## Zero Order Hold Filtering and Recovery

lets assume a signal x(t) is given by:

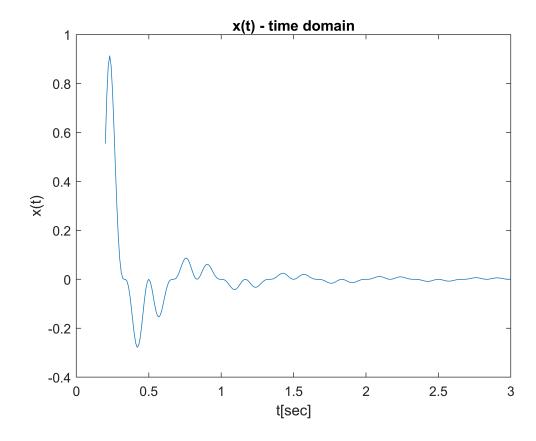
$$x(t) = \frac{4}{\omega_m \pi t^2} \cdot \sin^2(\omega_m t) \cos(\omega_m t) \sin(2\omega_m t)$$

where:

 $\omega_m = 3\pi$ 

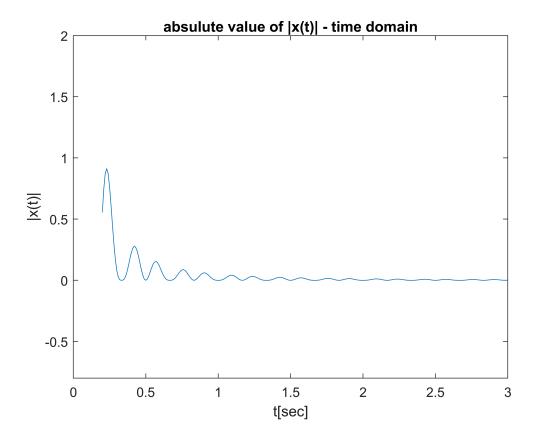
```
wm = 3*pi;
t = 0.2:1/100:3;
x = @(t) 4./(wm*pi*t.^2).*(sin(wm*t)).^2.*(cos(wm*t)).*(sin(2*wm*t));
ylim([-0.8 1.6])
```

let'stake a look on x(t) in time:



we are asked to display the absulute value |x(t)|:

```
plot(t,abs(x(t))); title ('absulute value of |x(t)| - time domain'); ylabel('|x(t)|'); xlabel('tylim([-0.8 2])
```



$$X(\omega) = \mathcal{F}\{x(t)\}$$

$$= \mathcal{F}\left\{\frac{4}{\omega_m \pi t^2} \cdot \sin^2(\omega_m t) \cos(\omega_m t) \sin(2\omega_m t)\right\}$$

here it is apropriate to use the convolution quality:

$$=\frac{4}{\omega_m\pi}\mathcal{F}\left\{\frac{\sin(\omega_mt)}{t}\right\}*\mathcal{F}\left\{\frac{\sin(\omega_mt)}{t}\right\}*\mathcal{F}\left\{\sin(2\omega_mt)\right\}*\mathcal{F}\left\{\cos(\omega_mt)\right\}$$

$$=\frac{4}{\pi\omega_{m}}\left[\Pi\left(\frac{\omega}{2\omega_{m}}\right)*\Pi\left(\frac{\omega}{2\omega_{m}}\right)\right]*\left[\pi(\delta(\omega-2\omega_{m})-\delta(\omega+2\omega_{m}))*\pi(-i)(\delta(\omega-\omega_{m})+\delta(\omega+\omega_{m}))\right]$$

notice that convolving with dirac's deltas is equivalent to time shifting:

$$= -\frac{4}{\pi\omega_m}i\left[\Lambda\left(\frac{\omega}{2\omega_m}\right)\right]*\left[-\pi i(\delta(\omega-2\omega_m)-\delta(\omega+2\omega_m))\right]*\left[\pi(\delta(\omega-\omega_m)+\delta(\omega+\omega_m))\right]$$

$$=\frac{1}{\omega_{m}}\left[\Lambda\left(\frac{\omega-2\omega_{m}}{2\omega_{m}}\right)-\Lambda\left(\frac{\omega+2\omega_{m}}{2\omega_{m}}\right)\ \right]*\pi\left(\delta(\omega-\omega_{m})+\delta(\omega+\omega_{m})\right)$$

$$X(\omega) = \frac{\pi}{\omega_m} \left[ \Lambda \left( \frac{\omega - 3\omega_m}{2\omega_m} \right) + \Lambda \left( \frac{\omega - \omega_m}{2\omega_m} \right) - \Lambda \left( \frac{\omega + 3\omega_m}{2\omega_m} \right) - \Lambda \left( \frac{\omega + \omega_m}{2\omega_m} \right) \right]$$

we got four lambda signals in frequency domain.

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from the above expression it is clear that  $X(\omega)$  ranges from  $\omega = -5\pi$  up to  $5\pi$  in frequency domain.

meaning  $X(\omega) = 0 \ \forall |\omega| \ge 5\pi$ .

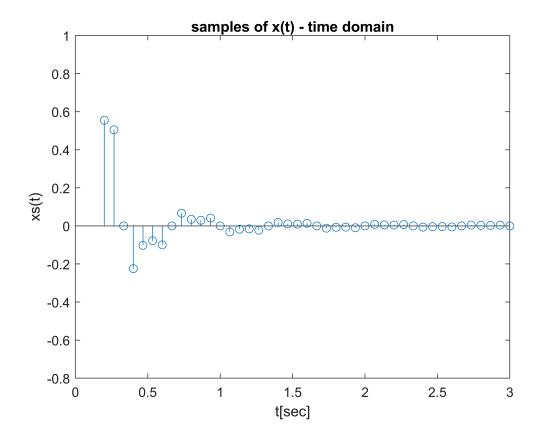
said realizaton provides us with sufficient terms for sampling in Nyquist rate.

here we will use:  $\omega_s = \omega_{\rm nyquist} = 2\omega_{\rm max} = 10\omega_{\rm m}$ 

let's sample the signal

```
ws = 10 * wm;
Ts = 2*pi/(ws);
pT = 0.2:Ts:3;
N = 0:(floor(3/Ts)+1);

xs = x(pT);
stem(pT,xs); title ('samples of x(t) - time domain'); ylabel('xs(t)'); xlabel('t[sec]');
ylim([-0.8 1])
```

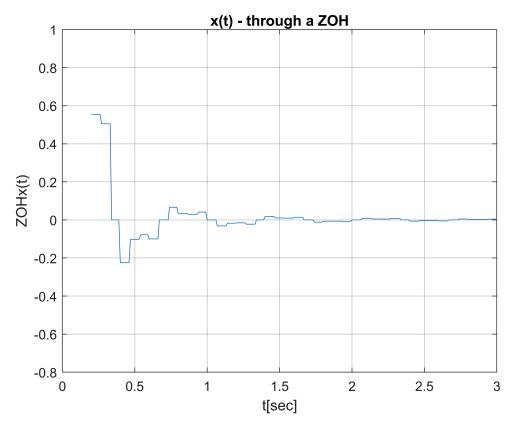


now we will implement a zero order hold:

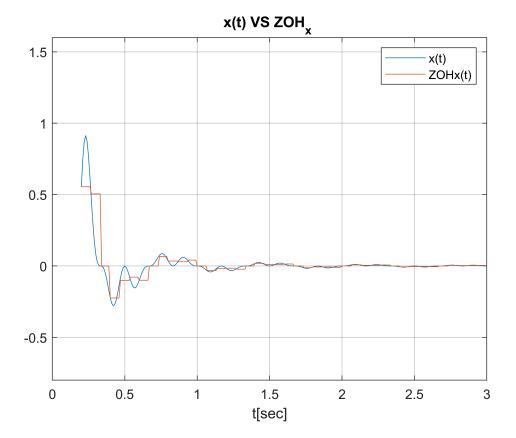
```
      x_ZOH = xs(floor((t-0.2)/Ts)+1); \\ plot(t,x_ZOH); title ('x(t) - through a ZOH'); ylabel('ZOH{x(t)}'); xlabel('t[sec]'); \\ ylim([-0.8 1])
```

we will display them both on one grid:

```
grid on;
```



```
plot(t,x(t));
hold on
plot(t,x_ZOH);
title ('x(t) VS ZOH_x '); legend('x(t)','ZOH{x(t)}'); xlabel('t[sec]');
ylim([-0.8 1.6])
grid on;
hold off;
```



$$X_{\rm ZOH}(\omega) = \mathcal{F}\big\{x_{\rm ZOH}(t)\big\}$$

$$= \mathcal{F}\left\{\sum_{n=0}^{\infty} x(n\cdot T_s)\cdot \Pi\left(\frac{t-\left(n+\frac{1}{2}T_s\right)}{T_s}\right)\right\}$$

$$= \sum_{n=0}^{\infty} x(n \cdot T_s) \cdot \mathcal{F} \left\{ \Pi \left( \frac{t - \left( \left( n + \frac{1}{2} \right) T_s \right)}{T_s} \right) \right\}$$

$$= \sum_{n=0}^{\infty} x(n \cdot T_s) \cdot e^{-i \, n \omega} \, \mathcal{F} \left\{ \prod \left( \frac{t - \frac{1}{2} \, T_s}{T_s} \right) \right\}$$

$$= \mathcal{F}\left\{ \prod \left( \frac{t - \frac{1}{2} T_s}{T_s} \right) \right\} \sum_{n=0}^{\infty} x(n \cdot T_s) \cdot e^{-i n \cdot T_s \omega}$$

here we may recognize a familliar form:

 $\sum_{n=0}^{\infty} x(n \cdot T_s) \cdot e^{-i (n \cdot T_s) \omega}$  is no other then then the DTFT of the sampled x(t) signal!

$$= X_P(\omega)$$

$$X_{\rm ZOH}(\omega) = \mathcal{F}\left\{\Pi\left(\frac{t-\frac{1}{2}T_s}{T_s}\right)\right\}X_P(\omega)$$

$$= e^{-i \cdot \frac{1}{2} T_s \cdot \omega} \mathscr{F} \left\{ \Pi \left( \frac{t}{T_s} \right) \right\} X_P(\omega)$$

$$= e^{-i \cdot \frac{1}{2} T_s \cdot \omega} T \cdot \operatorname{sinc}\left(\frac{\omega}{\omega_s}\right) X_P(\omega)$$

$$= e^{-i \cdot \frac{1}{2} T_s \cdot \omega} T \cdot \operatorname{sinc}\left(\frac{\omega}{\omega_s}\right) \frac{1}{T} \sum_{n=-\infty}^{\infty} X(\omega - n\omega_s)$$

$$= e^{-i \cdot \frac{1}{2} T_s \cdot \omega} \cdot \operatorname{sinc}\left(\frac{\omega}{\omega_s}\right) \sum_{n=-\infty}^{\infty} X(\omega - n\omega_s)$$

without further a due, we shall represent  $X_{\text{ZOH}}(\omega)$  in our code.we already have most of the needed input, so it will be a good idea to use this form:

$$X_{\mathrm{ZOH}}(\omega) = = e^{-i \cdot \frac{1}{2} T_s \cdot \omega} \cdot \operatorname{sinc}\left(\frac{\omega}{\omega_s}\right) \sum_{n=-\infty}^{\infty} X(\omega - n\omega_s)$$

$$w = -17*pi:1/100:17*pi;$$

```
TRI = @(w,s) triangularPulse((w+s*wm)/(2*wm));

X = @(w) -1i*(TRI(w,-3)+TRI(w,-1)-TRI(w,1)-TRI(w,3));
%plot(w,X(w)); title ('X(w)- frequancy domain'); ylabel('X(w)'); xlabel('w[sec-1]');
```

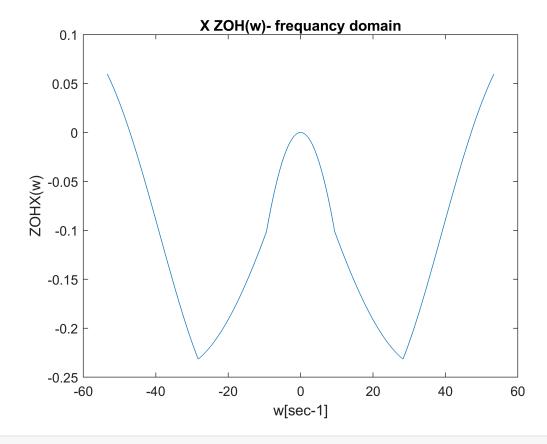
for now we have  $X(\omega)$  correctly,  $X_p(\omega)$  would be duplications of it:

```
X_P = @(w) (pi/wm)*(X(w-ws) +X(w) +X(w+ws));
time_shift = @(w) exp(-1i*0.5*Ts*w);

X_ZOH = time_shift(w).*sinc(w/ws).*X_P(w);
plot(w,X_ZOH); title ('X ZOH(w)- frequency domain'); ylabel('ZOH{X(w)}'); xlabel('w[sec-1]');
```

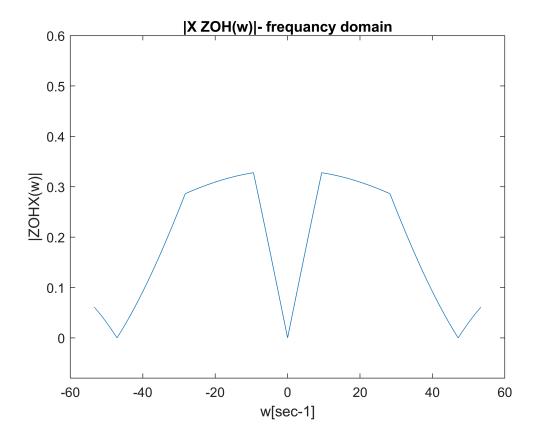
Warning: Imaginary parts of complex X and/or Y arguments ignored

hold off



now let's see the absulute value of  $X_{\text{ZOH}}(\omega)$ 

```
plot(w,abs(X_ZOH)); title ('|X|ZOH(w)|- frequancy domain'); ylabel('|ZOH\{X(w)\}|'); xlabel('|X|ZOH(w)|- frequancy domain'); ylabel('|X|ZOH(w)|- ylabel
```



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HEA

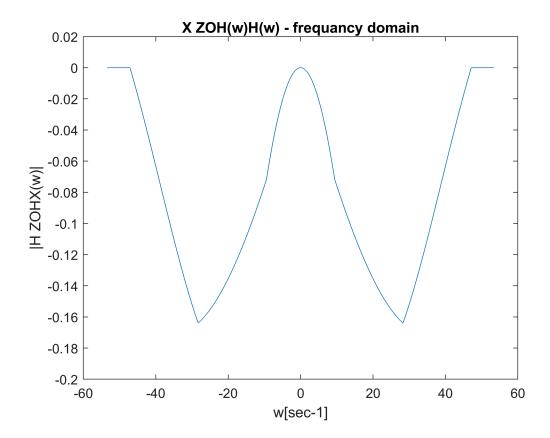
let us define  $H(\omega)$ :

```
filter = @(w) exp(1i*pi*w/ws)/(sinc(w/ws));
H = @(w) (filter(w))*(abs(w) < (ws/2));</pre>
```

now is the time to multiply  $H(\omega)$  with  $X_{\text{ZOH}}(\omega)$  to get the filterd signal:

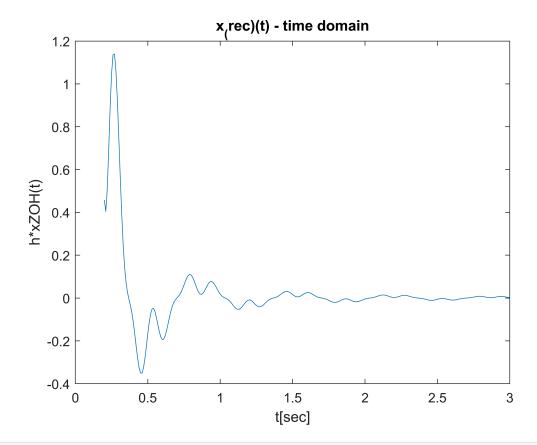
```
X_rec = H(w).*X_ZOH;
plot(w, X_rec);title ('X ZOH(w)H(w) - frequency domain'); ylabel('|H {ZOH{X(w)}}|'); xlabel('w)
Warning: Imaginary parts of complex X and/or Y arguments ignored

ylim([-0.2 0.02])
```



as we can see here, all other clones of  $X(\omega)$  are eliminated, and only the one around  $\omega = 0$  has an effect. now, the inverse fourier transfom shuld return the original signal x(t)

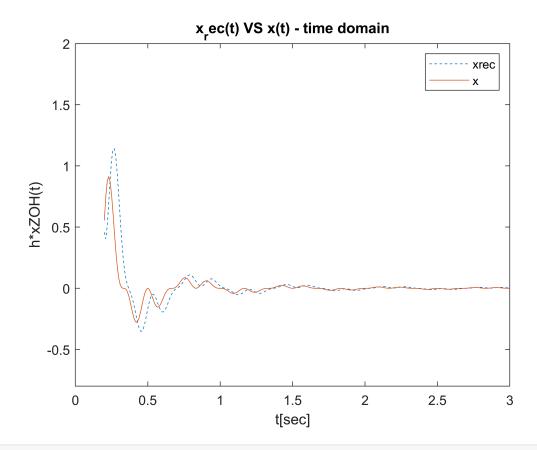
Warning: Imaginary parts of complex X and/or Y arguments ignored



lets plot  $x_{rec}(t)$  and the original x(t) toether:

```
plot(t,x_rec,'--', t, x(t)); title ('x_rec(t) VS x(t) - time domain'); ylabel('h*xZOH(t)'); xla
Warning: Imaginary parts of complex X and/or Y arguments ignored

legend('xrec' , 'x')
ylim([-0.8 2])
```



a little comment deep from the heart might be in place here. the signal *shuld have* mached percisely its recoverd form. both theory, and specific calculations show that, and we brought them bothhere in our work. but after hours and hours of debugging i can say that I lost my left and right trying to find the bug causing the mismatch, without success. my hatred for matlab has meanwhile grew to a ranging fire, that burns with the fierce passion of a million suns. we know what to expact, and are **confident** enough in the studied material to say: there is an error, and it suld be confessed, matlab is not my favourite coding platform.

of curse x(t) cannot be recoverd from equally spaced samples at sampling rate  $\omega_s$  because such sub-nyquist rate would cause *aliasing* in  $X_p(\omega)$ 

```
ws = 9*wm;
X_P = @(w) (1/3)*(X(w-ws) + X(w) + X(w+ws));
time\_shift = @(w) exp(-1i*0.5*(2*pi/ws)*w);
```

```
legend('xrec' , 'x')
ylim([-0.8 2])
```

