

Multi-criteria performance assessment based on closed-loop system identification

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Abstract

A method to assess the performance of control loops, based on closed-loop system identification is proposed. This method allows to take into account the trade-off between process variable and manipulated variable energy, thus overcoming one of the most important criticisms to Harris'index. To illustrate the proposed approach, a numerical example is given, for which the proposed index is 8% greater than Harris'index.

Keywords: Control Loop, Performance Assessment

1. Introduction

2 In the process industry sector, companies need to extract actionable information from sensor-based data collected on a daily basis through their integrated
4 IT/OT systems [1]. It is estimated that each operator typically needs to assess the performance of 90-180 control loops [2], and therefore the interest in
6 developing computer-aided tools to facilitate their work. In this framework, timely and accurate assessment may enable the implementation of appropriate
8 corrective actions as soon as required.

A common approach for control loop performance assessment (CPLA) is
10 based on selecting an *ideal benchmark* to which the actual loop can be compared [3]. As an example, if the Minimum Variance (MV) control loop is selected, it
12 is possible to determine the well-known *Harris' index* (HI), using only routine

data, collected during normal operation, without the need of *ad hoc* invasive experiments [4]. This index has been extensively used in commercial applications and extended to assess complex processes [5].

Although numerous of similar techniques are available [6], due to real-world phenomena like non-linearity, unknown disturbances and dynamics, unstable or multivariable loops, etc., this can still be a challenging problem. In this paper we propose an approach a multi-criteria approach for performance assessment. The rationales behind this approach are two:

- It may be necessary to consider different aspects related to loop performance, possibly in contradiction, which is not possible using the classic HI. It is well-known that this index is not realistic because the MV controller is too aggressive. Therefore, in practice we may be largely underestimating loop performance.
- Companies are currently implementing, or willing to implement, computing infrastructure able to run more complex analysis algorithms, compared to past situation.

The structure of this paper is as follows. In section II, problem statement is presented. In section III, the proposed approach is described. In section IV a case study is presented.

2. Problem statement

Let \mathcal{S} be a commissioned stable single-input single-output closed-loop LTI system, as in Fig. 1, where $\varepsilon(k)$ is a disturbance assumed to be unit *white noise*. Based on the data collected during the lapse $k = 0, 1, 2, \dots, N - 1$, we define the loop performance \mathcal{P} to be a discrete variable

$$\mathcal{P} \in \{\text{Good, Fair, Poor}\} \quad (1)$$

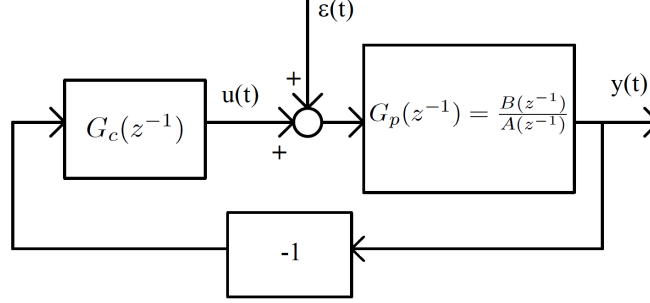


Figure 1: Commissioned control system

Thus, the plant under control is described by the following model

$$y(k) = \frac{G_p(z^{-1})}{1 + G_c(z^{-1})G_p(z^{-1})} \varepsilon(k) \quad (2)$$

$$G_p(z^{-1}) = \frac{B(z^{-1})}{A(z^{-1})} \quad (3)$$

38 We assume that $A(z^{-1})$, $B(z^{-1})$ are unknown polynomials, and n , m their
corresponding degrees. The controller model $G_c(z^{-1})$ is assumed known. We
40 are interested in the following problem: *given \mathcal{S} determine its performance \mathcal{P} ,
by using only the available measurements*

$$y(-N_H) \dots, y(-2), y(-1), y(0), y(1), y(2), \dots, y(N-1) \quad (4)$$

42 where $N_H + N$ is the total number of available measurements, with $N_H \gg N$.

As discussed in the previous section, the classical approach to solve this
44 problem is to determine the *Harris' index* (HI), denoted hereafter η_{MV} . If we
denote σ_{MV}^2 the minimum variance achievable by any linear controller, η_{MV}
46 will be given by the ratio:

$$\eta_{MV} = \frac{\sigma_{MV}^2}{s_y^2} \quad (5)$$

where s_y^2 is the measured output variance

$$s_y^2 = \frac{1}{N-1} \sum_{k=0}^{N-1} [y(k) - m_y]^2 \quad (6)$$

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$$m_y = \frac{1}{N} \sum_{k=0}^{N-1} y(k) \quad (7)$$

The estimation of minimum variance σ_{MV}^2 requires the identification of an
 50 ARMA model (see for example [3])

$$\hat{A}_{cl}(z^{-1})y(k) = \hat{B}_{cl}(z^{-1})\varepsilon(k) \quad (8)$$

from historical data $y(-N_H)\dots, y(-2), y(-1), y(0)$, and then the closed-loop
 52 transfer function is given by

$$\hat{G}_{cl}(z^{-1}) = \frac{\hat{B}_{cl}(z^{-1})}{\hat{A}_{cl}(z^{-1})} \quad (9)$$

Finally, the loop performance \mathcal{P} is given by

$$\mathcal{P} = \begin{cases} \text{Good, if } \eta_{MV} \geq \eta_G \\ \text{Fair, if } \eta_P < \eta_{MV} < \eta_G \\ \text{Poor, if } \eta_{MV} \leq \eta_P \end{cases} \quad (10)$$

54 where η_G, η_P define performance ranges, usually determined by human experts
 according to their past experience and *a priori* information.

56 The previous approach is well-known and statistical confidence intervals can
 be calculated for η_{MV} [7]. However, as discussed in the previous section, in
 58 practice we may be largely underestimating loop performance.

3. Proposed approach

60 We focus on the trade-off between process variable $y(k)$ and manipulated
 variable $u(k)$ energy. For this, let's denote

$$m_u = \frac{1}{N} \sum_{k=0}^{N-1} u(k) \quad (11)$$

$$s_u^2 = \frac{1}{N-1} \sum_{k=0}^{N-1} [u(k) - m_u]^2 \quad (12)$$

$$s_{y,u} = (s_y^2, s_u^2) \quad (13)$$

As in the previous section, we start by estimating $\hat{A}_{cl}(z^{-1})$, $\hat{B}_{cl}(z^{-1})$ (see equation 9) from which we obtain the open-loop transfer function

$$\hat{G}_p(z^{-1}) = \frac{\hat{G}_{cl}(z^{-1})}{1 - G_c(z^{-1})\hat{G}_{cl}(z^{-1})} = \frac{\hat{B}(z^{-1})}{\hat{A}(z^{-1})}$$

64 Then, let $\hat{y}(k + \tau)$ be the one-step-ahead output prediction:

$$\hat{y}(k + 1) = \frac{F(z^{-1})}{\hat{B}(z^{-1})} y(k) + E(z^{-1})u(k) \quad (14)$$

where $E(z^{-1})$ and $F(z^{-1})$ are obtained solving the Diophantine equation

$$\hat{B}(z^{-1}) = E(z^{-1})\hat{A}(z^{-1}) + z^{-1}F(z^{-1}) \quad (15)$$

66 For each value $\alpha = \frac{1}{N_P}, \frac{2}{N_P}, \dots, 1$, with $N_P \gg 1$, we determine the solution $u^*(k)$ to the following optimization problem

$$\min_{u(k)} \left[\alpha \hat{y}(k + 1)^2 + (1 - \alpha) u(k)^2 \right] \quad (16)$$

68 Here, it is possible to show that the solution is $u^*(k)$ such that

$$u^*(k) = -K_\alpha^*(z^{-1})y(k) \quad (17)$$

with

$$K_\alpha^*(z^{-1}) = \frac{\alpha F(z^{-1})E(z^{-1})}{\alpha E(z^{-1})^2 B(z^{-1}) + (1 - \alpha)B(z^{-1})} \quad (18)$$

70 Note that this result (18) was found independently by the author of this paper, however an equivalent result was previously published in [8]. Now, denote

72 S^* the set of points $(s_{u,\alpha}^{*2}, s_{y,\alpha}^{*2})$

$$s_{u,\alpha}^{*2} = \left\| \frac{G_p(z^{-1})K_\alpha^*(z^{-1})}{1 + K_\alpha^*(z^{-1})G_p(z^{-1})} \right\|_2^2 \quad (19)$$

$$s_{y,\alpha}^{*2} = \left\| \frac{G_p(z^{-1})}{1 + K_\alpha^*(z^{-1})G_p(z^{-1})} \right\|_2^2 \quad (20)$$

74 for $\alpha = \frac{1}{N_P}, \frac{2}{N_P}, \dots, 1$. Note that S^* is a numerical *approximation of the Pareto front* [9] corresponding to the problem

$$\min_{u(k)} \begin{pmatrix} u(k)^2 \\ \hat{y}(k + 1)^2 \end{pmatrix} \quad (21)$$

76 Now, denote $(p_{u,\alpha}^{*2}, p_{y,\alpha}^{*2})$ the point in S^* which is *nearest to* $s_{y,u} = (s_y^2, s_u^2)$
but dominates $s_{y,u}$, which means

$$p_{u,\alpha}^{*2} \leq s_u^2 \quad (22)$$

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$$p_{y,\alpha}^{*2} \leq s_y^2 \quad (23)$$

The performance will be assessed based on the following indicator:

$$\eta_{MC} = \frac{p_{y,\alpha}^{*2}}{s_y^2} \quad (24)$$

Note immediately that

$$p_{y,\alpha}^{*2} \geq \sigma_{MV}^2$$

80 and therefore

$$\eta_{MC} \geq \eta_{MV} \quad (25)$$

4. Numerical results

82 Consider the first-order system described by

$$y(k) = \frac{1}{1 + a_1 z^{-1}} [u(k) + \varepsilon(k)] \quad (26)$$

with $a_1 = -0.9$ (also considered in [3]) and controller

$$G_c(z^{-1}) = \frac{0.2z^{-1}}{1 - 0.8999z^{-1}} \quad (27)$$

84 As explained in the previous section, first we generate the data

$$y(-N_H) \dots, y(-2), y(-1) \quad (28)$$

and identify the following third-order ARMA model

$$\hat{G}_{cl}(z^{-1}) = \frac{\hat{B}_{cl}(z^{-1})}{\hat{A}_{cl}(z^{-1})} = \frac{1 - 0.7863z^{-1} - 0.0629z^{-2} - 0.0298z^{-3}}{1 - 1.743z^{-1} + 0.620z^{-2} + 0.09022z^{-3}} \quad (29)$$

86 and generate S^* showed in Fig. 2, with $N_P = 30$, and

$$K_\alpha^*(z^{-1}) = \alpha \frac{0.904 - 1.504z^{-1} + 0.505z^{-2} + 0.1020z^{-3}}{1 - 0.7863z^{-1} - 0.0629z^{-2} - 0.0298z^{-3}} \quad (30)$$

Note that the distribution of points (black dots) is not uniform. The red line corresponds to the interpolation curve corresponding to the Pareto approximation obtained by using the real polynomials $A(z^{-1})$ and $B(z^{-1})$.

Next, we generate the data

$$y(0), y(1), y(2), \dots, y(N-1)$$

which corresponds to the window we want to assess, obtaining the following results

$$(s_y^2, s_u^2) = (2.74, 0.5) \quad (31)$$

$$(p_{u,\alpha}^{*2}, p_{y,\alpha}^{*2}) = (1.22, 0.28) \quad (32)$$

$$\eta_{MV}(\%) = 36.5 \quad (33)$$

$$\eta_{MC}(\%) = 44.4 \quad (34)$$

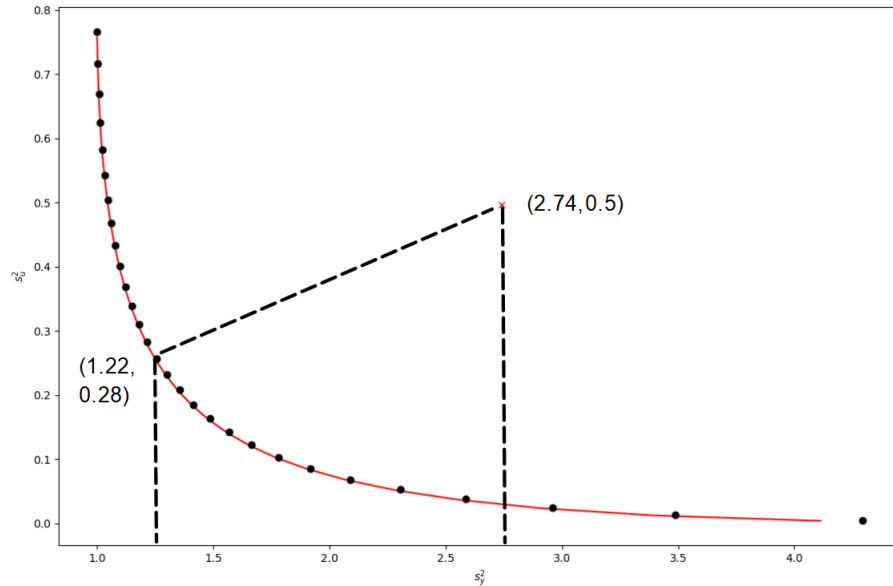
Note, that, as expected, the proposed multi-criteria index is almost 8% greater than Harris' index. The code to obtain these results is available in the following link: https://github.com/multiopti/research/blob/main/MC_CLPA.ipynb

5. Conclusion

A multi-criteria approach for control loop performance assessment was proposed, which allows to take into account the trade-off between process variable and manipulated variable energy, thus overcoming one of the most important criticisms to Harris' index. In future, we plan to extend this method to more complex models, including unstable multivariable processes.

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Figure 2: Pareto approximation

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