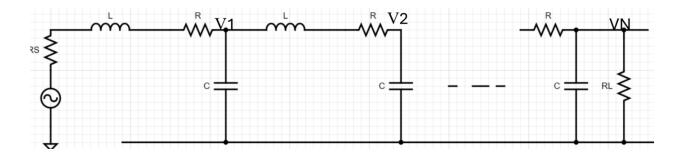
Compare RLC ladder approximation to NILT0 and NILTcv

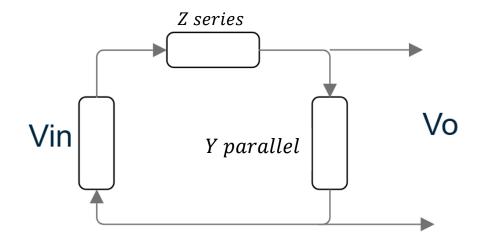
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Date: 11/11/2024

• A transmission line can be approximated using the following exact solution:



This can be represented as the following to obtain the exact solution:



Then,

$$Z series = Zo sinh(\gamma l)$$

$$Zo = \sqrt{\frac{z}{y}}$$
, $Y = G + SC$, $Z = (R + SL)$, $\gamma = \sqrt{ZY}$

$$\textit{Y parallel} = \textit{Yo} \tanh \left(\frac{\gamma l}{2}\right), \quad \textit{Yo} = \frac{1}{\textit{Zo}} \;, \quad \textit{Z parallel} = \frac{1}{\textit{Y parallel}}$$

$$T(s) = \frac{Vo}{Vin} = \frac{Z Parallel}{Zseries + Z parallel}$$

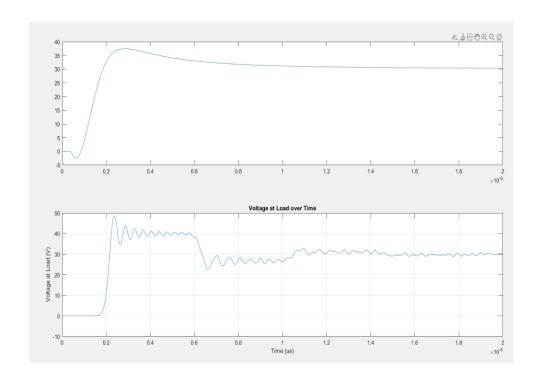
$$T(s) = \frac{Vo}{Vin} = \frac{\frac{\sqrt{R+SC}}{\tan h\left(l*\frac{\sqrt{(R+SC)(G+SC)}}{2}\right)}}{\sqrt{\frac{R+SC}{G+SC}}\sinh\left(l*\sqrt{(R+SC)(G+SC)}\right) + \frac{\sqrt{\frac{R+SC}{G+SC}}}{\tanh\left(l*\frac{\sqrt{(R+SC)(G+SC)}}{2}\right)}}}$$

$$T(s) = \frac{Vo}{Vin} = \frac{\frac{1}{\tanh\left(l*\frac{\sqrt{(R+SC)(G+SC)}}{2}\right)}}{\frac{\sinh\left(l*\sqrt{(R+SC)(G+SC)}\right) + \frac{1}{\tanh\left(l*\frac{\sqrt{(R+SC)(G+SC)}}{2}\right)}}}{\tanh\left(l*\frac{\sqrt{(R+SC)(G+SC)}}{2}\right)}}$$

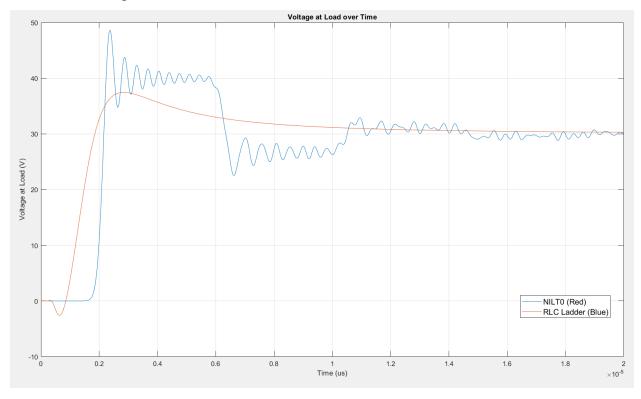
$$\tanh\left(\frac{x}{2}\right) = \frac{\cosh(x) - 1}{\sinh(x)}$$

$$T(s) = \frac{Vo}{Vin} = \frac{1}{\cosh\left(l*\sqrt{(R+SC)(G+SC)}\right)}$$

Using NILTO, at M=2 and compare it with RLC ladder (not state space)



Or in one plot,



This with NILT0 without time stepping.

• RLC ladder using space state representation.

Telegrapher equations are known as.

$$\frac{dV(x,t)}{dx} = -R(x) * i(x,t) - L(x) \frac{di(xi,t)}{dt}$$
$$\frac{di(x,t)}{dx} = -G(x) * V(x,t) - C(x) \frac{dV(xi,t)}{dt}$$

in state space,

$$\frac{d}{dx} \begin{bmatrix} V(x,t) \\ i(x,t) \end{bmatrix} = \begin{bmatrix} 0 & -R(x) \\ -G(x) & 0 \end{bmatrix} * \begin{bmatrix} V(x,t) \\ i(x,t) \end{bmatrix} - \begin{bmatrix} 0 & -L(x) \\ -C(x) & 0 \end{bmatrix} * \frac{d}{dt} \begin{bmatrix} V(x,t) \\ i(x,t) \end{bmatrix}$$

Moving this to Laplace domain we get,

$$\frac{d}{dx}\begin{bmatrix} V(x,s) \\ i(x,s) \end{bmatrix} = \begin{bmatrix} 0 & -Z(x,s) \\ -Y(x,s) & 0 \end{bmatrix} * \begin{bmatrix} V(x,s) \\ i(x,s) \end{bmatrix} + \begin{bmatrix} 0 & L(x) \\ C(x) & 0 \end{bmatrix} * \begin{bmatrix} V(x,0) \\ i(x,0) \end{bmatrix}$$

Where Z(x, s) = R(x) + sL(x), and Y(x, s) = G(x) + sC(x, s) are series impedance.

Let,

$$W(x,s) = \begin{bmatrix} V(x,s) \\ I(x,s) \end{bmatrix}, M = \begin{bmatrix} 0 & -Z(x,s) \\ -Y(x,s) & 0 \end{bmatrix}, and N = \begin{bmatrix} 0 & L(x) \\ C(x) & 0 \end{bmatrix}$$

So,

$$\frac{dW(x,s)}{dx} = M * W(x,s) + N * W(x,0)$$

Then, W at the end of the transmission line (W (1, s)) should equal to,

Let
$$\Phi = e^{Ml}$$
, so $\phi = \begin{bmatrix} \Phi_{11} & \Phi_{12} \\ \Phi_{21} & \Phi_{22} \end{bmatrix}$, $W(l,s) = \Phi W(0,s) + \int_0^l e^{M(l-x)} *N *W(x,0) dx$

If the initial conditions are zero, (i.e. at W(x,0) = 0) then,

$$W(l,s) = \Phi W(0,s)$$

Or matrix format,

$$W(l,s) = \begin{bmatrix} V(l,s) \\ I(l,s) \end{bmatrix} = \begin{bmatrix} \Phi_{11} & \Phi_{12} \\ \Phi_{21} & \Phi_{22} \end{bmatrix} \cdot \begin{bmatrix} V(0,s) \\ I(0,s) \end{bmatrix}$$

If I(1,s) = 0, Then,

$$\begin{bmatrix} V(l,s) \\ 0 \end{bmatrix} = \begin{bmatrix} \Phi_{11} & \Phi_{12} \\ \Phi_{21} & \Phi_{22} \end{bmatrix} \cdot \begin{bmatrix} V(0,s) \\ I(0,s) \end{bmatrix}$$

$$0 = \Phi_{21}V(0,s) + \Phi_{22}I(0,s)$$

$$I(0,s) = \Phi_{21}V(0,s) - \Phi_{22}^{-1}$$

$$V(l,s) = \Phi_{11}V(0,s) + \Phi_{12}I(0,s)$$

$$V(l,s) = \Phi_{11}V(0,s) - \Phi_{12}\Phi_{21}\Phi_{22}^{-1}V(0,s)$$

This is in the S domain, one can implement an NILT to solve it and get V(l,t) and to do this the following steps can be followed:

1. Find e^Ml, MATLAB has a function can do that very easily called expm(M*l).

Testing expm with the following example.

$$\frac{d}{dt}[x] = \begin{bmatrix} 0 & 2 \\ -3 & -5 \end{bmatrix} * [x] + \begin{bmatrix} 0 \\ 1 \end{bmatrix} * xin$$
$$y = \begin{bmatrix} 1 & 0 \end{bmatrix} * [x]$$

The answer to this is,

$$y(t) = C * e^{at} * B$$
$$e^{at} = M * e^{\Omega t} * M^{-1}$$

These can be obtained using the model matrix method, then y(t),

$$y(t) = 2 * e^{-2t} - 2e^{-3t}$$

Using the following code in MATLAB, we get the same result easily,

```
A = [0 2; -3 -5];
B = [0;1];
C= [1 0];
eat = expm(A*t);
y = mtimes(mtimes(C,eat),B);
```

Implemtning this to get an expression for V(l,s) we get,

$$V(l,s) = \frac{2 * Vs(s) * e^{l*\sqrt{Y*Z}}}{e^{(2*l*\sqrt{Y*Z})} + 1}$$

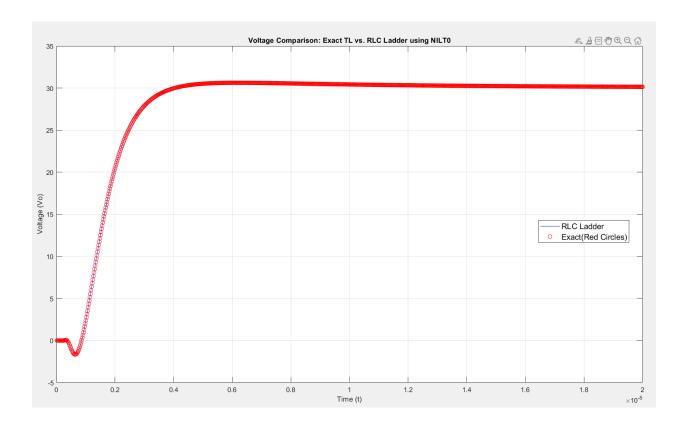
Where Vs(s) is the input in the s domain.

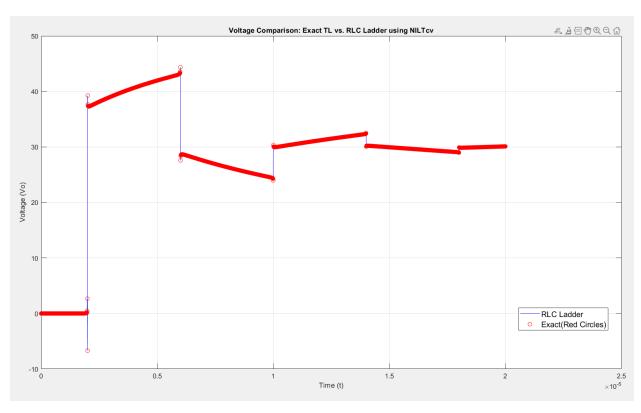
Let
$$x = l * \sqrt{Y * Z}$$
, so $V(l, s) = \frac{2 * Vs(s) * e^x}{e^{(2x)} + 1}$

Using the fact that,

$$cosh(x) = \frac{(e^x + e^{-x})}{2}$$
, then

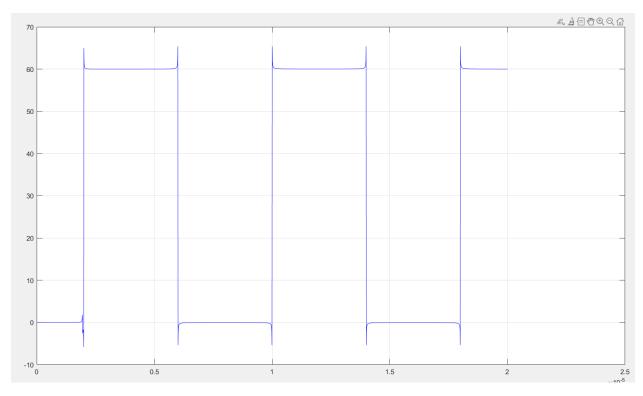
$$V(l,s) = \frac{Vs(s)}{\cosh(l*\sqrt{Y*Z}) + 1}$$
 which is the same as the exact solution above.



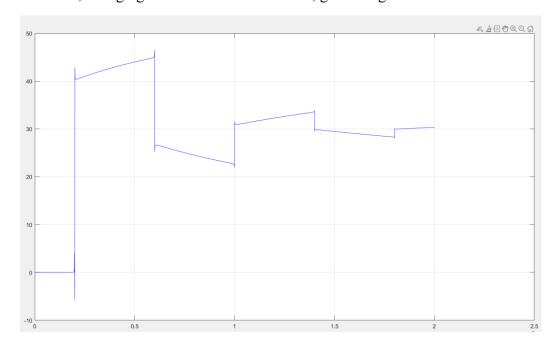


• NILTev

When passing the exact solution to NILTcv with R = G = 0, L = 2.5e-7, C = 1e-10 and a unit step input, the output was like the following.



However, changing the value of either G or R, get me a good result as below:



- Further reading includes
- 1. Understand how Niltev works.
- 2. Look at the derived expression for the transmission Line in MATLAB for Engineers Applications in Control, Electrical Engineering, IT and Robotics.

$$V(s,x) = Vi(s) \frac{Z_c(s)}{Z_c(s) + Z_i(s)} * \frac{e^{-\gamma(s)x} + \rho_2(s)e^{-\gamma(s)[2l-x]}}{1 - \rho_1(s)\rho_2(s)e^{-\gamma(s)l}}$$

Where,

$$Z_c(s) = \sqrt{\frac{z}{y}}, \qquad Y = G + SC, \qquad z = (R + SL),, \quad \gamma = \sqrt{ZY}$$

$$\rho_1(s) = \frac{Z_i(s) - Z_c(s)}{Z_i(s) + Z_c(s)}, \qquad \rho_2(s) = \frac{Z_L(s) - Z_c(s)}{Z_L(s) + Z_c(s)}$$