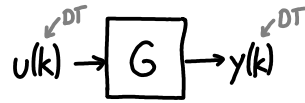


# System identification

## Notation differences to signals+systems:

- DT notation:  $x[n]$  is now  $x(k)$
- DT-fourier-transform:  $X(\Omega)$  is now  $X(e^{j\omega})$ ;  $\Omega \rightarrow \omega$
- Discrete-fourier-series:  $X[k]$  is now  $X(e^{j\omega_n})$ ;  $\Omega = \frac{2\pi k}{N} \rightarrow \omega_n = \frac{2\pi n}{N}$

## Fundamental Concept:



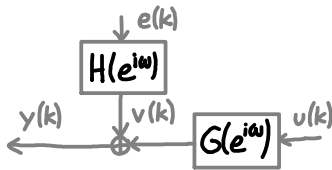
- 1) Choose input  $u(k), k=1,2,\dots$
- 2) Apply  $u(k)$  to plant  $G$
- 3) Measure output  $y(k), k=1,2,\dots$
- 4) Get model of  $G$  from  $u(k), y(k) \Rightarrow y = G \cdot u \rightarrow \hat{G} \approx y/u$

- frequency domain (DT-Fourier Transform)  
Estimate  $U(e^{j\omega}), Y(e^{j\omega}) \rightarrow$  get  $\hat{G}(e^{j\omega})$   
(requires more data, has less noise problems)
- time domain (Z-Transform)  
get  $\hat{G}(z)$  directly from  $u(k), y(k)$   
(requires less data, has more noise problems)

- open-loop:  $\hat{y} = \hat{G}u$   
  $\hookrightarrow \|y - \hat{y}\|$  low!

- closed-loop:  $\hat{y} = \frac{\hat{G}C}{1 + \hat{G}C} \cdot r$  (if  $G$  unstable)  
  $\hookrightarrow \frac{\hat{G}C}{1 + \hat{G}C}$  stable!,  $\|\frac{\hat{G}C}{1 + \hat{G}C} - \frac{\hat{G}C}{1 + \hat{G}C}\|$  low!

## Open-loop configuration

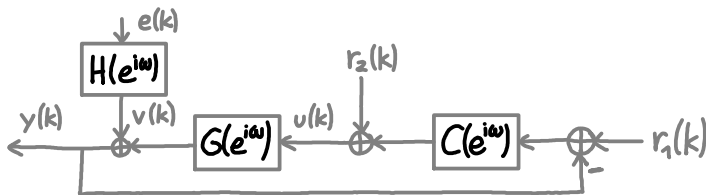


$G$ : plant to identify  
 $H$ : noise filter  
 $e$ : white noise  
 $v$ : filtered noise  
 $U$ : input  
 $y$ : output

real:  $y(k) = \sum_{l=0}^{\infty} g(l)u(k-l) + v(k)$   
assumed:  $y(k) = \sum_{l=0}^{N-1} g(l)u(k-l)$   
 $Y(e^{j\omega_n}) = \hat{G}(e^{j\omega_n}) \cdot U(e^{j\omega_n})$

- $u$  periodic  $\rightarrow$  no transient in  $y$
- $N$  with many periods of  $u \rightarrow$  lower noise effect
- many samples per period  $\rightarrow$  higher resolution  $\hat{G}$

## Closed-loop configuration



$r_1$ : reference  
 $r_2$ : excitation  
 $r = r_2 + C r_1$  different from usual  
 $r = r_1, r_2 = 0$  !  
 $y = SGr + Sv$   
 $u = Sr - SCv$   
 $S = 1/(1 + CG)$  : sensitivity

## Quality of model estimation $\hat{G}(e^{j\omega})$

- Bias:  $\text{Bias}(\hat{G}) = G - E\{\hat{G}\}$
- Variance:  $\text{var}(\hat{G}) = E\{|\hat{G} - E\{\hat{G}\}|^2\}$
- Mean-square error:  $\text{MSE}(\hat{G}) = E\{|G - \hat{G}|^2\}$
- coherency spectrum:  $\hat{K}_{yy}(e^{j\omega_n}) = \sqrt{\hat{\Phi}_{yy}/\hat{\Phi}_y \hat{\Phi}_u}$

$\hat{G}$ : many estimations of plant (all different due to noise)  
 $E\{[\cdot]\}$ : expected value = mean  
 $\rightarrow \text{MSE}(\hat{G}) = \text{var}(\hat{G}) + \text{Bias}(\hat{G})^2$   
 $\forall \omega_n$ : if output per. from noise ( $\kappa=0$ ) or input ( $\kappa=1$ )

## Discrete Fourier Series (DFS)

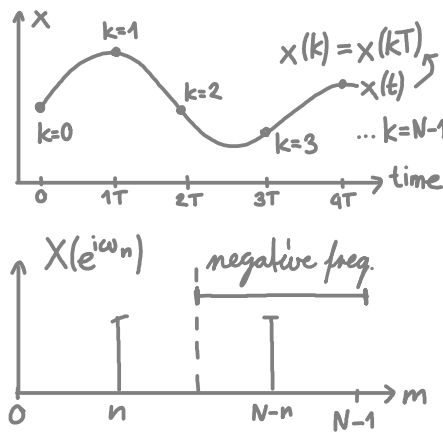
$$X(e^{i\omega_n}) = \sum_{k=0}^{N-1} x(k) e^{-i\omega_n k} \quad \left( \begin{array}{l} \omega_n = \omega_{\text{irl}} \cdot T \\ \omega_n = 2\pi n/N \end{array} \right)$$

neg. freq:  $X(e^{i\omega_n}) = X(e^{i\omega_{N-n}})$

amplitude:  $\begin{cases} n < \frac{N}{2}: a_n = \frac{2}{N} \cdot \|X(e^{i\omega_n})\| \\ n = \frac{N}{2}: a_{N/2} = \frac{1}{N} \cdot \|X(e^{i\omega_{N/2}})\| \end{cases}$

offset:  $n=0: a_0 = \frac{1}{N} \cdot \|X(e^{i\omega_0})\|$

phase:  $\varphi_n = \angle(X(e^{i\omega_n}))$



$T$ : sampling time  
 $x(t)$ : continuous signal  
 $x(k)$ : sampled signal  
 $k$ : sample # from 0 to  $M$   
 $N$ : # of samples  $k$  measured  
 $\tau_p$ : meas. time ( $\tau_p = TN$ )

eg.  $x(t) = \sin(4 \cdot 2\pi \cdot t)$   $N=120$   $T=1/30$   $\left\{ \begin{array}{l} \text{peak at} \\ \omega_n = 4 \cdot 2\pi \cdot \frac{1}{30} = 2\pi \frac{n}{120} \rightarrow n=16 \end{array} \right\}$

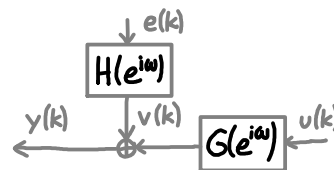
! DFS assumes signal repeats every  $N$ . If it does not (eg.  $x(k)=0 \forall k < 0$ ) transient error is introduced!

## Other Fourier function properties

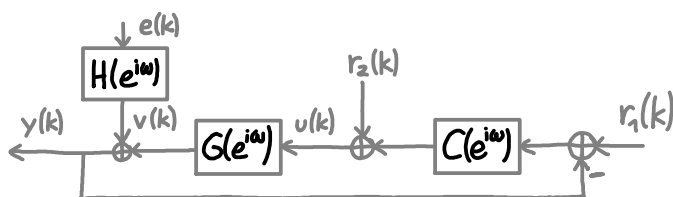
$$x(k), k=0, \dots, N-1 \rightarrow X(e^{i\omega_n}), \omega_n = \frac{2\pi n}{N}, n=0, \dots, N/2$$

	finite discrete (assumes periodic)	infinite discrete	continuous
transform:	$X(e^{i\omega_n}) = \sum_{k=0}^{N-1} x(k) e^{-i\omega_n k}$ $\frac{1}{N} \sum_{k=0}^{N-1} X(e^{i\omega_n}) e^{i\omega_n k}$	$\sum_{k=-\infty}^{\infty} x(k) e^{-i\omega k}$ $\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{i\omega}) e^{i\omega k} d\omega$	$\int_{-\infty}^{\infty} x(t) e^{-i\omega t} dt$ $\int_{-\infty}^{\infty} X(\omega) e^{i\omega t} d\omega$
autocorrelation:	$R_x(\tau) = \frac{1}{N} \sum_{k=0}^{N-1} x(k) \cdot x(k-\tau)$	$\sum_{k=-\infty}^{\infty} x(k) \cdot x(k-\tau)$	?
spectral dens.:	$\Phi_x(e^{i\omega_n}) = \frac{1}{N} \ X(e^{i\omega_n})\ ^2 = \mathcal{F}(R_x(\tau))$	$\ X(e^{i\omega})\ ^2 = \mathcal{F}(R_x(\tau))$	?
energy:	$E = \sum_{k=0}^{N-1}  x(k) ^2 = \sum_{n=0}^{N-1} \Phi_x(e^{i\omega_n})$	?	?
cross-correlation:	$R_{xy}(\tau) = \frac{1}{N} \sum_{k=0}^{N-1} x(k) \cdot y(k-\tau)$	?	?
cross-spectral:	$\Phi_{xy}(e^{i\omega_n}) = \frac{1}{N} X(e^{i\omega_n}) \cdot Y^*(e^{i\omega_n}) = \mathcal{F}(R_{xy}(\tau))$	?	?

$$\begin{aligned} R_x(-\tau) &= R_x(\tau) & \Phi_x(e^{i\omega}) &\in \mathbb{R} \\ R_x(0) &\geq R_x(\tau) \quad \forall \tau > 0 & \Phi_x(e^{i\omega}) &\geq 0 \quad \forall \omega \\ \Phi_{xy}(e^{i\omega}) &= \Phi_{yx}(e^{i\omega}) & \Phi_x(e^{i\omega}) &= \Phi_y(e^{-i\omega}) \end{aligned}$$



$$\begin{aligned} \Phi_y &= |G|^2 \cdot \Phi_v \\ \Phi_{yv} &= G \cdot \Phi_v \\ \Phi_v &= \Phi_y - |\Phi_y|^2 / \Phi_v \end{aligned}$$



$$\begin{aligned} r &= r_2 + C r_1 \\ S &= 1/(1+CG) \\ T &= CG/(1+CG) \\ L &= CG \end{aligned}$$

$$\begin{aligned} \Phi_{yv} &= |S|^2 G \Phi_r - |S|^2 C^* \Phi_v \\ \Phi_v &= |S|^2 \Phi_r + |S|^2 |C|^2 \Phi_v \end{aligned}$$

unusual  $r$ ! with  $r=r_1, r_2=0$ :

$$\begin{aligned} y &= SGr + Sv \\ u &= Sr - SCv \end{aligned}$$

$$\begin{aligned} y &= Tr + Sv \\ u &= SCr - SCv \end{aligned}$$

## Input signals

### steps/doublets:

$$u(k) = \begin{cases} k \geq 0 : A \\ \text{else} : 0 \end{cases} \rightarrow U(e^{j\omega}) = 1/j\omega$$

$$u(k) = \begin{cases} 0 \leq k \leq \varepsilon/2 : A \\ \varepsilon/2 \leq k \leq \varepsilon : -A \\ \text{else} : 0 \end{cases} \rightarrow U(e^{j\omega}) = \dots$$

### sinusoids:

$$u_n(k) = \sin(\omega_n k) \rightarrow U(e^{j\omega}) = \text{peak at } \omega_n$$

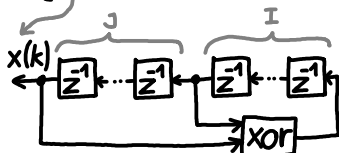
### filtered white noise:

$$U(e^{j\omega}) = L(e^{j\omega}) \cdot e$$

filter  $\uparrow$  white noise

### PRBS: (periodic with $2^{I+J}-1$ )

$$u(k) = \begin{cases} x(k)=1 : +A \\ x(k)=0 : -A \end{cases} \rightarrow |U(e^{j\omega})| \approx \text{const.}$$



I+J	J
3	1 or 2
4	1 or 3
5	2 or 3
6	1 or 5
7	1 or 6
8	—
9	4 or 5
⋮	⋮

### multi-sinusoidal

$$u(k) = \sum_{s=1}^S \sqrt{2\alpha_s} \cos(\omega_s kT + \phi_s) \quad (S < N/2L)$$

$$\sum_{s=1}^S \alpha_s = 1, \quad \omega_s = \frac{2\pi l_s}{N}, \quad l_s \in \mathbb{N}, \quad \phi_s = 2\pi \sum_{n=1}^s n \alpha_n$$

sig. per = 1      harmonic frequencies      minimize peak  $u(k)$

## Improve model estimation

### averaging: divide experiment data into R parts and estimate $\hat{G}_r(e^{j\omega_n})$ of each $\rightarrow$ average to $\tilde{G}(e^{j\omega_n})$

$$\tilde{G}(e^{j\omega_n}) = \sum_{r=1}^R \alpha_r \hat{G}_r(e^{j\omega_n}) \quad (\text{average: } \alpha_r = 1/R \quad \text{min variance: } \alpha_r(e^{j\omega_n}) = |U_r(e^{j\omega_n})|^2 / \sum_{r=1}^R |U_r(e^{j\omega_n})|^2)$$

$\rightarrow R \uparrow \Rightarrow \text{var}(\tilde{G}) \downarrow, \text{bias}(\tilde{G}) \uparrow \Rightarrow \text{tradeoff!}$  find R with lowest MSE( $\tilde{G}$ ) (with ref. dataset!)  
 $\uparrow$  (due to transient in each r. not the case if u periodic)

### smoothing:

$$\tilde{G}(e^{j\omega_n}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} W_y(e^{j(\xi-\omega_n)}) \frac{1}{N} |U_n(e^{j\xi})|^2 \hat{G}(e^{j\xi}) d\xi \quad / \quad \frac{1}{2\pi} \int_{-\pi}^{\pi} W_y(e^{j(\xi-\omega_n)}) \frac{1}{N} |U(e^{j\xi})|^2 d\xi$$

$$\approx \sum_{k=0}^N W_y(e^{j\omega_{k-n}}) \cdot |U(e^{j\omega_k})|^2 \cdot \hat{G}(e^{j\omega_k}) / \sum_{k=0}^N W_y(e^{j\omega_{k-n}}) \cdot |U(e^{j\omega_k})|^2$$

$$\approx \sum_{\tau=-\gamma}^{\gamma} w_y(\tau) \hat{R}_y(\tau) e^{-j\tau\omega_n} \quad / \quad \sum_{\tau=-\gamma}^{\gamma} w_y(\tau) \hat{R}_y(\tau) e^{-j\tau\omega_n}$$

$W_y(e^{j\omega})$ : FD window function

$W_y(e^{j\omega_n})$ : DFD window function

$w_y(\tau)$ : TD window function

$\rightarrow \gamma \downarrow \Rightarrow \text{var}(\tilde{G}) \downarrow, \text{bias}(\tilde{G}) \uparrow \Rightarrow \text{tradeoff!}$  find  $\gamma$  with lowest MSE( $\tilde{G}$ ) (with ref. dataset!)

e.g. Bartlett window:  $w_y(\tau) = 1 - |\tau|/\gamma, -\gamma \leq \tau \leq \gamma \leftrightarrow W_y(e^{j\omega}) = \frac{1}{\gamma} \left( \frac{\sin \gamma \omega/2}{\sin \omega/2} \right)^2$  or Hann, Hamming, Welch, ...

### TD windowing:

$$U_w(e^{j\omega_n}) = \sum_{k=0}^{N-1} w_{\text{data}}(k) \cdot u(k) \cdot e^{-jk\omega_n} \quad (w_{\text{data}}(k) = w_y(k - N/2) \big|_{\gamma=N/2} : \text{some window over whole } N \text{ centered at } N/2)$$

$$E_{\text{scL}} = \sum_{k=0}^{N-1} |w_{\text{data}}(k) u(k)|^2 / \sum_{k=0}^{N-1} |u(k)|^2 \quad (\text{scale factor to preserve pow. in periodogram}) \quad \Phi_{u_w} = \frac{1}{E_{\text{scL}}} \cdot \frac{1}{N} \cdot |U_w(e^{j\omega_n})|^2$$

Welch's method:  $\tilde{\Phi}_0(e^{j\omega_n}) = \frac{1}{NL} \sum_{l=1}^L \frac{1}{E_{\text{scL}}} |U_l(e^{j\omega_n})|^2$

$U_l = U_w$  for  $u = U_l$  with  $l = 1 \dots L$

$\rightarrow$  advantages: reduce transient effect       $\rightarrow$  disadvantages: "smears" frequencies

### TD detrending: $u_d(k) = u(k) - (\alpha k + \beta)$ ( $\alpha k + \beta$ : best linear fit) (only do this when drift/offset unwanted)

## Frequency-Domain open-loop identification

- Sinusoidal correlation: (sine test signal with  $\omega_0 \rightarrow$  measure output phase+amplitude  $\rightarrow$  reconstruct  $\hat{G}$ )

input:  $u(k) = \alpha \cos(\omega_0 k)$  ( $k=0, \dots, N-1$ )

output:  $y(k) = \alpha \underbrace{|G(e^{i\omega_0})|}_{\text{unknowns}} \cdot \underbrace{\cos(\omega_0 k + \angle(G(e^{i\omega_0}))}_{\text{unknowns}} + \underbrace{\text{noise } v(k)}_{\text{disturbances}} + \underbrace{\text{transient}}_{\text{disturbances}}$

- record many periods
- start when transient gone

correlation functions

$$I_c = \frac{1}{N} \sum_{k=0}^{N-1} y(k) \cos(\omega_0 k) \Rightarrow \begin{matrix} \text{expected val. for } N \rightarrow \infty \\ E\{I_c\} = \frac{\alpha}{2} |G(e^{i\omega_0})| \cos(\angle G(e^{i\omega_0})) \end{matrix}, \begin{matrix} \text{exp. variance for } N \rightarrow \infty \\ \text{var}\{I_c\} = 0 \end{matrix}$$

$$I_s = \frac{1}{N} \sum_{k=0}^{N-1} y(k) \sin(\omega_0 k) \Rightarrow \begin{matrix} \text{expected val. for } N \rightarrow \infty \\ E\{I_s\} = \frac{-\alpha}{2} |G(e^{i\omega_0})| \sin(\angle G(e^{i\omega_0})) \end{matrix}, \begin{matrix} \text{exp. variance for } N \rightarrow \infty \\ \text{var}\{I_s\} = 0 \end{matrix}$$

$$\hookrightarrow \hat{G}(e^{i\omega_0}) = \frac{I_c^2 + I_s^2}{\alpha/2}, \quad |G(e^{i\omega_0})| = \frac{\sqrt{I_c^2 + I_s^2}}{\alpha/2}, \quad \angle G(e^{i\omega_0}) = -\arctan(I_s/I_c)$$

- Empirical transfer function estimation (ETFE):

input  $u(k)$ : anything  $\rightarrow$  fourier:  $U(e^{i\omega_n})$   
output  $y(k)$ : measured  $\rightarrow$  fourier:  $Y(e^{i\omega_n})$

$$\hat{G}(e^{i\omega_n}) = Y(e^{i\omega_n})/U(e^{i\omega_n})$$

(real:  $Y/U = G + V/U$ )

best results if  $\rightarrow$  high signal/noise ratio ( $U \gg V$ )  $\Rightarrow u(k)$  as large as possible  
 $\rightarrow$  no transient/periodic signal  $U$   $\Rightarrow$  take periodic  $u(k)$  and record multiple periods!  
 $\Rightarrow$  ignore first couple of periods

- Spectral estimation:

opt. 1  $\left\{ \begin{array}{l} \text{input } u(k): \text{ anything} \rightarrow \text{fourier: } U(e^{i\omega_n}) \\ \text{output } y(k): \text{ measured} \rightarrow \text{fourier: } Y(e^{i\omega_n}) \end{array} \right\} \begin{array}{l} \Phi_{uu}(e^{i\omega_n}) = \frac{1}{N} |U(e^{i\omega_n})|^2 \\ \Phi_{uy}(e^{i\omega_n}) = \frac{1}{N} U(e^{i\omega_n}) \cdot Y^*(e^{i\omega_n}) \end{array} \Rightarrow \hat{G}(e^{i\omega_n}) = \Phi_{yu}(e^{i\omega_n})/\Phi_{uu}(e^{i\omega_n})$

opt. 2  $\left\{ \begin{array}{l} \text{input } u(k): \text{ anything} \rightarrow R_{uu}(\tau) = \frac{1}{N} \sum_{k=0}^{N-1} u(k)u(k-\tau) \rightarrow \text{fourier: } \Phi_{uu}(e^{i\omega_n}) \\ \text{output } y(k): \text{ measured} \rightarrow R_{uy}(\tau) = \frac{1}{N} \sum_{k=0}^{N-1} u(k)y(k-\tau) \rightarrow \text{fourier: } \Phi_{uy}(e^{i\omega_n}) \end{array} \right\} \hat{G}(e^{i\omega_n}) = \Phi_{yu}(e^{i\omega_n})/\Phi_{uu}(e^{i\omega_n})$

! same best result tips as for ETFE!

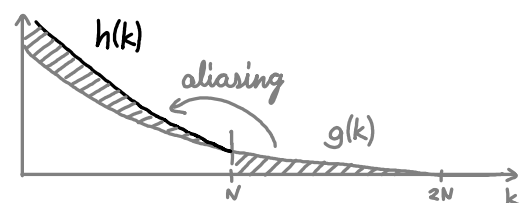
$\uparrow$  all formula without assumption of periodicity?

FD subspace identification (get state-space  $\hat{A}, \hat{B}, \hat{C}, \hat{D}$  from  $\hat{G}(e^{i\omega_n}) = \hat{G}(n)$ )

$$\begin{array}{l} \hat{G}(e^{i\omega_n}) \text{ (e.g. from ETFE)} \\ \downarrow \\ \begin{array}{l} x(k+1) = \hat{A}x(k) + \hat{B}u(k) \\ y(k) = \hat{C}x(k) + \hat{D}u(k) \end{array} \end{array} \quad \begin{array}{l} \text{impulse resp: } g(k) = \begin{cases} 0 & k < 0 \\ \hat{D} & k = 0 \\ \hat{C}\hat{A}^{k-1}\hat{B} & k > 0 \end{cases} = \mathcal{F}^{-1}(\hat{G}(e^{i\omega})) \\ \text{aliased imp. resp: } h(k) = \hat{C}\hat{A}^{k-1}(\sum_{l=0}^N \hat{A}^{N-l})\hat{B} = \mathcal{F}^{-1}(\hat{G}(e^{i\omega_n})) \approx \mathcal{F}^{-1}(\hat{G}(e^{i\omega})) \end{array}$$

non trivial, because  $\mathcal{F}^{-1}(\hat{G}(e^{i\omega_n})) \neq g(k)$  impulse response. actually aliased impulse resp.  $h(k)$

$\hookrightarrow$  solution: subspace identification algorithm (lecture 7)  
(+ alternative for when  $\omega_n$  non-uniformly spaced)



## Frequency-Domain close-loop identification (estimate $\hat{G}$ while it is being controlled by $C$ )

usefull when: plant unstable or need to stay at operating point.

**! watch out,  $r \neq$  usual  $r$  !**

• direct method: (like for open-loop ETFE and spectral estimate)

(worked well in open-loop because  $\Phi_{uv}=0$ , but in closed loop  $\Phi_{uv} \neq 0$  so worse estimate!)

$$\hat{G}(e^{i\omega_n}) \approx Y(e^{i\omega_n})/U(e^{i\omega_n}) = \frac{SGr + Sv}{Sr + Scv} \quad \hat{G}(e^{i\omega_n}) \approx \Phi_{yu}(e^{i\omega_n})/\Phi_{uu}(e^{i\omega_n}) = \frac{|S|^2 G \Phi_r - |S|^2 C^* \Phi_v}{|S|^2 \Phi_r + |S|^2 C^* \Phi_v} \quad (S = 1/(1+CG))$$

only works if:  $S$  not too low (if  $C$  is too good it "hides"  $G$ ) ;  $r \gg v$  resp.  $\Phi_r \gg \Phi_v$  (more signal than noise)

• indirect method:  $\frac{\hat{Y}(e^{i\omega_n})}{\hat{R}(e^{i\omega_n})} = \hat{T}_{yr}(e^{i\omega_n}) = \frac{\hat{G}}{1+CG} \rightarrow \hat{G}(e^{i\omega_n}) = \frac{\hat{T}_{yr}(e^{i\omega_n})}{1 - \hat{T}_{yr}(e^{i\omega_n})C(e^{i\omega_n})}$

(why better?)

• input-output method:  $\frac{\hat{Y}(e^{i\omega_n})}{\hat{R}(e^{i\omega_n})} = \hat{T}_{yr}(e^{i\omega_n}) = \frac{\hat{G}}{1+CG}$  ;  $\frac{\hat{U}(e^{i\omega_n})}{\hat{R}(e^{i\omega_n})} = \hat{T}_{ur}(e^{i\omega_n}) = \frac{1}{1+CG} \rightarrow \hat{G}(e^{i\omega_n}) = \frac{\hat{T}_{yr}(e^{i\omega_n})}{\hat{T}_{ur}(e^{i\omega_n})}$

(why better?)

• Youla method: in the above methods a  $\hat{G}$  can result that is not stabilized by  $C$ ! (bad, since real  $G$  is)  
 $\rightarrow$  use dual-Youla method to force  $\hat{G}$  to be stabilized by  $C$ . (lecture 8)

## Time-Domain open-loop identification

choose model structure + parameter  $\theta$

pulse response:  $g(k)$

$$\theta = (g(0), g(1), \dots)$$

transfer function:  $G(z) = \frac{b_1 z^{-1} + b_2 z^{-2} + \dots + b_m z^{-m}}{1 + a_1 z^{-1} + \dots + a_n z^{-n}}$

$$\theta = (a_1, \dots, a_n, b_1, \dots, b_m)$$

state space:  $x(k+1) = Ax(k) + Bu(k)$   
 $y(k) = Cx(k) + Du(k)$

$$\theta = (A_{ij}, B_{ij}, C_{ij}, D_{ij})$$

choose cost function  $J(\theta)$  or any other norm

residual error:  $J(\theta) = \|y - Gu\|_2^2, \|y - Gu\|_\infty, \|y - Gu\|_1$

parametric error:  $J(\theta) = \|\theta - \theta_0\|$  ( $\theta_0$ : ??)

prediction error:  $J(\theta) = |y(k+1) - \hat{y}(k+1)|^2$  prediction  
 $\uparrow$  meas. outside experiment

$\rightarrow$  solve optimization problem: minimize  $J(\theta) \hat{=} \hat{\theta} = \argmin(J(\theta))$  (usually numerical solver)

• correlation-based method (identify first  $M$  entries of  $g(k)$  with  $N$  measurements of  $u(k), y(k)$  and norm2 residual error)

$$R_{yu}(\tau) = g(k) * R_u(\tau) \quad (\text{accurate if } u, y: \text{periodic + noise free + } g(k) = 0 \forall k \geq M)$$

$$\begin{pmatrix} \hat{R}_{yu}(0) \\ \vdots \\ \hat{R}_{yu}(N-1) \end{pmatrix} = \begin{pmatrix} \hat{R}_u(0) & \dots & \hat{R}_u(-(M-1)) \\ \vdots & & \vdots \\ \hat{R}_u(N-1) & \dots & \hat{R}_u(?) \end{pmatrix} \cdot \begin{pmatrix} \hat{g}(0) \\ \vdots \\ \hat{g}(M-1) \end{pmatrix} \rightarrow \hat{g} = \hat{R}^{\pm} \cdot \hat{R}_{yu}$$

pseudoinverse: minimize  $\|\hat{R}\hat{g} - \hat{R}_{yu}\|_2$

only works if  $u(k)$  is: persistently exciting of order  $M$

• linear regression method (identify  $G(z)$  with prediction error)

$$y(k) = G(z)u(k) = -a_1 y(k-1) - \dots - a_n y(k-n) + b_1 u(k-1) + \dots + b_m u(k-m)$$

$$= \varphi(k)^T \cdot \theta = [-y(k-1), \dots, -y(k-n), u(k-1), \dots, u(k-m)] \cdot [a_1, \dots, a_n, b_1, \dots, b_m]^T$$

$$\rightarrow \begin{pmatrix} y(0) \\ \vdots \\ y(N-1) \end{pmatrix} = \begin{pmatrix} \varphi(0)^T \\ \vdots \\ \varphi(N-1)^T \end{pmatrix} \cdot \hat{\theta} \rightarrow \hat{\theta} = \Phi^{\pm} Y$$

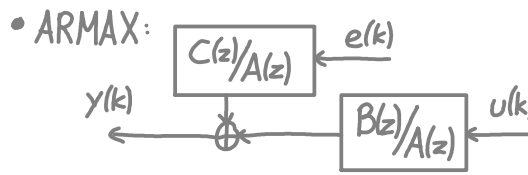
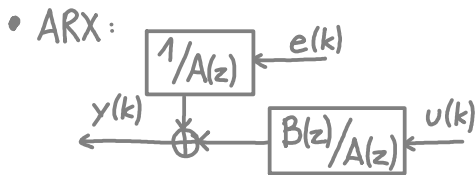
pseudoinverse: minimize  $\|\Phi\hat{\theta} - Y\|_2$

only works if  $u(k)$  is: persistently exciting of order  $n+m$  and  $y, u$  noise free...

general  $u(k)$ :  $\Phi_u(e^{i\omega_n}) \neq 0$  for at least  $M$  frequencies  $\omega_n$ .  
 step. func.:  $M \leq 1$  | PRBS:  $M \leq \text{period of PRBS}$   
 multi-sinusoids:  $M \leq \# \text{ of sin/cos with } 0 < \omega_s < \pi$

$M=n+m$   
here

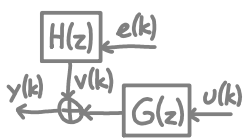
$\rightarrow$  no bias only if ARX model ; low variance in  $\theta$  with long measurement



ARMAX can be made to ARX if  $C(z)$  known + invertible with  $u_F = C^{-1}u$ ,  $y_F = C^{-1}y$

• pseudor linear regression + noise prediction (identify  $G(z)$  with known  $H(z)$  and error prediction)

Noise prediction: (predict  $y(k)$  when  $y, u$  for  $0..(k-1)$  is known and  $G(z)$  known and  $H(z)$  known)



$G(z)$ : stable  
 $H(z), H_{inv}(z)$ : stable  
 $e(k)$ : white noise

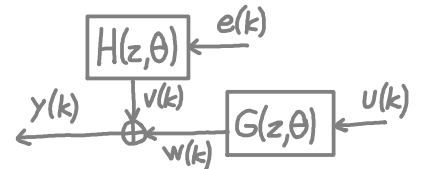
$$v(k) = \sum_{i=0}^k h(i)e(k-i) = \underbrace{e(k)}_{\text{noise}} + \underbrace{\sum_{i=1}^k h(i)e(k-i)}_{\text{observable } m(k-1)} \quad (H(z): \text{monic} \hat{=} h(0)=1)$$

if  $\text{mean}(e) \hat{=} \text{not probab. } e \rightarrow \hat{v}(k|k-1) = m(k-1)$

$$\hat{v}(k|k-1) = m(k-1) = \dots = (1 - H_{inv}(z))v(k) - \sum_{i=1}^k h_{inv}(i)v(k-i) \quad (H_{inv}(z) = 1/H(z))$$

$$\hat{y}(k|k-1) = G(z)u(k) + \hat{v}(k|k-1) = \dots = H_{inv}(z)G(z)u(k) + (1 - H_{inv}(z))y(k)$$

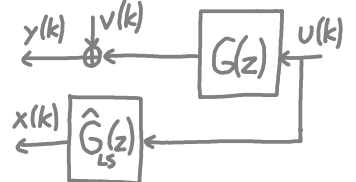
- 1) pick representation of  $\hat{G}(z, \theta), \hat{H}(z, \theta)$  ( $H(z, \theta)$  must be monic  $\hat{=} h(0)=1$ )
- 2) choose initial condition  $\theta = \theta_0$
- 3)  $\hat{w}(k) = \hat{G}(z) \cdot u(k) \rightarrow \hat{e}(k) = 1/\hat{H}(z) \cdot (y(k) - \hat{w}(k))$
- 4)  $\hat{y}(k|k-1) = \hat{w}(k) + (\hat{H}(z) - 1) \cdot \hat{e}(k)$  (since  $e(k) = y(k) - y(k|k-1)$ )
- 5) find  $\theta$  that minimizes  $\|\hat{e}(k)\|_2 \leftarrow$  (because  $H(z)$  is monic)



no bias if real  $H, \hat{H}$  stable  
 lower variance in  $\theta$  if measurements long

• instrumental variable method: (linear regression, avoiding bias)

- 1) estimate  $G(z)$  with linear regression as if  $v(k)=0 \forall k \rightarrow \hat{G}_{LS}(z), \theta_{LS}, \Phi$
- 2) calculate  $x(k) = \hat{G}_{LS}(z) \cdot u(k)$
- 3) build  $Z = \begin{pmatrix} \Phi(0)^T \\ \vdots \\ \Phi(N-1)^T \end{pmatrix}$  analogous to  $\Phi = \begin{pmatrix} \varphi(0)^T \\ \vdots \\ \varphi(N-1)^T \end{pmatrix}$  but with  $y(k)$  replaced by  $x(k)$
- 4) get  $\hat{\theta}_{iv} = (\hat{N} \sum_{k=0}^{N-1} S(k) \varphi^T(k))^{-1} \cdot \hat{N} \sum_{k=0}^{N-1} S(k) y(k)$



improves linear regression to be unbiased with non ARX model

• maximum likelihood method: (identify  $G(z)$  if  $v(k)$  statistics are known)

- 1) write down formula for  $Z_\theta$   
 $z(k, \theta) = y(k) - G(z, \theta) \cdot u(k)$   
 $Z_\theta = [z(1, \theta) \dots z(N, \theta)]^T$
- 2) calculate statistics of  $Z_\theta$  from  $v(k)$  ( $v$ : mean vector,  $\Sigma$ : covariance matrix)  
 $z(k, \theta) = v(k) = H(z)e(k)$  (if known)  $\rightarrow$  e.g.  $v \sim \mathcal{N}(\mu, \lambda) \rightarrow v = \mu \cdot \text{ones}(R^{N \times 1})$   
 $Z_\theta \sim \mathcal{N}(v, \Sigma)$   
 $\Sigma = \lambda \cdot \text{diag}(R^{N \times N})$

3) write down likelihood formula

$$f(\theta) = \frac{1}{(2\pi)^{N/2} \sqrt{\det \Sigma}} \cdot e^{-\frac{1}{2} (Z_\theta - v)^T \Sigma^{-1} (Z_\theta - v)}$$

4) calculate maximum likelihood  $\hat{\theta}$

$$\hat{\theta} = \arg \max_{\theta} f(\theta) = \arg \min_{\theta} (Z_\theta - v)^T \Sigma^{-1} (Z_\theta - v) \leftarrow \text{(could be brought to closed form)}$$

- with a priori knowledge about  $\theta \sim \mathcal{N}(\mu_\theta, \lambda_\theta)$ :  $f_{\text{new}}(\theta) = f_{\text{old}}(\theta) \cdot \frac{1}{\lambda_\theta \sqrt{2\pi}} \cdot e^{-\frac{1}{2} \left( \frac{\theta - \mu_\theta}{\lambda_\theta} \right)^2}$

TD subspace identification (get state-space  $\hat{A}, \hat{B}, \hat{C}, \hat{D}$  from  $\hat{G}(z)$ )

similar concept as for FD. (lecture 13)