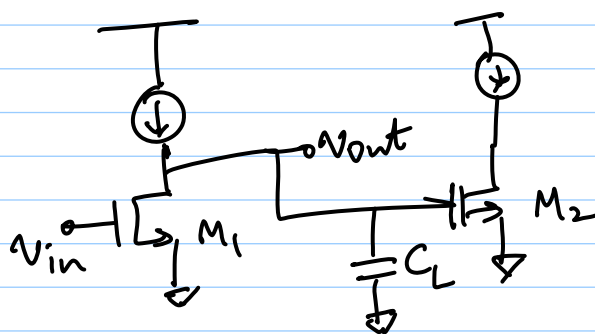


## Lecture 34 : Wideband Amplifiers - I

- Applications :
- PCB chip-to-chip links
  - Optical fibre communications
  - wideband wireline comm. (e.g. TV)
  - Measurement instrument front-ends (e.g. Oscilloscopes)
  - Wireless : UWB radios & multiband radios

### A BW product

\* C.S. amplifier driving identical stage



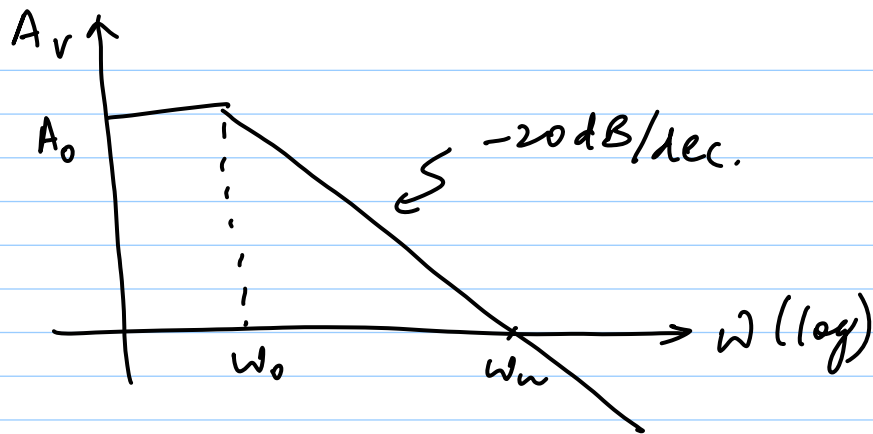
$$\frac{V_{out}}{V_{in}}(s) = A_V(s)$$

$$A_V(s) = \frac{g_m r_{ds}}{1 + s C_L r_{ds}}$$

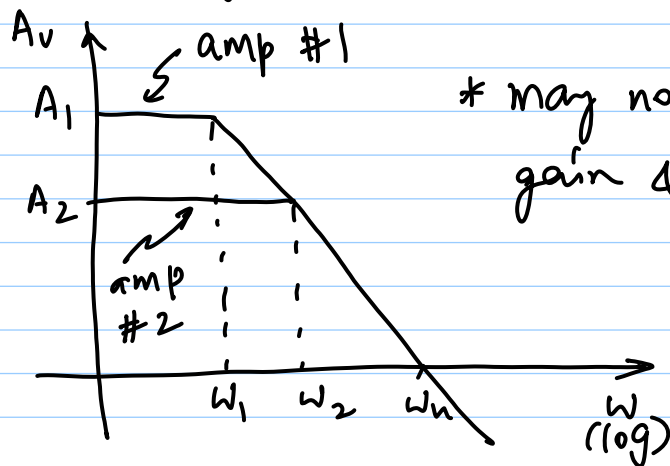
$$\text{DC gain } A_0 = g_m r_{ds}$$

$$\omega_{-3dB} = \omega_0 = \frac{1}{C_L r_{ds}}$$

$$\text{Unity gain freq. } \omega_u = A_0 \omega_0 = \frac{g_m}{C_L}$$



$\omega_u = f(\text{device size, bias point, load cap})$



\* may not be able to satisfy both gain & BW specs with one stage amp. without increased power diss.

\* increasing power cannot take us far:



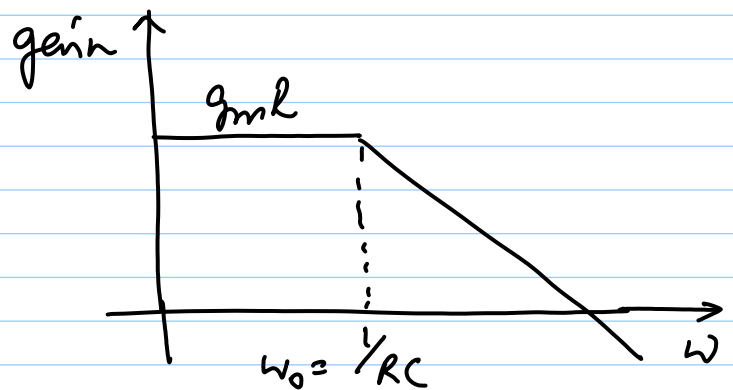
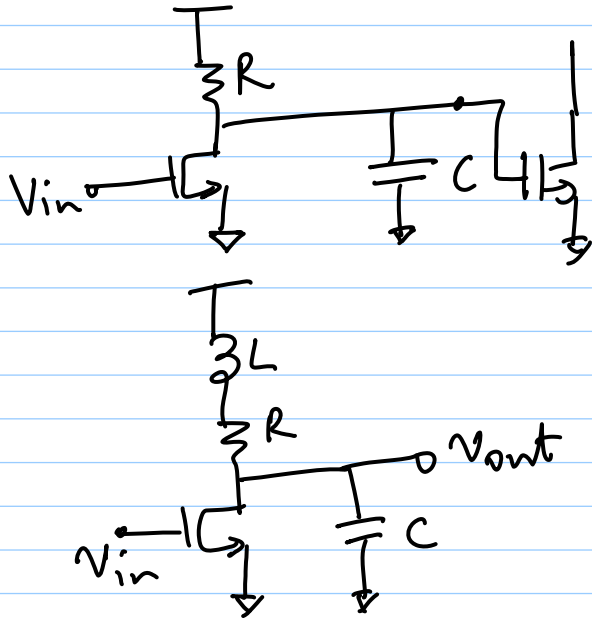
\* If  $\omega \uparrow \Rightarrow g_m \uparrow$   
but so does  $C_L$   
 $\Rightarrow \frac{g_m}{C_L} = \text{constant}$

\* cascade a number of single-pole amps can accomplish this, but  $\Rightarrow$  more delay

$\Rightarrow$  Gain - BW - Delay tradeoff is fundamental

## BW Enhancement Techniques

### 1) Shunt - Peaking :



- \*  $L$  introduces a zero  
( $z \uparrow$  with freq.)
- $\Rightarrow$  broader freq. range than  $\omega_0$  is possible

\* In the time domain: assume an input voltage step

$\rightarrow$  inductor delays current flow in branch containing  $R, L$

$\rightarrow$  more current available for charging  $C$

$\Rightarrow$  rise time is reduced

i.e. BW is increased

$$Z(s) = (sL + R) \parallel \frac{1}{sC}$$

$$= \frac{R \left( s \cdot \frac{L}{R} + 1 \right)}{s^2 LC + sRC + 1}$$

$$\left\{ \begin{array}{l} A_v(s) = g_m |Z(j\omega)| \\ \Rightarrow \text{study } |Z(j\omega)| \end{array} \right\}$$

$\Rightarrow$  2 poles (complex conjugate is possible)  
1 zero  $\omega_z = -R/L$

Define:

(i) original 3dB BW  $\omega_0 = \frac{1}{RC}$

(ii) time constant corresponding  
to zero  $\tau = L/R$

$$(iii) m = \frac{\text{original time constant}}{\text{new time constant}} = \frac{RC}{L/R} \\ = \frac{1}{\omega_0 \tau}$$

$$\Rightarrow \frac{Z(s)}{R} = \frac{s\tau + 1}{s^2 \tau^2 m + s\tau m + 1}$$

$$\frac{|Z(j\omega)|}{R} = \sqrt{\frac{(\omega\tau)^2 + 1}{(1 - \omega^2 \tau^2 m)^2 + (\omega\tau m)^2}}$$

$$(a) \quad \omega = \omega_{3dB}, \quad \frac{|Z(j\omega)|}{R} = \frac{1}{\sqrt{2}}$$

$$\Rightarrow \frac{1 + \omega^2 \tau^2}{(1 - \omega^2 \tau^2 m)^2 + (\omega^2 \tau^2 m)^2} = \frac{1}{2}$$

$$\text{Let } x = \omega^2 \tau^2$$

$$\Rightarrow \frac{1 + x}{(1 - mx)^2 + m^2 x} = \frac{1}{2}$$

$$\Rightarrow 2 + 2x = m^2 x^2 - 2mx + 1 + m^2 x$$

$$m^2 x^2 + (m^2 - 2m - 2)x - 1 = 0$$

$$\Rightarrow x = \frac{(2m + 2 - m^2) + \sqrt{(2m + 2 - m^2)^2 + 4m^2}}{2m^2}$$

only +ve root because  $x = \omega^2 \tau^2 > 0$

$$x = \frac{1}{m^2} \left\{ (m + 1 - \frac{m^2}{2}) + \sqrt{(m + 1 - \frac{m^2}{2})^2 + m^2} \right\}$$

$$\begin{aligned} x &= (\omega_{3dB} \cdot \tau)^2 = \left( \frac{\omega_{3dB}}{\omega_0} \right)^2 \cdot (\omega_0 \tau)^2 \\ &= \left( \frac{\omega_{3dB}}{\omega_0} \right)^2 \cdot \frac{1}{m^2} \end{aligned}$$

$$\Rightarrow \frac{\omega_{3dB}}{\omega_0} = \sqrt{\left(-\frac{m^2}{2} + m + 1\right) + \sqrt{\left(-\frac{m^2}{2} + m + 1\right)^2 + m^2}}$$

\* You can plot  $\frac{\omega_{3dB}}{\omega_0}$  as a function of  $m$

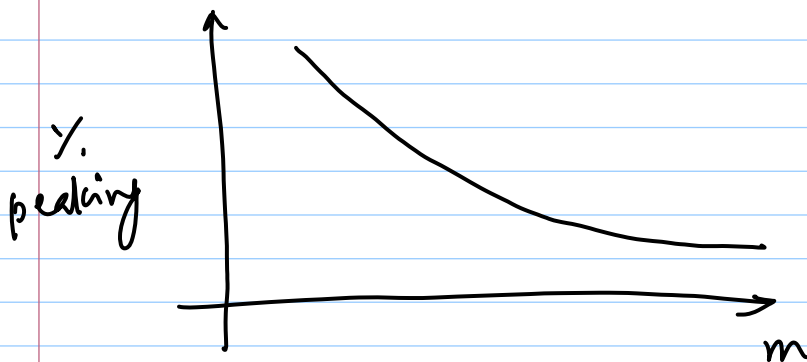
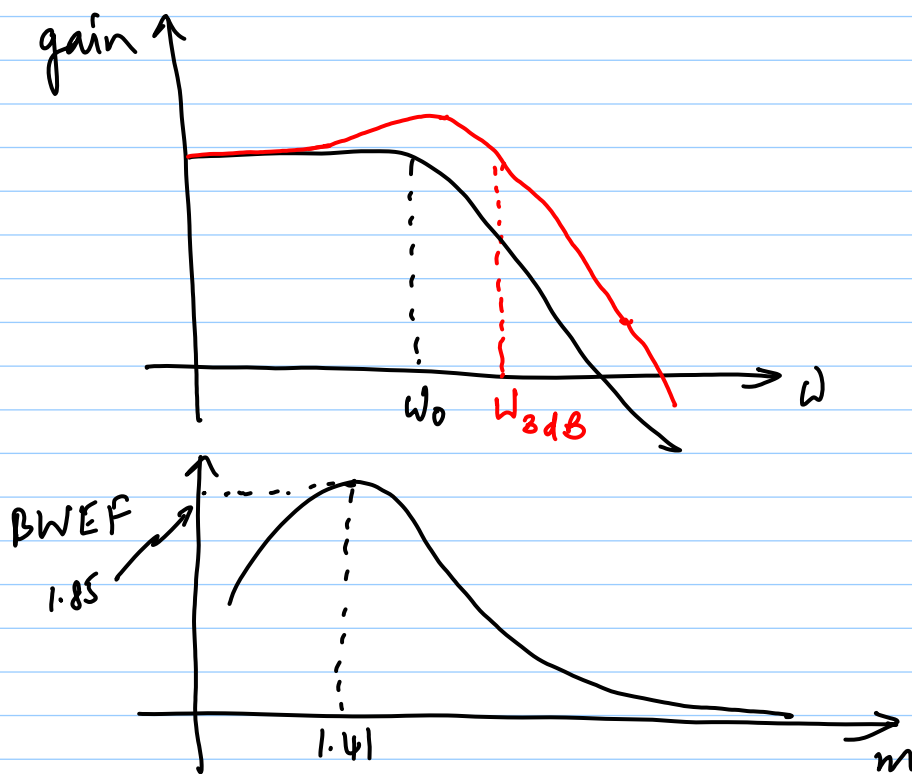
\* Also plot  $\frac{Z(j\omega)}{Z}$  as a function of  $m$

1) For max BW extension,

$$m = \sqrt{2} = 1.41$$

$$\begin{aligned} \Rightarrow \frac{\omega_{3dB}}{\omega_0} &= 1.85 \quad \text{at no increase in power!} \\ &= \text{BW extension factor (BWEF)} \end{aligned}$$

\* Problem : almost 20% peaking in freq. response



\* set  $|z| = R$  @  $\omega_0$  to moderate peaking  
 $\Rightarrow$  solving for  $m$  gives  
 $m = 2$

$$\omega_{3dB} = \omega_0 \sqrt{1 + \sqrt{5}} \approx 1.8 \omega_0$$

$\rightarrow$  BW extension almost the same

$\rightarrow$  peaking  $\sim 3\%$  { often-used optimum }

2) Maximally flat response (Butterworth)

⇒ no peaking

\*  $|Z(j\omega)|^2 \Rightarrow$  maximise # of derivatives  
where value is zero @ DC

$$\Rightarrow m = 1 + \sqrt{2} = 2.41$$

$$\Rightarrow \omega_{3dB} = 1.72 \omega_0$$

↖ BWF

3) Even with maximally flat response,  
phase distortion may occur. (ISI)

⇒ optimise for group-delay response  
e.g. optical comm. applications, UWB

\* ideal wide band amp  $\Rightarrow$  phase  $\uparrow$  linearly  
with freq. (i.e. same delay for all  
freq.)  $\Rightarrow \frac{d\phi}{d\omega} = \text{constant over freq.}$

\* Non-linear phase response  $\Rightarrow$  unequal  
delay of freq. components  
 $\Rightarrow$  group-delay distortion

\* Maximally flat group delay:

$$T_D(\omega) = - \frac{d\phi}{d\omega}$$

⇒ maximise # of derivatives of  $T_D(\omega)$   
where value is zero at DC.

after lots of algebra:

$$\Rightarrow m \approx 3.1$$

$$\Rightarrow \omega_1 \approx 1.6 \omega_0$$

↖ BWEF

\* conditions for maximally flat gain and delay do not coincide, so tradeoff is involved.

### Design

Given: DC gain, load cap  $C$ ,  $\omega_{3dB}$ ,  
constraint on max BW / mag. response /  
phase response

$$\omega_0 = \frac{\omega_{3dB}}{BW \times F} = \frac{1}{RC} \Rightarrow \textcircled{R}$$

$$A_{V_{dc}} = g_m R \Rightarrow \textcircled{g_m}$$

$$m = \frac{RC}{L/R} \Rightarrow L = \frac{R^2 C}{m} \Rightarrow \textcircled{L}$$

\*  $R$  is in series with  $L$

$\Rightarrow$  low- $Q$   $L$  is OK, absorb series  $r_s$  into  $R$

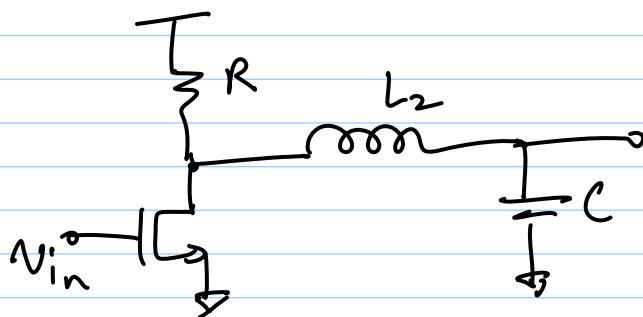
\* Emphasis more on area  $\Rightarrow$  max  $L$  in minimum area

$\rightarrow$  series stacked structures are popular



condition	$m = R^2 C / L$	BW X F	Normalised peak freq. response
max BW	1.41	1.85	1.19
$ z  = R @ \omega_0$	2	1.8	1.03
minimally flat	2.41	1.72	1
Group delay	3.0	1.6	1
No shunt peaking	$\infty$	1	1

### Series Peaking



again  $L_2 = \frac{R^2 C}{m}$

\* max BW  $\Rightarrow m = 2$  ; BW X F =  $\sqrt{2}$

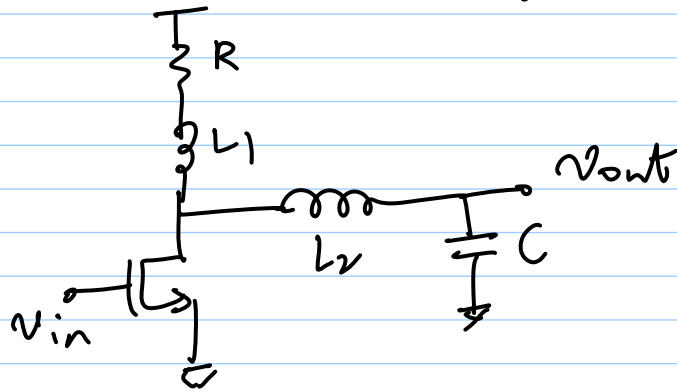
$\rightarrow$  minimally flat amplitude also

\* max flat group delay  $\Rightarrow m = 3$  ; BW X F = 1.36

\* Shunt peaking BW X F > series peaking BW X F

\* Why not use both?

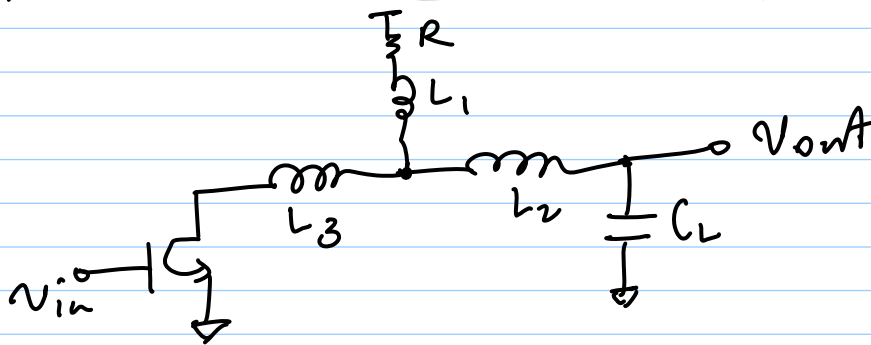
## Shunt-series peaking



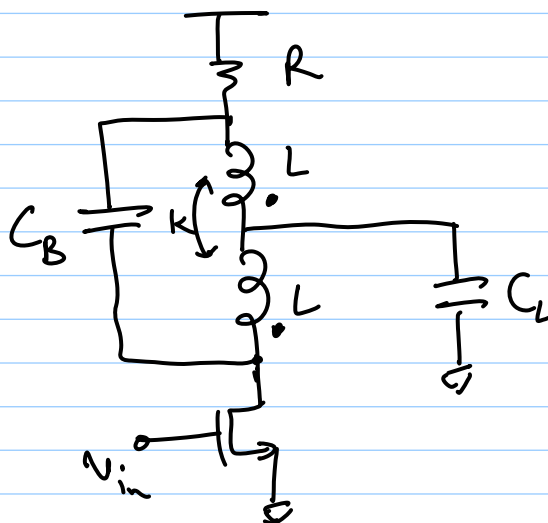
\* BW - delay  
tradeoff

\*  $C_{db}$  &  $C_L$  are  
charged serially  
in time

## Shunt-double series peaking



- \*  $L_1, L_2, L_3$  can be replaced by a single Xfmr to save area
- \* Add bridging cap to create parallel resonance (this helps with BW too)



Bridged  
T-coil

\* You can show that

$$L = \frac{R^2 C}{2(1+k)}$$

$$C_B = \frac{C}{4} \cdot \frac{(1-k)}{(1+k)}$$


\*  $k = 1/3 \Rightarrow$  Butterworth mag. response

$k = 1/2 \Rightarrow$  max. flat group delay

\* Used in oscilloscopes for a long time

$$\omega_{3dB} (\max) = 2\sqrt{2} \omega_0$$

$$\approx 2.83 \omega_0$$

 BWEF