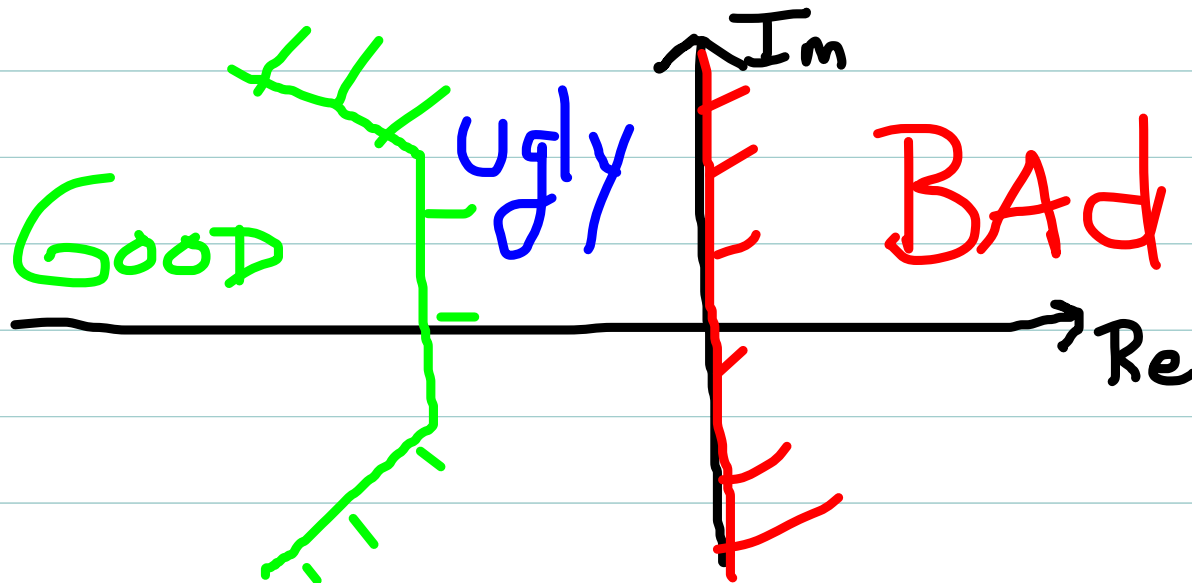


Fundamental Consideration: Closed-loop Stability

Most basic design consideration: ~~←~~ == !!

Closed-loop poles should be "good", and certainly must be stable.

Thus, sol'n's of CE: $1+L(s)=0$ must be in left half of complex plane, preferably in "good region" (far from imag Axis, relatively close to or on the real Axis).



A crucial Observation:

If $L(j\omega) = -1$ for some ω , then

$1 + L(s) = 0$ has a sol'n $s = j\omega$ for some ω

\Rightarrow closed-loop dynamics has poles at $\pm j\omega$, on imag Axis

\Rightarrow Such poles are on the boundary between bad and ugly

\Rightarrow This situation must be avoided!!!

Now if $L(j\omega) = -1$ for some $\omega > 0$, then:

\Rightarrow polar plot of $L(j\omega)$ passes through -1

$\Rightarrow \omega_a = \omega_\gamma$ (both crossover freqs same)

$\Rightarrow a = 0 \text{ dB}, \gamma = 0^\circ$ (both margins 0)

Any such feedback loop is bad!

Now, suppose $\exists \omega \geq 0 \ni L(j\omega) \approx -1$ (i.e. close to, but not exactly -1)

By continuity of $L(s)$, $1 + L(s) = 0$ would have a sol'n very near (but not exactly on) the imag axis.

Some poles of $T(s)$ would be in bad or ugly region
 \Rightarrow Also undesirable!

Now, if $L(j\omega) \approx -1$ for some $\omega \geq 0$

\Rightarrow polar plot of $L(j\omega)$ comes very close to -1
but doesn't pass exactly through it

\Rightarrow (typically) $|a_{dB}|$ and $|\gamma|$ very small
(small margins)

\Rightarrow This should also be avoided.

Thus, for $T(s)$ to have only good poles, we need conditions:

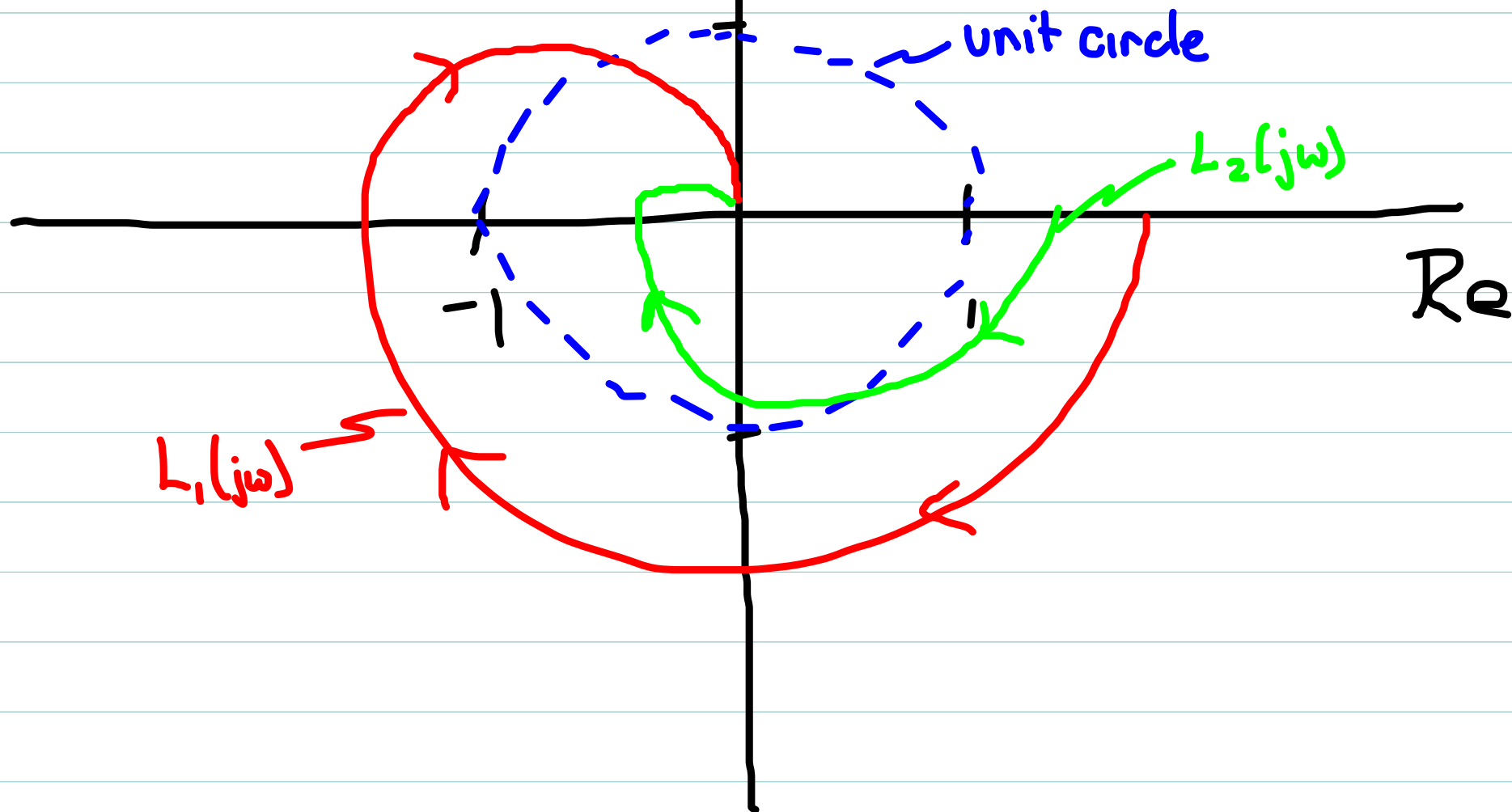
\Rightarrow Gain and phase margins of $L(s)$ \leftarrow !!!
to be large

\Rightarrow polar plot of $L(j\omega)$ avoids -1 by wide margins

Necessary, but not sufficient!

Both plots avoid -1 by
large margins

Is one better?
Yes! But criterion is
un-obvious!



Nyquist Stability Criterion

All roots of $1+L(s)=0$ are in LHP if.

the Nyquist diagram (a modified polar plot) of $L(j\omega)$

circles the -1 point the correct number of times).

\Rightarrow Major Theoretical result! Used extensively in
Control theory

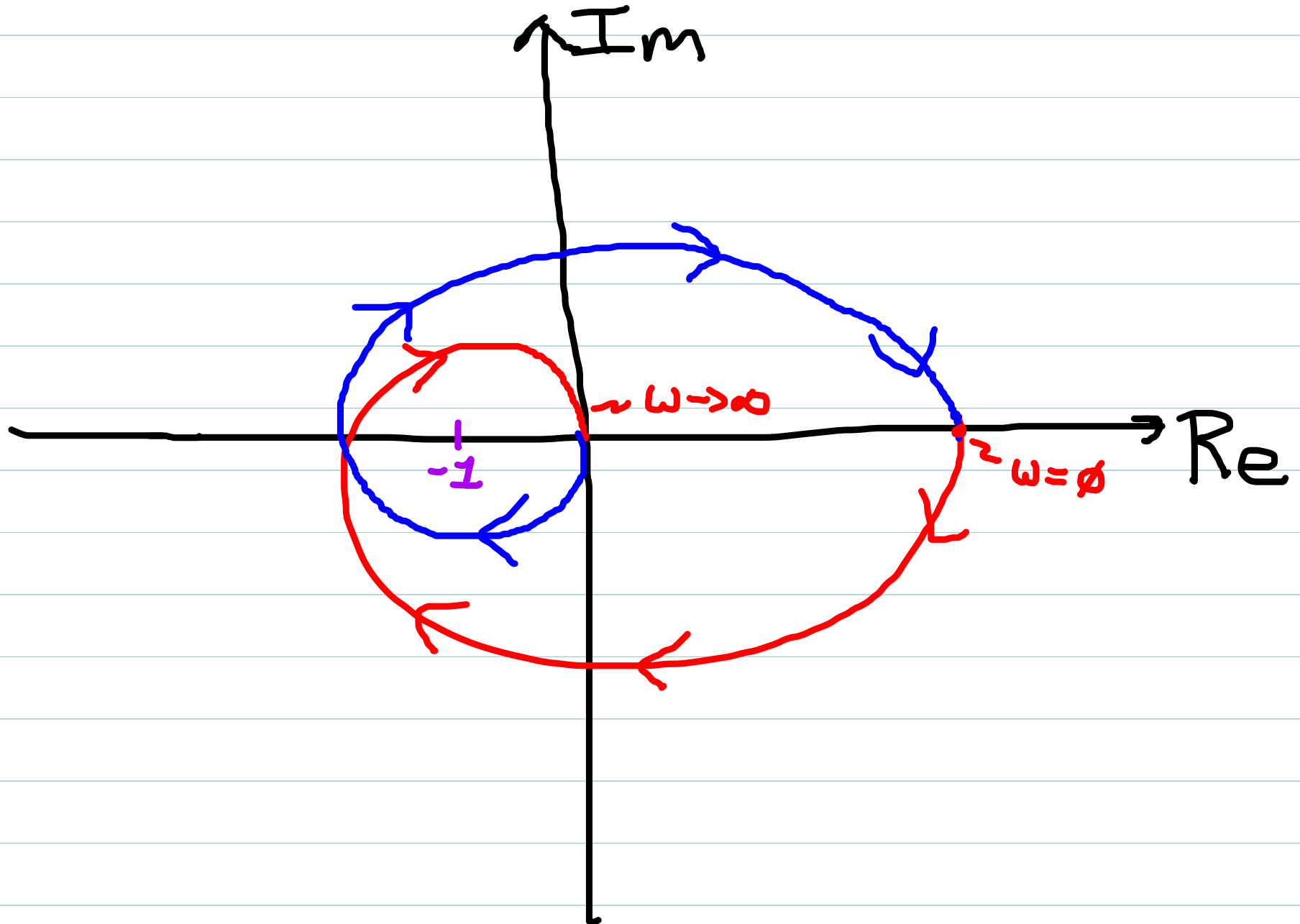
\Rightarrow Questions to answer

\Rightarrow How to create diagram from polar?

\Rightarrow How to count encirclements of -1?

\Rightarrow How many encirclements needed?

Example: $L(s) = \frac{K_D}{(\tau s + 1)^3} \quad K_D, \tau > 0$



Nyquist Diagram

When $L(s)$ is type $N \leq 0$ (no poles at origin)

\Rightarrow Draw polar of $L(j\omega)$

\Rightarrow "Flip" polar of L about real axis
(this is the polar of $L(-j\omega)$, i.e. for negative frequencies)

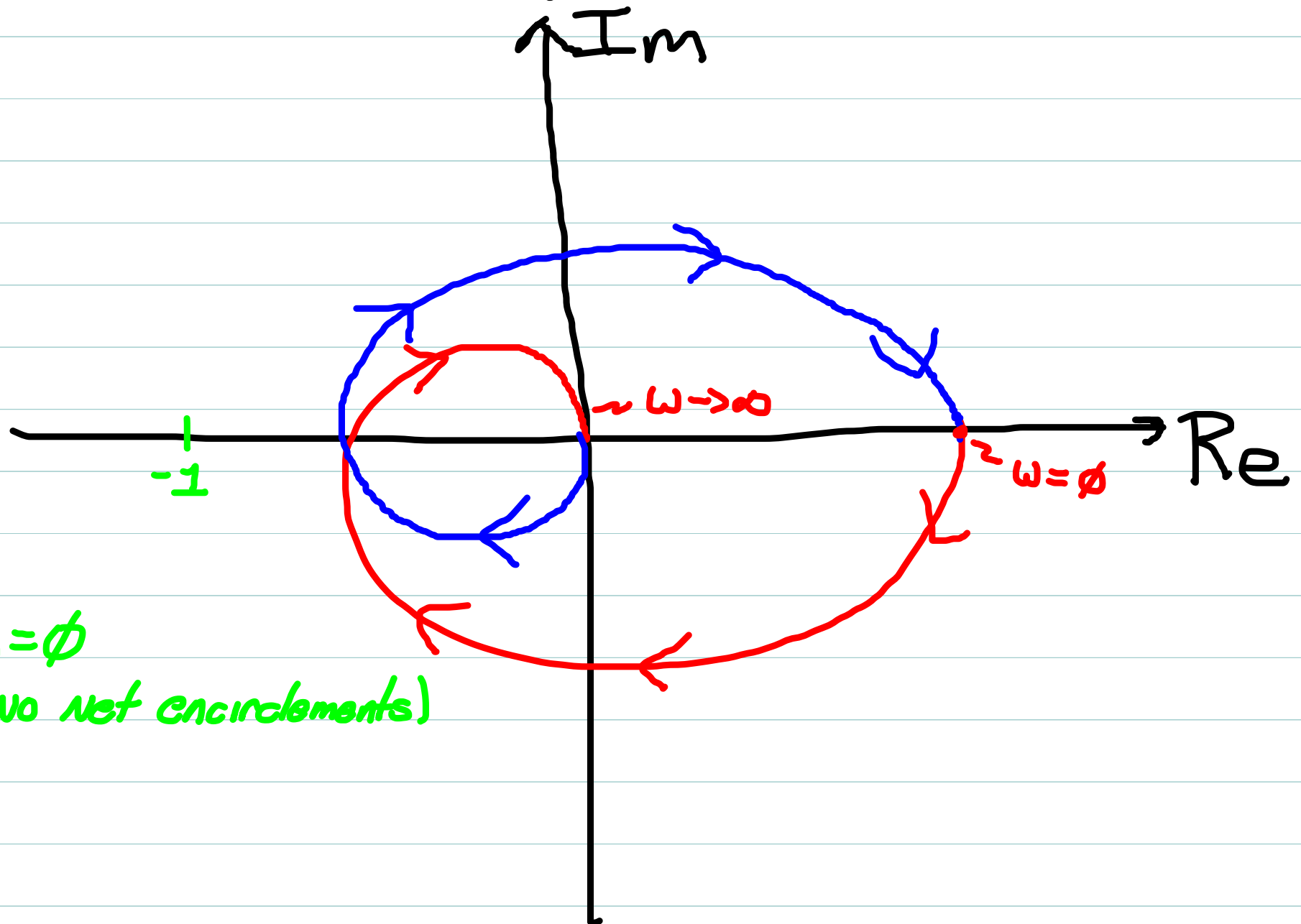
\Rightarrow Put arrows on flipped plot whose direction is consistent with direction of arrows on original polar plot
(i.e. arrows show direction of increasing frequency, from $\omega = -\infty$, through $\omega = 0$, to $\omega = \infty$).

(We will modify for $N > 0$ after we examine complete stability condition.)

Counting Encirclements

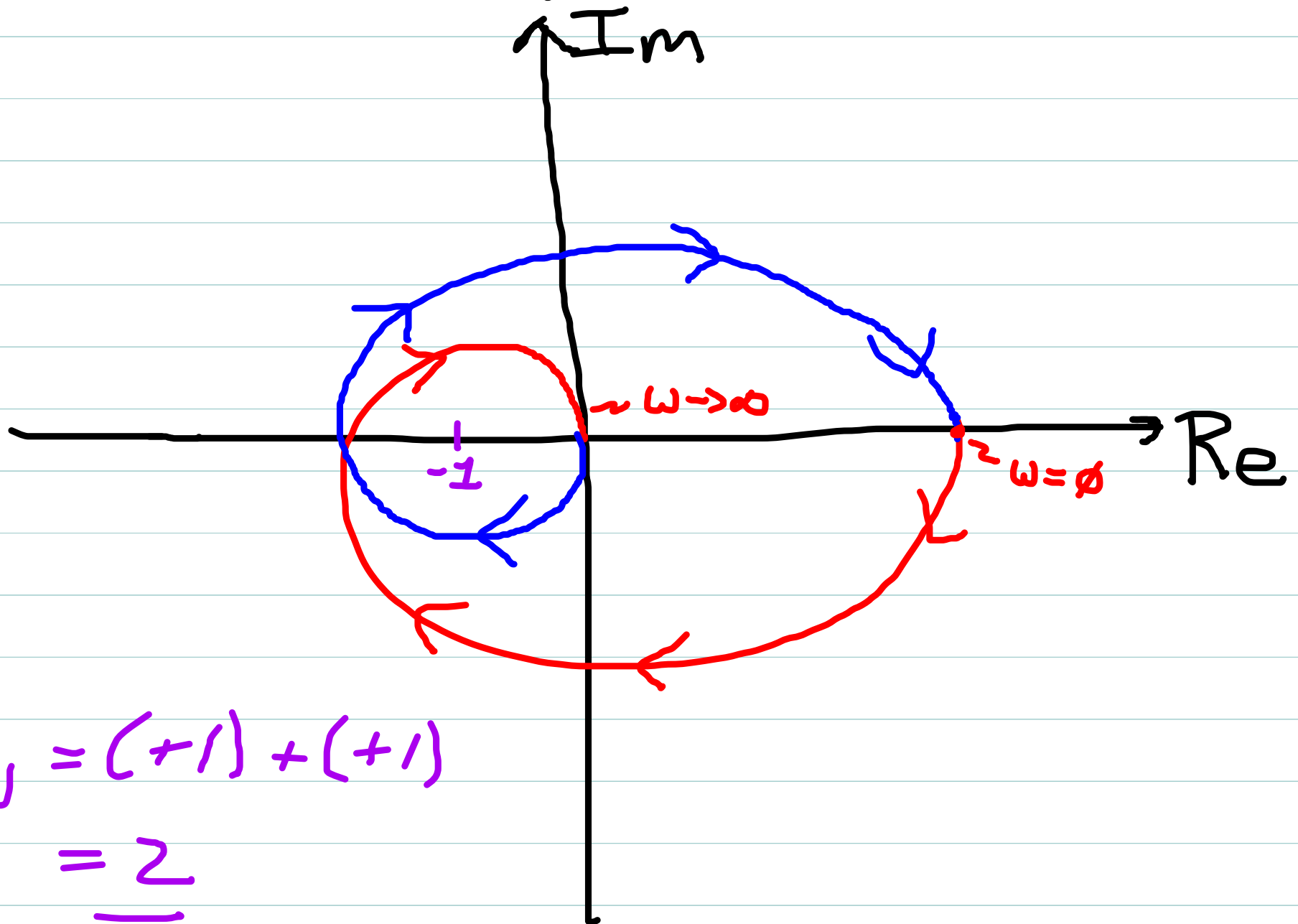
- \Rightarrow Count the number of complete loops the diagram makes around -1.
- \Rightarrow A clockwise loop counts as +1 encirclement
A counter-clockwise loop counts as -1 encirclement
- \Rightarrow Diagrams may have both CW or CCW loops around -1
- \Rightarrow Let $N_{cw}(L)$ be the net number of CW encirclements for Nyquist diagram of L
(i.e. result of adding contribution of each loop using the ± 1 convention above).

Example: $L(s) = \frac{K_B}{(\tau s + 1)^3}$ $K_B, \tau > 0$



$N_{CW} = 0$
(no net encirclements)

Example: $L(s) = \frac{K_B}{(\tau s + 1)^3}$ $K_B, \tau > 0$



$$N_{CW} = (+1) + (+1)$$

$$= \underline{2}$$

Easy Way to Count Encirclements

"Ray trick"

=> Draw a ray radially outward from -1 in any direction

=> Looking along the ray, away from -1

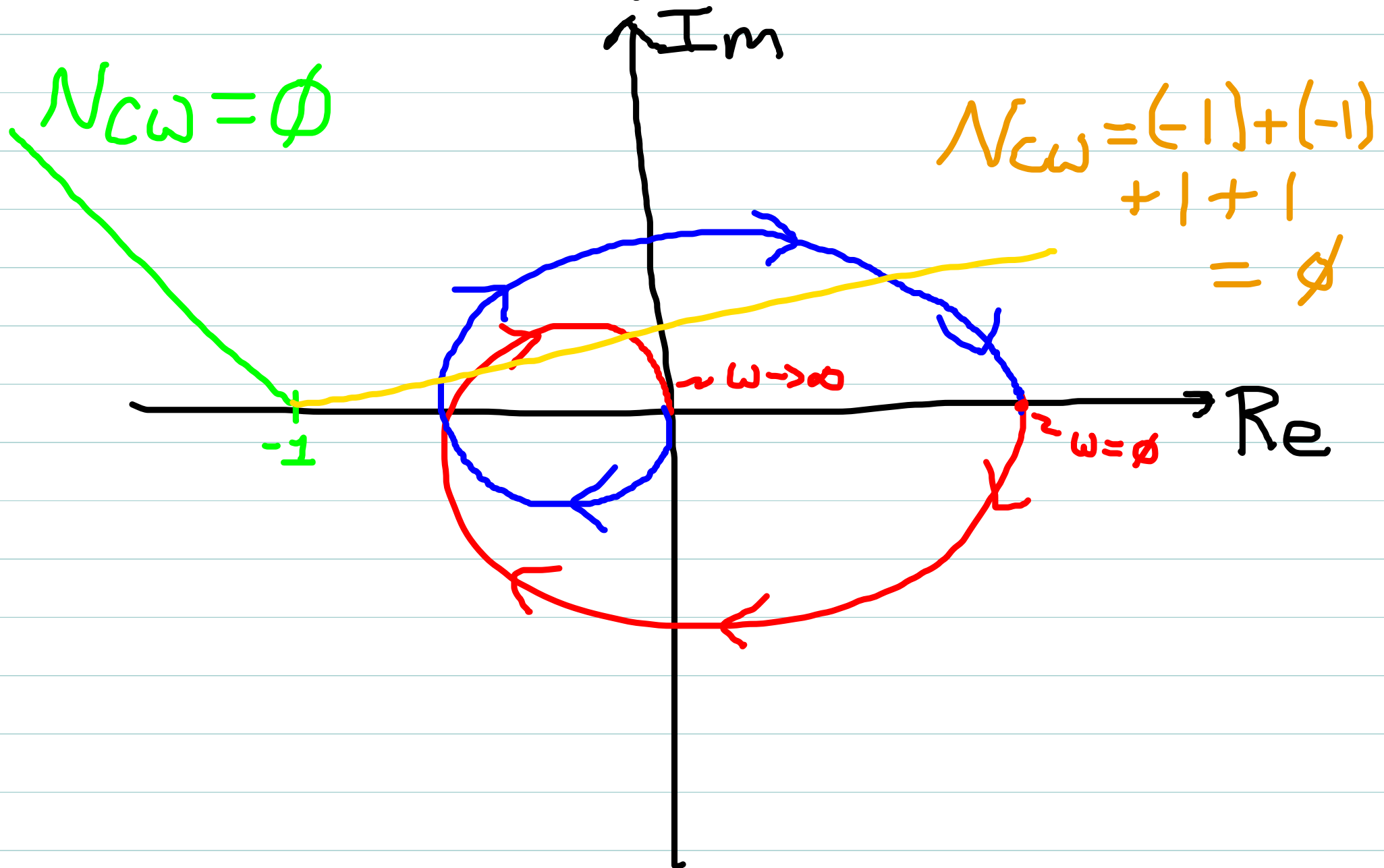
=> Count $+1$ each time diagram CROSSES
ray from left to right.

=> Count -1 each time ray is crossed right to left.

=> Same result regardless of ray direction

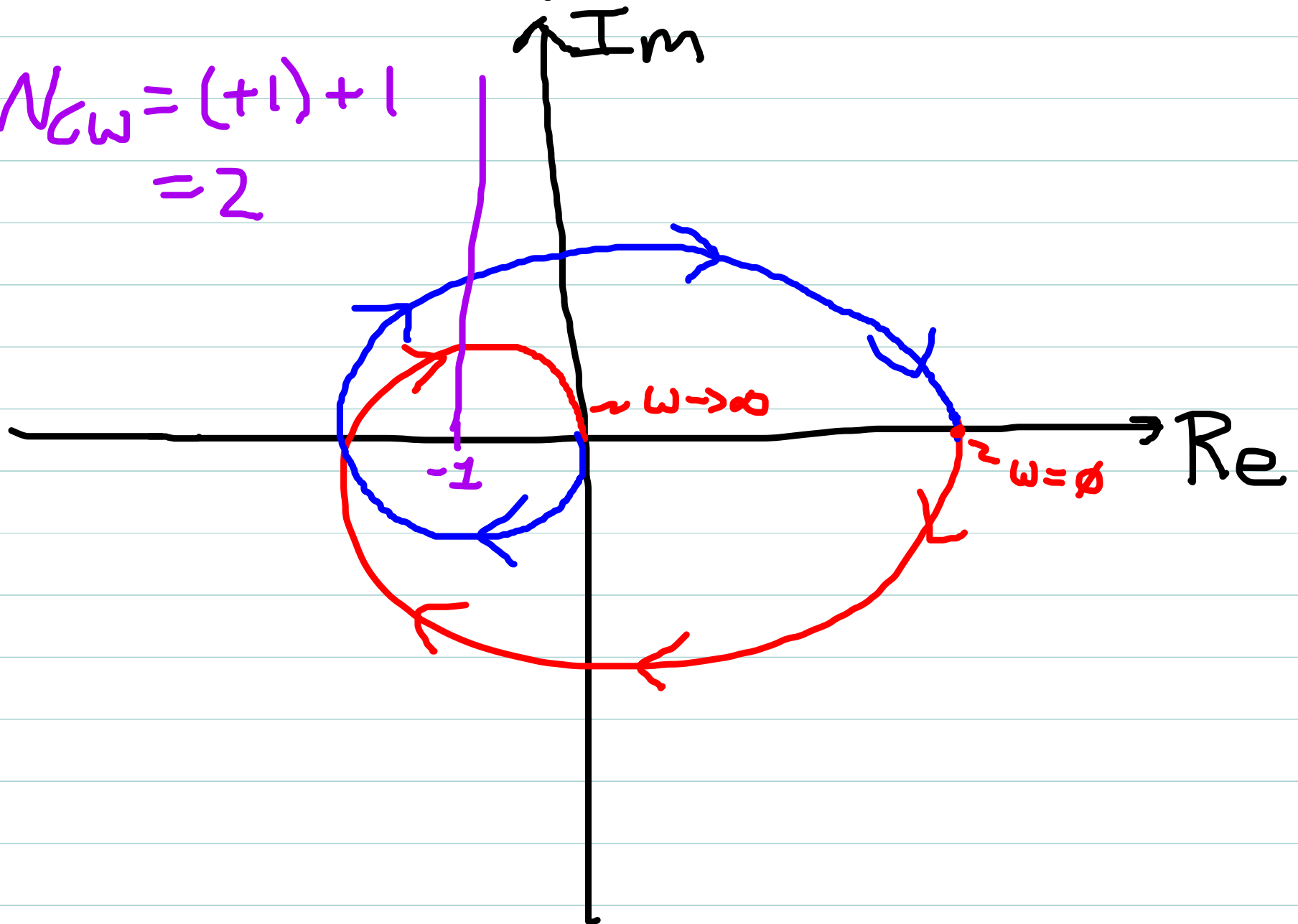
=> Choose direction with least number of intersections
for easiest counting.

Example: $L(s) = \frac{K_B}{(\tau s + 1)^3}$



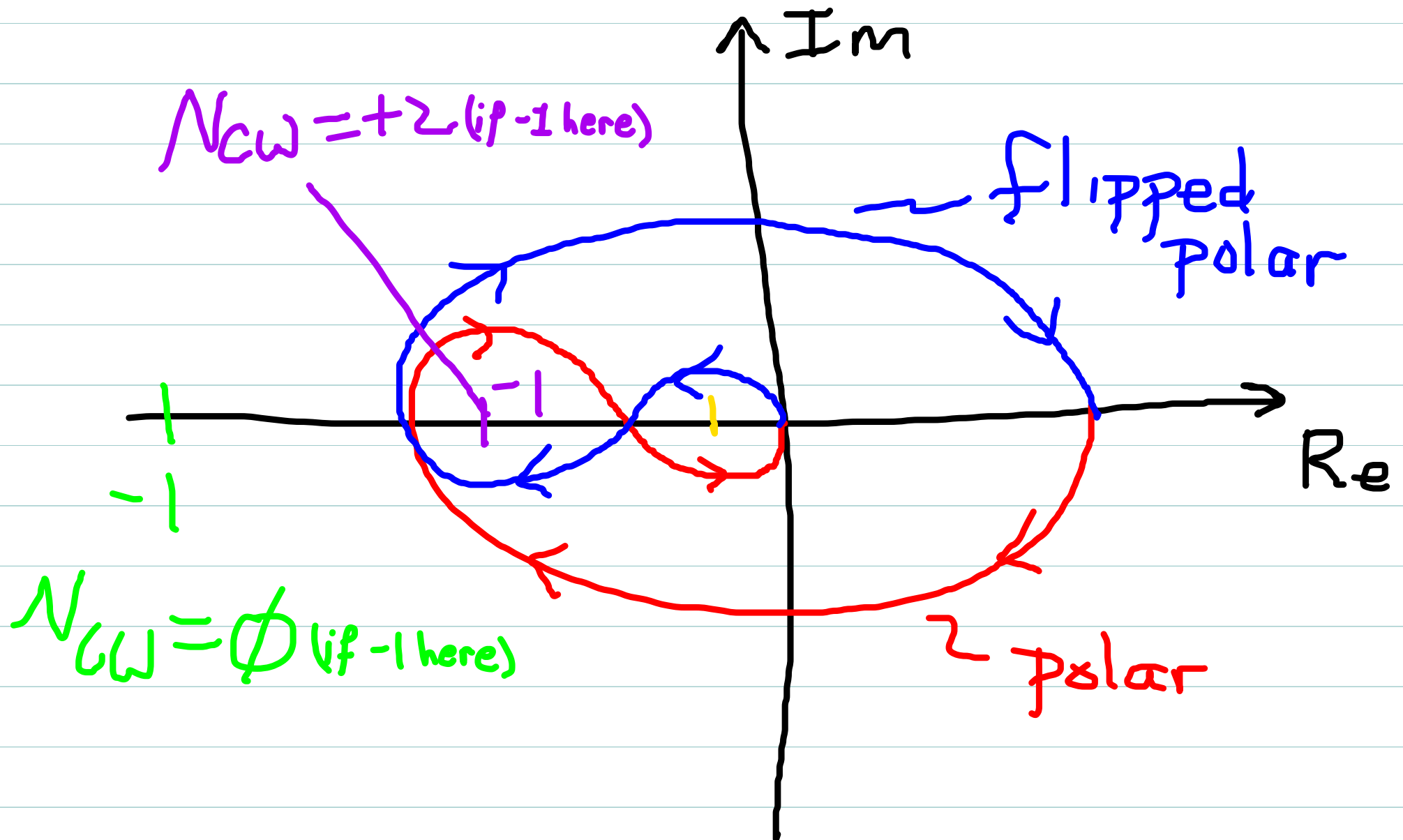
Example: $L(s) = \frac{K_B}{(\tau s + 1)^3}$

$$N_{CW} = (+1) + 1 = 2$$



A more complicated Example:

$$L(s) = \frac{K_B(\tau_1 s + 1)^2}{(\tau_2 s + 1)^3} \quad \tau_2 \gg \tau_1 > 0$$

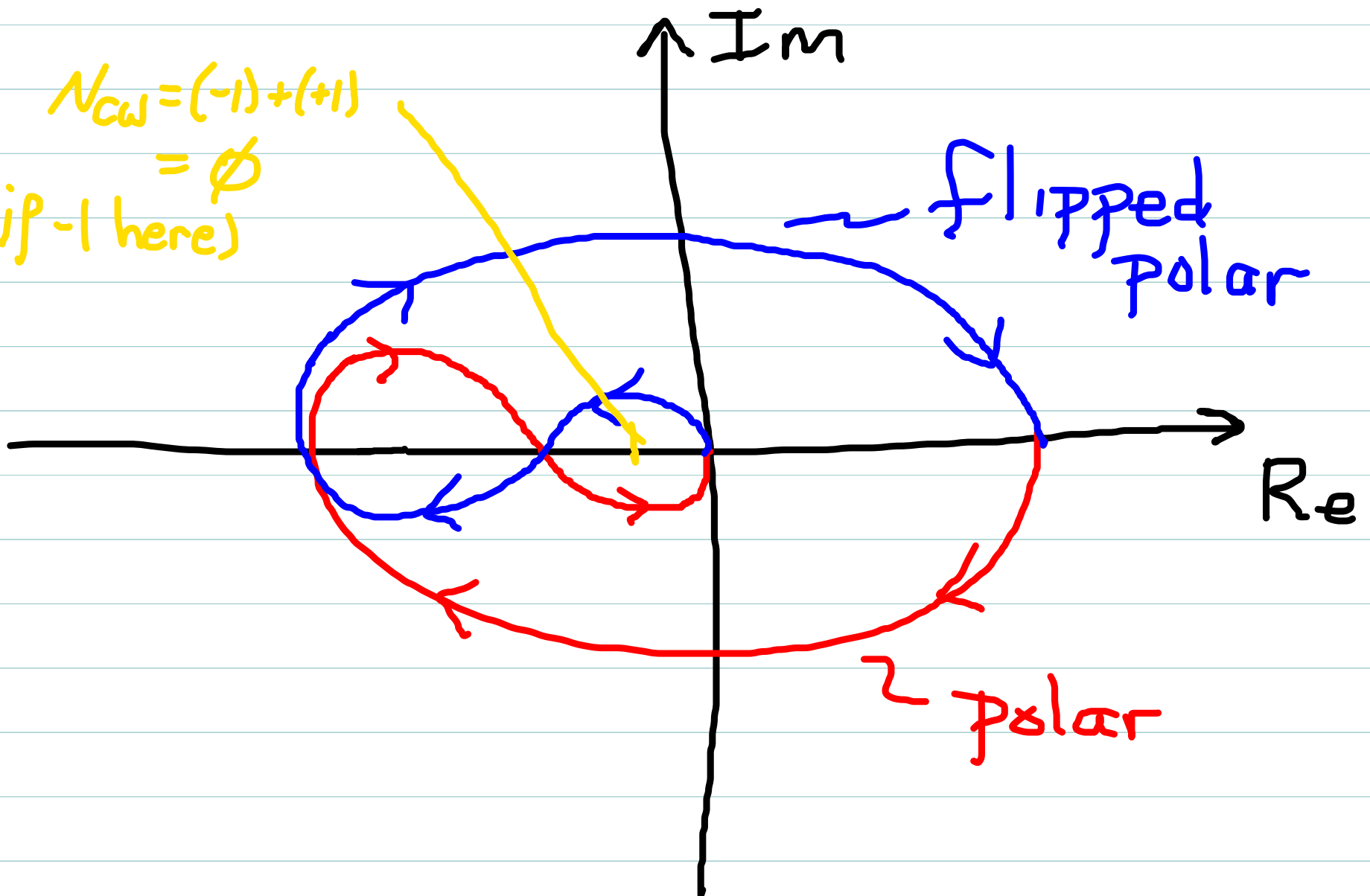


A more complicated Example:

$$L(s) = \frac{K_B(\tau_1 s + 1)^2}{(\tau_2 s + 1)^3} \quad \tau_2 \gg \tau_1 > 0$$

$$N_{cw} = (-1) + (+1) = 0$$

(if -1 here)



Nyquist Stability Theorem

For an arbitrary transfer function $G(s)$, define

$$P_R(G) = \# \text{RHP (unstable) poles of } G(s)$$

Nyquist showed:

$$N_{cw}(L) = P_R(T) - P_R(L)$$

Re-arranging:

$$P_R(T) = P_R(L) + N_{cw}(L)$$

want to predict

— Known

Note: $P_R(T) \geq 0$ always. If you compute $P_R(T) < 0$

\Rightarrow you have drawn the diagram incorrectly, or
 \Rightarrow you have counted encirclements incorrectly.

Implication

\Rightarrow We must have $P_R(T) = \emptyset$ (Stable closed-loop system)

\Rightarrow $N_{CW}(L) = -P_R(L)$ (Stability Condition)

i.e. Nyquist diagram must show a net negative number of encirclements, equal to number of unstable poles of $L(s)$.

Recall negative CW encirclements are CCW encirclements

\Rightarrow Nyquist diagram must show a net number of CCW encirclements equal to $\#$ unstable poles of $L(s)$

Note: if $P_R(L) = \emptyset$ ($L(s)$ is stable) then the diagram must show no (\emptyset) net encirclements

A more complicated Example:

$$L(s) = \frac{k_B(\tau_1 s + 1)^2}{(\tau_2 s + 1)^3} \quad \tau_2 \gg \tau_1 > 0$$

$$P_R(L) = \emptyset$$

