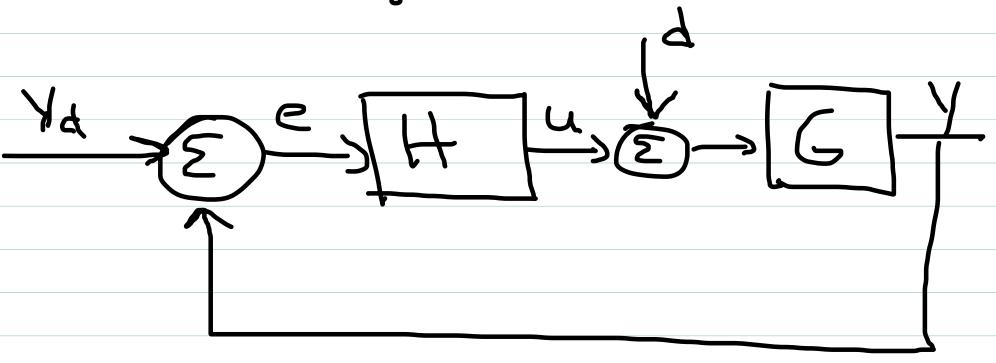
Input disturbance rejection



det external "disturbance" input to the system: Not under our direct control, and cannot be predicted or measured during operation of the system.

What affect will this have on stability or accuracy?

"rejection": ability to maintain ess(1) small even when d(t) # Ø.

Re-derive feedback loop equations:

$$=G(s)H(s)E(s)+G(s)D(s)$$

or 
$$Y(s) = \frac{L(s)}{1 + L(s)}Y_{d}(s) + \frac{G(s)}{1 + L(s)}D(s)$$

$$T(s)$$

Si(s) "input sensituity" TF.

Added term due to disturbance

Note: poles of Si(s) same as T(s) => Si(s) is stable

If T(s) is.

=> Disturbonce Cannot destabilize system!

Distribunce can however, worsen tracking:

$$Y(s) = T(s)Y_{d}(s) + S_{i}(s)D(s)$$

$$E(s) = Y_{d}(s) - Y(s) = (1 - T(s))Y_{d}(s) - S_{c}(s)D(s)$$

Want to quantify the added errors due to distribunce

Can analyze similarly to above, but need a bit more core:

$$\int_{c} (s) = \frac{G(s)}{1 + L(s)}$$

Let 
$$G(s) = \frac{N_G(s)}{D_G(s)}$$
,  $H(s) = \frac{N_H(s)}{D_H(s)}$  so  $L(s) = \frac{N_G(s) N_H(s)}{D_G(s) D_H(s)}$ 

$$S_{c}(s) = \frac{G(s)}{1 + L(s)} = \frac{N_{c}(s)D_{H}(s)}{D_{c}(s)D_{H}(s) + N_{c}(s)N_{H}(s)}$$

Then additional error:

$$S_{i}(s)D(s) = \left[\frac{N_{c}(s)D_{H}(s)}{D_{c}(s)D_{H}(s) + N_{c}(s)N_{H}(s)}\right] \left[\frac{\alpha(s)}{b(s)}\right]$$

Internal model principle again!

If  $N_6(s)D_H(s)$  cancels non-stable roots of b(s)then in steady-state  $J''\{5;D\}=\emptyset$ 

i.e. distribance creates No additional error!

## Implications:

If  $N_6(s)D_{H}(s)$  cancels non-stable roots of b(s), then cancellation is either due to:

=> No(s) cancelling (extremely rare)

=> DH(s) cancelling (can design for this)

So generally, external distribunces create NO Add'l error if compensator contains an internal movel of distribunce.

That is, if Compensator H(s) has some non-stable poles as the disturbance.

i.e. if d(t)=do (constant), No add'l tracking error
if H(s) has Pole at origin.

## Summary of error analysis

For perfect tracking of "typep" desired behaviors

L(s) must have pri poles at origin

For perfect rejection of type p disturbances d(4), but nonzero H(s) must have p+1 poles at origin errors

required poles

in Both cases, P Poles at orga

(one less) will

Note: tracking objectives can be satisfied if required poles come from plant, compensator, or a combination of both

But dist. rejection regt's can be satisfied only by poles in the compensator.

=> Above are special cases of IMP.

Good accuracy thus often requires H(s) to have at least one pole at origin.

=> This pole adds - 90° of phase at all frequencies!

=> Works against our stability/performance guidelines of increasing phase margin.

=> Even adding a LHP zero doesn't help here:

$$H(s) = K\left[\frac{s - \epsilon_c}{s}\right]$$
  $\epsilon_c < \phi$ 

has ¥H(jw)<ذ for all w≥o.

=) May be acceptable if  $\angle G(j\omega)$  already has "adequate" positive phase, so  $\angle L = \angle G + \angle H$  can tolerate a Small reduction.

| More go<br>to still<br>despite | enerally, | me, 9    | Pequire | exta | LHP    | <del>Ze</del> ro(s) |
|--------------------------------|-----------|----------|---------|------|--------|---------------------|
| to \$711                       | Provide   | Position | e Phase | cha  | nges t | 6 L(s)              |
| despite                        | required  | Doke o   | forigin |      |        |                     |
|                                | 7         |          | · 7     |      |        |                     |

Implementability requires an additional LHP pole:

$$H(s) = \left\{ \frac{(s-2c_1)(s-2c_2)}{5(s-2c_1)} \right\}$$

4 degrees of freedom total!

Things get even more complicated if H(s) needs

2 poles at origin to achive bracking
objectives!

Remember: Tracking of Ya(+) depends on properties of L(s)

Disturbance rejection depends on proporties of H(s)