

## Laboratory Work #8

### FOURIER ANALYSIS AND SPECTRAL ANALYSIS OF DETERMINISTIC SIGNALS

#### 1. Goal of the lab:

The goal of the lab is to develop an intuitive feel and an appreciation for the use of spectral density in identification.

#### 2. Introduction

##### FOURIER SERIES

A continuous-time periodic signal (function) with fundamental period  $T_p = \frac{1}{F_0}$  is expressed as a (linear) weighted combination of (positive and negative) harmonics.

$$x(t) = \sum_{n=-\infty}^{\infty} c_n e^{j2\pi n F_0 t} \quad (\text{Synthesis equation})$$

This equation is the Fourier series expansion of  $x(t)$  in terms of complex sinusoids that have the same period (not fundamental).

The coefficient  $c_n$  in the Synthesis equation is calculated as

$$c_n = \frac{1}{T_p} \int_{T_p} x(t) e^{-j2\pi n F_0 t} dt \quad (\text{Analysis equation})$$

The result is predominantly useful in theoretical analysis of signals and systems.

##### FOURIER TRANSFORM

Aperiodic signals can be treated as a limiting case of periodic signal with period  $T_p \rightarrow \infty$ . Then, the spacing of harmonics on the frequency axis  $\Delta F = \frac{1}{T_p} \rightarrow 0$ . The family of basis functions then belongs to the continuous-domain.

The synthesis equation takes the integral form

$$x(t) = \int_{-\infty}^{\infty} X(F) e^{j2\pi F t} dF \quad (\text{Fourier Synthesis})$$

Its dual, the analysis equation is now evaluated over the entire time axis,

$$X(F) \triangleq \mathcal{F}\{x(t)\} = \int_{-\infty}^{\infty} x(t) e^{-j2\pi F t} dt \quad (\text{Fourier Transform})$$

The quantity  $X(F)$  has a similar interpretation as that of  $c_n$  in the Analysis equation. It is a complex-valued continuous function of the frequency  $F$ .

## SPECTRUM AND SPECTRAL DENSITY

A vast variety of signals that we encounter are either finite-energy aperiodic or stochastic (or mixed) signals, which are characterized by energy and power spectral density, respectively. However, the practical situation is that we have a finite-length signal  $x_N = \{x[0], x[1], \dots, x[N-1]\}$ . Computing the N-point DFT amounts to treating the underlying infinitely long signal  $\tilde{x}[k]$  as periodic with period  $N$ . Thus, strictly speaking we obtain a power spectrum (line spectrum) regardless of the nature of underlying signal.

In order to be able to use the spectrum as a tool for estimating spectral density of finite-length observations of random signals and to bring the deterministic and stochastic signals into a single framework of estimation, an artificial spectral density known as the periodogram is constructed as follows.

The power spectrum  $P_{xx}(f_n)$  for the finite-length signal  $\{x[k]\}_{k=0}^{N-1}$  is obtained by recalling the definition of power for a periodic signal

$$\Rightarrow P_{xx}(f_n) = |c_n|^2 = \frac{|X[n]|^2}{N^2}$$

Subsequently, an **empirical power spectral density** (power per unit cyclic frequency) for the finite-length sequence is introduced as,

$$P_{xx}(f_n) \triangleq PSD(f_n) = \frac{P_{xx}(f_n)}{\Delta f} = N|c^2| = \frac{|X[n]|^2}{N}$$

where the units of frequency is, as usual, in cycles/sample. The quantity  $P_{xx}(f_n)$  is known as the periodogram, first introduced by Schuster (1897).

### 3. Tasks

#### 3.1. Estimating signal with Fourier transform.

The series of interest consists of  $N = 500$  samples of a simulated process

$$v[k] = \sin(0.2\pi k) + \sin(0.5\pi k) + e[k]$$

where  $e[k]$  is the GWN process with variance  $\sigma_e^2 = 2.25$ .

- Firstly, we need to generate the measurement. As the series of interest consists of 500 samples, define the discrete time series (**k**) as a vertical vector of (0:499).
- Generate the output signal using the equation above by representing the GWN process as normally distributed random numbers with the variance of  $\sigma_e^2 = 2.25$ . (*randn*)
- Plot the generated output signal:

**Label for X:** 'Series'

**Label for Y:** 'Amplitude'

**Title:** 'Generated signal'

- d. Compute periodogram (Use *periodogram* command). Specify the input and output parameters as below:

**Outputs:**

- Distribution of power per unit frequency:  $P_{xx}$
- Vector of normalized frequencies:  $W$

**Inputs:**

- Detrended output series: '*output – mean (output)*' or use *detrend* command.
- Window: *Default window ([ ])*
- The number of FFT points: *the length of simulated output series (N)*
- Sampling frequency:  $F_s = 1 \text{ Hz}$

- e. Plot the periodogram output using the vector of normalized frequencies ( $W$ ) and normalized PSD ( $\frac{P_{xx}}{\text{sum}(P_{xx})}$ )

**Title:** '*Periodogram*'

**Label for X:** '*Frequency (cycles/sample)*'

**Label for Y:** '*Spectral density*'

- f. Take Fourier Transform of simulated output series (Use *fft* command).
- g. Split the magnitude and phase part of the FFT result using *abs* and *phase* commands (**Hint:** Use the half of the data).
- h. Zero out the contributions (assumed to be) due to noise.
- i. Estimate the signal using the inverse Fourier transform function (*ifft* command).
- j. Plot the estimated signal against the true deterministic signal (the first 100 data). The true signal will be the simulated process represented above (part a.) removing the noise part.

**Label for X:** '*Series*'

**Label for Y:** '*Amplitude*'

**Legend:** '*True*', '*Estimated*'

**3.2.** Repeat the procedure using different value of the  $e[k]$  variance.

## 4. Questions

**4.1.** What is the basic difference between a Fourier series and a Fourier transform?

**4.2.** How is the Fourier transform useful in arriving at the frequency response functions for LTI systems?

**4.3.** Define periodogram and explain the reasons for introducing it.