

PROPERTIES OF LAPLACE TRANSFORM

Linearity

- Many parallel properties of the CTFT, but for Laplace transforms we need to determine implications for the ROC
- For example:

Linearity

$$ax_1(t) + bx_2(t) \longleftrightarrow aX_1(s) + bX_2(s)$$

ROC at least the intersection of ROCs of $X_1(s)$ and $X_2(s)$

ROC can be bigger (due to pole-zero cancellation)

E.g. $x_1(t) = x_2(t)$ and $a = -b$

Then $ax_1(t) + bx_2(t) = 0 \longrightarrow X(s) = 0$

\Rightarrow ROC entire s -plane

Time Shifting

- If

$$x(t) \xleftrightarrow{\mathcal{L}} X(s), \quad ROC: R1$$

- then

$$\boxed{x(t - t_0) \xleftrightarrow{\mathcal{L}} e^{-st_0} X(s), \quad ROC: R1}$$

$$x(t - T) \longleftrightarrow e^{-sT} X(s), \text{ same ROC as } X(s)$$

Time Shifting

$$\frac{e^{3s}}{s+2}, \quad \Re\{s\} > -2 \quad \longleftrightarrow \quad ?$$

Time Shifting

$$\frac{e^{-sT}}{s+2}, \quad \Re\{s\} > -2 \quad \longleftrightarrow \quad e^{-2t}u(t)|_{t \rightarrow t-T}$$

$$\downarrow T = -3$$

$$\frac{e^{3s}}{s+2}, \quad \Re\{s\} > -2 \quad \longleftrightarrow \quad e^{-2(t+3)}u(t+3)$$

Shift in s-domain

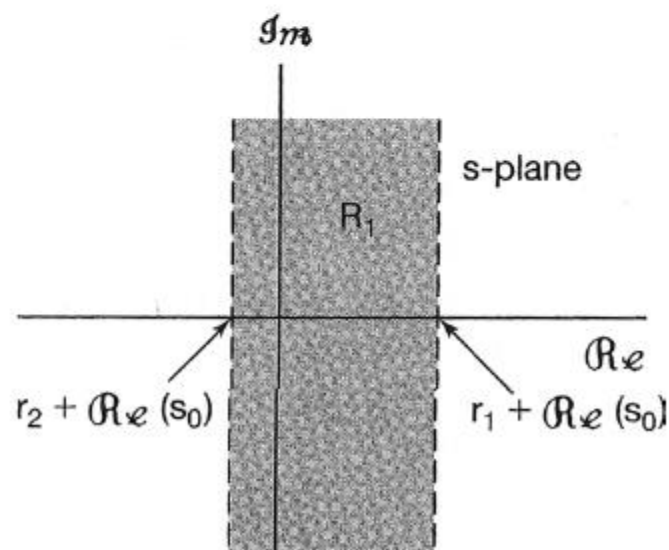
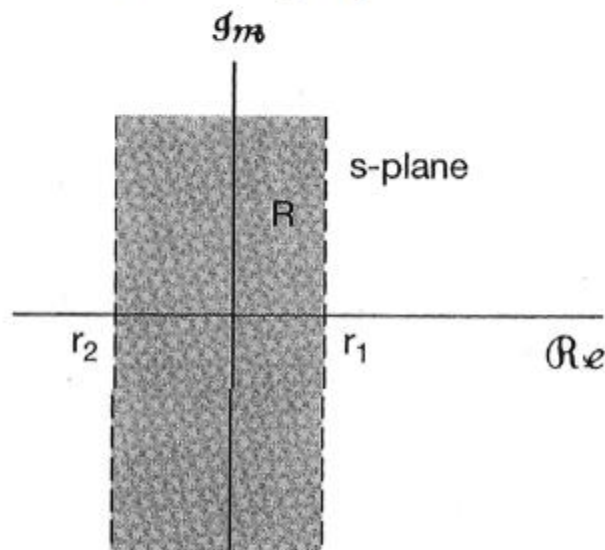
- If

$$x(t) \xleftrightarrow{\mathcal{L}} X(s), \quad \text{ROC} : R$$

- then

$$e^{s_0 t} x(t) \xleftrightarrow{\mathcal{L}} X(s - s_0), \quad \text{ROC} : R + \text{Re}\{s_0\}$$

- ROC associated with $X(s - s_0)$ is that of $X(s)$, shifted by $\text{Re}\{s_0\}$



Shift in s-domain

$$x(t) \xleftrightarrow{\mathcal{LT}} X(s); \quad \text{with ROC} = R$$

$$e^{s_0 t} x(t) \xleftrightarrow{\mathcal{LT}} X(s - s_0); \quad \text{with ROC} = R + \text{Re}\{s_0\}$$

- Special Case: $s_0 = j\omega_0$ gives:

$$e^{j\omega_0 t} x(t) \xleftrightarrow{\mathcal{LT}} X(s - j\omega_0); \quad \text{with ROC} = R$$

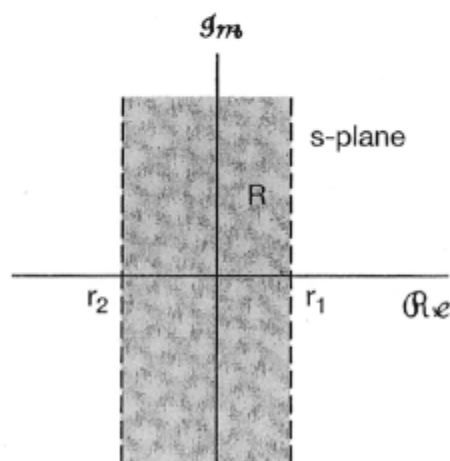
Time Scaling

$$x(t) \xleftrightarrow{\mathcal{L}} X(s); \text{ with ROC } = R$$

$$x(at) \xleftrightarrow{\mathcal{L}} \frac{1}{|a|} X\left(\frac{s}{a}\right); \text{ with ROC } R_1 = \frac{R}{a}$$

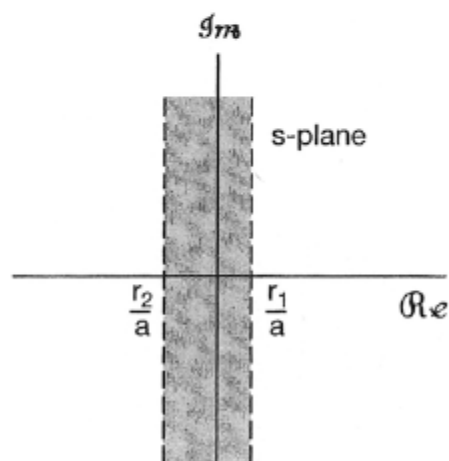
- special case: $a = -1$

$$x(-t) \xleftrightarrow{\mathcal{L}} X(-s); \text{ with ROC } R_1 = -R$$



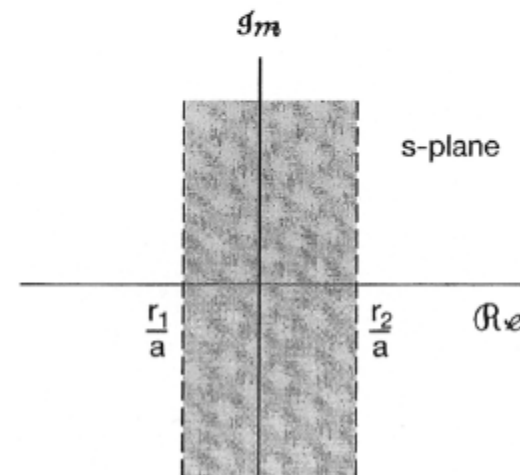
(a)

ROC of $X(s)$



(b)

ROC of $(1/|a|)X(s/a)$, $a > 1$



(c)

ROC of $(1/|a|)X(s/a)$, $0 > a > -1$

Conjugation Property

- If

$$x(t) \xleftrightarrow{\mathcal{L}} X(s), \text{ with } ROC = R$$

- then:

$$x^*(t) \xleftrightarrow{\mathcal{L}} X^*(s^*), \text{ with } ROC = R$$

- and:

$$X(s) = X^*(s^*), \text{ when } x(t) \text{ is real}$$

- Consequently, if $x(t)$ is real and if $X(s)$ has a pole or zero at $s = s_0$, then $X(s)$ also has a pole or zero at the complex conjugate point $s = s_0^*$.
- Example: $X(s)$ has a pole at $s = 1 + 3j$; it must also have a pole at $s = 1 - 3j$

Convolution Property

- If

$$x_1(t) \xleftrightarrow{\mathcal{L}} X_1(s), \quad ROC = R_1$$

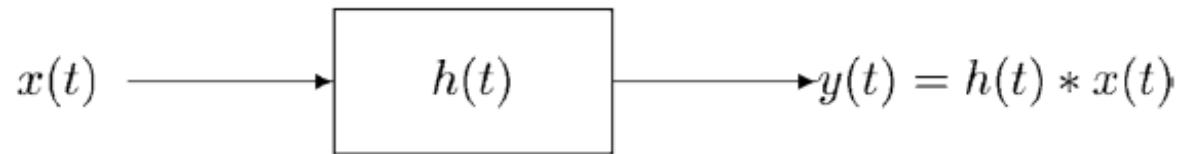
$$x_2(t) \xleftrightarrow{\mathcal{L}} X_2(s), \quad ROC = R_2$$

- Then

$$\boxed{x_1(t) * x_2(t) \xleftrightarrow{\mathcal{L}} X_1(s)X_2(s), \quad ROC = R_1 \cap R_2}$$

- The ROC of $X_1(s)X_2(s)$ includes the intersection of the ROC s of $X_1(s)$ and $X_2(s)$ and can be larger if pole-zero cancellation occurs in the product.

Convolution Property



For $x(t) \longleftrightarrow X(s), y(t) \longleftrightarrow Y(s), h(t) \longleftrightarrow H(s)$
Then $Y(s) = H(s) \cdot X(s)$

- ROC of $Y(s) = H(s)X(s)$: at least the overlap of the ROCs of $H(s)$ & $X(s)$

Convolution Property

- ROC could be empty if there is no overlap between the two ROCs
e.g.

$$x(t) = e^t u(t), \text{ and } h(t) = -e^{-t} u(-t)$$

- ROC could be larger than the overlap of the two. e.g.

Example:

$$X_1(s) = \frac{s+1}{s+2}, \quad \text{Re}\{s\} > -2; \quad X_2(s) = \frac{s+2}{s+1}, \quad \text{Re}\{s\} > -1$$

$$X_1(s) \cdot X_2(s) = 1, \quad \text{ROC entire } s\text{-plane}$$

Time Domain Differentiation

$$x(t) = \frac{1}{2\pi j} \int_{\sigma-j\omega}^{\sigma+j\omega} X(s)e^{st} ds, \quad \frac{dx(t)}{dt} = \frac{1}{2\pi j} \int_{\sigma-j\omega}^{\sigma+j\omega} sX(s)e^{st} ds$$

↓

$\frac{dx(t)}{dt} \longleftrightarrow sX(s),$ with ROC containing the ROC of $X(s)$

ROC could be bigger than the ROC of $X(s)$, if there is pole-zero cancellation. e.g.,

$$\begin{aligned} x(t) &= u(t) \leftrightarrow \frac{1}{s}, & \Re\{s\} > 0 \\ \frac{dx(t)}{dt} &= \delta(t) \leftrightarrow 1 = s \cdot \frac{1}{s} & \text{ROC} = \text{entire } s\text{-plane} \end{aligned}$$

s-Domain Differentiation

s-Domain Differentiation

$$-tx(t) \leftrightarrow \frac{dX(s)}{ds}, \text{ with same ROC as } X(s)$$

(Derivation is
similar to $\frac{d}{dt} \leftrightarrow s$)

$$\text{E.g., } te^{-at}u(t) \leftrightarrow ?$$

s-Domain Differentiation

$$\text{E.g., } te^{-at}u(t) \leftrightarrow -\frac{d}{ds} \left[\frac{1}{s+a} \right] = \frac{1}{(s+a)^2},$$

$$\Re\{s\} > -a$$

Time Domain Integration

- If

$$x(t) \xleftrightarrow{\mathcal{L}} X(s), \quad ROC = R$$

- then

$$\int_{-\infty}^t x(\tau) d\tau \xleftrightarrow{\mathcal{L}} \frac{1}{s} X(s), \quad ROC = R \cap \{\operatorname{Re}\{s\} > 0\}$$

- Inverse of differentiation property
- Property can be derived from convolution property, i.e.,

$$\int_{-\infty}^t x(\tau) d\tau = u(t) * x(t) \xleftrightarrow{\mathcal{L}} U(s)X(s) = \frac{1}{s} X(s)$$

END