

# EE101

## Tutorial 1 and Tutorial 2 (16-AUG-2013 and 20-AUG-2013) Solutions

1. This problem has been done in class assuming ideal diodes. Here, we want you to redo the calculations when the diode is not ideal, i.e. has a forward voltage drop of 0.7 volts.

For  $0 \leq V_i \leq 2.7$  V BOTH D1 and D2 are OFF as they are reverse biased  $V_O = V_i$

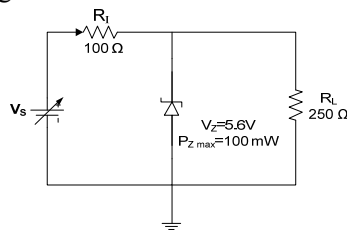
For  $V_i > 2.7$  V D2 is OFF but D1 turns ON  $i_{R1} = \frac{V_i - (2 + 0.7)}{R1 + R2} = \frac{V_i - 2.7}{20}$  mA

$$V_O = i_{R1} R_2 + 2.7 = \left( \frac{V_i - 2.7}{20} \right) 10 + 2.7 \quad V_O = 0.5V_i + 1.35$$

For  $-4.7 \leq V_i \leq 0$  V BOTH D1 and D2 are OFF as they are reverse biased  $V_O = V_i$

For  $V_i < -4.7$  V D2 is ON but D1 stays OFF  $V_O = -4.7$  V

2. (a) For the circuit shown below, what is the maximum value of the source voltage  $V_S$  for which the voltage across the load resistance  $R_L$  can be maintained at 5.6V?



Since  $P_Z = V_Z I_Z$  where  $I_Z$  is the current through the zener diode, we have  $I_{Z, MAX} = 100/5.6 = 17.86$  mA

Since  $I_{RL} = 5.6/0.250 = 22.4$  mA, we have  $I_{RI MAX} = I_{Z MAX} + I_{RL} = 17.86 + 22.4 = 40.26$  mA

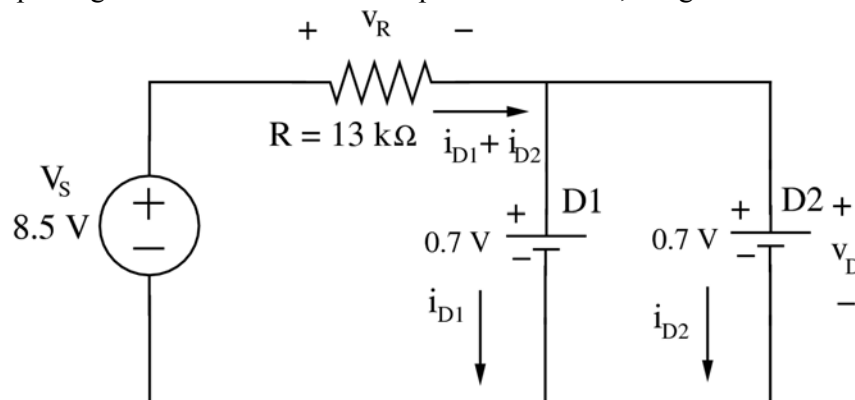
Therefore,  $V_{S MAX} = I_{RI MAX} R_I + V_Z = 40.26 * 0.1 + 5.6 = 9.63$  V

- (b) If the Zener diode is such that a minimum current of 1 mA is required for the Zener action to take place, what is the minimum source voltage  $V_S$  that can be used?

When the zener diode is drawing minimum current, we have  $I_{RI MIN} = 1 + I_{RL} = 23.4$  mA

Therefore,  $V_{S MIN} = I_{RI MIN} R_I + V_Z = 23.4 * 0.1 + 5.6 = 7.94$  V

3. On replacing the diodes with their simple on/off model, the given circuit reduces to as below:



Now, as the diodes are in parallel, the voltage across each of the diode would be

$$v_D = 0.7 \text{ V}$$

The voltage drop across the resistor (R) be

$$v_R = V_S - v_D = 8.5 - 0.7 = 7.8 \text{ V}$$

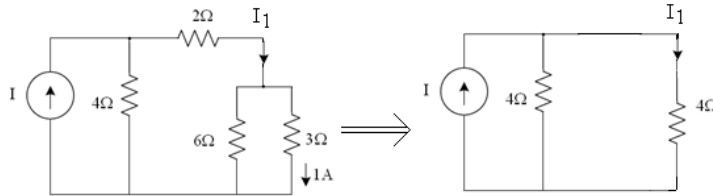
The total diode current (= resistor current) be

$$i_{D1} + i_{D2} = i_R = v_R / R = 7.8 / 13 \text{ k}\Omega = 0.6 \text{ mA}$$

Note that in the simple on/off model, the forward resistance of diode is assume to be zero so we cannot find the individual diode current for diodes connected in parallel.

4. Considering the division of current,  $1 = \frac{6}{6+3} I_1 \quad \therefore I_1 = \frac{3}{2} \text{ A}$

Further, the circuit can be simplified as,



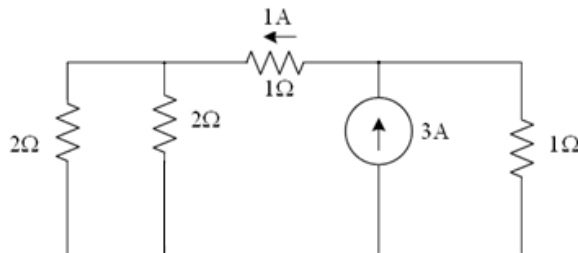
Therefore,  $I = 2I_1 = 3 \text{ A}$

Power dissipated in 4 ohm register,  $P = ((I/2)^2) \cdot 4 = 9 \text{ watt}$

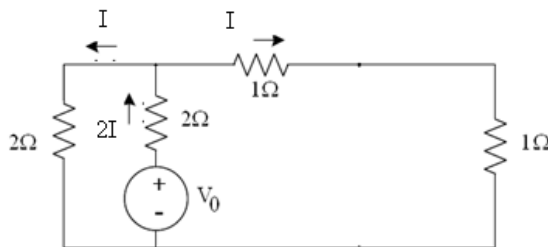
5. We apply superposition.

When the voltage source is short circuited, a current of 1 A as shown flows

$$\left( 3 \times \frac{1}{(2 \parallel 2 + 1) + 1} = 1 \text{ A} \right).$$



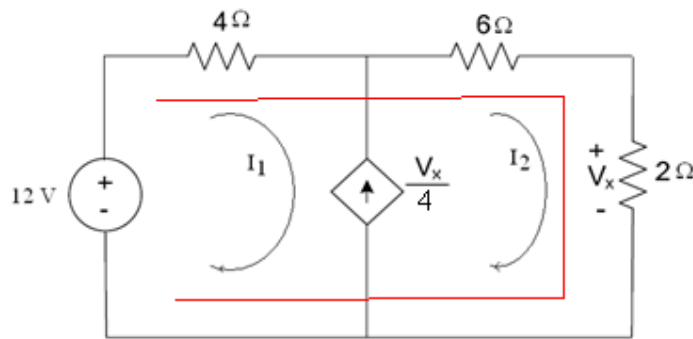
When the current source is open circuited, the situation is shown in the figure below



When both the sources are present, if the net current in the 1 Ohm resistance is to be 1 A as stated in the problem, I has to be 2 A.

Therefore, applying KVL in the figure above,  $V_0 = 2 \times 4 + 2 \times 2 = 12 \text{ V}$

6. Since we have common current source between the meshes, we apply the concept of super-mesh as shown



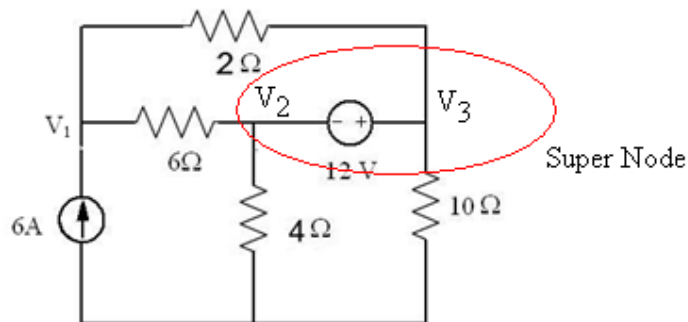
Applying KVL,  $12 - 4I_1 - 6I_2 - V_x = 0$

We also have  $V_x = 2I_2$  and  $I_2 - I_1 = \frac{V_x}{4}$

Reducing further,  $12 = 4I_1 + 8I_2$  and  $2I_2 - 4I_1 = 0$

Solving the above equations,  $I_1 = \frac{3}{5} \text{ A}$  and  $I_2 = \frac{6}{5} \text{ A}$

7. Since the voltage source appear between two nodes, we form a super node



The equations are as follows:

$$V_2 - V_3 = -12$$

At the node 1 (where node voltage is  $V_1$ ),

$$\frac{V_1 - V_2}{6} + \frac{V_1 - V_3}{2} = 6$$

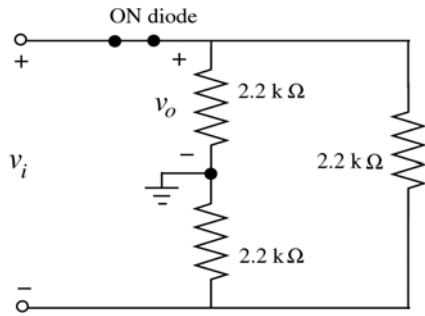
At super node

$$\frac{V_2 - V_1}{6} + \frac{V_2}{4} + \frac{V_3}{10} + \frac{V_3 - V_1}{2} = 0$$

Eliminating  $V_3$  we get,

$$4V_1 - 4V_2 = 72 \text{ and } 40V_1 - 61V_2 = 432. \text{ Solving for } V_1, \text{ we get } V_1 = \frac{666}{21} = 31.714 \text{ V}$$

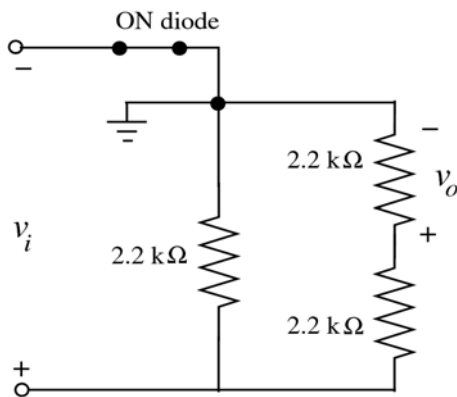
8. For positive half-cycle of  $v_i$ , the given circuit could be redrawn as



On applying the voltage divider rule, we have

$$v_{o_{\max}} = \frac{2.2\text{k}\Omega \times v_{i_{\max}}}{2.2\text{k}\Omega + 2.2\text{k}\Omega} = \frac{v_{i_{\max}}}{2} = \frac{100}{2} = 50\text{V}$$

For negative half-cycle of  $v_i$ , the given circuit could be redrawn as



Again applying the voltage divider rule, we have

$$v_{o_{\max}} = \frac{2.2\text{k}\Omega \times v_{i_{\max}}}{2.2\text{k}\Omega + 2.2\text{k}\Omega} = 50\text{V}$$

Note the polarity of  $v_o$  across  $2.2\text{ k}\Omega$  resistor acting as load is the same for both the positive and negative cycles of the input waveform.

Thus the output voltage  $v_o$  be a full-wave rectified version of the input waveform but with peak value  $V_m$  of 50 V.

For the full-wave rectified waveform, the dc voltage  $V_{dc} = 0.636 V_m = 0.636(50) = 31.8\text{V}$

