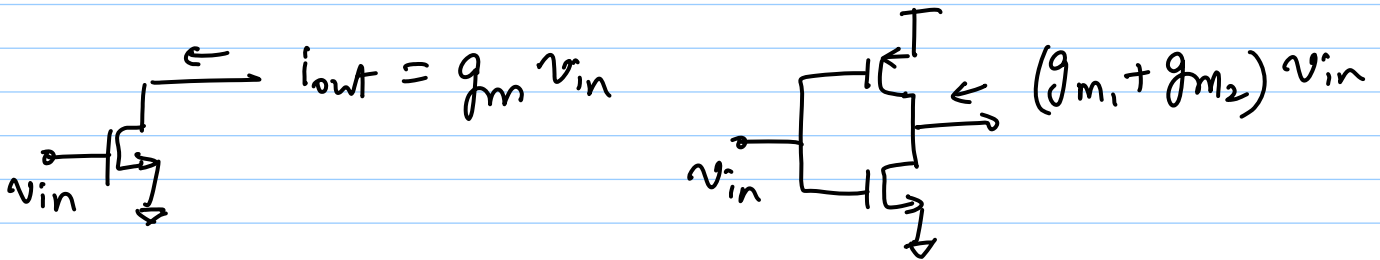


## Lecture 19: LNA Design (final) : Introduction to Mixers

### Other ideas to explore (LNAs)

1) Current re-use:



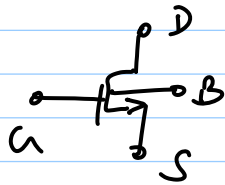
2) Feedback: \* to match  $R_{in} = 50 \Omega$   
\* to improve  $11P_3$  (linearity)

issues:  $\rightarrow$  NF may increase substantially  
 $\rightarrow$  Stability may worsen

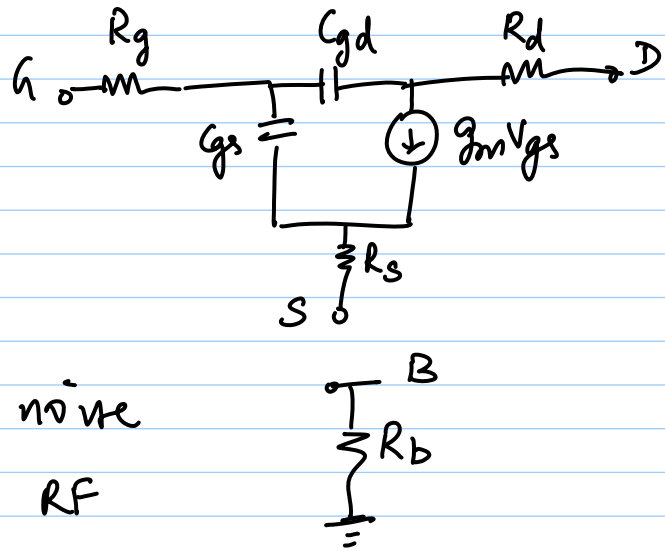
### Other general rules of thumb

- \* Start off with same  $w$  &  $L$  for cascode device; usually cascode gate =  $V_{DD}$
- \* Start with  $f_T \geq 10 \times$  operating freq.
- \* Model noise accurately (e.g. model may use  $\delta = 2/3$  &  $\delta = 0$ ).
- \* Hand calculations are approximations only, so use simulation tools effectively:
  - $\rightarrow$  Plot  $g_m$ ,  $f_T$  and  $\Gamma$  vs.  $I_{bias}$  and  $w$
  - $\rightarrow$  Plot NF vs. current density
  - $\rightarrow$  Usually,  $NF_{opt.} \neq S_{21, opt.} \neq 11P_{3, opt.}$

# Extrinsic sources of noise (mos)



$\equiv$



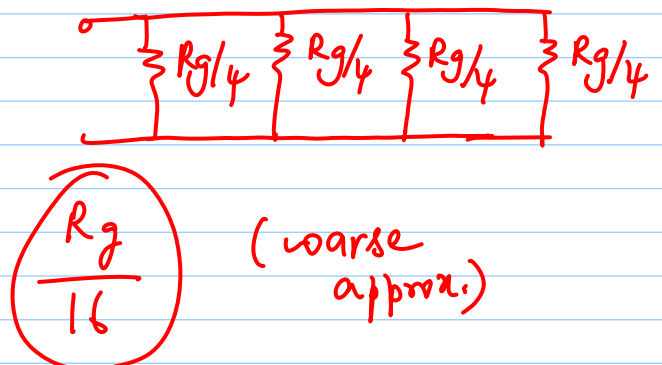
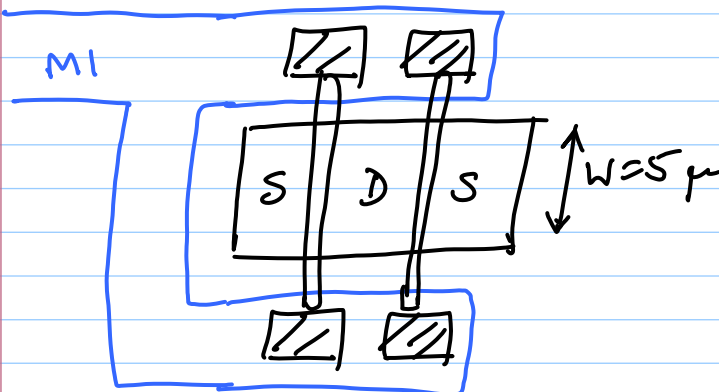
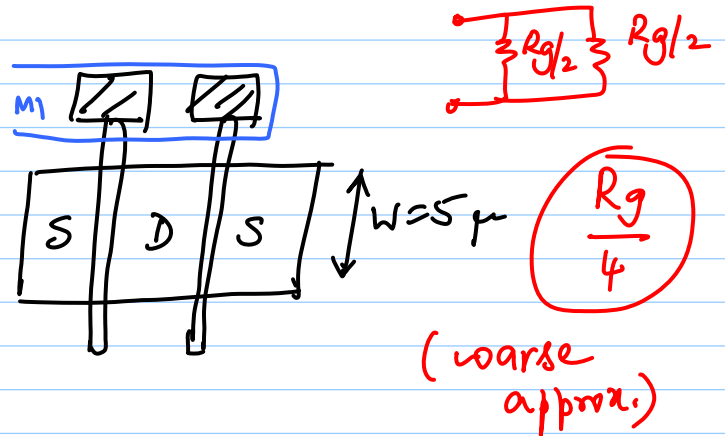
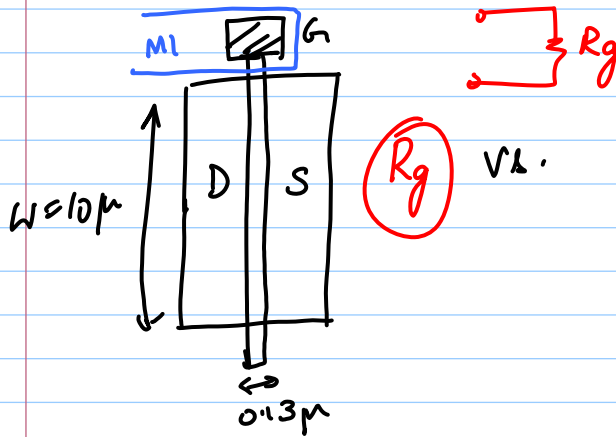
\*  $R_g, R_d, R_s, R_b$   
contribute thermal noise

solution: use sound RF

layout techniques

- 1)  $R_g$  — use multifinger devices to reduce  $R_g$   
— contacts on both sides of gate

e.g.  $\frac{W}{L} = \frac{10\mu}{0.13\mu}$  ;  $R_{metal} \ll R_{poly}$



Accurate expression: (takes distributed effects into account)

$$R_g = K \cdot \frac{R_{sh} \cdot W}{n^2 L} \quad \left. \begin{array}{l} K = \frac{1}{3} \text{ for 1-side gate contact} \\ K = \frac{1}{12} \text{ for 2-sided gate contact} \end{array} \right\}$$

→  $R_{sh}$  = sheet resistance

→  $n$  = number of fingers

- 2)  $R_b$  — use more number of substrate contacts  
— thermal noise of  $R_b$  can modulate backgate of mos  
— can impact gain & NF

$$\overline{i_{nd,sub}^2} = 4kT R_b \cdot g_{mbs}^2 \Delta f$$

## Introduction to Mixers:

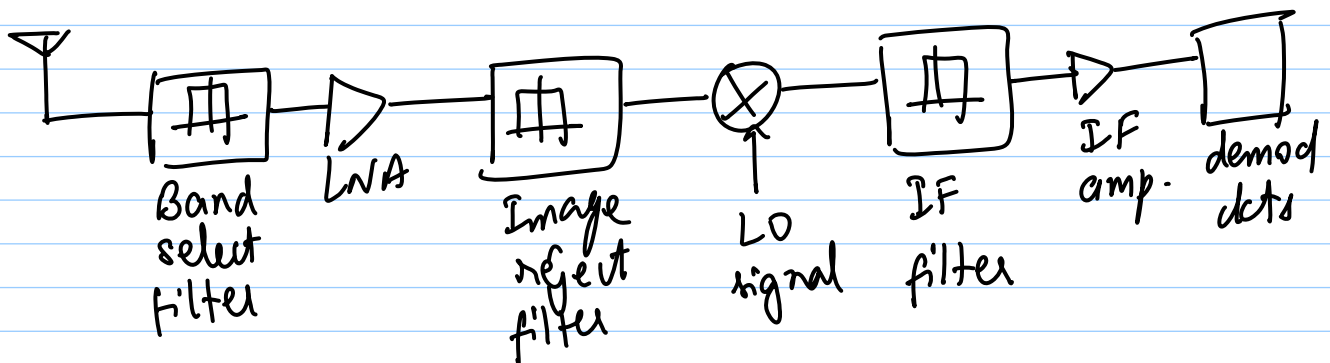
- \* mixers are used for freq. translation

$T_x \rightarrow$  up conversion of BB signals to IF/RF

Rx  $\rightarrow$  down-conversion of RF signals to IF/BB

## Down-conversion mixers

Superhet. receiver



Simple mixer : RF  $\rightarrow$   $\otimes$   $\rightarrow$  IF  
                                 $\uparrow$   
                                LO

$$\left. \begin{aligned} x_{RF}(t) &= A_{RF} \cos \omega_{RF} t \\ x_{LO}(t) &= A_{LO} \cos \omega_{LO} t \end{aligned} \right\} x_{IF}(t) = x_{RF}(t) \cdot x_{LO}(t)$$

$$x_{IF}(t) = (A_{RF} \cos \omega_{RF} t) \cdot (A_{LO} \cos \omega_{LO} t)$$

$$= \frac{A_{RF} A_{LO}}{2} \left[ \underbrace{\cos(\omega_{LO} - \omega_{RF})t}_{\text{desired IF term}} + \cos(\omega_{LO} + \omega_{RF})t \right]$$

Metrics

1)  $G_c \equiv$  conversion gain =  $\frac{\text{IF amplitude (desired IF)}}{\text{RF amplitude}}$

$$G_c = \frac{A_{RF} A_{LO} / 2}{A_{RF}} = \frac{A_{LO}}{2}$$

$G_c < 1$  may be ok in several cases  
(depending on NF)

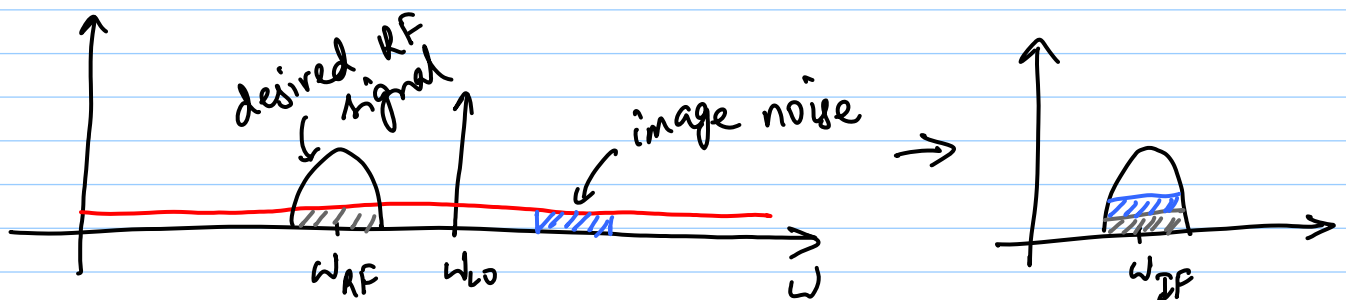
$\Rightarrow$  Passive Mixers

2) Noise Figure NF

$$NF = \frac{\text{SNR @ RF port}}{\text{SNR @ IF port}}$$

\* Beware of image frequencies when computing NF of mixers

$\rightarrow$  Even if there is no image signal @  $\omega_{IM}$ , noise from  $\omega_{IM}$  still contributes to NF!



\*  $NF_{DSB}$  is measured with useful RF signal at both  $\omega_{RF}$  &  $\omega_{IF}$  (i.e.  $\omega_{LO} \pm \omega_{IF}$ )

\*  $NF_{SSB}$  is measured with useful RF signal only at one freq. (either  $\omega_{LO} + \omega_{IF}$  or  $\omega_{LO} - \omega_{IF}$ )

$$\Rightarrow NF_{SSB} = NF_{DSB} + 3dB \quad (\text{noise power adds as mean-square})$$

$\Rightarrow$  Check carefully to see which NF is quoted!

$NF_{SSB}$  is the accurate metric in most cases

\* Note that even if mixer is noiseless,  
 $NF_{SSB} = 3dB!$

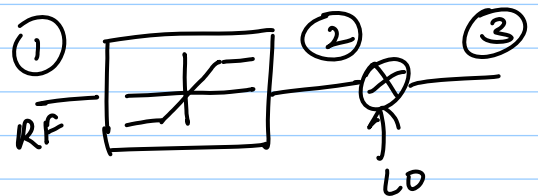
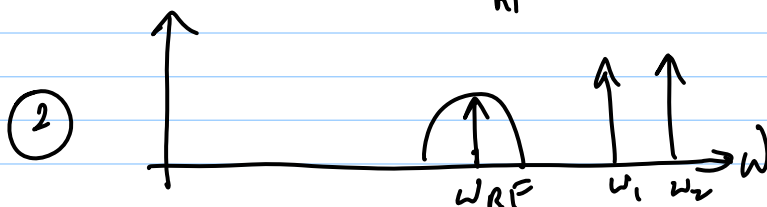
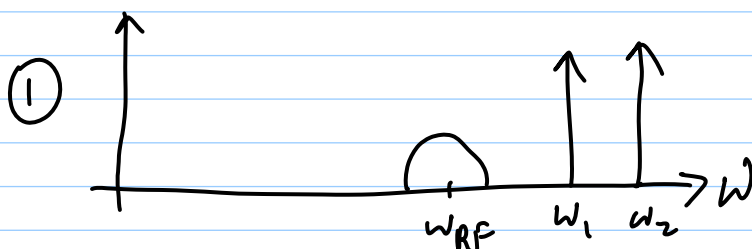
typical values are

$$NF_{SSB} \sim 10-15dB$$

$\Rightarrow$  LNA needs to precede mixer to keep  $NF_{tot.}$  low.

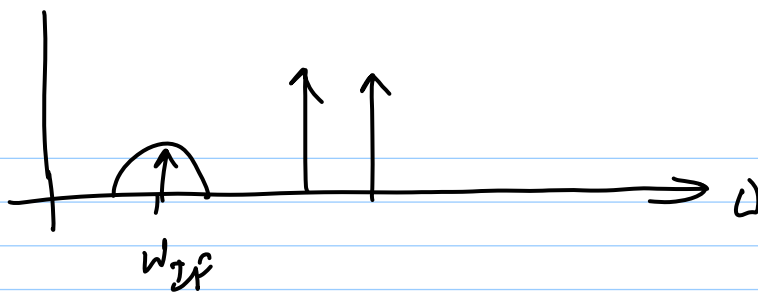
3) Linearity : measured as  $1IP_3/OIP_3$  at appropriate frequency (RF/IF)

RF path non-linearity :



$IM_3$  component falls at desired  $\omega_{RF}$

③



\* Cubic non-linearity can cause problems even with a single input!

e.g. AM radio

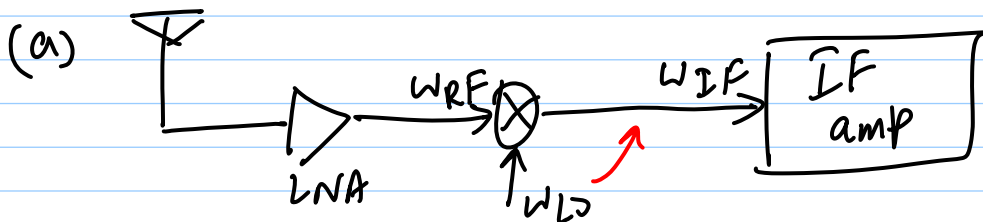
$$IF = 455 \text{ kHz} ; f_{ch} = 910 \text{ kHz}$$

$$\Rightarrow f_{LO} = 1365 \text{ kHz}$$

cubic nonlinearity - generates  $(2\omega_{RF} - \omega_{LO})$  term

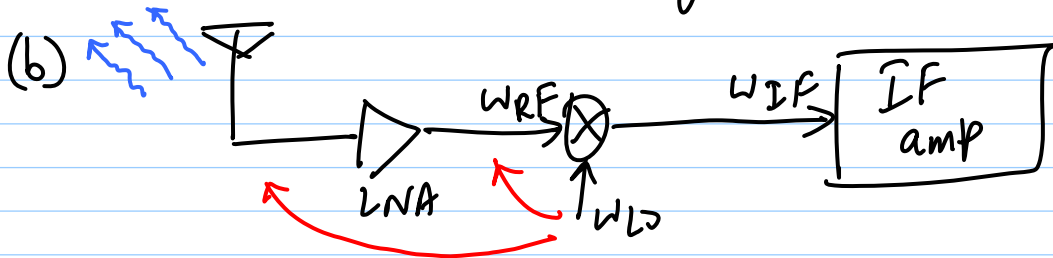
In this case,  $2\omega_{RF} - \omega_{LO} = 455 \text{ kHz} = \omega_{IF}!$

4) Isolation:



\*  $\omega_{LO} \gg \omega_{RF}, \omega_{IF}$

\* LO-IF feed through would saturate IF amp.



LO-RF feedthrough (reverse isolation)

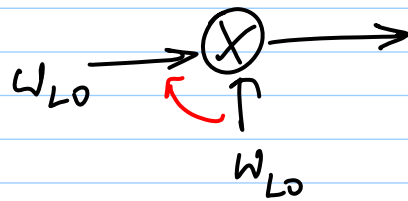
\* Re-radiation of LO & its harmonics

\* Usually,  $A_{LO} \gg A_{RF}$

Direct-conversion (Homodyne) receiver:

$$\omega_{RF} = \omega_{LO}, \quad \omega_{IF} = 0$$

we want only  $\omega_{RF}$  @ BB



"self-mixing"

→ low-freq. term that depends on LO only

→ can be larger than desired RF signal

## 5) Spurs

Mixers, by nature produce a bunch of freq. components

\* Undesired freq. components @ mixer output are called spurious signals or 'spurs'

\* Highly tedious in practice:

if there are  $m$  &  $n$  harmonic numbers @ RF & LO frequencies respectively,

$$f_{spur} = m f_{RF} + n f_{LO} \quad \text{for all combinations of signs of } m \text{ \& } n$$