

EE101
Solution Tutorial 4 (30-AUG-2013)

1. (a)

(i) Assuming the transistor to be in the active region –

$$V_{CC} = R_C(I_C + I_B) + I_B R_B + 0.7 + I_E R_E$$

$$I_B = \frac{10 - 0.7}{R_C(\beta + 1) + R_B + (\beta + 1)R_E}$$

$$= \frac{9.3}{101 \times 4.7 + 250 + 101 \times 1.2} = 0.011 \text{ mA}$$

$$I_C = 1.1 \text{ mA}$$

$$V_C = 10 - 4.7 \times (1.1 + 0.011) = 4.78 \text{ V}$$

$$V_E = 1.2 \times 101 \times (0.011) = 1.33 \text{ V}$$

$$\text{Therefore } V_{CE} = V_C - V_E = 3.45 \text{ V}$$

Note that $V_B = 2.03 \text{ V}$ implying that the C-B junction is reverse biased as it should be for the transistor to operate in the active region.

(ii) Assuming the transistor to be in the active region –

Thevenin's Equivalent of the Base Voltage supply gives

$$V_{BB} = \frac{10}{3} \text{ V} \quad R_B = R_1 \parallel R_2 = 13.333 \text{ K}\Omega$$

$$V_{BB} = I_B R_B + 0.7 + I_E R_E$$

$$I_B = \frac{3.333 - 0.7}{13.333 + (101)1.2} = 0.0196 \text{ mA}$$

$$I_C = 1.96 \text{ mA}$$

$$V_C = 10 - 2.8 \times 1.96 = 4.51 \text{ V}$$

$$V_E = 1.2 \times (101) \times 0.0196 = 2.38 \text{ V}$$

$$V_B = V_E + 0.7 = 3.08 \text{ V}$$

$$\text{Therefore } V_{CE} = V_C - V_E = 2.13 \text{ V}$$

Note that B-C junction will be reverse biased so transistor is indeed in the active region

(a) In this case, $V_{BB} = 5 \text{ V}$ and $R_B = 10 \text{ K}\Omega$.

If transistor is assumed to be in the active region, then –

$$I_B = \frac{5 - 0.7}{10 + 101 \times 1.2} = 0.0328 \text{ mA} \quad I_C = 3.28 \text{ mA} \quad I_E = 3.313 \text{ mA}$$

$$\text{and } V_E = 3.976 \text{ V } V_B = 4.676 \text{ V } V_C = 0.816 \text{ V}$$

But this would make B-C forward biased which is clearly impossible. Therefore, **transistor cannot be in the active region.**

If transistor is assumed to be in the saturation region, then –

$$5 - 0.7 = 10 I_B + 1.2(I_B + I_C) \quad 11.2 I_B + 1.2 I_C = 4.3$$

$$10 - 0.1 = 2.8 I_C + 1.2(I_B + I_C) \quad 1.2 I_B + 4 I_C = 9.9$$

$$\text{Solving, we get } I_B = 0.122 \text{ mA } I_C = 2.44 \text{ mA } I_E = 2.562 \text{ mA}$$

Note that $I_C < \beta I_B$, therefore **the transistor is indeed in saturation**

2. (a) Time constant $\tau = RC = (56\text{k}\Omega)(0.1\mu\text{F}) = 5.6\text{ms}$

(b) $\frac{T}{2} = \frac{1\text{ms}}{2} = 0.5\text{ms} \ll 5\tau = 28\text{ms}$ and the ratio is 1:56

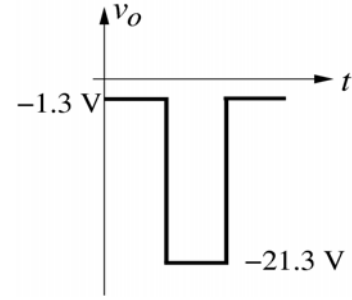
(c) For positive pulse of v_i :

The diode is "ON" and $v_o = -2\text{V} + 0.7\text{V} = -1.3\text{V}$

The capacitor charges to $10\text{V} + 2\text{V} - 0.7\text{V} = 11.3\text{V}$

For negative pulse of v_i :

The diode is "OFF" and $v_o = -10\text{V} - 11.3\text{V} = -21.3\text{V}$



3. (a) $I_E = \frac{V_E}{R_E} = \frac{2.1\text{V}}{0.68\text{k}\Omega} = 3.09\text{mA}$, assuming $I_C \approx I_E$ $\beta = \frac{I_C}{I_B} = \frac{3.09\text{mA}}{20\mu\text{A}} = 154.5$

(b) $V_{CC} = V_{R_C} + V_{CE} + V_E = (2.7\text{k}\Omega)(3.09\text{mA}) + 7.3\text{V} + 2.1\text{V} = 17.74\text{V}$

(c) $R_B = \frac{V_{R_B}}{I_B} = \frac{V_{CC} - V_{BE} - V_E}{I_B} = \frac{17.74\text{V} - 0.7\text{V} - 2.1\text{V}}{20\mu\text{A}} = \frac{14.94\text{V}}{20\mu\text{A}} = 747\Omega$

4. (a) $P_1 = \frac{v_m i_{1m}}{2} \cos \theta_1 = \frac{141 \times 7.07}{2} \cos 60^\circ = 250 \text{ W}$,

$P_2 = \frac{v_m i_{2m}}{2} \cos \theta_2 = \frac{141 \times 10}{2} \cos 30^\circ = 610.5 \text{ W}$

Total Power = $250 + 610.5 = 860.5 \text{ W}$

(b) $V = (141/\sqrt{2})\angle 0^\circ = 100\angle 0^\circ \text{ V}$,

$I_2 = (10/\sqrt{2})\angle 30^\circ = 7.07\angle 30^\circ \text{ A}$;

Apparent power, $S = 100 \times 8.66 = 866 \text{ VA}$

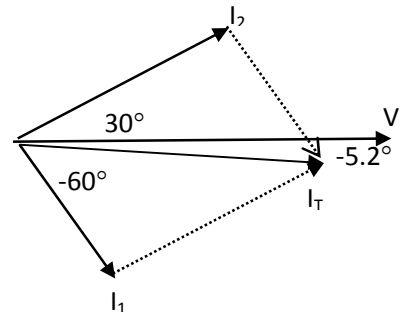
Power factor = $\cos(-5.2^\circ) = 0.996$ (lag)

$I_1 = (7.07/\sqrt{2})\angle -60^\circ = 5\angle -60^\circ \text{ A}$,

$I_T = 5\angle -60^\circ + 7.07\angle 30^\circ = 8.66\angle -5.2^\circ \text{ A}$

(c) Expression for total current is: $i_T = 8.66 \times \sqrt{2} \sin(\omega t - 5.2^\circ) \text{ A}$

(d) The vector diagram is as shown.



5. $Z = R + j\omega L + \frac{1}{j\omega C}$. Therefore, $|Z| = \sqrt{R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2}$

The current maximum occurs at $\omega = \omega_0$, when $\left(\omega_0 L - \frac{1}{\omega_0 C}\right) = 0$ or $\omega_0 = \frac{1}{\sqrt{LC}}$

$$\cos \theta = \frac{R}{|Z|} = \frac{R}{\sqrt{R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2}} = \frac{R}{\sqrt{R^2 + \omega L \left(1 - \frac{\omega_0^2}{\omega^2}\right)^2}}$$

Given that at $\omega = 2\omega_0$, $\cos \theta = \frac{1}{\sqrt{2}}$. Therefore, $R^2 + 2\omega_0 L \left(1 - \frac{1}{4}\right)^2 = 2R^2$

$$R^2 = 2 \frac{L}{\sqrt{LC}} \left(\frac{3}{4}\right)^2 = 2 \sqrt{\frac{L}{C}} \left(\frac{3}{4}\right)^2 = 4 \left(\frac{3}{4}\right)^2 \Rightarrow R = \frac{3}{2} \Omega$$