# EE518 LAB Experiment 7

# Design and implement a matrix convolution circuit with padding and N-bit parameterised integer ALU using Verilog



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March 19, 2023

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## 1 Objective:

Design and implement Verilog modules to:

- a. Modify the convolution module for inclusion of the padding function. The convolution block should contain a flag to enable/disable padding for convolution.
- b. Design of a N bit parameterized integer arithmetic and logic unit, to perform addition, subtraction, right shift, left shift, arithmetic right shift, arithmetic left shift and logical operations as not, and, or and xor

## 2 Theory of Convolution:

Convolution is the treatment of a matrix by another one which is called "kernel". A convolution is a type of matrix operation, consisting of a kernel, a small matrix of weights, that slides over input data performing element-wise multiplication with the part of the input it is on, then summing the results into an output.

Intuitively, a convolution allows for weight sharing - reducing the number of effective parameters - and image translation (allowing for the same feature to be detected in different parts of the input space).

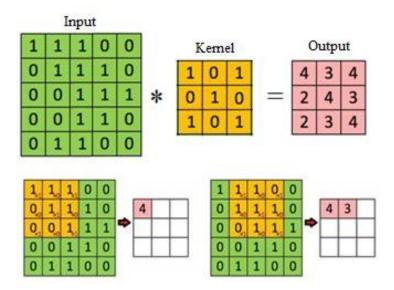


Figure 1: Basic Convolution Operation

#### 2.1 Padding:

Padding is a term relevant to convolutional neural networks as it refers to the amount of pixels added to an image when it is being processed by the kernel of a CNN. For example, if the padding in a CNN is set to zero, then every pixel value that is added will be of value zero. If, however, the zero padding is set to one, there will be a one pixel border added to the image with a pixel value of zero.

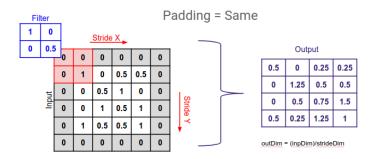


Figure 2: Basic Convolution Operation with padding

## 3 Matrix Convolution with padding:

#### 3.1 Design Approach:

The design only has one input which accepts each input at the positive edge of clock. So, both the matrix elements are inserted to the design one by one. Initially the kernel inputs are send which is a mxm matrix (m=2) so first  $m^2$  (4) elements are stored in a shift register which is our kernel. Now the main matrix input are fed one by one. We have such an arrangement of shift registers that we get a  $m^*m$  window of the main  $n^*n$  matrix at any instant of time and the convolution output is produced directly. The figure gives a rough representation of our algorithm. The output signals out\_valid and end\_conv indicate us which output values are correct and when the operation has ended.

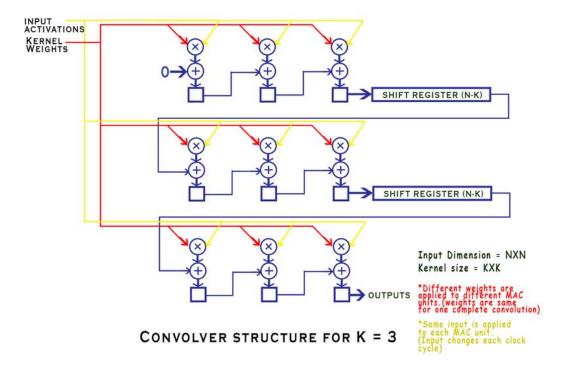


Figure 3: Convolution Algorithm Schematic

The code is kept same just a border of pixel with 0 values are added to have the padding implementation. However we couldnot add a flag to enable or disable padding which can also be implemented with some modifications. Code is given for fixed point representation only. Floating point has same code with some modification in MAC unit.

### 3.2 Verilog Code:

```
module conv_fix(
    input clock,
    input rst,
    input [31:0] a_in,
    output [31:0] result,
    output reg out_valid,
    output reg end_conv
```

```
);
   parameter n=3, m=2;
   reg [31:0] k [m*m-1:0];
   reg [31:0] w [m*m-1:0];
   reg [31:0] u [m*m-1:0];
   reg [31:0] v [(n+2)*(n+2)-1:0];
   reg [31:0] p [(m-1)*(n+2)-1:0];
   reg [5:0] count, count1, count2, count3;
   reg [5:0] y, i1=0, i2=0;
   always@(posedge clock) begin
   if(rst) count <= 'b0;</pre>
   else if(end_conv == 'b1) count <= 'b0;</pre>
   else count <= count + 1'b1;</pre>
   end
   always@(posedge clock)
   k[m*m-1] \le (count \le m*m) ? a_in : k[m*m-1];
   genvar i;
   generate
   for(i=0;i<m*m-1;i=i+1)
   always@(posedge clock)
   k[m*m-1-i-1] \le (count< m*m) ? k[m*m-1-i] : k[m*m-1-i-1] ;
   endgenerate
//code change for padding
   always@(posedge clock) begin
   if(count<m*m) w[m*m-1] <= 'b0;
   else if(count-m*m<n+3) begin
   v[i1] <= a_in;
   i1 = i1 + 1;
   w[m*m-1] \le b0; end
   else if(count>m*m+n+3-1+n*(n+2)) w[m*m-1] <= 'b0;
   else if((count-m*m-n+3-1)\%(n+2)<n) begin
   w[m*m-1] \le v[i2];
   i2 = i2 + 1;
   v[i1] <= a_in;
```

```
i1 = i1 + 1;
    end
    else if((count-m*m-n+3-1)\%(n+2)>=n) begin
    v[i1] <= a_in;
    i1 = i1 + 1;
    w[m*m-1] \le b0; end
    else w[m*m-1] \le b0;
    end
//padding code changes till here
    genvar j,1;
    generate
    for(i=0;i< m-1;i=i+1) begin
    for(j=0;j<m-1;j=j+1)
    always@(posedge clock)
    w[m*m-1-i*m-j-1] \le w[m*m-1-i*m-j];
    always@(*)
    p[0+i*m] \le w[m*m-1-i*m];
    for(1=0;1<(n+2)-1;1=1+1)
    always@(posedge clock) begin
    p[1+i*m+1] <= p[1+i*m];
    w[m*m-1-i*m-m] \le p[(n+2)-1];
    end
    end
    for(j=0; j< m-1; j=j+1)
    always@(posedge clock)
    w[m-1-j-1] \le w[m-1-j];
    endgenerate
    //inputs at w and k
    always0(*) u[m*m-1] \le w[m*m-1]*k[m*m-1];
    generate
    for(i=m*m-2;i>=0;i=i-1)
    always@(*) u[i] \le u[i+1] + w[i]*k[i];
    endgenerate
```

```
assign result = u[0];
// logic to generate valids
//n-->n+2 for padding logic
always@(posedge clock) begin
if(rst) begin out_valid <= 'b0; end_conv <= 'b0; end
else if(count < (m-1)*(n+2) + m - 'b1 + m*m );
else if(count == (m-1)*(n+2) + m - b1 + m*m) begin
out_valid <= 'b1; count1<='b1; count2<='b1; count3<='b1; end</pre>
else if(end_conv == 'b1) begin out_valid <= 'b0; end_conv <= 'b0; end
else if((count2 == (n+2)-m) && (count3 == (n+2)-m+'b1)) begin
end_conv <= 'b1; count1<='b0; count2<='b0; count3<='b0; end
else if(count2 == (n+2)-m+'b1) begin
out_valid <= 'b0; count1<=count1+'b1; count2<='b0; end
else if(count1\frac{1}{(n+2)} == 'b0) begin
out_valid <= 'b1; count1<=count1+'b1; count2<='b1; count3<=count3+'b1; end
else begin count1<=count1+'b1; count2<=count2+'b1; end
end
```

#### endmodule

Note: Decimal point is not defined in code. Just take proper care in testbench to have x bits after decimal in input and 2x bits in output. Here we take 5 bits after decimal in input and 10bits in output for simulation. Code works for all cases.

#### 3.3 Test bench;

```
module tbconv_fix( );
reg clock, rst;
reg [31:0] a;
wire [31:0] result;
wire out_valid, end_conv;
always #50 clock = ~clock;
```

```
conv_fix DUT (clock, rst, a, result, out_valid, end_conv);
always@(negedge clock) a = a+6'b100000;
initial begin
clock = 0; rst = 1;
#100 rst = 0; a = 6'b000000;
end
endmodule
```

# 3.4 RTL Schematic:

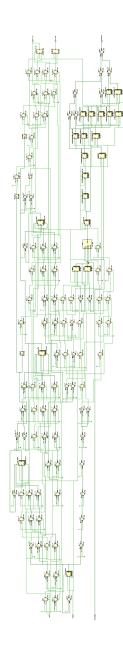


Figure 4: Schematic of convolution circuit with padding

#### 3.5 Simulation:

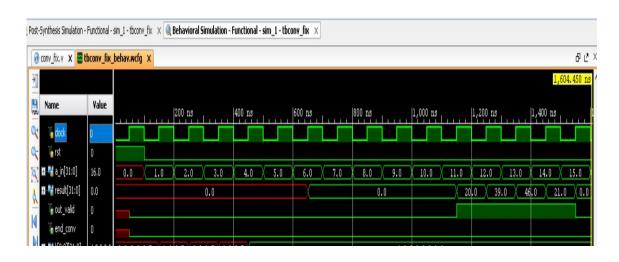


Figure 5: Behavioural simulation waveform 0-50

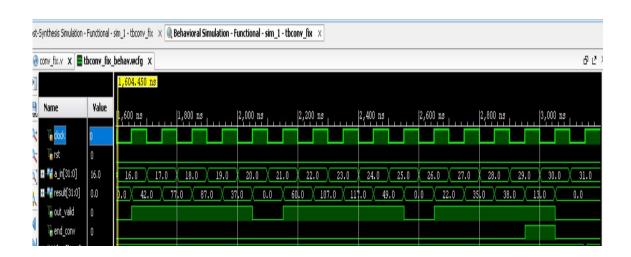


Figure 6: Behavioural simulation waveform 50-100



Figure 7: Post-synthesis functional simulation waveform 0-50

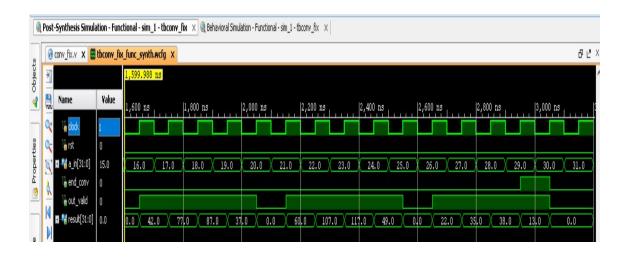


Figure 8: Post-synthesis functional simulation waveform 50-100

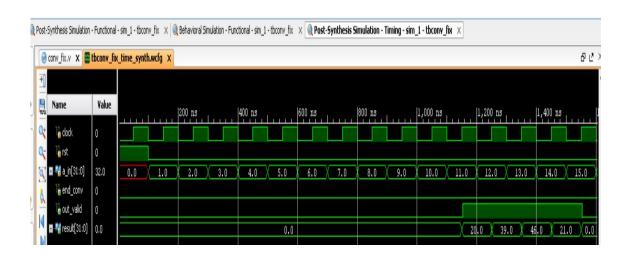


Figure 9: Post synthesis timing simulation waveform 0-50



Figure 10: Post synthesis timing simulation waveform 50-100

**Note:** The representation of the input and output are kept in fixed point. The kernal matrix is 1,2,3,4 and main matrix is 5,6,7,8,9,10,11,12,13 which is padded with 0s in the border, so the outputs should be 20, 39, 46, 21, 42, **77, 87,** 37, 60, **107, 117,** 49, 22, 35, 38, 13. The results are as expected. The bold outputs were available without padding and others are added with padding.

#### 3.6 Synthesis Reports:

The values have been found after synthesis of the corresponding designs. I have done the experiment on Vivado 2014.1. The FPGA board selected is Artix-7. The LUTs and Flops have been found from the utilization report. The delay has been found from the timing report and the power has been found from the power report.

We have added proper constraints and the synthesis results are shown below.

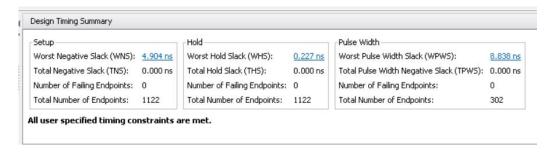


Figure 11: Timing summary

Site Type						Available	0.0		
Slice LUTs*	I	347			1				
LUT as Logic	1	291	1	0	1	134600	I	0.21	1
LUT as Memory	L	56	1	0	1	46200	1	0.12	1
LUT as Distributed RAM	1	24	1	0	1		1		١
LUT as Shift Register	U	32	1	0	1		1		1
Slice Registers	1	213	1	0	1	269200	1	0.07	1
Register as Flip Flop	ľ	213	1	0	1	269200	T	0.07	1
Register as Latch	1	0	1	0	1	269200	1	0.00	1
F7 Muxes		0	1	0	1	67300	1	0.00	1
F8 Muxes	1	0	1	0	1	33650	1	0.00	1
	-+-		-+		-+-		-+		-+

Figure 12: Utilization report

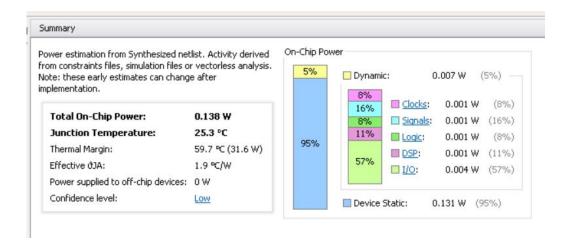


Figure 13: Power report

## 4 N-bit parameterised integer ALU:

#### 4.1 Design Approach:

A simple arithmetic and logical unit has been designed with muxing logic. The functions implemented are addition, subtraction, right shift, left shift, arithmetic right shift, arithmetic left shift and logical operations as not, and, or, xor and the corresponding control signals are 1, 2, 3, 4, 5, 6, 7, 8, 9, a. The code is shown below.

## 4.2 Verilog Code:

```
always@(posedge clock) outP <= out;</pre>
always@(rs1, rs2, ctrl) begin
case(ctrl)
4'h1 : out <= rs1 + rs2 ;
                          //add
4'h2 : out <= rs1 - rs2 ;
                          //sub
4'h3 : out <= rs1 >> rs2[4:0] ; //rightshift
4'h4 : out <= rs1 << rs2[4:0] ; //leftshift
4'h5 : out <= $signed(rs1) >>> rs2[4:0] ; //rightshift arithmetic
4'h6 : out <= $signed(rs1) <<< rs2[4:0] ; //leftshift arithmetic
4'h7 : out <= ~rs1 ;
                      //not
4'h8 : out <= rs1 & rs2 ;
                            //and
4'h9 : out <= rs1 | rs2 ;
                            //or
4'ha : out <= rs1 ^ rs2 ; //xor
default : out <= 32'hxxxxxxxx ;</pre>
endcase
end
endmodule
4.3
      Test bench;
module tb_alu(
reg clock;
reg [31:0] rs1, rs2;
reg [3:0] ctrl;
wire [31:0] outP;
always #100 clock = ~clock;
alu_n DUT (rs1, rs2, clock, ctrl, outP);
initial begin
clock = 0;
#200 ctrl = 4'h1; rs1=$random; rs2=$random;
```

```
#200 ctrl = 4'h2; rs1=$random; rs2=$random;

#200 ctrl = 4'h3; rs1=$random; rs2=5'h4;

#200 ctrl = 4'h4; rs1=$random;

#200 ctrl = 4'h5; rs1=$random;

#200 ctrl = 4'h6; rs1=$random;

#200 ctrl = 4'h7; rs1=$random; rs2=$random;

#200 ctrl = 4'h8; rs1=$random; rs2=$random;

#200 ctrl = 4'h9; rs1=$random; rs2=$random;

#200 ctrl = 4'h9; rs1=$random; rs2=$random;

#200 ctrl = 4'ha; rs1=$random; rs2=$random;
```

#### 4.4 RTL Schematic:

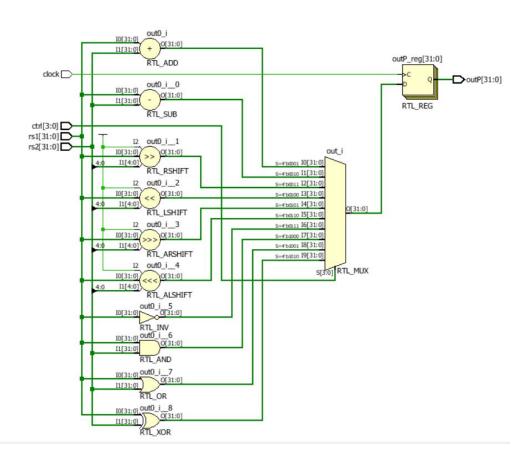


Figure 14: Schematic of ALU

#### 4.5 Simulation:

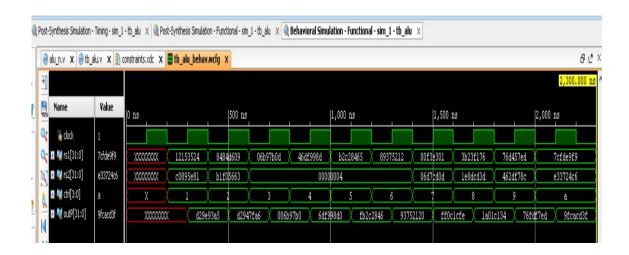


Figure 15: Behavioural simulation waveform

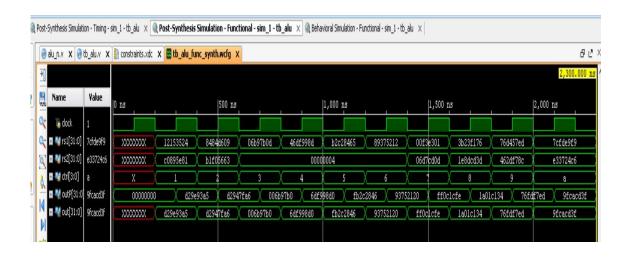


Figure 16: Post-synthesis functional simulation waveform

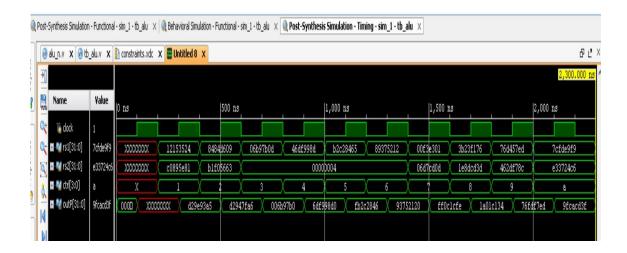


Figure 17: Post synthesis timing simulation waveform

#### 4.6 Synthesis Reports:

The values have been found after synthesis of the corresponding designs. I have done the experiment on Vivado 2014.1. The FPGA board selected is Artix-7. The LUTs and Flops have been found from the utilization report. The delay has been found from the timing report and the power has been found from the power report.

We have added proper constraints and the synthesis results are shown below.

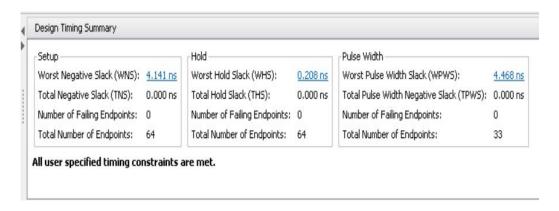


Figure 18: Timing summary

Site Type	1		ı,			Available	•		
Slice LUTs*	1	388	63	0	10	134600		0.28	
LUT as Logic	1	388	1	0	1	134600	1	0.28	
LUT as Memory	1	0	1	0	1	46200	1	0.00	
Slice Registers	1	32	1	0	1	269200	1	0.01	
Register as Flip Flop	1	32	1	0	1	269200	1	0.01	
Register as Latch	1	0	1	0	1	269200	1	0.00	
F7 Muxes	1	0	1	0	1	67300	1	0.00	
F8 Muxes		0	1	0	1	33650	1	0.00	
	-+-		-+-		-+-		+-		

Figure 19: Utilization report

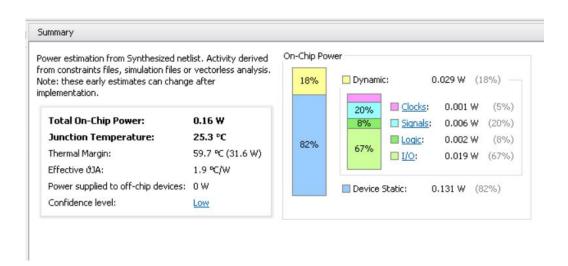


Figure 20: Power report

# 5 Conclusions:

• We have designed the circuit as per the requirement.

- The functionality of our design have been verified . The functionality are showing as expected.
- The different parameters of the design such as LUTs, delay and power have been calculated from the synthesis and tabulated.