

# Modeling CrIS Nonlinearity

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We consider the definition of reference truth for measurements made with a Michelson interferometer, taking into account relatively small (or possibly non-existent) effects such as filter position and potential mathematical artifacts. The immediate application is defining reference truth and a corresponding calibration equation for the CrIS instrument.

## 1 Michelson Interferometer

Figure 1 shows a basic Michelson interferometer. Let  $r_{\text{in}}(\nu)$  be incoming radiance as a function of frequency  $\nu$ ,  $\delta$  mirror displacement, and  $r_{\text{out}}(\nu, \delta)$  radiance on the path to the detector. In practice the signal from the detector is the product of incoming radiance, beam-splitter efficiency, and detector responsivity. But suppose for the moment that we have a perfect beam splitter and mirrors, and that there are no filters. Then radiance  $r_{\text{out}}$  on the path to the detector can be represented as

$$r_{\text{out}}(\nu, \delta) = r_{\text{in}}(\nu)(1 + \cos 2\pi\nu\delta)/2 \quad (1)$$

including a term  $r_{\text{in}}(\nu)/2$  that depends only on  $\nu$ . Integrating over frequency, we have

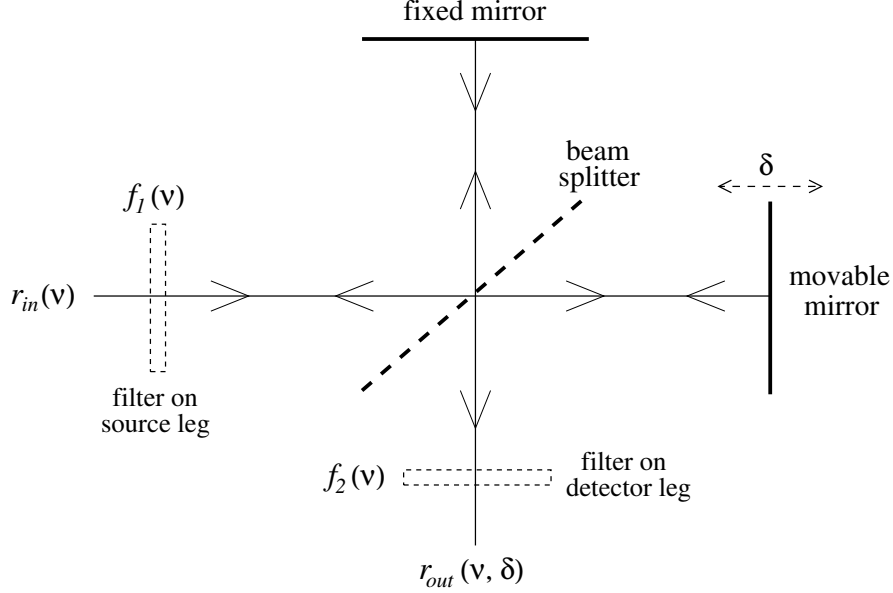


Figure 1: Michelson interferometer with filters

$$r_{\text{out}}(\delta) = \frac{1}{2} \int_{\nu=0}^{\text{inf}} r_{\text{in}}(\nu) d\nu + \frac{1}{2} \int_{\nu=0}^{\text{inf}} r_{\text{in}}(\nu) \cos(2\pi\nu\delta) d\nu \quad (2)$$

This is the continuous interferogram as a function of displacement  $\delta$ . We want to consider the AC and DC components separately.

$$r_{\text{dc}}(\delta) = \frac{1}{2} \int_{\nu=0}^{\text{inf}} r_{\text{in}}(\nu) d\nu, \quad r_{\text{ac}}(\delta) = \frac{1}{2} \int_{\nu=0}^{\text{inf}} r_{\text{in}}(\nu) \cos(2\pi\nu\delta) d\nu \quad (3)$$

$r_{\text{dc}}$  is a function of the scene, while  $r_{\text{ac}}$  is a function of both the scene and mirror position. For  $\delta = 0$  (zero mirror displacement) the two terms are identical.

Figure 2 shows sample transfer functions, taking radiance to voltage.  $G_1(r)$  and  $G_2(r)$  are linear, of the form  $G(r) = br + c$ , for some  $b$  and  $c$ . For CrIS  $c$  is Vinst, a DC bias value, while  $b$  depends on analog gain and detector properties.  $G_3(r)$  is a plausible arbitrary nonlinear transfer function.

