



## **Department of Electronics & Electrical Engineering**





# Lecture 8

Network Theorems for Sinusoidal  
Steady State Analysis

By

Dr. Praveen Kumar

Associate Professor

Department of Electronics & Electrical Engineering





# Steps to Analyze AC Circuits

- The basic steps involved in applying network theorems to AC circuits are

**Step1:** Transform the circuit to the phasor or frequency domain

**Step 2:** Solve the problem using the circuit techniques such as nodal, analysis, mesh analysis, superposition theorem, etc.

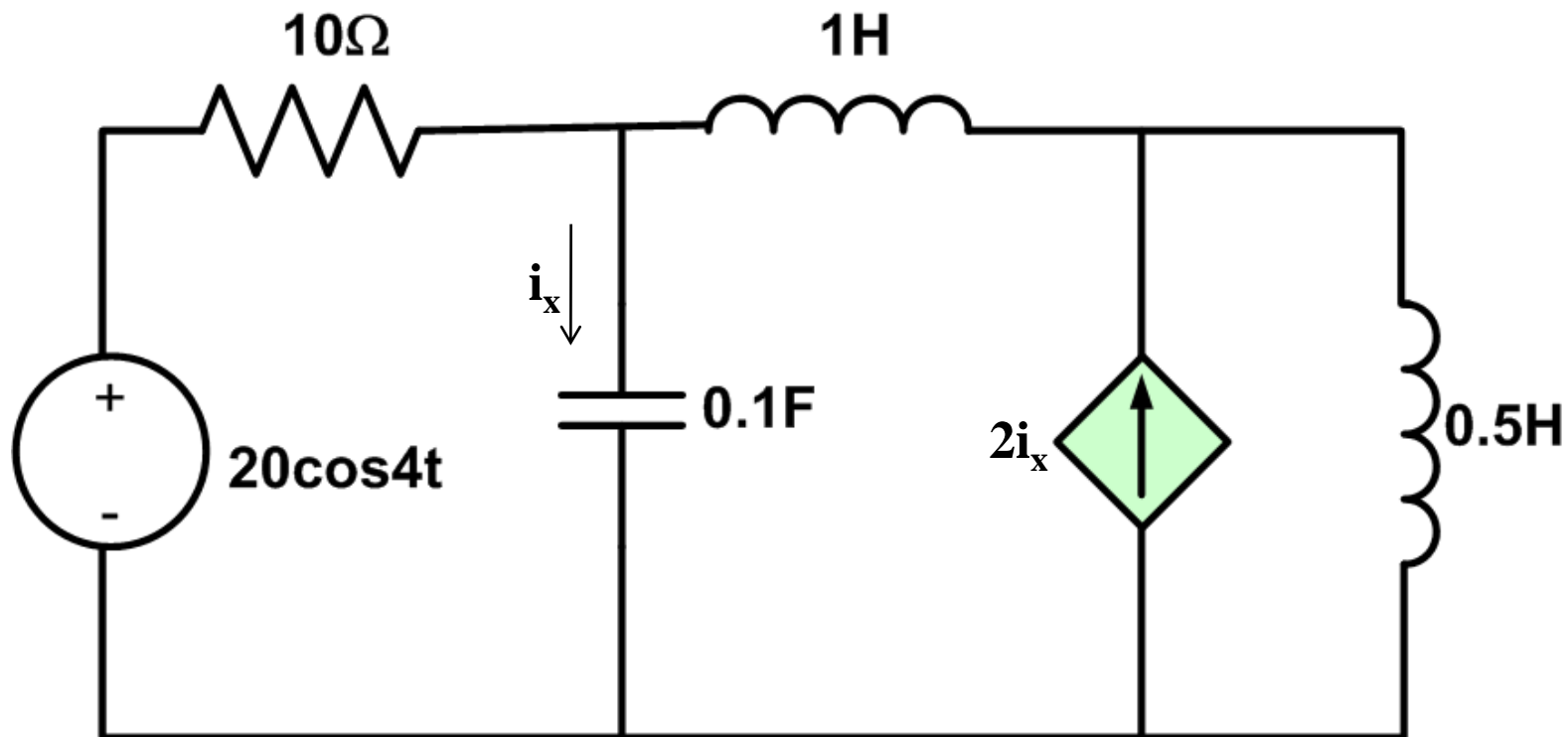
**Step 3:** Transform the resulting phasor to the time domain.

- The Step 1 is not necessary if the problem is specified in the frequency domain.
- In Step 2, the analysis is performed in the same manner as dc circuit analysis except that complex numbers are involved.





# Nodal Analysis



**Fig.1:** Network for Nodal Analysis



# Nodal Analysis

- Convert the entire circuit to the frequency domain

$$20 \cos 4t \Rightarrow 20 \angle 0^\circ, 1H \Rightarrow j\omega L = j4$$

$$0.5H \Rightarrow j\omega L = j2, 0.1F \Rightarrow \frac{1}{j\omega L} = -j2.5$$

where  $\omega = 4 \text{ rad / s}$

- The frequency domain equivalent circuit is shown in Fig.2.

- Applying KCL at node 1

$$\frac{20 - V_1}{10} = \frac{V_1}{-j2.5} + \frac{V_1 - V_2}{j4} \quad (1)$$

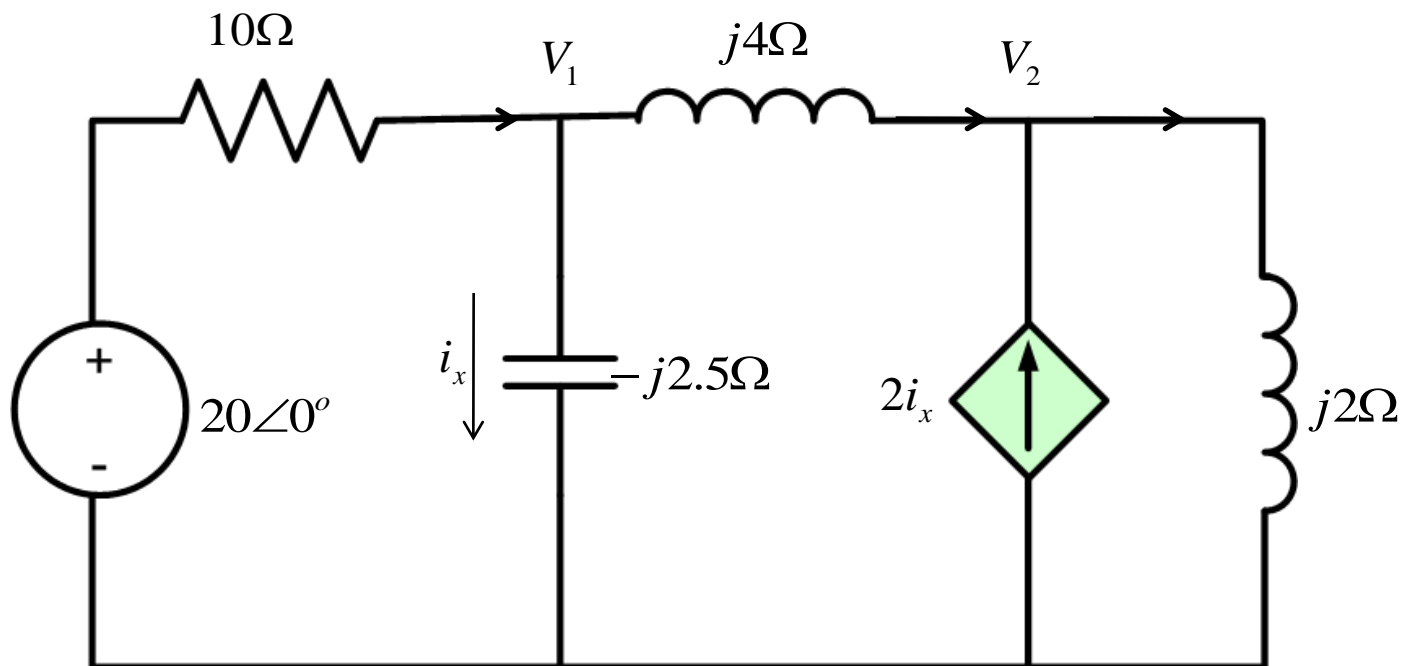
- The KCL at node 2

$$2I_x + \frac{V_1 - V_2}{j4} = \frac{V_2}{j2} \quad (2)$$





# Nodal Analysis



**Fig.2:** Frequency domain equivalent of the circuit in Fig.1



# Nodal Analysis

- Solution of eq.1 and eq.2 gives

$$V_1 = 18.97 \angle 18.43^\circ \text{ V}$$

$$V_2 = 13.91 \angle 198.3^\circ \text{ V}$$

- The current

$$I_x = \frac{V_1}{-j2.5} = \frac{18.97 \angle 18.43^\circ}{2.5 \angle -90^\circ} = 7.59 \angle 108.4^\circ \text{ A}$$

- Transforming  $i_x$  to the time domain gives

$$i_x = 7.59 \cos(4t + 108.4^\circ) \text{ A} \quad (3)$$



# Mesh Analysis

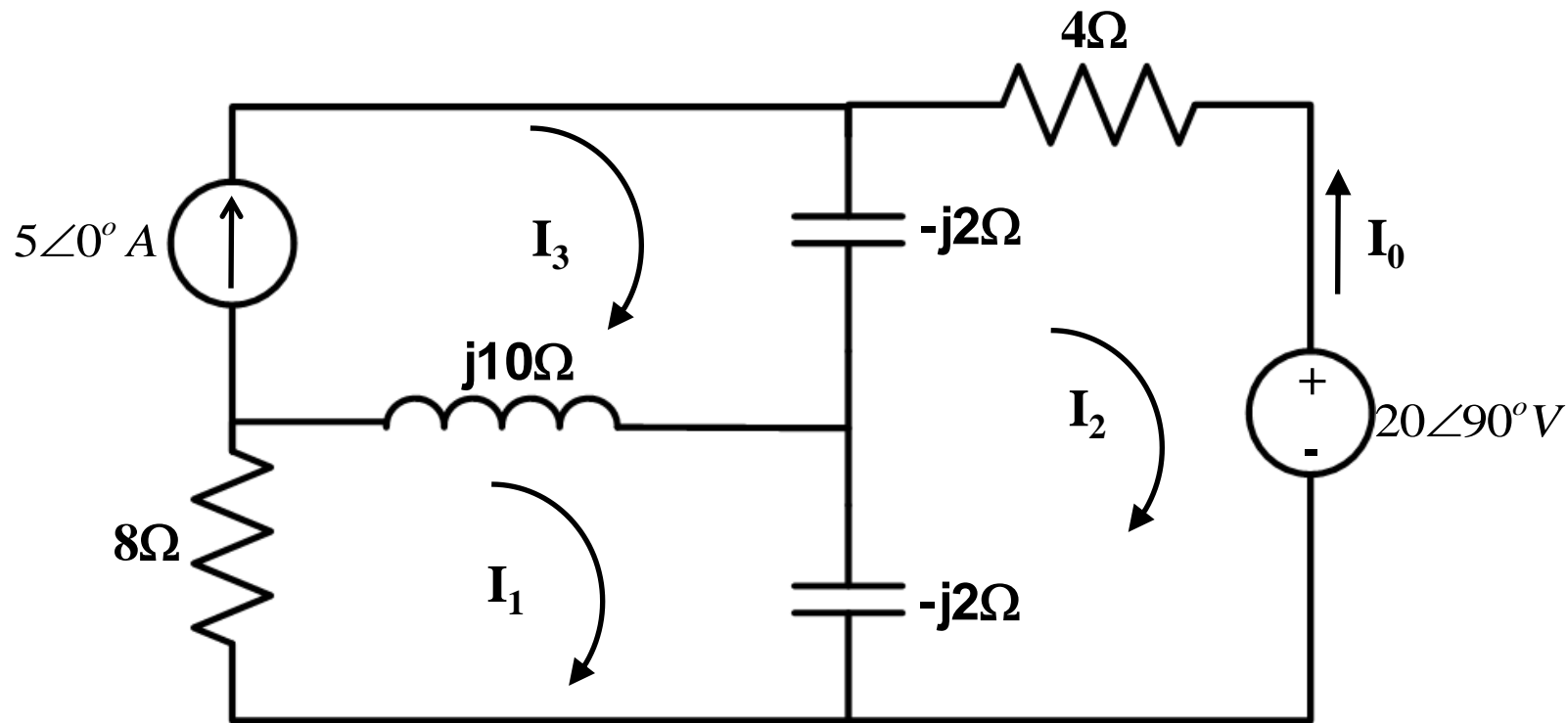


Fig.3: The network for nodal analysis





# Mesh Analysis

- Applying KVL to mesh 1

$$(8 + j10 - j2)I_1 - (-j2)I_2 - j10I_3 = 0 \quad (3)$$

- The KVL for mesh 2 is

$$(4 - j2 - j2)I_2 - (-j2)I_1 - (-j2)I_3 + 20\angle 90^\circ = 0 \quad (4)$$

- For mesh 3

$$I_3 = 5 \quad (5)$$

- Substituting eq.5 in eq.3 and eq.4 gives

$$(8 + j8)I_1 + j2I_2 = j50 \quad (6)$$

$$j2I_1 + (4 - j4)I_2 = -j20 - j10 \quad (7)$$

- Solving eq.6. and eq.7 gives

$$I_2 = 6.12\angle -35.22^\circ \text{ A}$$

$$\Rightarrow I_0 = -I_2 = 6.12\angle 144.78^\circ \text{ A}$$





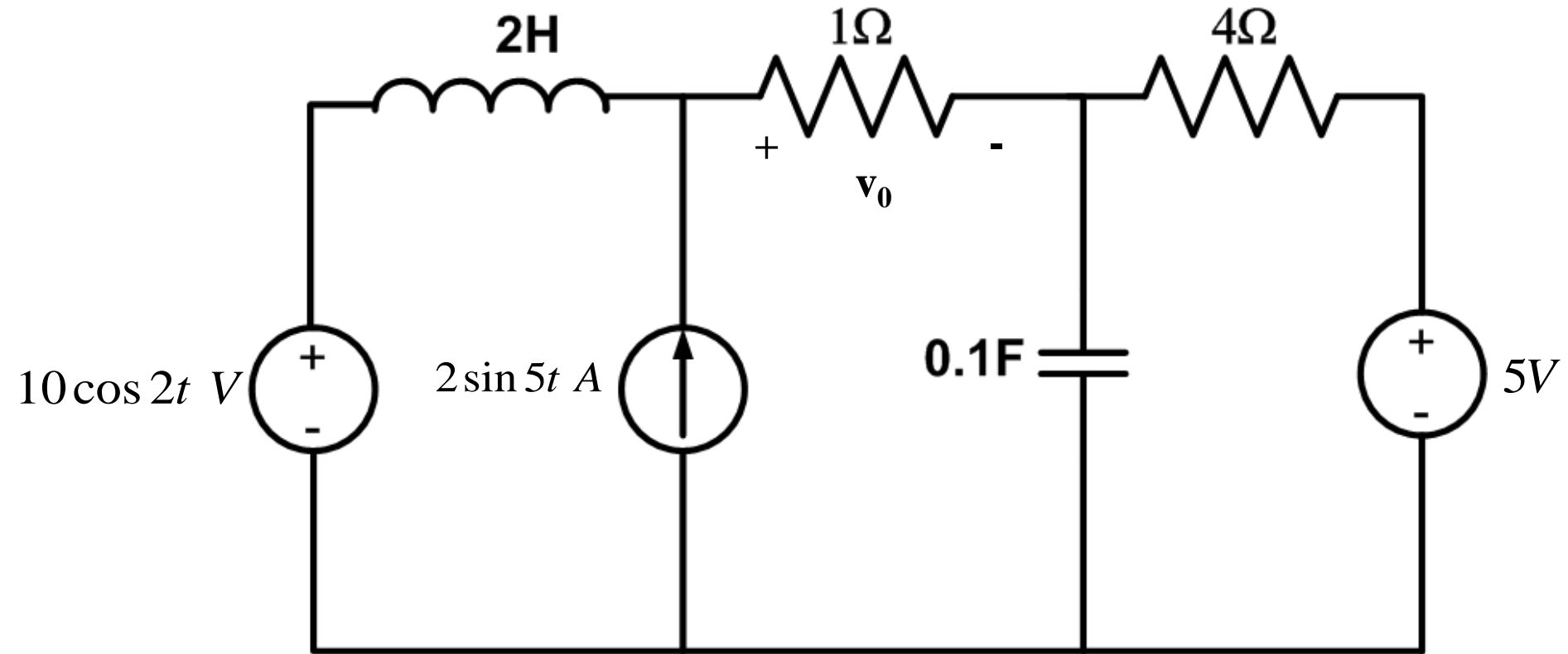
# Superposition Theorem

- Since AC circuits are linear, the superposition theorem applies to AC circuits the same way it applies to dc circuits.
- *The theorem becomes important if the circuit has sources operating at different frequencies. In this case, since the impedances depend in frequency, it is required to have a different frequency domain circuit for each frequency.*
- In case of sources with different frequencies, the total response must be obtained by adding the individual responses in time domain. It is incorrect to add the responses in the phasor or frequency domain





# Superposition Theorem



**Fig.4:** The Network for Superposition Theorem



# Superposition Theorem

- Since the circuit operates at three different frequencies, the problem is divided into single frequency problems. Let

$$v_0 = v_1 + v_2 + v_3 \quad (8)$$

where

$v_1$  is due to 5 V dc voltage source

$v_2$  is due to the  $10\cos 2t$  V voltage source

$v_3$  is due to the  $2\sin 5t$  A current source

- To find  $v_1$ , set all sources except the 5V dc source. In steady state, a capacitor is an open circuit to dc while an inductor is a short circuit to dc. The equivalent circuit is shown Fig.5a. By voltage division

$$-v_1 = \frac{1}{1+4} \times 5 = 1V \quad (9)$$





# Superposition Theorem

- To find  $v_2$ , the 5v voltage source is open circuited and the current source is short circuited. The equivalent circuit is shown in Fig.5b.

$$10\cos 2t \Rightarrow 10\angle 0^\circ, \omega = 2\text{rad/s}$$

$$2H \Rightarrow j\omega L = j4\Omega$$

$$0.1F \Rightarrow \frac{1}{j\omega C} = -j5\Omega$$

The parallel combination of  $-j5\Omega$  and  $4\Omega$  is

$$Z = \frac{-j5 \times 4}{4 - j5} = 2.439 - j1.951$$

- By voltage division

$$V_2 = \frac{1}{1 + j4 + Z} \times (10\angle 0^\circ) = \frac{10}{3.439 + j2.049} = 2.498\angle -30.79^\circ$$

In time domain

$$v_2 = 2.498\cos(2t - 30.79^\circ)$$

(10)





# Superposition Theorem

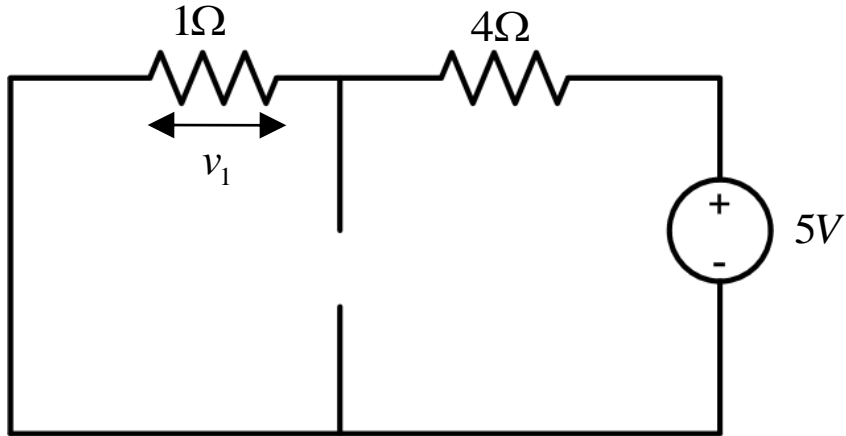


Fig.5a:

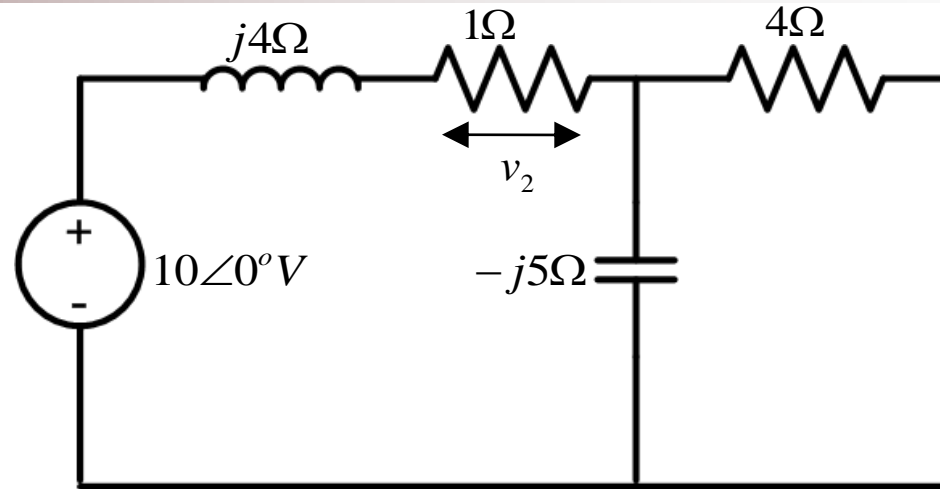


Fig.5b:

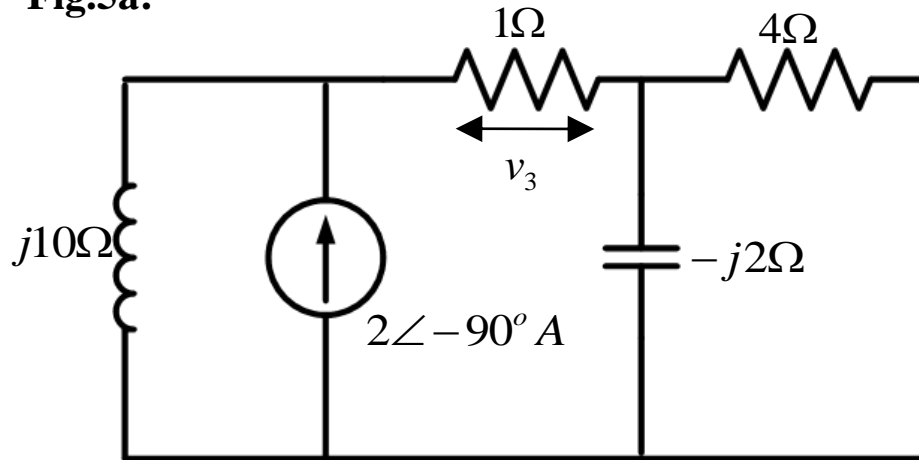


Fig.5c:



# Superposition Theorem

- To obtain  $v_3$ , set the voltage sources to zero (Fig.5c) and transform what is left to the frequency domain

$$2 \cos 5t \Rightarrow 2 \angle -90^\circ, \omega = 5 \text{ rad/s}$$

$$2H \Rightarrow j\omega L = j10\Omega$$

$$0.1F \Rightarrow \frac{1}{j\omega C} = -j2\Omega$$

The parallel combination of  $-j2\Omega$  and  $4\Omega$  is

$$Z = \frac{-j2 \times 4}{4 - j2} = 0.8 - j1.6$$

- By current division

$$I_3 = \frac{j10}{j10 + 1 + Z} (2 \angle -90^\circ) \text{ A}, \quad v_3 = I_1 \times 1 = \frac{j10}{1.8 + j8.4} (-j2) = 2.328 \angle -80^\circ \text{ V} \quad (11)$$

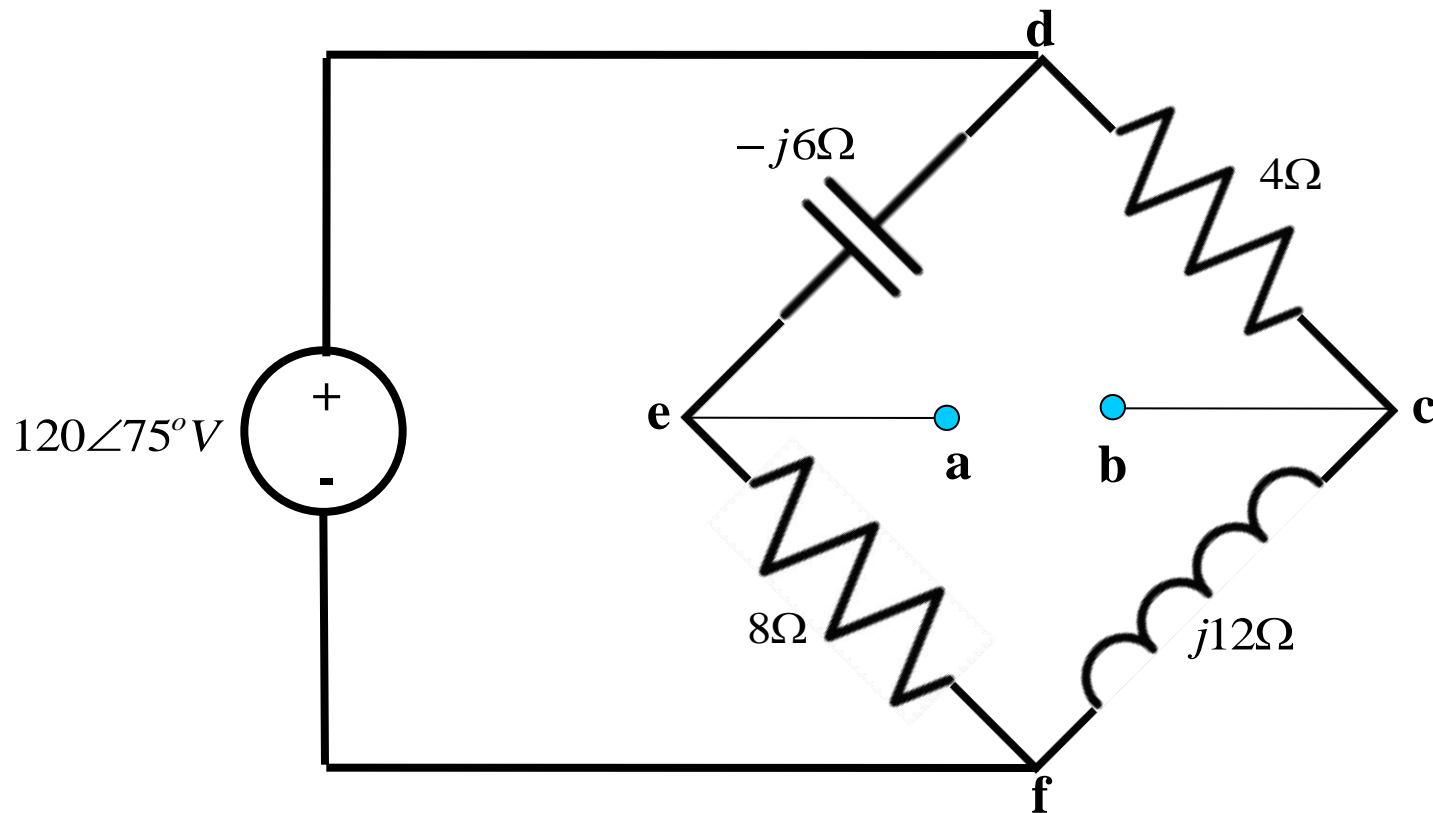
- The final output is summation of eq.9, eq.10 and eq.11

$$v_0(t) = -1 + 2.498 \cos(2t - 30.79^\circ) + 2.33 \sin(5t + 10^\circ) \text{ V} \quad (12)$$





# Thevenin and Norton Equivalent Circuits



**Fig.6:** The Network for Thevenin and Norton's Theorem





# Thevenin and Norton's Theorem

- The value of Thevenin's impedance  $Z_{th}$  is obtained by setting the voltage source to zero, Fig.7a. From Fig.7a it is seen that  $8\Omega$  and  $-j6\Omega$  are in parallel:

$$Z_1 = \frac{-j6 \times 8}{8 - j6} = 2.88 - j3.84\Omega \quad (13)$$

- The  $4\Omega$  resistance is in parallel with the  $j12$  reactance:

$$Z_2 = \frac{j12 \times 4}{4 + j12} = 3.6 + j1.2\Omega \quad (14)$$

- The Thevenin impedance is the series combination of  $Z_1$  and  $Z_2$ , i.e.

$$Z_{th} = Z_1 + Z_2 = 6.48 - j2.64 \quad (15)$$

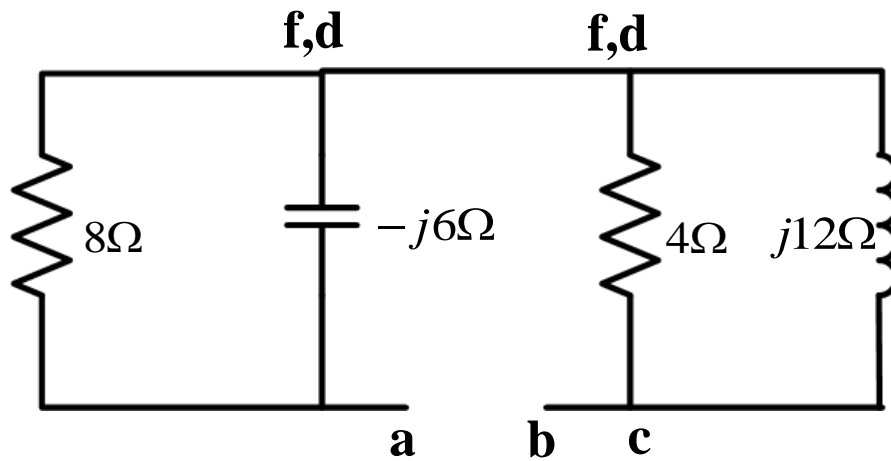
- To find  $V_{th}$ , consider the circuit in Fig.7b, currents  $I_1$  and  $I_2$  are obtained as

$$I_1 = \frac{120\angle 75^\circ}{8 - j6} \text{ A}, \quad I_2 = \frac{120\angle 75^\circ}{4 + j12} \quad (16)$$

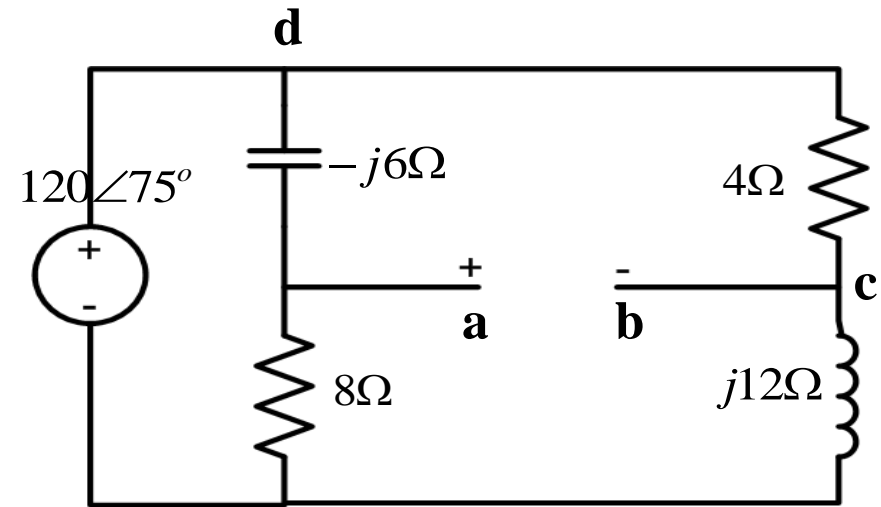




# Thevenin and Norton's Theorem



**Fig.7a:** The Network for Thevenin Theorem for finding  $Z_{th}$



**Fig.7b:** The Network for Thevenin Theorem for finding  $V_{th}$



# Thevenin and Norton's Theorem

- Applying KVL around loop **bcdeab** in Fig.7b gives

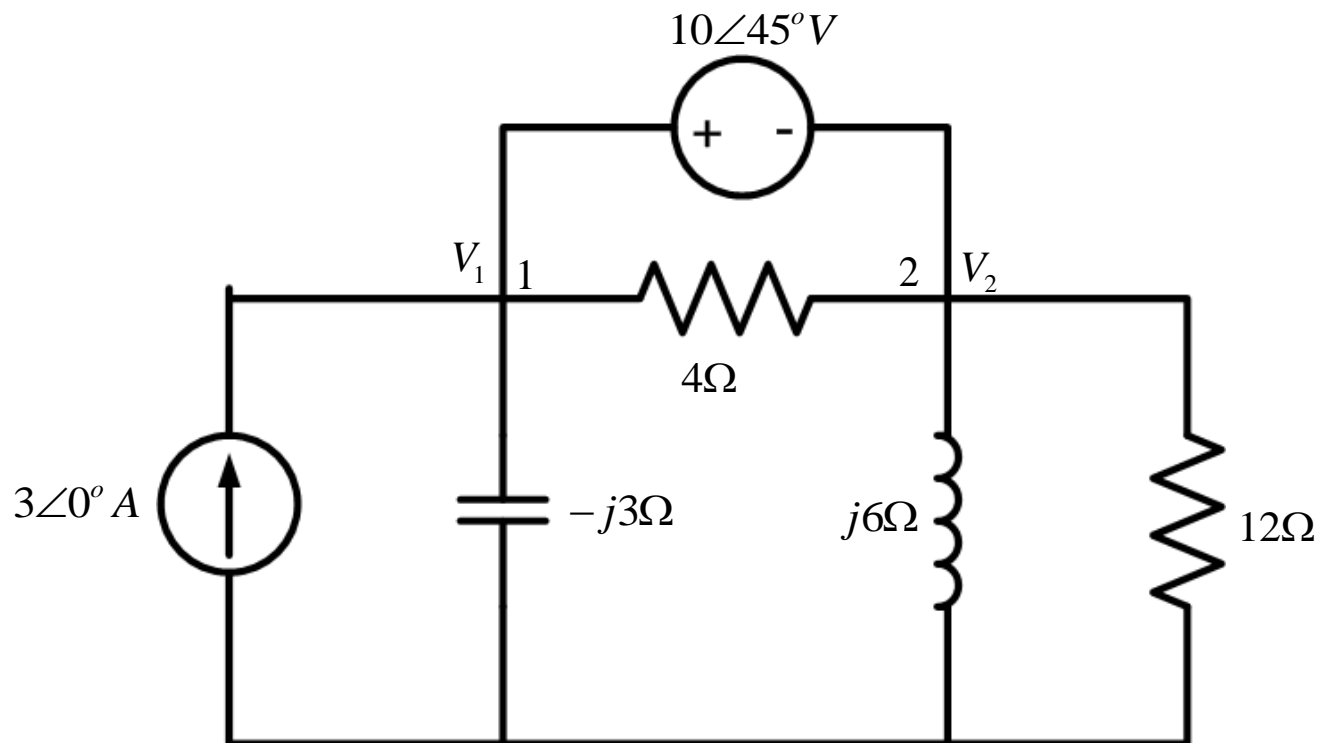
$$V_{th} - 4I_2 + (-j6)I_1 = 0 \quad (17)$$

$$\begin{aligned} V_{th} &= 4I_2 + j6I_1 = \frac{480\angle 75^\circ}{4 + j12} + \frac{720\angle 75^\circ + 90^\circ}{8 - j6} \\ &= 37.95\angle 3.43^\circ + 72\angle 201.87^\circ = 37.95\angle 220.31^\circ \text{ V} \end{aligned}$$



# Example 1

- Compute  $V_1$  and  $V_2$  in the network shown in Fig.8



**Fig.8:** The Network for Example 1



# Example 1

- Node 1 and Node 2 form a supernode, Fig.9.

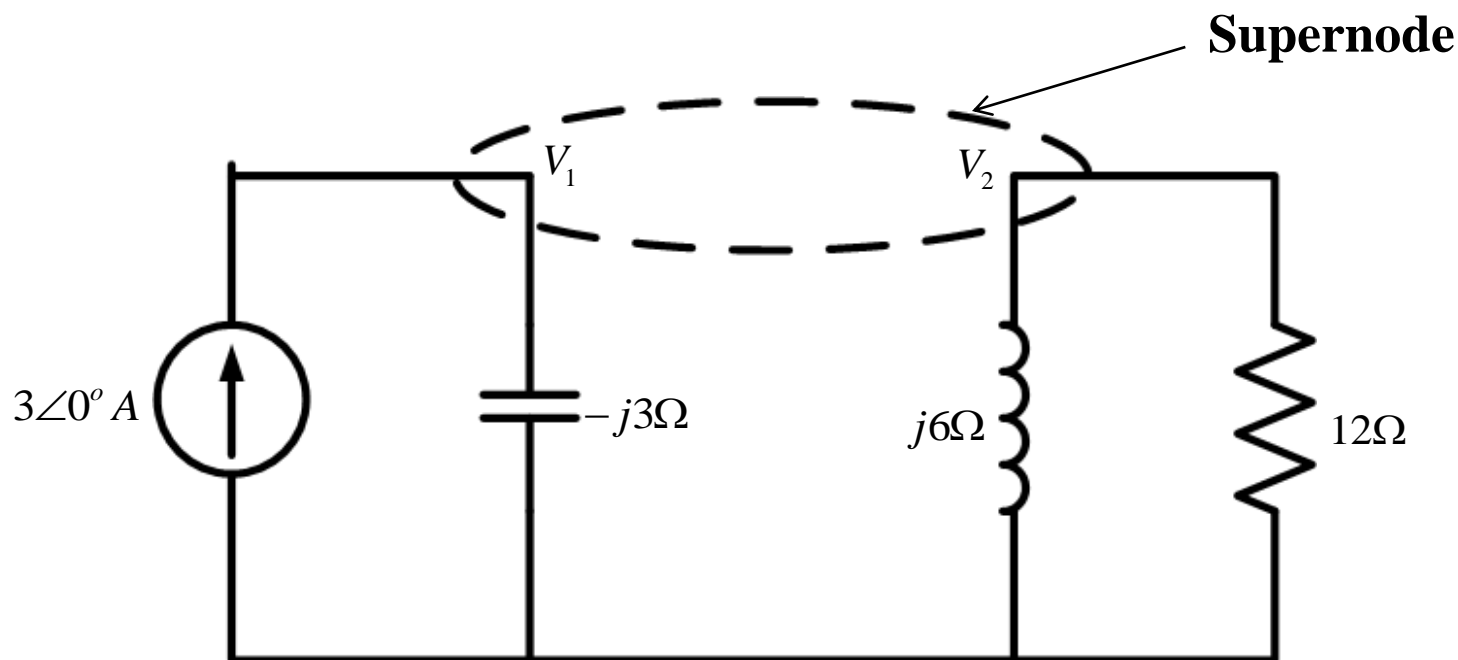


Fig.9: Supernode



# Example 1

- Applying KCL to supernode gives

$$3 = \frac{V_1}{-j3} + \frac{V_2}{j6} + \frac{V_2}{12}$$
$$36 = j4V_1 + (1 - j2)V_2 \quad (18)$$

- A voltage source is connected between nodes 1 and 2, hence

$$V_1 = V_2 + 10\angle 45^\circ \quad (19)$$

- Substituting eq.19 into eq.18 gives

$$36 - 40\angle 135^\circ = (1 + j2)V_2$$
$$V_2 = 31.41\angle -87.18^\circ \text{ V}$$
$$V_1 = V_2 + 10\angle 45^\circ = 25.78\angle -70.48^\circ$$