

Z- TRANSFORM

Introduction

- Developed Laplace transform as an extension of the CT Fourier transform.
- Laplace transform can be applied to a broader class of signals than the Fourier transform.
- In this lecture we develop the z-transform as the discrete-time counterpart of the Laplace transform.

Z - Transform

$$\begin{aligned}
 x(t) = e^{st} &\longrightarrow \boxed{h(t)} \longrightarrow y(t) = \int_{-\infty}^{\infty} h(\tau) e^{s(t-\tau)} d\tau \\
 &= \left[\int_{-\infty}^{\infty} h(\tau) e^{-s\tau} d\tau \right] e^{st} \\
 &= \underbrace{H(s)}_{\text{eigenvalue}} \underbrace{e^{st}}_{\text{eigenfunction}}
 \end{aligned}$$

$$\begin{aligned}
 x[n] = z^n &\longrightarrow \boxed{h[n]} \longrightarrow y[n] = \sum_{m=-\infty}^{\infty} h[m] z^{n-m} \\
 &= \left[\sum_{m=-\infty}^{\infty} h[m] z^{-m} \right] z^n \\
 &= \underbrace{H(z)}_{\text{eigenvalue}} \underbrace{z^n}_{\text{eigenfunction}}
 \end{aligned}$$

$$\begin{aligned}
 H(s) &= \int_{-\infty}^{\infty} h(\tau) e^{-s\tau} d\tau \quad \text{-- CT} \\
 H(z) &= \sum_{n=-\infty}^{\infty} h[n] z^{-n} \quad \text{-- DT}
 \end{aligned}$$

Z - Transform

- For a discrete-time LTI system with impulse response $h[n]$, the response $y[n]$ of the system to a complex exponential of the form z^n is:

$$y[n] = H(z) z^n$$

$$H(z) = \sum_{n=-\infty}^{\infty} h[n] z^{-n}$$

- For the case $z = e^{j\omega}$ with ω real, the equation above corresponds to the discrete-time FT of $h[n]$. In the more general case it is called the z -transform, and is defined as:

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

z is a complex variable

- and we refer to the z -transform pair as:

$$x[n] \xleftrightarrow{z} X(z)$$

Z - Transform

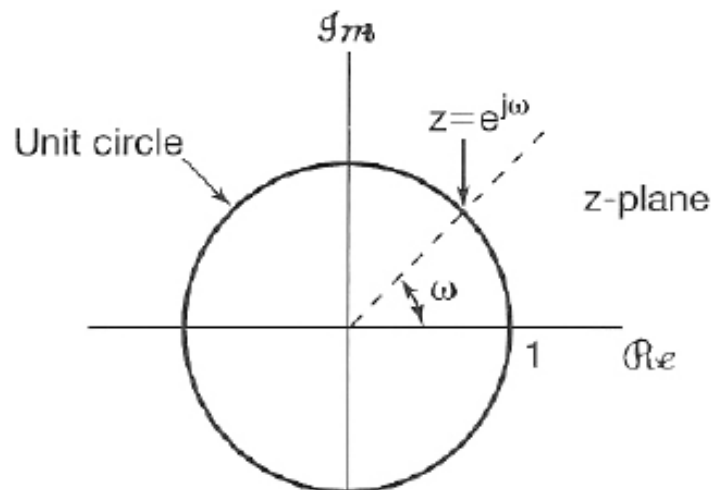
- To explore the relationships between the z – transform and the DTFT we express the complex variable z in polar form as:

$$z = re^{j\omega}$$

- with r the magnitude of z and ω the angle of z . We now express the z – transform as

$$X(re^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n](re^{j\omega})^{-n} = \sum_{n=-\infty}^{\infty} \{x[n]r^{-n}\}(e^{-j\omega})^n$$

- i.e., as the FT of the sequence $x[n]$ multiplied by a real exponential r^{-n}



Note that for $r = 1$, or equivalently $|z| = 1$, we have the relation:

$$X(z)|_{z=e^{j\omega}} = X(e^{j\omega}) = \mathcal{F}\{x[n]\}$$

i.e., the z -transform is the same as the Fourier transform.

Z - Transform and ROC

$$z = re^{j\omega}, r = |z|$$

$$\begin{aligned} X(re^{j\omega}) &= \sum_{n=-\infty}^{\infty} x[n] (re^{j\omega})^{-n} = \sum_{n=-\infty}^{\infty} (x[n]r^{-n}) e^{-j\omega n} \\ &= \mathcal{F}\{x[n]r^{-n}\} \end{aligned}$$

- $\text{ROC} = \left\{ z = re^{j\omega} \text{ at which } \sum_{n=-\infty}^{\infty} |x[n]r^{-n}| < \infty \right\}$

— depends only on $r = |z|$, just like the ROC in s -plane only depends on $\text{Re}\{s\}$

- Unit circle ($r = 1$) in the ROC \Rightarrow DTFT $X(e^{j\omega})$ exists

Example - 1

- Find the z-transform and plot the ROC

$$x[n] = a^n u[n] \text{ - right-sided}$$

Example - 1

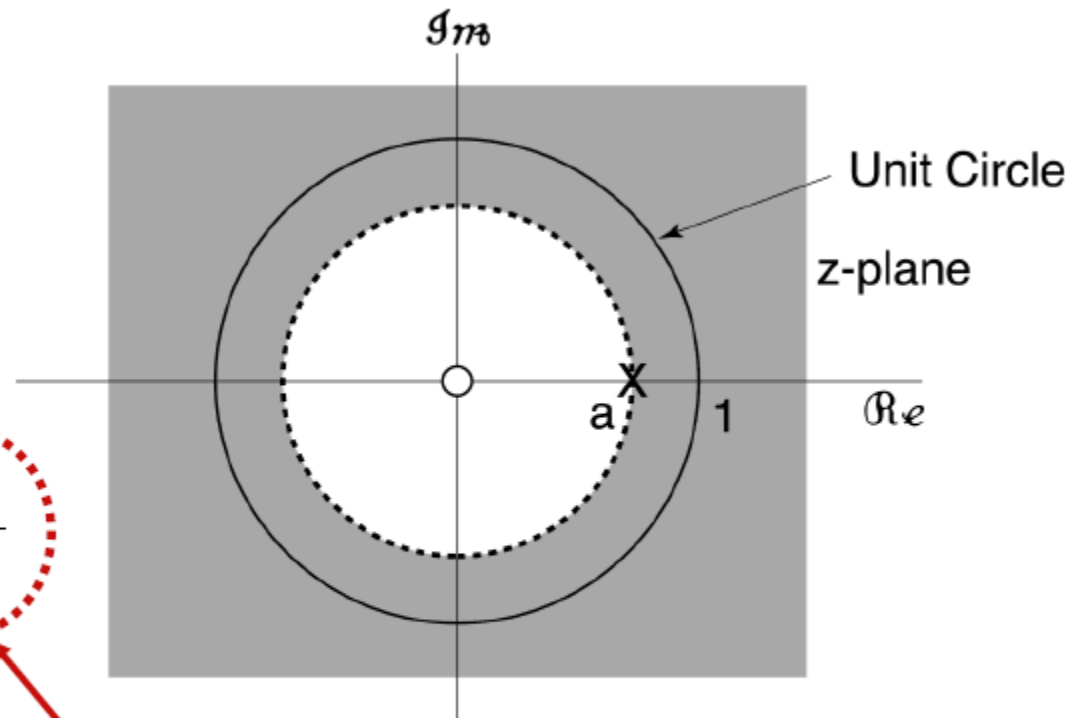
$$X(z) = \sum_{n=-\infty}^{\infty} a^n u[n] z^{-n}$$

$$= \sum_{n=0}^{\infty} (az^{-1})^n$$

$$= \frac{1}{1 - az^{-1}} = \frac{z}{z - a}$$

If $|az^{-1}| < 1$, i.e., $|z| > |a|$

That is, **ROC** $|z| > |a|$,
outside a circle



This form to find
pole and zero locations

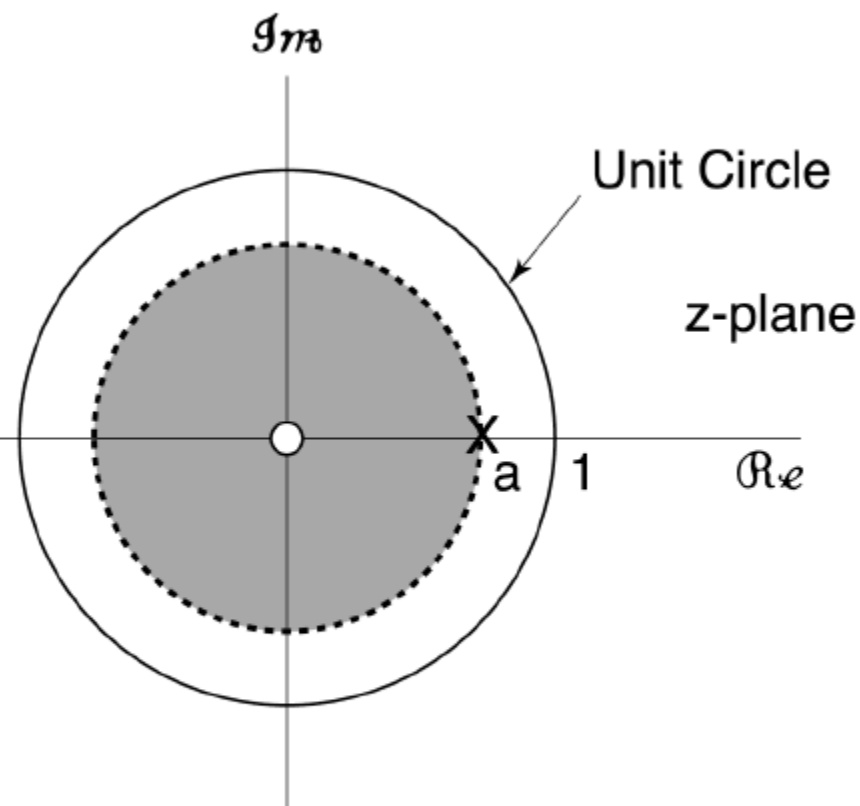
Example - 2

- Find the z-transform and plot the ROC

$$x[n] = -a^n u[-n - 1] \text{ - left-sided}$$

Example - 2

$$\begin{aligned}X(z) &= \sum_{n=-\infty}^{\infty} \{-a^n u[-n-1] z^{-n}\} \\&= - \sum_{n=-\infty}^{-1} a^n z^{-n} \\&= - \sum_{n=1}^{\infty} a^{-n} z^n = 1 - \sum_{n=0}^{\infty} (a^{-1} z)^n \\&= 1 - \frac{1}{1 - a^{-1} z} = \frac{a^{-1} z}{a^{-1} z - 1} \\&= \frac{z}{z - a},\end{aligned}$$



If $|a^{-1} z| < 1$, i.e., $|z| < |a|$

Same $X(z)$ as in **Example .1**, but different ROC.

Example - 3

- Consider a signal that is the sum of two real exponentials:

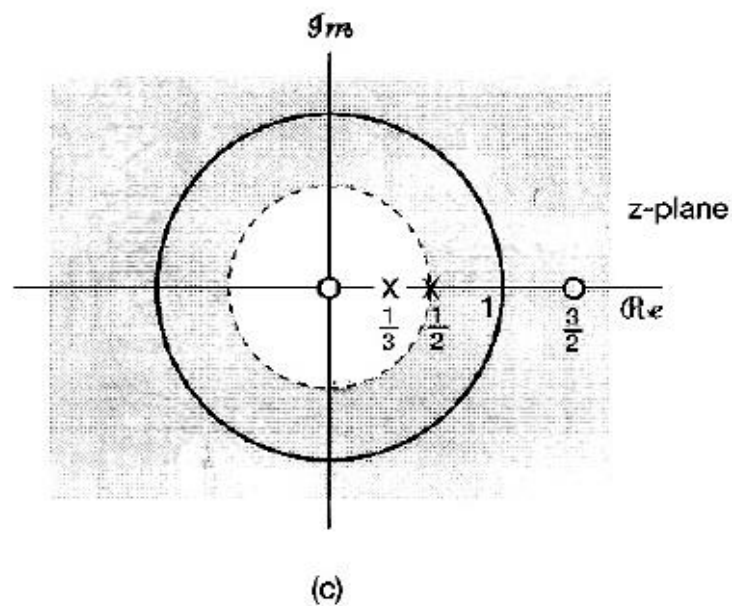
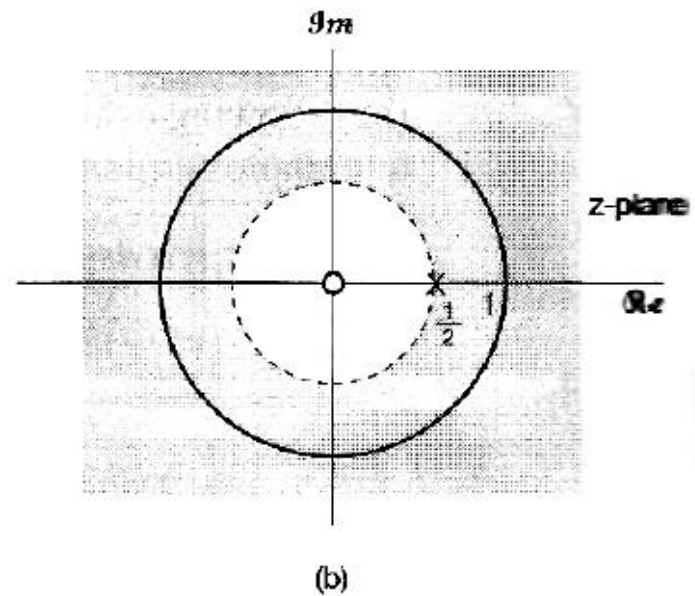
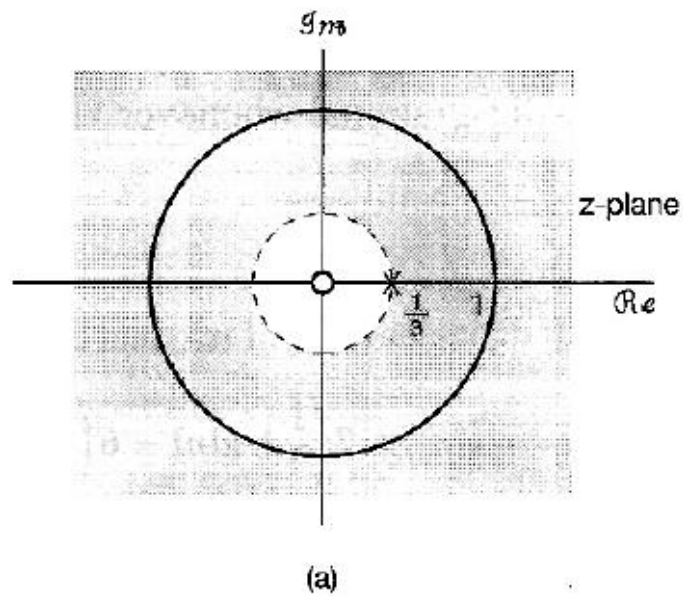
$$x[n] = 7\left(\frac{1}{3}\right)^n u[n] - 6\left(\frac{1}{2}\right)^n u[n]$$

Example - 3

$$\begin{aligned} X(z) &= \sum_{n=-\infty}^{\infty} \left\{ 7 \left(\frac{1}{3} \right)^n u[n] - 6 \left(\frac{1}{2} \right)^n u[n] \right\} z^{-n} \\ &= 7 \sum_{n=0}^{\infty} \left(\frac{1}{3} z^{-1} \right)^n - 6 \sum_{n=0}^{\infty} \left(\frac{1}{2} z^{-1} \right)^n \\ &= \frac{7}{1 - \frac{1}{3} z^{-1}} - \frac{6}{1 - \frac{1}{2} z^{-1}} \\ &= \frac{1 - \frac{3}{2} z^{-1}}{\left(1 - \frac{1}{3} z^{-1} \right) \left(1 - \frac{1}{2} z^{-1} \right)} = \frac{z \left(z - \frac{3}{2} \right)}{\left(z - \frac{1}{3} \right) \left(z - \frac{1}{2} \right)}; \end{aligned}$$

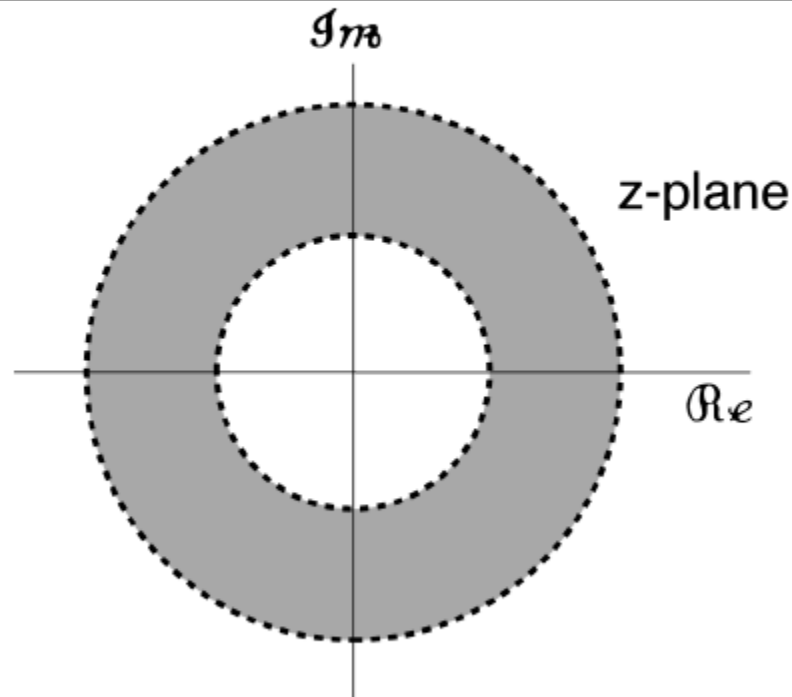
$$|z| > \max(1/3, 1/2) \Rightarrow |z| > 1/2$$

Example - 3



Properties of ROC

Property (1) -- The ROC of $X(z)$ consists of a ring in the z -plane centered about the origin (equivalent to a vertical strip in the s -plane)



Property (2) -- The ROC does *not* contain any poles (same as in LT).

Properties of ROC

Property (3) -- If $x[n]$ is of finite duration, then the ROC is the entire z -plane, except possibly at $z = 0$ and/or $z = \infty$.

Why?

$$X(z) = \sum_{n=N_1}^{N_2} x[n]z^{-n}$$

Examples:

CT counterpart

$$\delta[n] \longleftrightarrow 1 \quad \text{ROC all } z \quad \left| \quad \delta(t) \longleftrightarrow 1 \quad \text{ROC all } s$$

$$\delta[n-1] \longleftrightarrow z^{-1} \quad \text{ROC } z \neq 0 \quad \left| \quad \delta(t-T) \longleftrightarrow e^{-sT} \quad \Re\{s\} \neq -\infty$$

$$\delta[n+1] \longleftrightarrow z \quad \text{ROC } z \neq \infty \quad \left| \quad \delta(t+T) \longleftrightarrow e^{sT} \quad \Re\{s\} \neq \infty$$

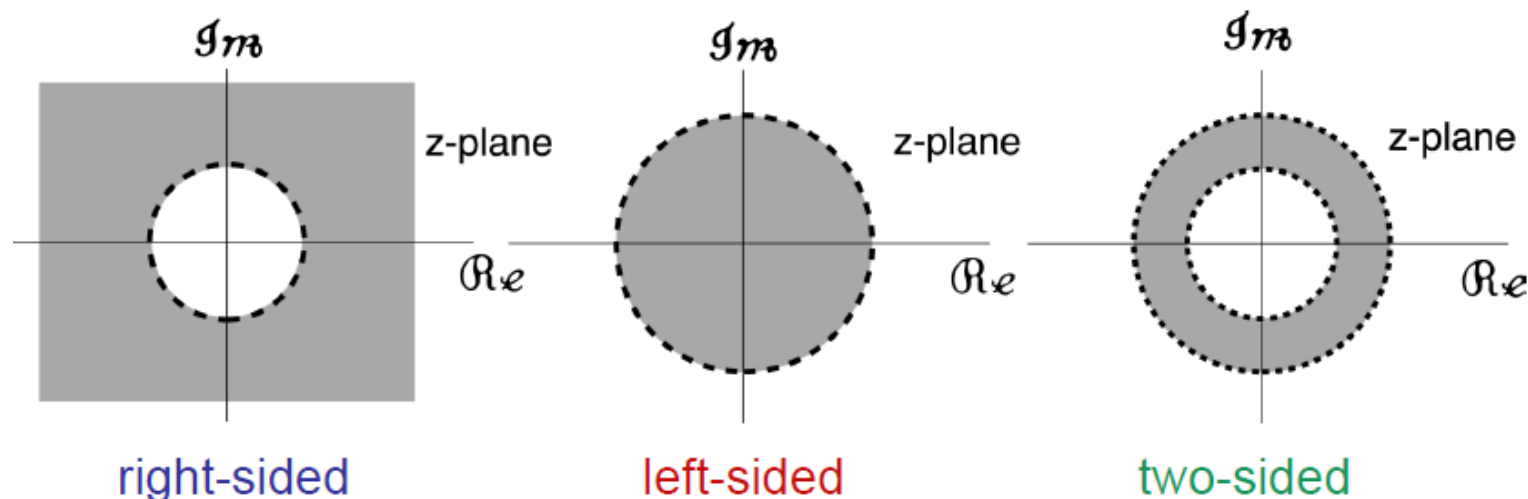
Properties of ROC

Property (4) -- If $x[n]$ is a right-sided sequence, and if $|z| = r_o$ is in the ROC, then all finite values of z for which $|z| > r_o$ are also in the ROC.

Property (5) -- If $x[n]$ is a left-sided sequence, and if $|z| = r_o$ is in the ROC, then all finite values of z for which $0 < |z| < r_o$ are also in the ROC.

Property (6) -- If $x[n]$ is two-sided, and if $|z| = r_o$ is in the ROC, then the ROC consists of a ring in the z -plane including the circle $|z| = r_o$.

What types of signals do the following ROC correspond to?



Properties of ROC

Property (7) -- If the z -transform $X(z)$ of $x[n]$ is rational, then its ROC is bounded by poles or it extends to infinity.

Property (8) -- If the z -transform $X(z)$ of $x[n]$ is rational, and if $x[n]$ is right-sided, then the ROC is the region in the z -plane outside the outermost pole -- i.e., outside the circle of radius equal to the largest magnitude of the poles of $X(z)$.
Furthermore, if $x[n]$ is causal (i.e., right-sided and equal to 0 for $n < 0$), then the ROC also includes $z = \infty$.

Property (9) -- If the z -transform $X(z)$ of $x[n]$ is rational, and if $x[n]$ is left-sided, then the ROC is the region in the z -plane inside the innermost pole -- i.e., inside the circle of radius equal to the smallest magnitude of the poles of $X(z)$ other than any at $z = 0$ and extending inward to and possibly including $z = 0$.
In particular, if $x[n]$ is anti-causal (i.e., left-sided and equal to 0 for $n > 0$), then the ROC also includes $z = 0$.

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