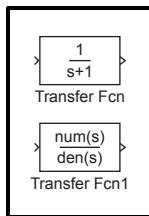
- It is very important **NOT** to use matrix operations with TFs.
- The **series** function multiplies (cascades) TFs, do not use *****.
- The **feedback** functions calculates the overall closed-loop feedback TF in MATLAB; do not use *****, ****.

Example

```
sysCtf = tf([1],[1 2]); %%
sysGtf = tf([4],[1 3]); sysHtf = [4]; %%
sysCGtf = series(sysCtf, sysGtf); %%
%          4
%  -----
%  s^2 + 5s + 6
sysCLtf = feedback(sysCGtf, sysHtf);
%          4
%  -----
%  s^2 + 5s + 22
sysWrong = (sysGtf*sysCtf)/(1+sysGtf*sysCtf*sysHtf)
%          4s^2 + 20s + 24
%  -----
%  s^4 + 10s^3 + 53s^2 + 140s + 132
```

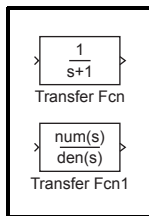
3.15 SIMULINK for LTI Models 1

- **Transfer Fcn block** (in SIMULINK continuous) - takes the numerator and denominator polynomials.

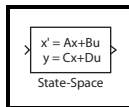


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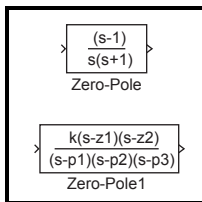
- **State-Space block** (in SIMULINK continuous) - takes the state space model matrices.



- Avoid typing model parameters directly into the SIMULINK blocks.

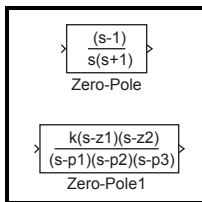
3.16 SIMULINK for LTI Models 2

- **Zero-Pole block** (in SIMULINK continuous) - takes the zeros, poles and gain of the model.

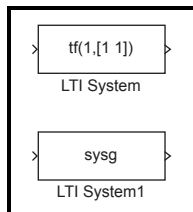


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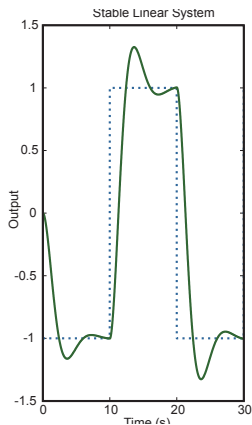
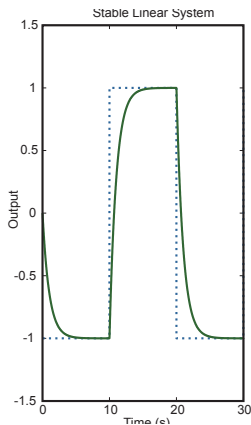


- **LTI block** (in SIMULINK Control Systems Toolbox) - takes any of the system models described above (tf, ss, zpk).



3.17 LTI Stability Responses

- An LTI system is **asymptotically stable** if all states decay to zero. An LTI system is **stable** if the response of the system is bounded for every bounded input (**BIBO**).



3.18 LTI Stability Examples

- A **stable system** has poles (roots of the characteristic equation) exclusively in the left-half s -plane.

Example

$$G_1(s) = \frac{10}{(s+3)(s+5)}$$

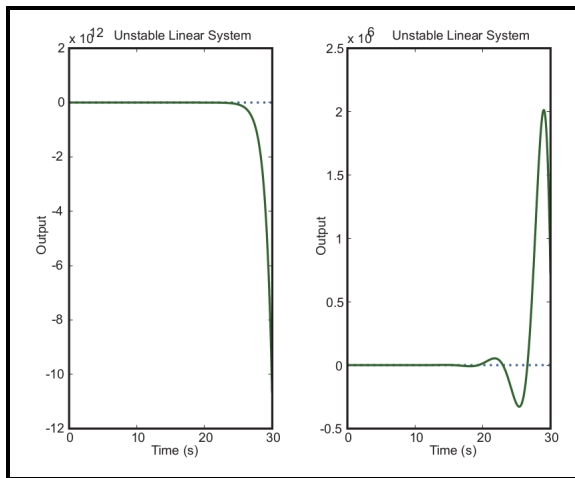
$$G_2(s) = \frac{s+4}{s^3 + 4s^2 + 4s + 3}$$

$$G_3(s) = \frac{s-4}{s+4}$$

- Only **poles** dictate system stability, **zeros** in the right-half s -plane do *NOT* cause instability.

3.19 LTI Unstable Response

- An LTI system is **unstable** if the response of the system is unbounded for *any* bounded input.



3.20 LTI Unstable Examples

- An **unstable system** has at least one pole in the right-half s -plane and/or imaginary axis poles of multiplicity greater than one.

Example

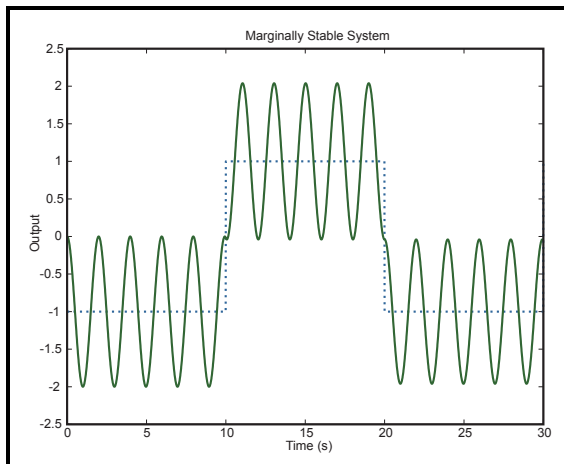
$$G_1(s) = \frac{s + 3}{(s + 2)(s - 3)}$$

$$G_2(s) = \frac{10}{s^2 - 2s + 10}$$

$$G_3(s) = \frac{10}{(s^2 + 4)^2}$$

3.21 LTI Marginally Stable Response

- A LTI system is **marginally/critically stable** if the natural response to a bounded input is offset or oscillating.



3.22 LTI Marginally Stable Examples

- A **marginally stable** system has distinct complex conjugate pole pairs on the imaginary axis and all other poles in the left-half s -plane

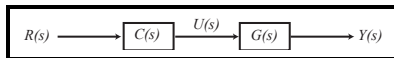
Example

$$G_1(s) = \frac{4}{(s^2 + 4)(s^2 + 16)}$$

$$G_2(s) = \frac{s - 4}{(s + 2)(s^2 + 1)}$$

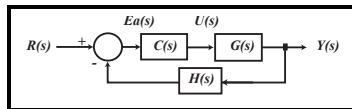
3.23 Linear Feedback Control Systems

- A **Control System** is a set of interconnected systems designed to provide a desired system response.
- **Open Loop System**



- Has no feedback.
- The output is generated directly by the input.

- **Closed Loop System**



- Has a feedback loop; is error based.
- Compares the output with the desired response.
- Often $H(s) = 1$, called a **unity feedback** system.
- Feedback reduces the **sensitivity** of the control system to model inaccuracies and disturbances.