

# Lab-2: DC Motor Control – PD Position Control

## Topics Covered

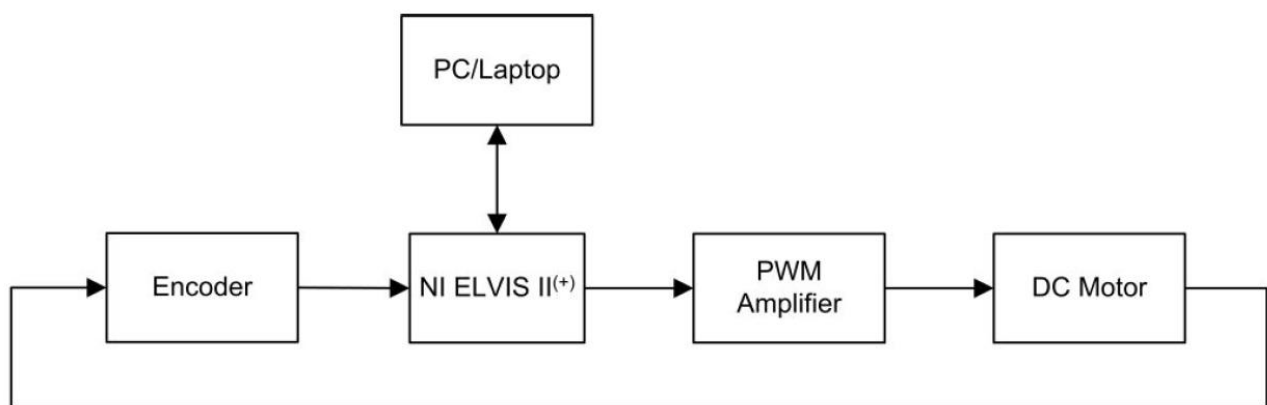
- QNET DC Motor position control.
- Proportional-derivative (PD) compensator design.
- Designing control according to specifications.

## Prerequisites

- QNET DC Motor Qualitative PD Position Control laboratory experiment.
- Modeling and Qualitative PI Speed Control

### Necessary Equipment:

- Labview software
- NI ELVIS II+ board
- Quanser DC Motor
- 1 optical encoder
- Power module



# 1 Background

Control of motor position is a natural way to introduce the benefits of derivative action. In this experiment a proportional-integral-derivative controller is designed according to specifications. The closed-loop PID control block diagram is shown in Figure 1.1.

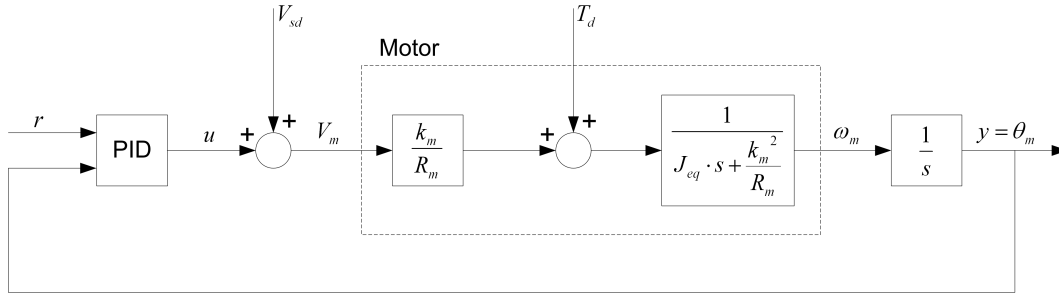


Figure 1.1: DC Motor PID closed-loop block diagram

The two-degree of freedom PID transfer function inside the PID block in Figure 1.1 is

$$u = k_p (b_{sp} r(t) - y(t)) + k_i \int_0^t (r(\tau) - y(\tau)) d\tau + k_d (b_{sd} \dot{r}(t) - \dot{y}(t)), \quad (1.1)$$

where  $k_p$  is the position proportional control gain,  $k_d$  is the derivative control gain,  $k_i$  is the integral control gain,  $b_{sp}$  is the set-point weight on the reference position  $r(t)$ , and  $b_{sd}$  is the set-point weight on the velocity reference of  $r(t)$ .

The dotted box labeled Motor in Figure 1.1 is the motor model in terms of the back-emf motor constant  $k_m$ , the electrical motor armature resistance  $R_m$ , and the equivalent moment of inertia of the motor pivot  $J_{eq}$ . The direct disturbance applied to the inertial wheel is represented by the disturbance torque variable  $T_d$  and the simulated disturbance voltage is denoted by the variable  $V_{sd}$ .

## 1.1 PD Control Design

The behavior of the controlling the motor position is first analyzed using a PD control. By setting  $k_i = 0$  in the PID control equation Equation 1.1 and taking its Laplace transform, the PD transfer function is

$$u = k_p (r - y) + k_d s (b_{sd} r - y). \quad (1.2)$$

Combining the position process model

$$\frac{\Theta_m(s)}{V_m(s)} = \frac{K}{s(\tau s + 1)} \quad (1.3)$$

with the PD control Equation 1.2 gives the closed-loop transfer function of the motor position system

$$G_{\theta,r}(s) = \frac{K(k_p + b_{sd} k_d s)}{\tau s^2 + (1 + K k_d)s + K k_p}. \quad (1.4)$$

Similarly to PI speed control, the standard characteristic equation  $s^2 + 2s\zeta\omega_n + \omega_n^2$  can be achieved by setting the proportional gain to

$$k_p = \frac{\omega_0^2 \tau}{K}, \quad (1.5)$$

and the derivative gain to

$$k_d = \frac{-1 + 2\zeta\omega_0\tau}{K}. \quad (1.6)$$

## 1.2 Peak Time and Overshoot

The standard second-order transfer function has the form

$$\frac{Y(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}, \quad (1.7)$$

where  $\omega_n$  is the natural undamped frequency and  $\zeta$  is the damping ratio. The properties of its response depend on the values of the  $\omega_n$  and  $\zeta$  parameters. Consider when a second-order system, as shown in Equation 1.7, is subjected to a step input given by

$$R(s) = \frac{R_0}{s} \quad (1.8)$$

with a step amplitude of  $R_0 = 1.5$ . The system response to this input is shown in Figure 1.2, where the red trace is the response (output)  $y(t)$  and the blue trace is the step input  $r(t)$ .

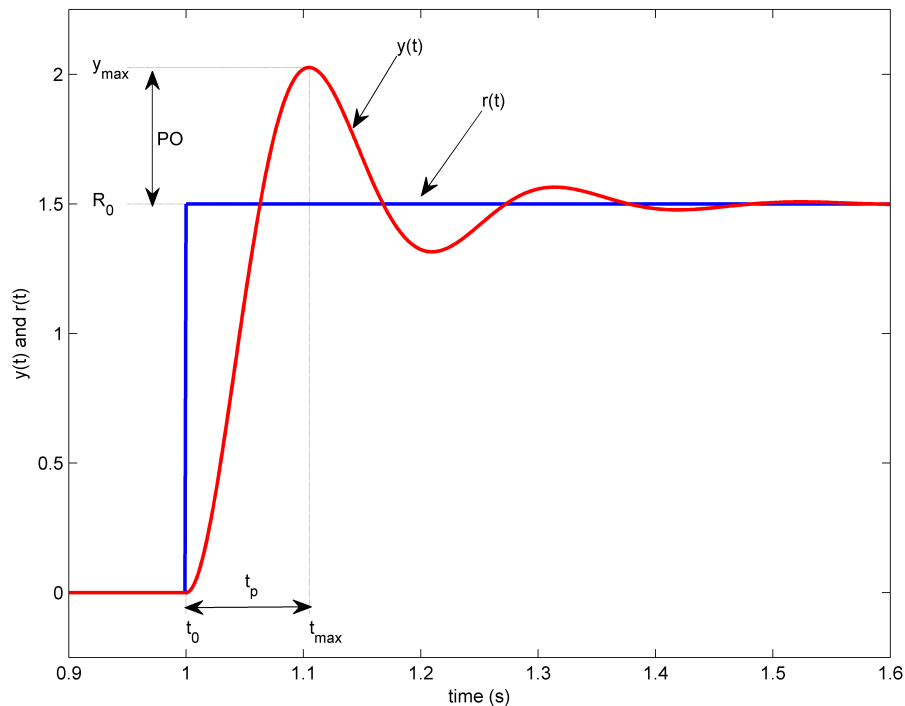


Figure 1.2: Standard second-order step response.

The maximum value of the response is denoted by the variable  $y_{max}$  and it occurs at a time  $t_{max}$ . For a response similar to Figure 1.2, the percent overshoot is found using

$$PO = \frac{100(y_{max} - R_0)}{R_0}. \quad (1.9)$$

In a second-order system, the amount of overshoot depends solely on the damping ratio parameter and it can be calculated using the equation

$$PO = 100e^{\left(-\frac{\pi\zeta}{\sqrt{1-\zeta^2}}\right)}. \quad (1.10)$$

From the initial step time  $t_0$ , the time it takes for the response to reach its maximum value is

$$t_p = t_{max} - t_0. \quad (1.11)$$

This is called the peak time of the system and it depends on both the damping ratio and natural frequency of the system. It can be derived analytically as

$$t_p = \frac{\pi}{\omega_n \sqrt{1 - \zeta^2}}. \quad (1.12)$$

Generally speaking, the damping ratio affects the shape of the response while the natural frequency affects the speed of the response.

## 1.3 Position Control Virtual Instrument

Tracking a reference position square wave using PID control is first examined in this laboratory. Then, disturbance effects using PD and PID are studied through direct manual interaction or a simulated using a control switch in the VI. The LabVIEW™ virtual instrument for position control is shown in Figure 1.3.

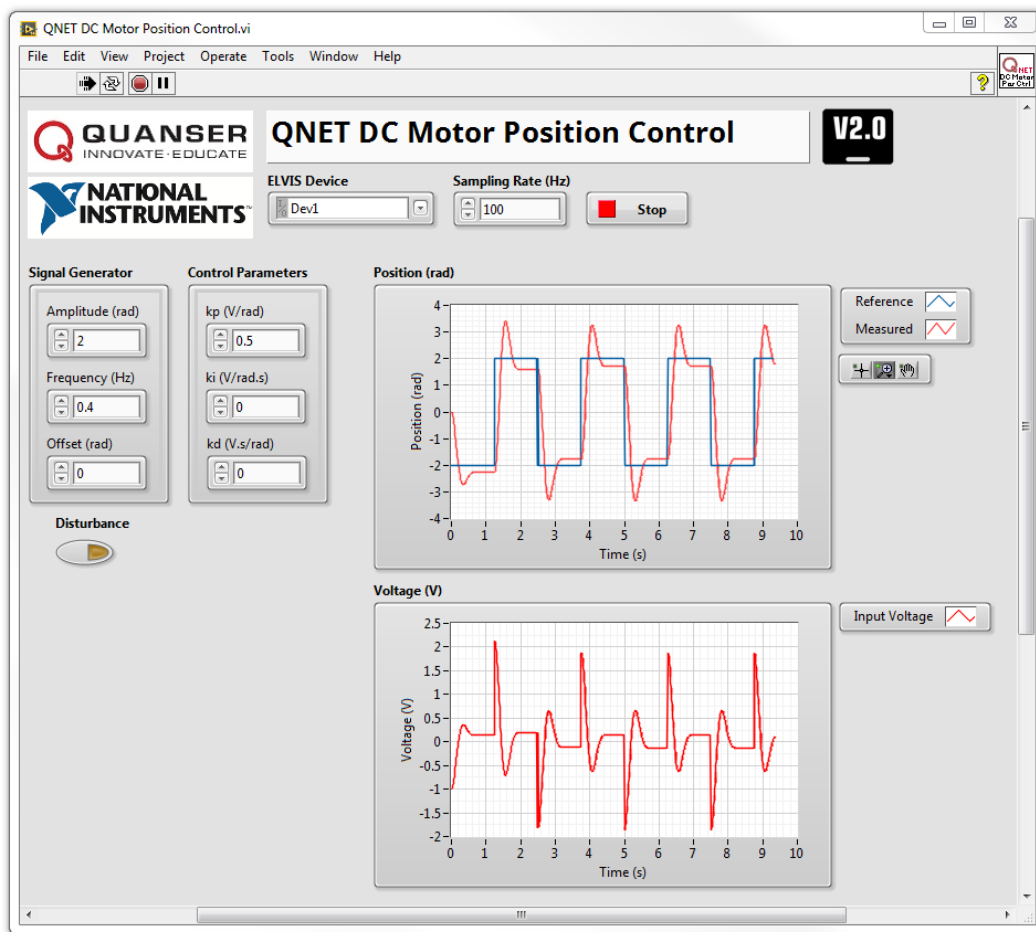


Figure 1.3: Virtual instrument for DC motor position control

See Wikipedia for more information on [motion control](#), [control theory](#) and [PID](#).

## 2 In-Lab Exercise

**R** 1. Calculate the expected peak time  $t_p$  and percentage overshoot  $PO$  given

- $\zeta = 0.60$ ,
- $\omega_0 = 25.0$  rad/s.

**Optional:** You can also design a VI that simulates the DC motor first-order model with a PD control and have it calculate the peak time and overshoot.

**R** 2. Assuming the model steady-state gain is  $K = 26$  V/rad and time constant is  $\tau = 0.145$  s, calculate the proportional and derivative control gains  $k_p$  and  $k_d$ , respectively, to meet the specifications above.

3. Open the QNET DC Motor Position Control.vi. **Make sure the correct Device is chosen.**

4. Run the VI. You should see the DC motor rotating back and forth.

5. In the *Signal Generator* section set:

- Amplitude (rad) = 0.50
- Frequency (Hz) = 0.40
- Offset (rad) = 0.00

6. In the Control Parameters section, set the PD gains to the values found in Step 2. The PD controller is implemented with  $b_{sd} = 0$ .

**R** 7. Capture the position response found in the Position (rad) scope and control signal used in the Voltage (V) scope.

**R** 8. Measure the peak time and percentage overshoot of the measured position response. Are the specifications satisfied?

**R** 9. What effect does changing the specification zeta have on the measured position response and the generated control gains?

**R** 10. What effect does changing the specification  $\omega_0$  have on the measured position response and the generated control gains?

11. Stop the VI by clicking on the Stop button.

# 3 Lab Report

## I. PROCEDURE

### 1. PD Control According to Specifications

- Briefly describe the main goal of the experiment.
- Briefly describe the experimental procedure in Step 7 in Section 2.
- Effect of changing damping ratio and natural frequency specifications in Step 9-10 in Section 2.

## II. RESULTS

Do not interpret or analyze the data in this section. Just provide the results.

1. Position control response plot from step 7 in Section 2.
2. PD disturbance response plot.
3. Provide applicable data collected in this laboratory from Table 1.

Table 1: DC motor Position Control results summary

Description	Symbol	Value	Unit
<b>Section 2: PD Control Design</b>			
Model gain used	$K$		rad/s
Model time constant used	$\tau$		s
Proportional gain	$k_p$		V/(rad/s)
Integral gain	$k_i$		V/rad
Measured peak time	$t_p$		s
Measured percent overshoot	$M_p$		%

## III. ANALYSIS

Provide details of your calculations (methods used) for analysis for each of the following:

1. Provide position control analysis with steady-state error.
2. Peak time and percent overshoot of speed control response in Step 8 in Section 2.

## IV. CONCLUSIONS

Interpret your results to arrive at logical conclusions for the following:

1. Whether the speed controller meets the specifications in Step 8 in Section 4.4.
2. Does the measured steady-state error using a PD control match what is expected.
3. Does the measured steady-state error using a PD control match what is expected when there is a disturbance