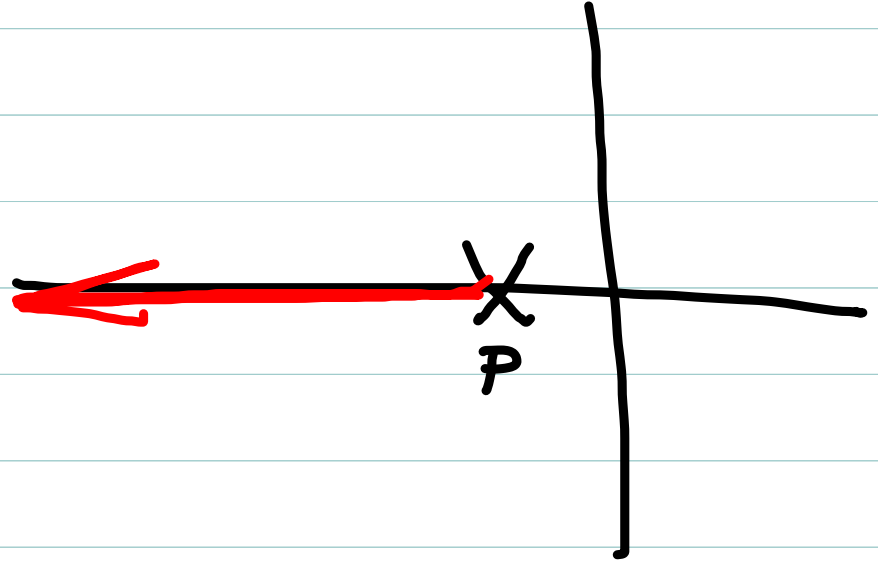
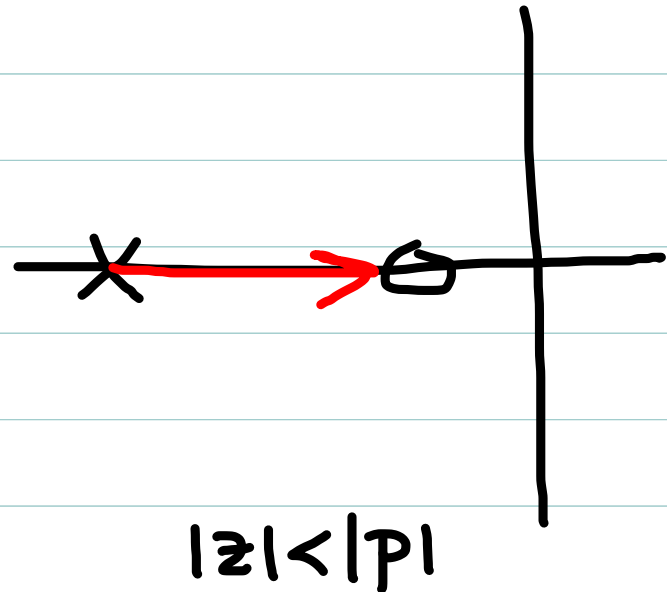
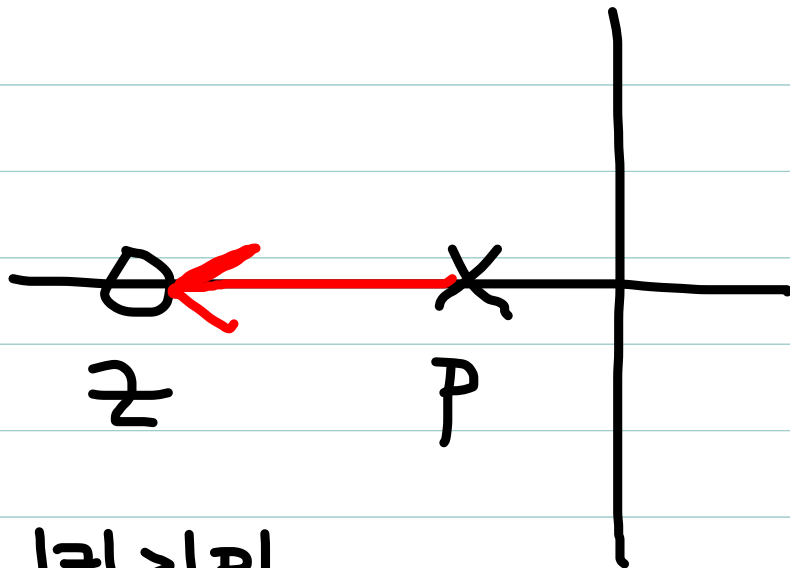


Simple Examples

$$\#1) \quad L(s) = \frac{K}{s-p}$$

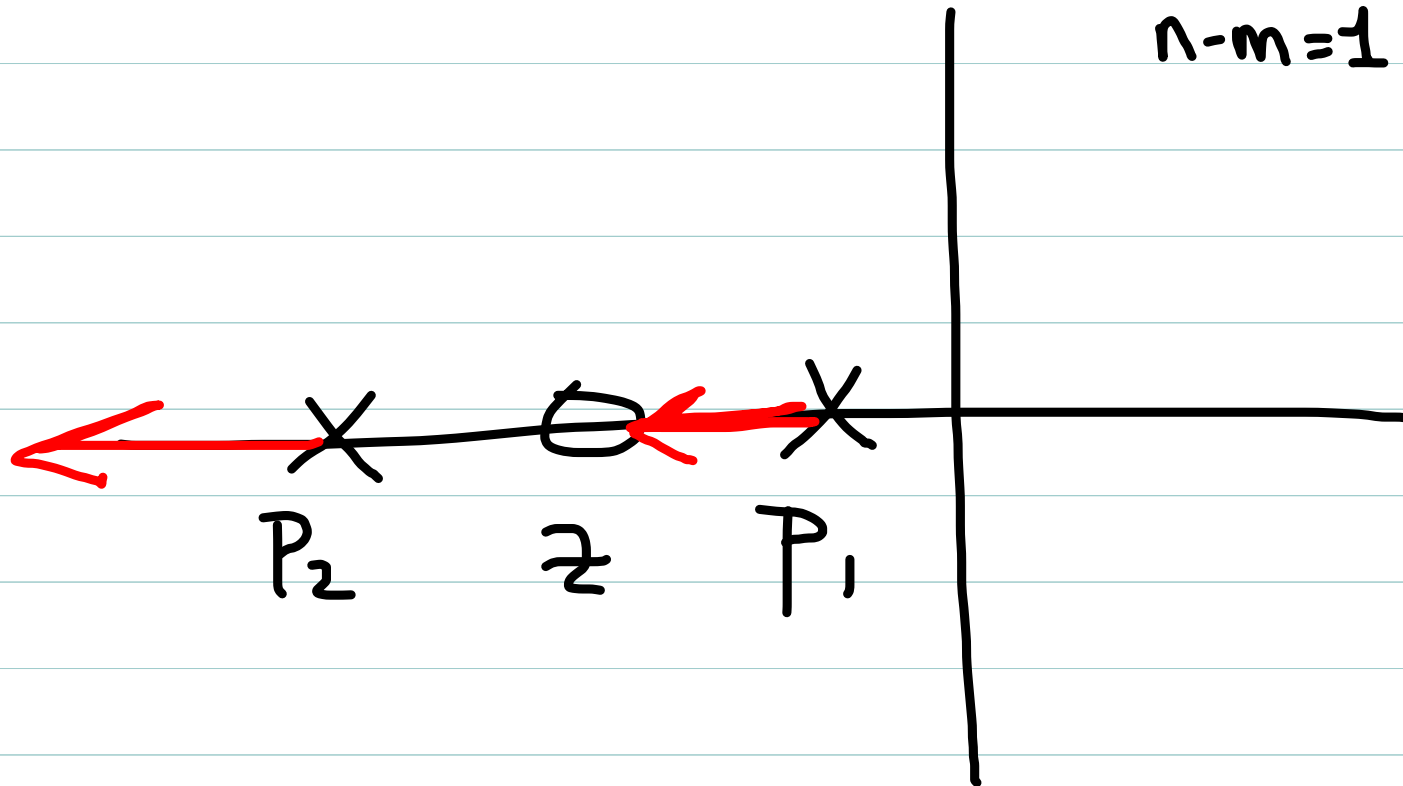


$$\#2) \quad L(s) = K \left[\frac{(s-z)}{(s-p)} \right]$$



$$*3] \quad L(s) = K \left[\frac{(s-z)}{(s-p_1)(s-p_2)} \right]$$

$n-m=1$: asymptote
along neg real Axis



$$|p_2| > |z| > |p_1|$$

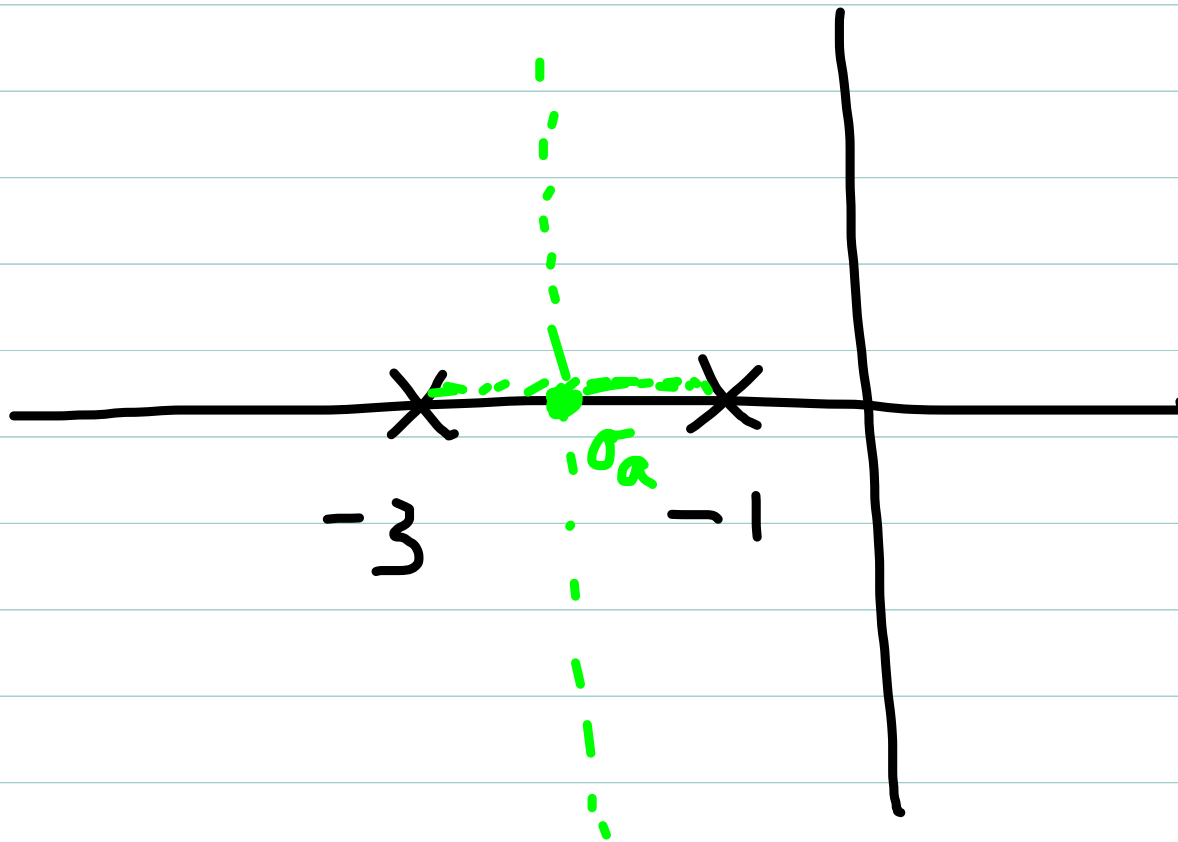
Case where $|z| < |p_1| \leq |p_2|$ is more complicated
-- see below.

$$\#4] \quad L(s) = \frac{K}{(s+1)(s+3)}$$

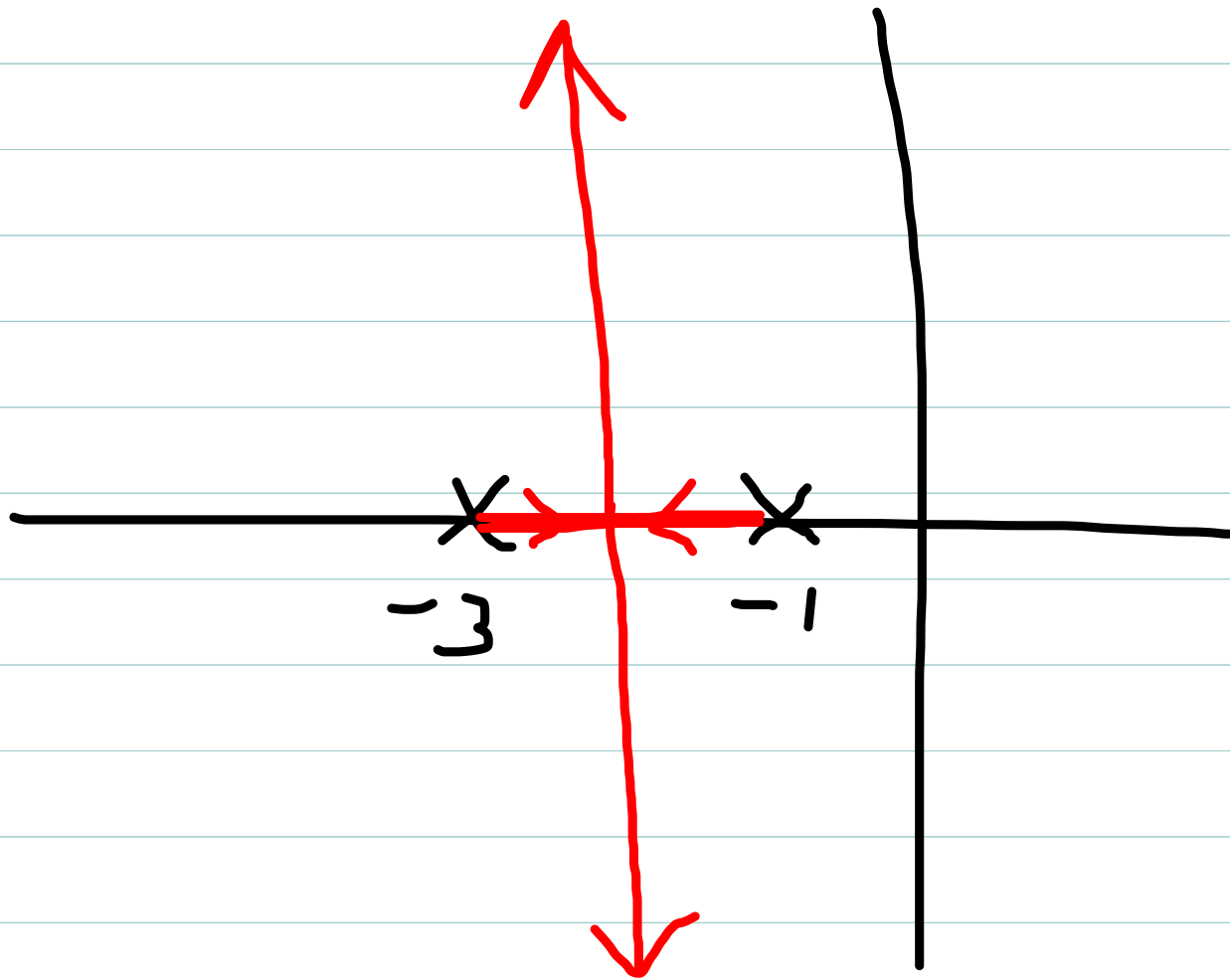
$$n-m=2 \Rightarrow \alpha_z = \pm 90^\circ$$

$$\sigma_a = \frac{(-1)+(-3)}{2} = -2$$

real axis locus from $-3 \rightarrow -1$



Actual locus:



Compare w/ exact sol'n for CL poles:

$$s^2 + 4s + (3+k) = 0 \Rightarrow s = -2 \pm \frac{\sqrt{16 - (3+k)^2}}{2}$$

Example #5

$$L(s) = \frac{K}{s(s+1)(s+2)}$$

$n=3, m=\emptyset, n-m=3 \Rightarrow 3$ branches go to ∞ along

asymptotes: $\alpha_e = \pm 60^\circ, 180^\circ$

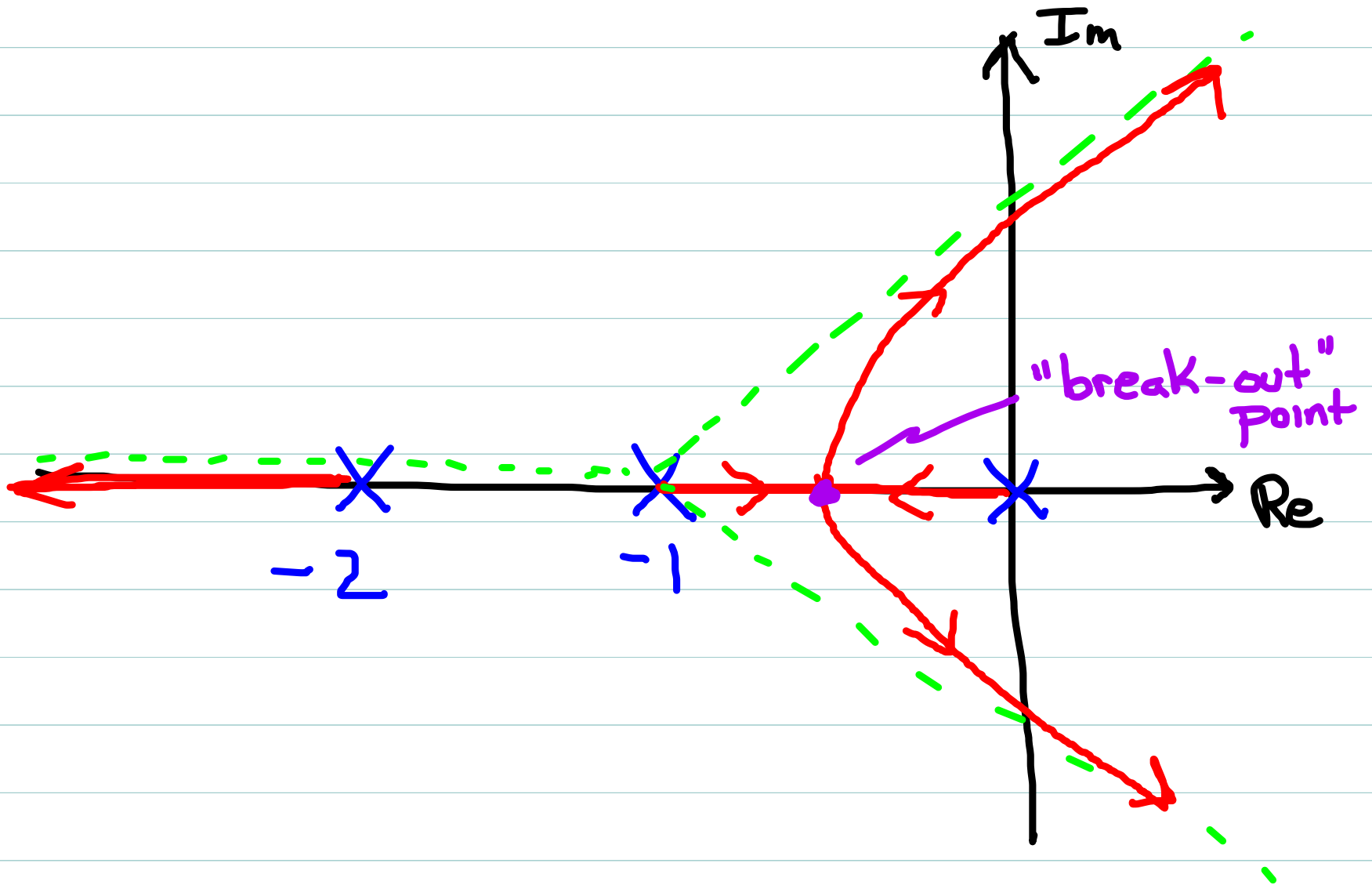
with intercept: $\sigma_a = \frac{0+(-1)+(-2)}{3} = -1$

Real axis branch locations.

\Rightarrow between -1 and \emptyset

\Rightarrow left of -2

Example #5, cont



Break-out Points

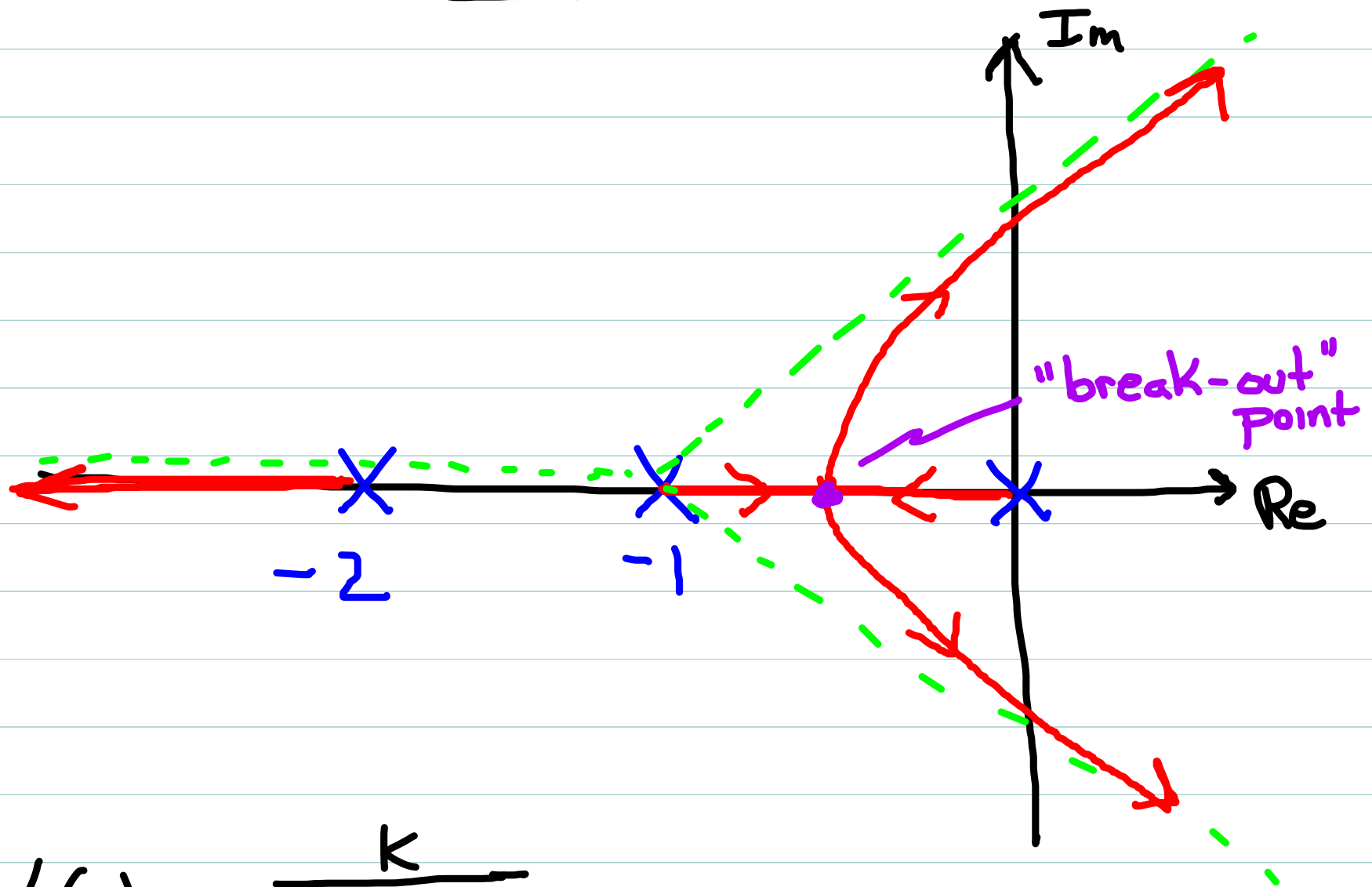
Break-out points occur for values of s satisfying

$$\frac{d}{ds} (L(s)) = 0$$

Since this (usually) leads to another high-order polynomial to factor, we often just approximate a break-out as occurring half-way along the branch

Use Matlab ("rlocus" command) to nail exact details when needed).

Example #5, cont



$$L(s) = \frac{K}{s(s+1)(s+2)}$$

High Gain Instability

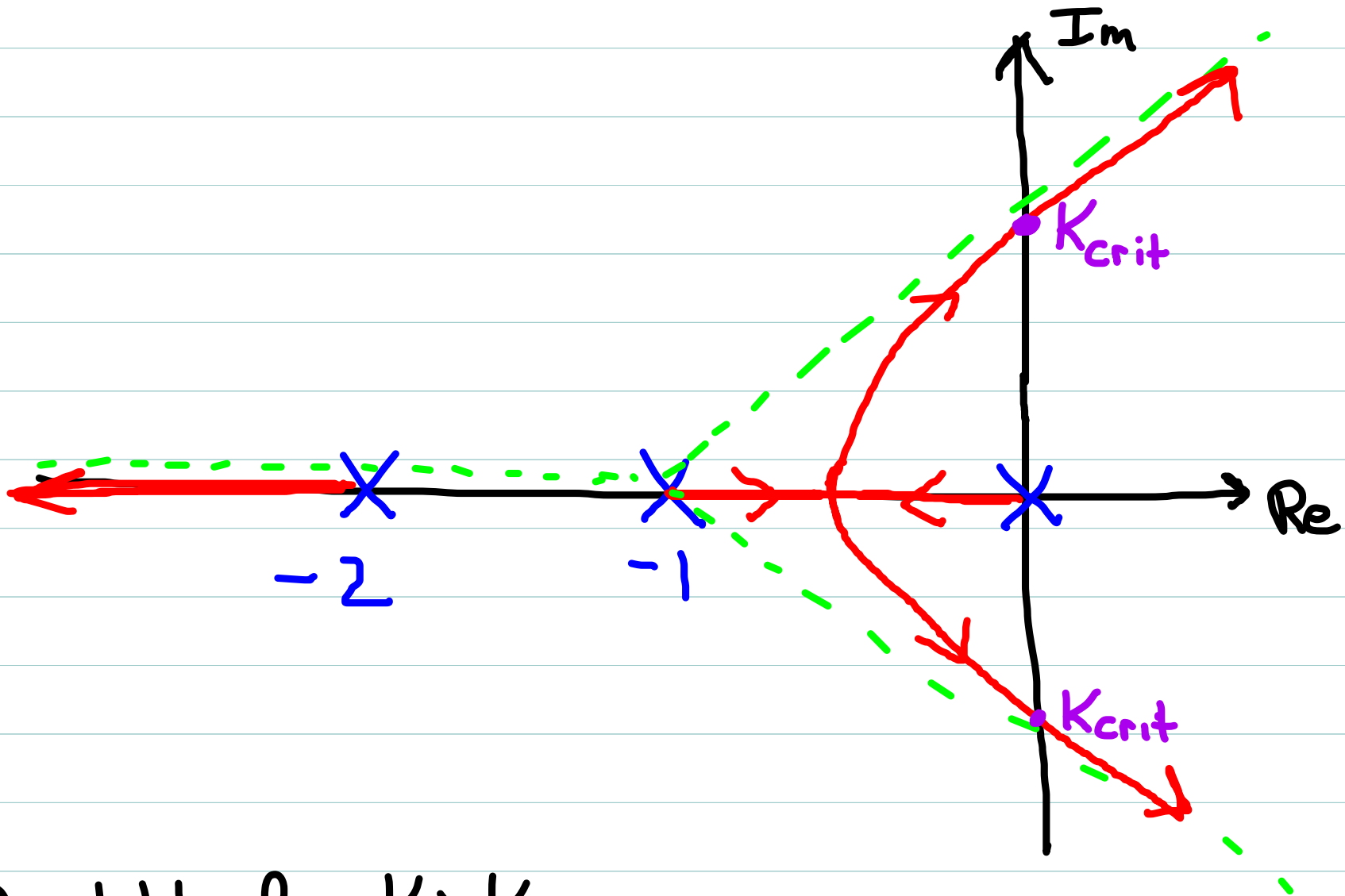
Since this example has Asymptotes in RHP, we can see the CL system will be unstable for sufficiently high gains K .

Whether this is a problem or not depends on the gain we want/need to get the desired CL poles

$T(s)$ is not automatically unstable b/c the root locus branches in RHP!

Such a locus only tells us $T(s)$ will be unstable for some values of K .

Example #5, cont



Unstable for $K > K_{crit}$

Example #6

$$L(s) = \frac{K(s+3)}{(s+1)(s+2)}$$

$$\Rightarrow n=2, m=1, n-m=1$$

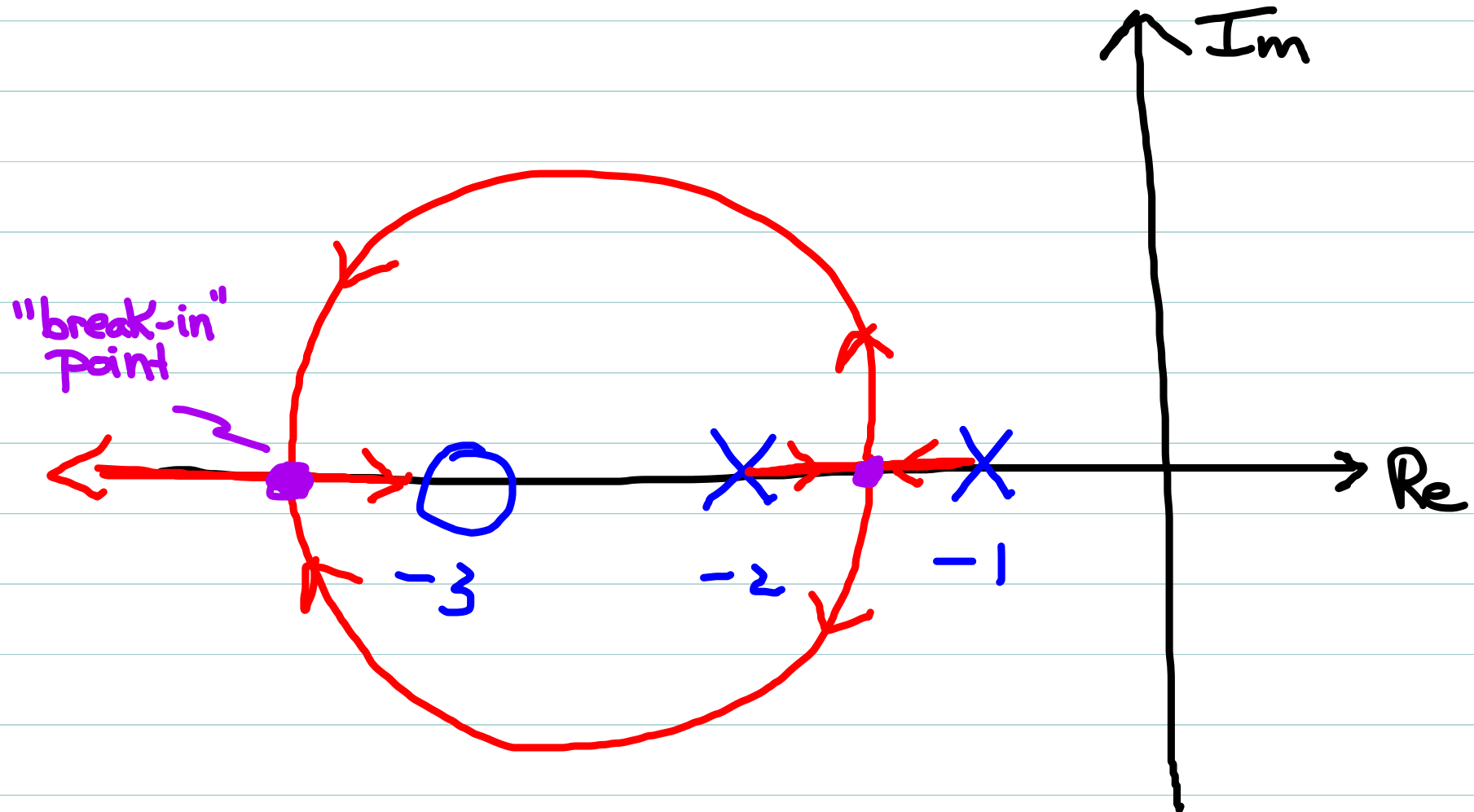
\Rightarrow One branch ends at -3 (OL zero). One branch goes to ∞ along asymptote $\alpha = 180^\circ$ (negative real Axis)

\Rightarrow Segments of branches lie on real Axis:

\Rightarrow Between -2 and -1

\Rightarrow left of -3

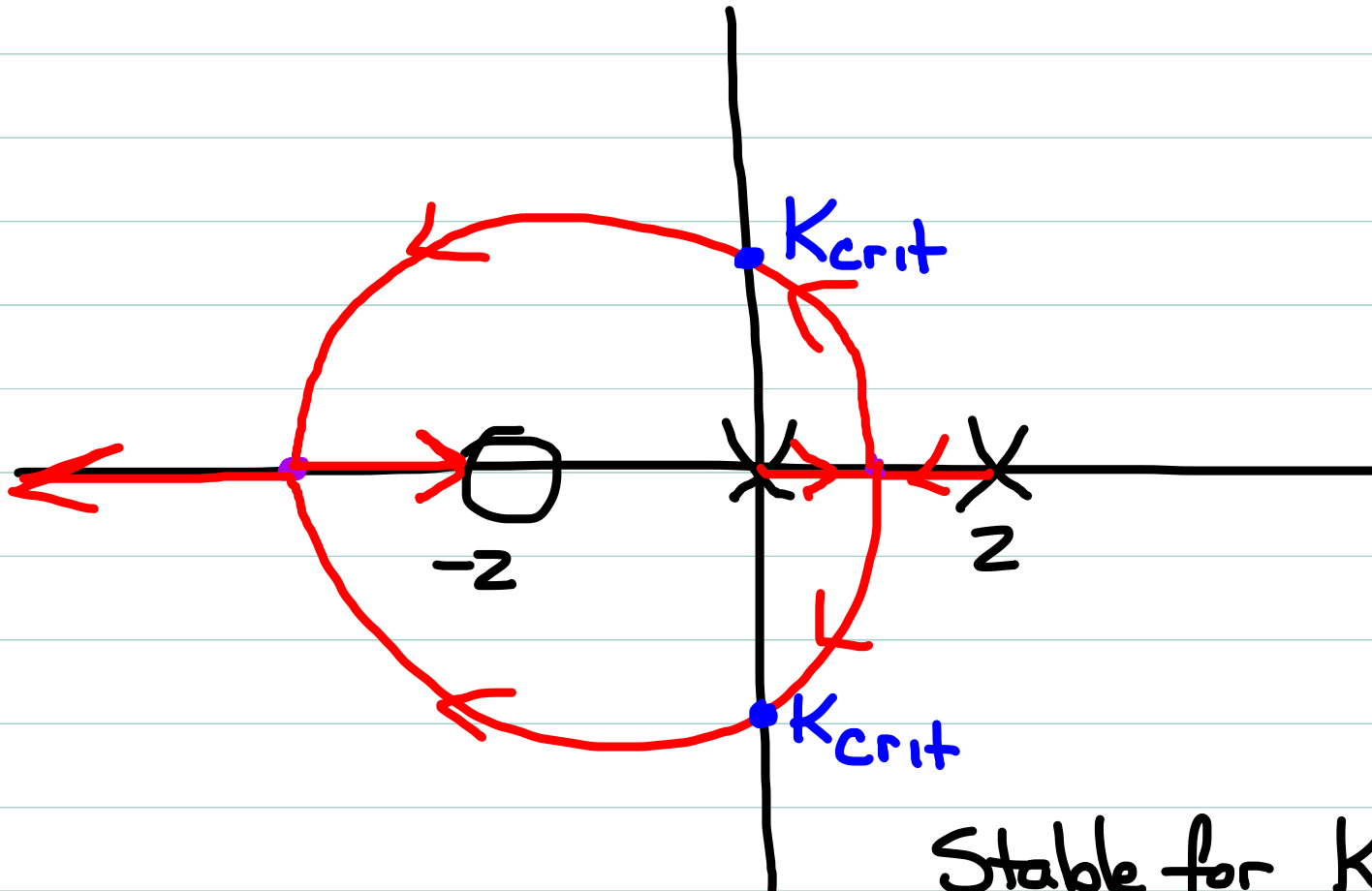
Example #6, cont



Example #7

$$L(s) = \frac{K(s+2)}{s(s-2)}$$

Similar analysis to above

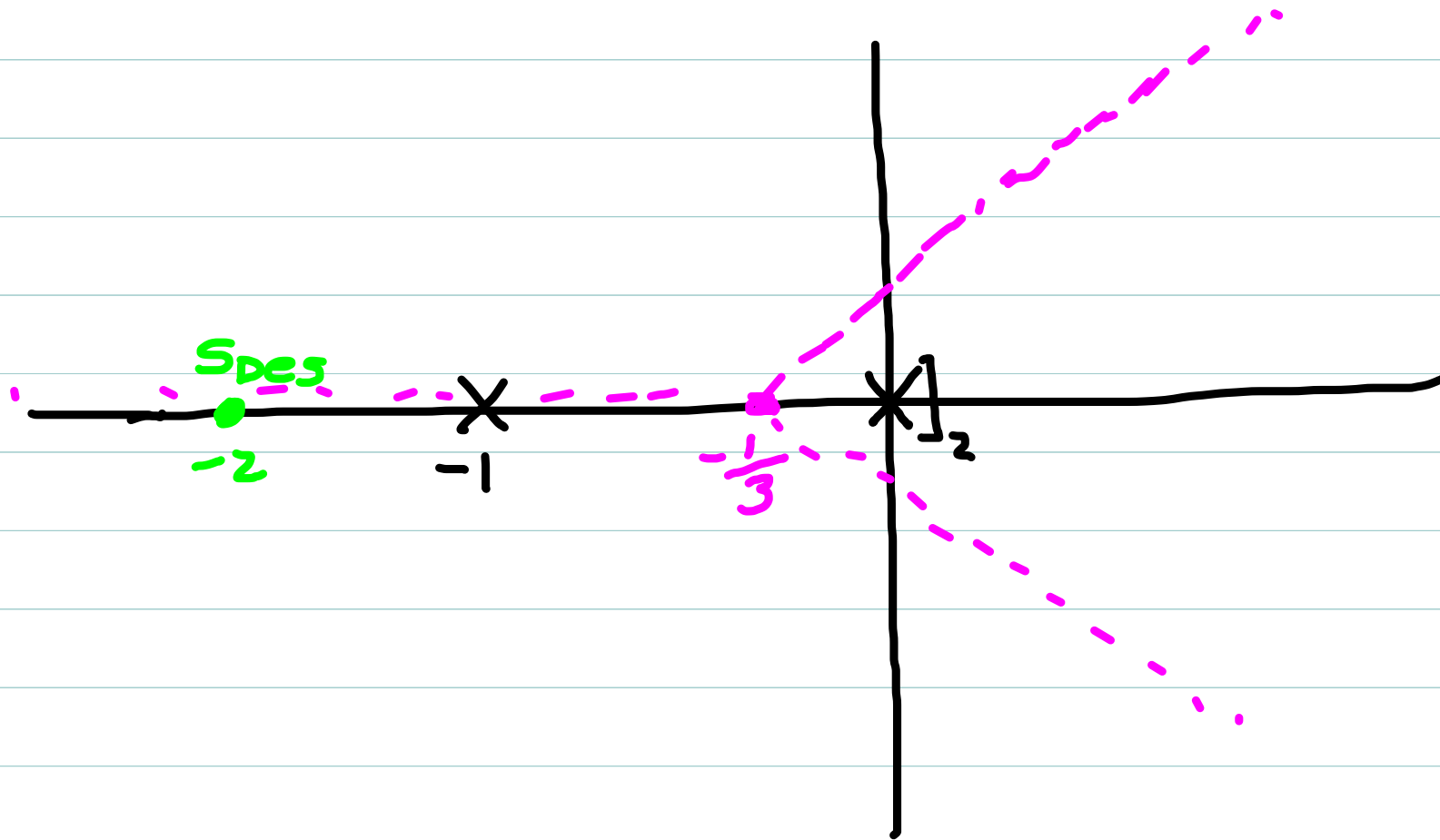


Stable for $K > K_{crit}$

Example #8

This is where we originally started our investigation

$$\text{with } H(s) = K, \quad L(s) = \frac{K}{s^2(s+1)}$$



Example #8

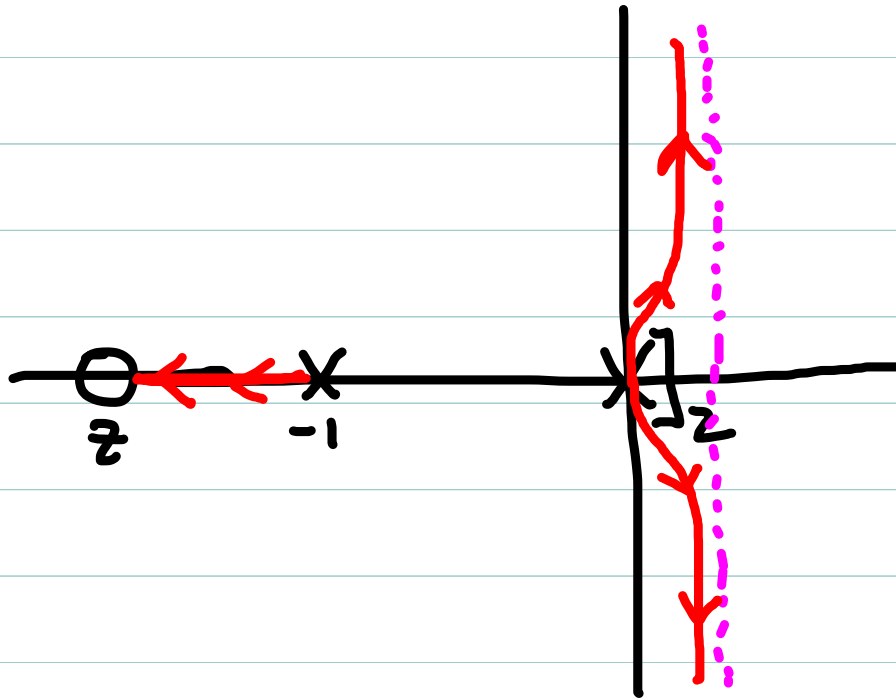
This is where we originally started our investigation

$$L(s) = \frac{K}{s^2(s+1)}$$



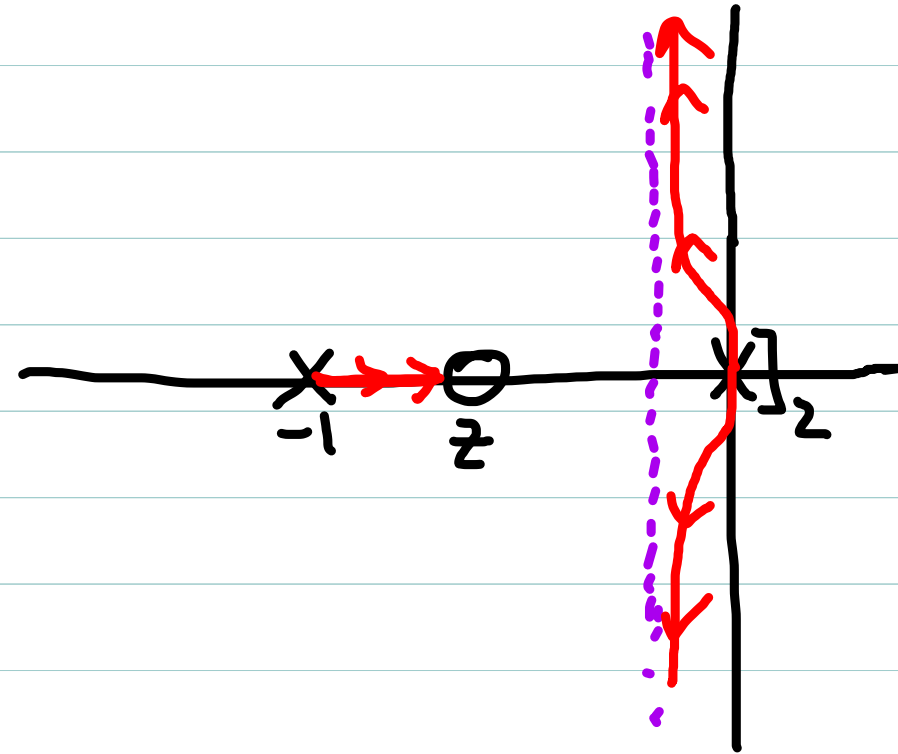
We can get the desired pole at -2, but will inevitably have poles of $T(s)$ in RHP

With instead $H(s) = K(s - z)$



$$z < -1 \Rightarrow \alpha_z = \pm 90^\circ$$

$$\sigma_a = -\frac{1}{2}(1 + z) > \phi$$



$$0 < z < -1$$

$$\Rightarrow \sigma_a < \phi$$

So, with $H(s) = K(s-z)$ we can stabilize the system as long as $|z| < 1$ (which would agree with a Nyquist/phase margin analysis)

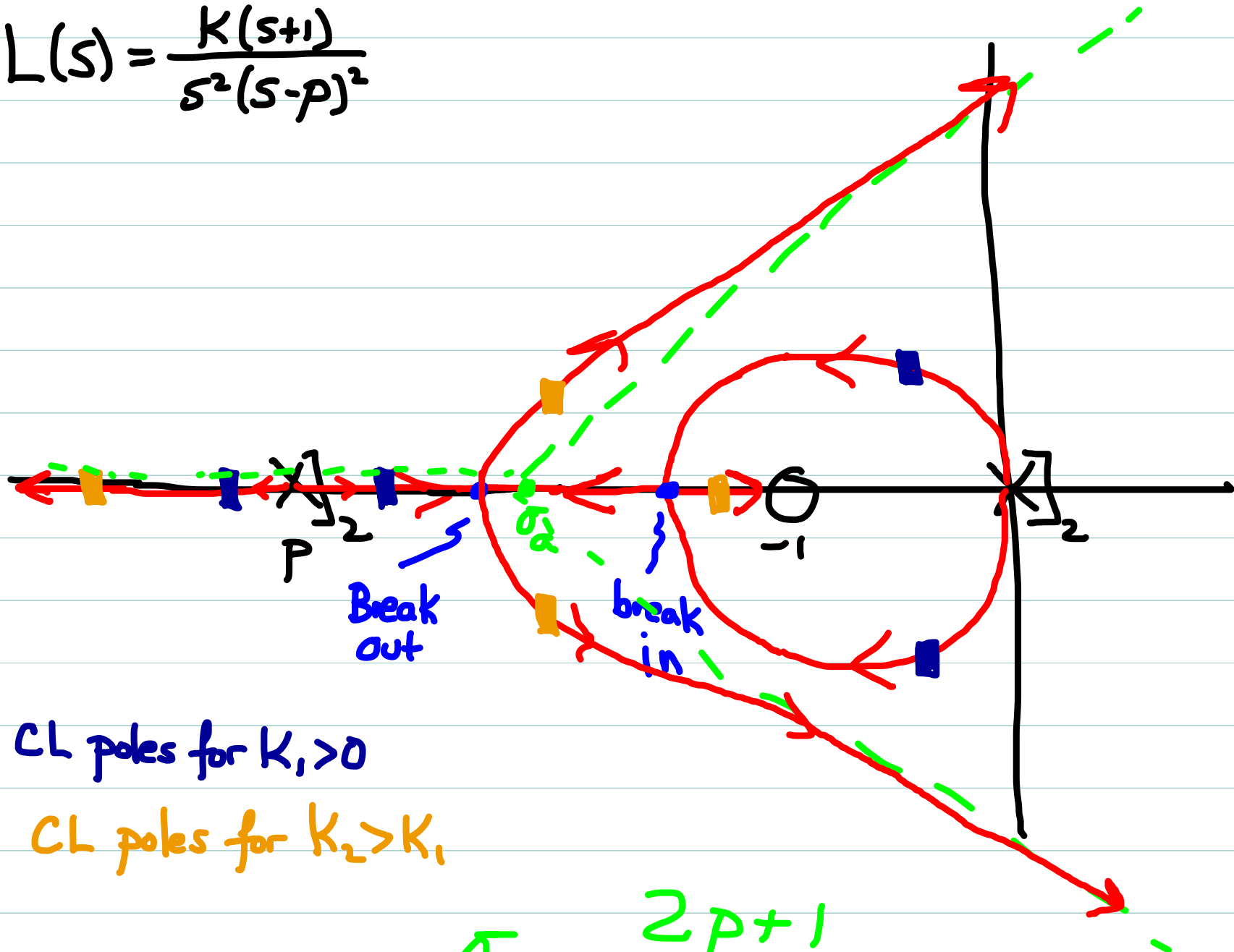
But we would have to accept a real pole > -1 , and moreover this pole would not be dominant

An implementable compensator which could allow a real dominant CL pole near -2 would be

$$H(s) = K \left[\frac{(s+1)^2}{(s-p)^2} \right]$$

which has an interesting locus (next page)

$$L(s) = \frac{K(s+1)}{s^2(s-p)^2}$$



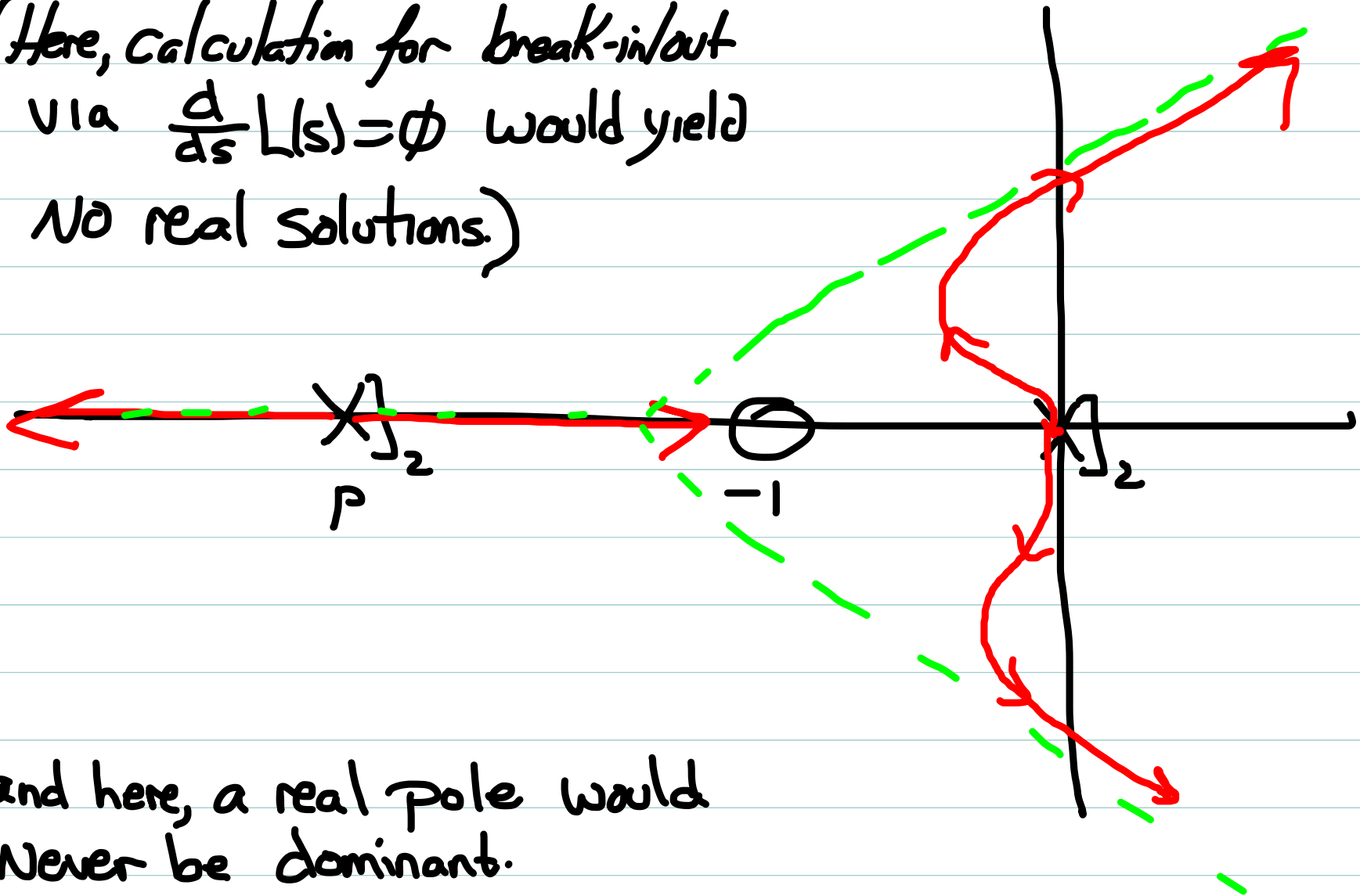
■ CL poles for $K_1 > 0$

■ CL poles for $K_2 > K_1$

$$\sigma_a = \frac{2p+1}{3}$$

However, depending on exact value of p , this is also possible:

(Here, calculation for break-in/out
via $\frac{d}{ds} L(s) = 0$ would yield
No real solutions.)



and here, a real pole would
Never be dominant.

Comments on root locus method

- ⇒ Rules are not determinative; there may be many locus shapes consistent with calculations (although Matlab rlocus command will show you an exact plot).
 - ⇒ Cannot adapt method to account for effects of time delay
 - ⇒ Can adapt method only for very simple kinds of robustness analysis.
 - ⇒ Bode/Nyquist methods preferred in professional practice.
-
- ⇒ But root locus does provide useful additional insights which are not available using freq. methods
 - ⇒ Familiarity with both gives "best of both worlds"