



The magnitude-squared response of Chebyshev-I filter

$$|H_a(j\Omega)|^2 = \frac{1}{1 + \varepsilon^2 T_N^2\left(\frac{\Omega}{\Omega_c}\right)}$$

N is the order of the filter,

Epsilon is the passband ripple factor

$$T_N(x) = \begin{cases} \cos(N \cos^{-1}(x)), & 0 \leq x \leq 1 \\ \cosh(\cosh^{-1}(x)), & 1 < x < \infty \end{cases}$$

Nth-order Chebyshev polynomial

(a) For $0 < x < 1$, $T_N(x)$ oscillates between -1 and 1 , and

(b) For $1 < x < \infty$, $T_N(x)$ increases monotonically to infinity

Figure on P.314 (two possible shapes)

Observations: P.315



Observations

At $x=0$ (or $\Omega=0$); $|H_a(j0)|^2 = 1$; for N odd;
 $= 1/(1+\epsilon^2)$; for N even

At $x=1$ (or $\Omega = \Omega_c$); $|H_a(j1)|^2 = 1/(1+\epsilon^2)$ for all N .

For $0 \leq x \leq 1$ (or $0 \leq \Omega \leq \Omega_c$)
 $|H_a(jx)|^2$ oscillates between 1 and $1/(1+\epsilon^2)$

For $x > 1$ (or $\Omega > \Omega_c$), $|H_a(jx)|^2$ decreases monotonically to 0 .

At $x = \Omega_r$, $|H_a(jx)|^2 = 1/(A^2)$.




Causal and stable $H_a(s)$

To determine a *causal and stable* $H_a(s)$, we must find the poles of $H_a(s)H_a(-s)$ and select the *left half-plane* poles for $H_a(s)$.

The poles of $H_a(s)H_a(-s)$ are obtained by finding the roots of

$$1 + \varepsilon^2 T_N^2 \left(\frac{s}{j\Omega_c} \right)$$

It can be shown that if $p_k = \sigma_k + j\Omega_k, k = 0, 1, \dots, N-1$ are the (left half-plane) roots of the above polynomial, then



$$p_k = \sigma_k + j\Omega_k, k = 0, 1, \dots, N-1$$

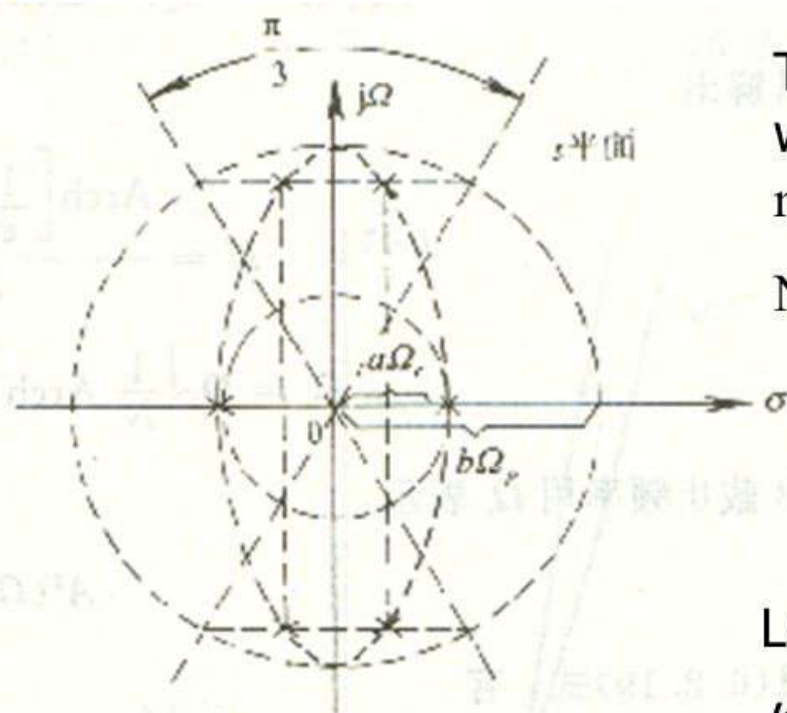
$$\sigma_k = (a\Omega_c) \cos\left[\frac{\pi}{2} + \frac{(2k+1)\pi}{2N}\right]$$

$$\Omega_k = (b\Omega_c) \sin\left[\frac{\pi}{2} + \frac{(2k+1)\pi}{2N}\right]$$

$$a = \frac{1}{2} \left(\sqrt[N]{\alpha} - \sqrt[N]{1/\alpha} \right), b = \frac{1}{2} \left(\sqrt[N]{\alpha} + \sqrt[N]{1/\alpha} \right),$$

$$\alpha = \frac{1}{\varepsilon} + \sqrt{1 + \frac{1}{\varepsilon^2}}$$

The poles of $H_a(s)H_a(-s)$



The poles fall on an **ellipse** with major axis $b \Omega_c$ and minor axis $a \Omega_c$.

Now the **system function** is

$$H_a(s) = \frac{K}{\prod_k (s - p_k)}$$

Left half-plane

K is a normalizing factor



Chebyshev-II filter

Related to the Chebyshev-I filter through a simple transformation.

It has a **monotone** passband and an **equiripple** stopband, which implies that this filter has **both poles and zeros** in the s-plane.

Therefore the **group delay** characteristics are **better** (and the phase response more **linear**) in the passband than the Chebyshev-I prototype.

$$|H_a(j\Omega)|^2 = \frac{1}{1 + \left(\varepsilon^2 T_N^2(\Omega_c / \Omega)\right)^{-1}}$$

The magnitude-squared response

$$|H_a(j\Omega)|^2 = \frac{1}{1 + \varepsilon^2 U_N^2\left(\frac{\Omega}{\Omega_c}\right)}$$

N: the order; epsilon: passband ripple; $U_N()$ is the Nth order **Jacobian** elliptic function

Typical responses for odd and even N are shown on **P.323**

Computation of filter order N:

$$N = \frac{K(k)K(\sqrt{1-k_1^2})}{K(k_1)K(\sqrt{1-k^2})}, k = \frac{\Omega_p}{\Omega_s}, k_1 = \frac{\varepsilon}{\sqrt{A^2-1}}, K(x) = \int_0^{\pi/2} \frac{d\theta}{\sqrt{1-x^2 \sin^2 \theta}}$$