EE518 LAB Experiment 6

Design and implement a matrix convolution circuit using Verilog



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1 Objective:

Design and implement Verilog modules to:

- a. Perform matrix convolution between two matrices of size nxn and mxm (n>m) using 2's complement fixed point representation.
- b. Perform matrix convolution between two matrices of size nxn and mxm (n>m) using IEEE754 format.

2 Theory:

Convolution is the treatment of a matrix by another one which is called "kernel". A convolution is a type of matrix operation, consisting of a kernel, a small matrix of weights, that slides over input data performing element-wise multiplication with the part of the input it is on, then summing the results into an output.

Intuitively, a convolution allows for weight sharing - reducing the number of effective parameters - and image translation (allowing for the same feature to be detected in different parts of the input space).

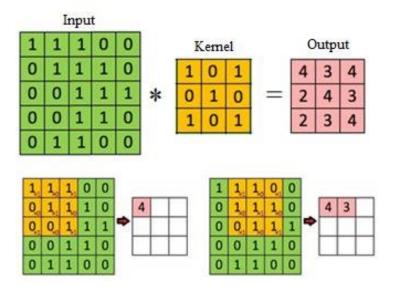


Figure 1: Basic Convolution Operation

3 Matrix Convolution with 2's complement representation:

3.1 Design Approach:

The design only has one input which accepts each input at the positive edge of clock. So, both the matrix elements are inserted to the design one by one. Initially the kernel inputs are send which is a mxm matrix (m=2) so first m^2 (4) elements are stored in a shift register which is our kernel. Now the main matrix input are fed one by one. We have such an arrangement of shift registers that we get a m^*m window of the main n^*n matrix at any instant of time and the convolution output is produced directly. The figure gives a rough representation of our algorithm. The output signals out_valid and end_conv indicate us which output values are correct and when the operation has ended.

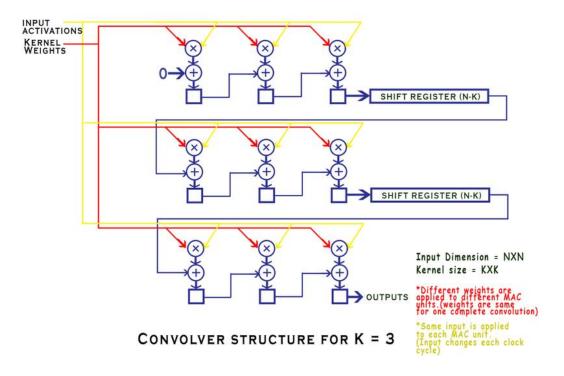


Figure 2: Convolution Algorithm Schematic

3.2 Verilog Code:

```
module conv_fix(
    input clock,
    input rst,
    input [31:0] a_in,
    output [31:0] result,
    output reg out_valid,
    output reg end_conv
    );
    parameter n=3, m=2;
    reg [31:0] k [m*m-1:0];
    reg [31:0] w [m*m-1:0];
    reg [31:0] u [m*m-1:0];
    reg [31:0] p [(m-1)*n-1:0];
    reg [5:0] count, count1, count2, count3;
    reg [5:0] y;
    always@(posedge clock) begin
    if(rst) count <= 'b0;</pre>
    else if(end_conv == 'b1) count <= 'b0;</pre>
    else count <= count + 1'b1;</pre>
    end
    always@(posedge clock)
    k[m*m-1] \le (count \le m*m) ? a_in : k[m*m-1];
    genvar i;
    generate
    for(i=0;i<m*m-1;i=i+1)
    always@(posedge clock)
    k[m*m-1-i-1] \le (count< m*m) ? k[m*m-1-i] : k[m*m-1-i-1] ;
    endgenerate
    always@(posedge clock)
    w[m*m-1] \le a_{in};
    genvar j,1;
```

```
generate
for(i=0;i< m-1;i=i+1) begin
for(j=0; j< m-1; j=j+1)
always@(posedge clock)
w[m*m-1-i*m-j-1] \le w[m*m-1-i*m-j];
always@(*)
p[0+i*m] \le w[m*m-1-i*m];
for(l=0;l< n-1;l=l+1)
always@(posedge clock) begin
p[1+i*m+1] <= p[1+i*m];
w[m*m-1-i*m-m] <= p[n-1];
end
end
for(j=0;j<m-1;j=j+1)
always@(posedge clock)
w[m-1-j-1] \le w[m-1-j];
endgenerate
//inputs at w and k
always@(*) u[m*m-1] \le w[m*m-1]*k[m*m-1];
generate
for(i=m*m-2;i>=0;i=i-1)
always@(*) u[i] \le u[i+1] + w[i]*k[i];
endgenerate
assign result = u[0];
// logic to generate valids
always@(posedge clock) begin
if(rst) begin out_valid <= 'b0; end_conv <= 'b0; end</pre>
else if(count < (m-1)*n + m - b1 + m*m);
else if(count == (m-1)*n + m - 'b1 + m*m) begin
out_valid <= 'b1; count1<='b1; count2<='b1; count3<='b1; end
else if(end_conv == 'b1) begin out_valid <= 'b0; end_conv <= 'b0; end</pre>
else if((count2 == n-m) && (count3 == n-m+'b1)) begin
```

```
end_conv <= 'b1; count1<='b0; count2<='b0; count3<='b0; end
else if(count2 == n-m+'b1) begin
out_valid <= 'b0; count1<=count1+'b1; count2<='b0; end
else if(count1%n == 'b0) begin
out_valid <= 'b1; count1<=count1+'b1; count2<='b1; count3<=count3+'b1; end
else begin count1<=count1+'b1; count2<=count2+'b1; end
end</pre>
```

endmodule

Note: Decimal point is not defined in code. Just take proper care in testbench to have x bits after decimal in input and 2x bits in output. Here we take 5 bits after decimal in input and 10bits in output for simulation. Code works for all cases.

3.3 Test bench;

```
module tbconv_fix( );

reg clock, rst;
reg [31:0] a;
wire [31:0] result;
wire out_valid, end_conv;

always #50 clock = ~clock;

conv_fix DUT (clock, rst, a, result, out_valid, end_conv);

always@(negedge clock) a = a+6'b100000;

initial begin
clock = 0; rst = 1;
#100 rst = 0; a = 6'b000000;
end
endmodule
```

3.4 RTL Schematic:

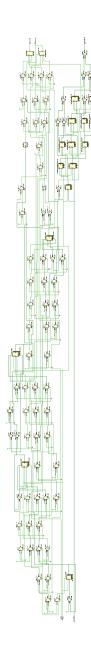


Figure 3: Schematic of 2's complement convolution circuit

3.5 Simulation:

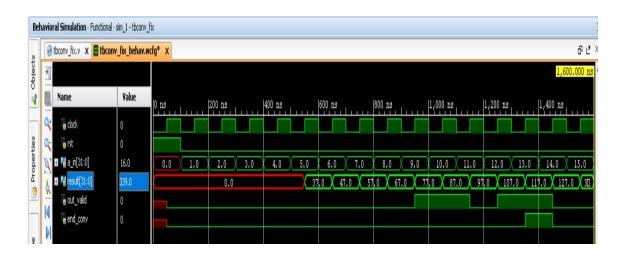


Figure 4: Behavioural simulation waveform

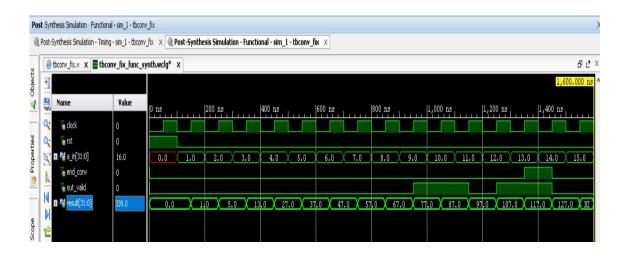


Figure 5: Post-synthesis functional simulation waveform

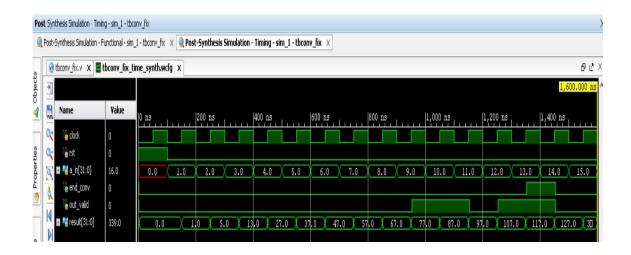


Figure 6: Post synthesis timing simulation waveform

Note: The representation of the input and output are kept in fixed point. The kernal matrix is 1,2,3,4 and main matrix is 5,6,7,8,9,10,11,12,13 so the outputs should be 77, 87, 107, 117. The results are as expected.

3.6 Synthesis Reports:

The values have been found after synthesis of the corresponding designs. I have done the experiment on Vivado 2014.1. The FPGA board selected is Artix-7. The LUTs and Flops have been found from the utilization report. The delay has been found from the timing report and the power has been found from the power report.

We have added proper constraints and the synthesis results are shown below.

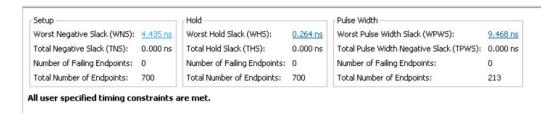


Figure 7: Timing summary

			1	посеа	Į.	Available	1	OTILE	-
	-+-		101		2550				57
lice Lurs*	1	153	1	U	1	134600	J	0.11	-
LUT as Logic	1	153	1	0	1	134600	1	0.11	1
LUT as Memory	1	0	1	0	1	46200	1	0.00	
lice Registers	1	203	1	0	1	269200	1	0.07	1
Register as Flip Flop		203	1	0	1	269200	1	0.07	
Register as Latch	1	0	1	0	1	269200	1	0.00	1
7 Muxes	1	0	1	0	1	67300	1	0.00	1
3 Muxes	1	0	1	0	1	33650	1	0.00	1
1	LUT as Memory Lice Registers Register as Flip Flop Register as Latch 7 Muxes	LUT as Logic LUT as Memory Lice Registers Register as Flip Flop Register as Latch Muxes	LUT as Logic 153 LUT as Memory 0 Lice Registers 203 Register as Flip Flop 203 Register as Latch 0 7 Muxes 0	LUT as Logic 153 LUT as Memory 0 Lice Registers 203 Register as Flip Flop 203 Register as Latch 0 7 Muxes 0	LUT as Logic 153 0 LUT as Memory 0 0 Lice Registers 203 0 Register as Flip Flop 203 0 Register as Latch 0 0 7 Muxes 0 0	LUT as Logic 153 0 LUT as Memory 0 0 Lice Registers 203 0 Register as Flip Flop 203 0 Register as Latch 0 0 Muxes 0 0	LUT as Logic 153 0 134600 LUT as Memory 0 0 46200 Lice Registers 203 0 269200 Register as Flip Flop 203 0 269200 Register as Latch 0 0 269200 7 Muxes 0 0 67300	LUT as Logic 153 0 134600 LUT as Memory 0 0 46200 Lice Registers 203 0 269200 Register as Flip Flop 203 0 269200 Register as Latch 0 0 269200 7 Muxes 0 0 67300	LUT as Logic 153 0 134600 0.11 LUT as Memory 0 0 46200 0.00 Lice Registers 203 0 269200 0.07 Register as Flip Flop 203 0 269200 0.07 Register as Latch 0 0 269200 0.00 7 Muxes 0 0 67300 0.00

Figure 8: Utilization report

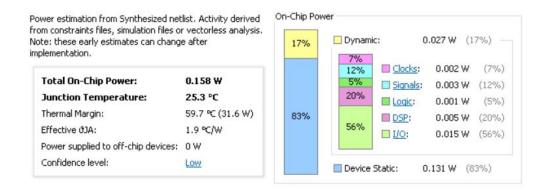


Figure 9: Power report

4 Matrix Convolution with IEEE754 representation:

4.1 Design Approach:

It is same as previous only the multipliers and adders are made by using the modules of previous experiments. Rest of the logic is all same.

4.2 Verilog Code:

```
module conv_flt(
    input clock,
    input rst,
    input [31:0] a_in,
    output [31:0] result,
    output reg out_valid,
    output reg end_conv
    );
    parameter n=3, m=2;
    reg [31:0] k [m*m-1:0];
    reg [31:0] w [m*m-1:0];
    wire [31:0] u [m*m-1:0];
    wire [31:0] v [m*m-2:0];
    reg [31:0] p [(m-1)*n-1:0];
    reg [5:0] count, count1, count2, count3;
    reg [5:0] y;
    always@(posedge clock) begin
    if(rst) count <= 'b0;</pre>
    else if(end_conv == 'b1) count <= 'b0;</pre>
    else count <= count + 1'b1;</pre>
    end
    always@(posedge clock)
    k[m*m-1] \le (count \le m*m) ? a_in : k[m*m-1];
    genvar i;
    generate
```

```
for(i=0;i<m*m-1;i=i+1)
always@(posedge clock)
k[m*m-1-i-1] \le (count< m*m) ? k[m*m-1-i] : k[m*m-1-i-1] ;
endgenerate
always@(posedge clock)
w[m*m-1] \le a_{in};
genvar j,1;
generate
for(i=0;i< m-1;i=i+1) begin
for(j=0; j< m-1; j=j+1)
always@(posedge clock)
w[m*m-1-i*m-j-1] \le w[m*m-1-i*m-j];
always@(*)
p[0+i*m] \le w[m*m-1-i*m];
for(l=0;l< n-1;l=l+1)
always@(posedge clock) begin
p[1+i*m+1] \le p[1+i*m];
w[m*m-1-i*m-m] <= p[n-1];
end
end
for(j=0;j<m-1;j=j+1)
always@(posedge clock)
w[m-1-j-1] \le w[m-1-j];
endgenerate
//inputs at w and k
fpmul_32b mul(w[m*m-1], k[m*m-1], u[m*m-1]);
generate
for(i=m*m-2; i>=0; i=i-1) begin
fpmul_32b mul(w[i],k[i],v[i]);
fpaddsub_32b adsb(u[i+1], v[i], 'b0, u[i]);
end
endgenerate
```

```
assign result = u[0];
    // logic to generate valids
    always@(posedge clock) begin
    if(rst) begin out_valid <= 'b0; end_conv <= 'b0; end
    else if(count < (m-1)*n + m - 'b1 + m*m );
    else if(count == (m-1)*n + m - b1 + m*m) begin
    out_valid <= 'b1; count1<='b1; count2<='b1; count3<='b1; end
    else if(end_conv == 'b1) begin out_valid <= 'b0; end_conv <= 'b0; end
    else if((count2 == n-m) && (count3 == n-m+'b1)) begin
    end_conv <= 'b1; count1<='b0; count2<='b0; count3<='b0; end
    else if(count2 == n-m+'b1) begin
    out_valid <= 'b0; count1<=count1+'b1; count2<='b0; end
    else if(count1%n == 'b0) begin
    out_valid <= 'b1; count1<=count1+'b1; count2<='b1; count3<=count3+'b1; end
    else begin count1<=count1+'b1; count2<=count2+'b1; end
    end
endmodule
module fpmul_32b(a_in, b_in, c_out );
parameter m = 8, n = 23;
input [m+n:0] a_in, b_in;
output [m+n:0] c_out;
wire sign, is_zero;
wire [m-1:0] exp, exp_c, exp_d;
wire [2*n+1:0] mant;
wire [2*n+1:0] mul;
assign sign = a_{in}[m+n] ^ b_{in}[m+n];
assign \ exp_c = a_in[m+n-1:n] \ + \ b_in[m+n-1:n] \ - \ \{(m-1)\{1'b1\}\};
assign mul = \{1'b1,a_{in}[n-1:0]\} * \{1'b1,b_{in}[n-1:0]\};
assign is_zero = (a_in[m+n-1:0] == 'b0) || (b_in[m+n-1:0] == 'b0);
assign exp = exp_c + mul[2*n+1];
assign mant = mul >> (mul[2*n+1]);
```

```
assign c_out = is_zero ? 'b0 : {sign,exp,mant[2*n-1:n]};
endmodule
module fpaddsub_32b(a_in, b_in, sub, c_out );
parameter m = 8, n = 23;
input [m+n:0] a_in, b_in;
input sub;
output [m+n:0] c_out;
wire sign, sub_b, a_is_big, b_is_big, both_equal, mant_abig;
wire [m-1:0] exp, exp_c, exp_d, shift_index;
wire [n+1:0] mant, mant_d, a_shifted, b_shifted, shifted;
assign a_{is}big = a_{in}[m+n-1:n] > b_{in}[m+n-1:n];
assign b_{is}big = a_{in}[m+n-1:n] < b_{in}[m+n-1:n];
assign both_equal = a_{in}[m+n-1:n] == b_{in}[m+n-1:n];
assign mant_abig = both_equal ?
                a_{in}[n-1:0] > b_{in}[n-1:0] : 'b0;
assign exp_c = a_is_big ? a_in[m+n-1:n] - b_in[m+n-1:n] :
                b_{is}big ? b_{in}[m+n-1:n] - a_{in}[m+n-1:n] : 'b0;
assign a_shifted = b_{is_big} ? {1'b1,a_in[n-1:0]}>>(exp_c) :
                     {1'b1,a_in[n-1:0]};
assign b_shifted = a_is_big ? \{1'b1,b_in[n-1:0]\} >> (exp_c) :
                     {1'b1,b_in[n-1:0]};
assign sub_b = a_in[m+n] ^ (b_in[m+n] ^ sub);
assign mant_d = sub_b ? (a_is_big || mant_abig) ?
                a_shifted - b_shifted :
                b_shifted - a_shifted :
                a_shifted + b_shifted;
assign sign = (a_is_big || mant_abig) ? a_in[m+n] : (b_in[m+n] ^ sub);
assign exp_d = a_is_big ? a_in[m+n-1:n] : b_in[m+n-1:n];
```

```
//corner case logic- not needed for normal operation
mantshift mant_shifter(exp_d, mant_d, shifted, shift_index );
assign mant = sub_b ? shifted : mant_d>>mant_d[n+1];
assign exp = sub_b ? exp_d - shift_index : exp_d + mant_d[n+1];
assign c_out = {sign,exp,mant[n-1:0]};
endmodule
module mantshift(exp, mant, shifted, shift_index );
parameter m = 8, n = 23;
    input [m-1:0] exp;
    input [n+1:0] mant;
    output reg [m-1:0] shift_index;
    output reg [n+1:0] shifted;
    reg [n+1:0] target;
    reg [$clog2(n)-1:0] cnt;
    always@(mant, exp) begin target = mant;
    shift_index = exp;
    for(cnt = 0; cnt < n+1; cnt = cnt + 1)begin</pre>
        if (target[cnt]) shift_index = n - cnt;
    end
    shifted = mant << shift_index;</pre>
    end
endmodule
4.3
      Test bench;
module tbconv_flt( );
    reg clock, rst;
    reg [31:0] a;
    wire [31:0] result;
    wire out_valid, end_conv;
    always #50 clock = ~clock;
```

```
conv_flt DUT (clock, rst, a, result, out_valid, end_conv);
initial begin
clock = 1; rst = 1; #100 rst = 0;
end
endmodule
```

4.4 RTL Schematic:

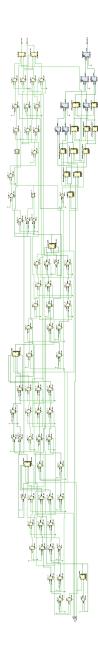


Figure 10: Schematic of IEEE754 convolution circuit

4.5 Simulation:

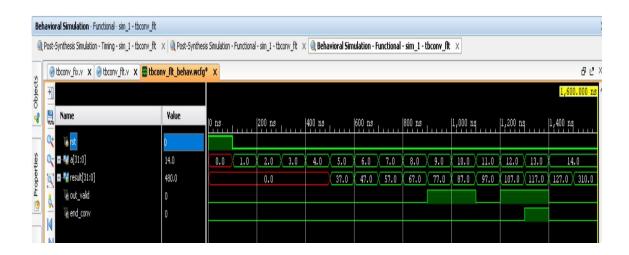


Figure 11: Behavioural simulation waveform

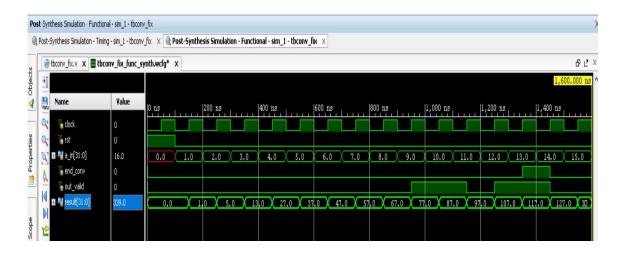


Figure 12: Post-synthesis functional simulation waveform

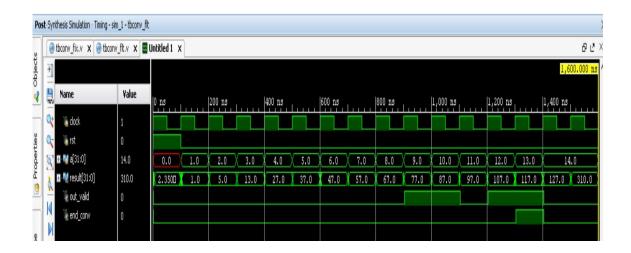


Figure 13: Post synthesis timing simulation waveform

Note: The representation of the input and output are kept in single precision floating point representation. The kernal matrix is 1,2,3,4 and main matrix is 5,6,7,8,9,10,11,12,13 so the outputs should be 77, 87, 107, 117. The results are as expected.

4.6 Synthesis Reports:

The values have been found after synthesis of the corresponding designs. I have done the experiment on Vivado 2014.1. The FPGA board selected is Artix-7. The LUTs and Flops have been found from the utilization report. The delay has been found from the timing report and the power has been found from the power report.

We have added proper constraints and the synthesis results are shown below.

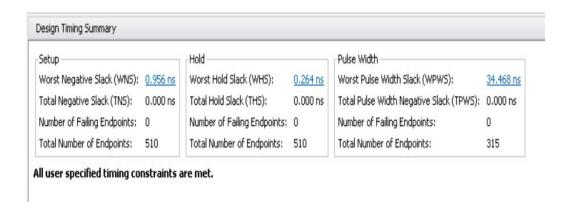


Figure 14: Timing summary

Site Type	I	Used	1	Loced	1	Available	1	Util%	1
 taka fasik (sinor) - pranci varberte kontroleskonto oliv - parto parta kanta en oli novi - pranci varberte transi pranci (sinor) - pranci parto pranci - part 	-+-				10.50	104600	-+-		
Slice LUTs*	ŀ	1756		0	2.5	134600	ł	1.30	
LUT as Logic		1756		0		134600		1.30	
LUT as Memory	1	0	1	0	1	46200	1	0.00	1
Slice Registers	1	314	1	0	1	269200	1	0.11	1
Register as Flip Flop	1	314	1	0	1	269200	1	0.11	1
Register as Latch	1	0	1	0	1	269200	1	0.00	1
F7 Muxes	t	0	1	0	1	67300	1	0.00	1
F8 Muxes	1	0	1	0	1	33650	1	0.00	1
+	-+-		-+-		-+		-+-		-+

Figure 15: Utilization report

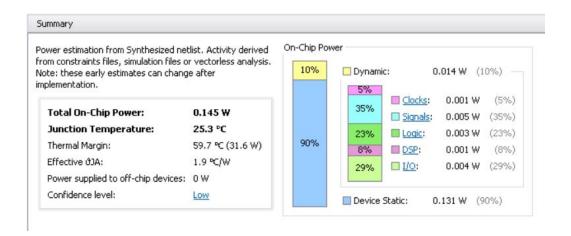


Figure 16: Power report

5 Conclusions:

- We have designed the circuit as per the requirement.
- The functionality of our design have been verified . The functionality are showing as expected.
- The different parameters of the design such as LUTs, delay and power have been calculated from the synthesis and tabulated.