

# Introduction to Electrical Circuits

## Final Term Lecture - 09

### Reference Book:

[1] Principles of Electrical Machines -V.K. Mehta, Rohit Mehta

[2] A Textbook of Electrical Technology , Volume- II, - B.L. Theraja, A.K. Theraja



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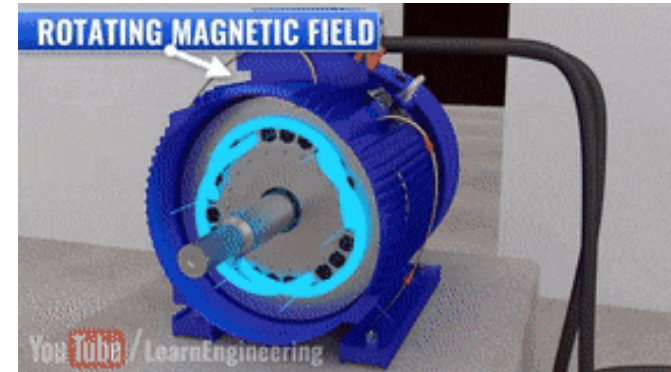


Week No.	Class No.	Chapter No.	Article No., Name and Contents	Example No.
W12	FC9		<b>Introduction to Induction Motor:</b> Basic principles of an induction motor, double field and cross field theory, making self-starting technique, difference between capacitor start and capacitor run motor with respect to applications (prob. 6.1, 6.2, 36.3). Draw the circuit and vector diagram.	36.1, 36.2, 36.3.



# Induction motor: Basic Principle

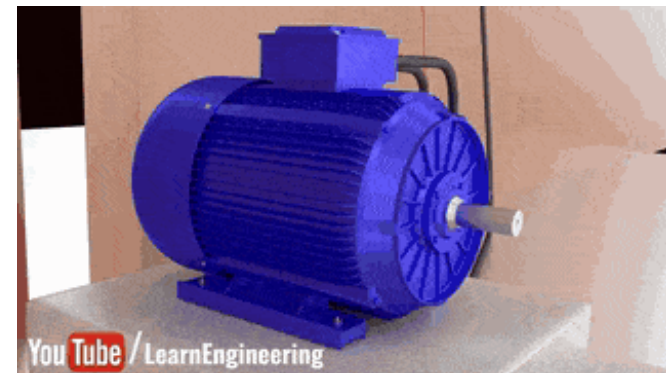
- The **Induction Motor**, the most versatile of the AC motors, has truly emerged as the prime mover in industrial applications.
- In A.C. motors, the rotor does receive electric power by **induction** in the same way as the secondary of a 2-winding transformer receives its power from the primary.
- They can be treated as **rotating transformer** i.e. one in which the primary winding is stationary, but the secondary is free to rotate.



An induction motor has two main parts -

- (i) Stator - The stationary section that contain the windings (magnetic field).
- (ii) Rotor - The rotating section that contains the conductors.

The rotor is separated from the stator by a small air-gap which ranges from 0.4 mm to 4 mm, depending on the power of the motor.



# Single-Phase Induction Motor

A single-phase induction motor is very similar to a 3-phase squirrel cage induction motor. It has -

- (i) a squirrel-cage rotor identical to a 3-phase motor and
  - (ii) a single-phase winding on the stator.
- A single-phase induction motor is not *self-starting*, it does not inherently develop any starting torque.
  - However, if the rotor is started by auxiliary means, it will continue to run in the direction of rotation and quickly accelerates until it reaches a speed slightly below the synchronous speed.
  - This strange behavior of the single-phase motor can be explained in two ways:
    - (i) by double-field revolving theory, and
    - (ii) by cross-field theory.

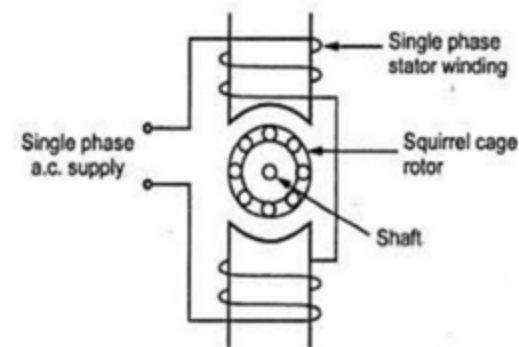


Fig: Single-phase induction motor



# Double-Field Revolving Theory

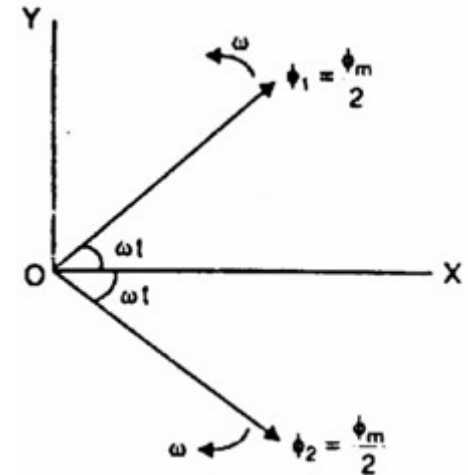
This theory based on the fact that the **alternating sinusoidal flux** produced by the stator winding can be represented by **two revolving fluxes**.

Each equal to one-half of the maximum value of alternating flux (i.e.,  $\Phi_m / 2$ ) and each rotating at synchronous speed ( $N_s = 120 f / P$ ,  $\omega = 2\pi f$ ) in opposite directions.

The instantaneous value of flux due to the stator current of a single-phase induction motor is given by;

$$\Phi = \Phi_m \cos \omega t$$

Consider two rotating magnetic fluxes, each be equal to  $\Phi_m/2$  start revolving from OX axis at  $t = 0$  in anti-clockwise and clockwise directions respectively, with angular velocity  $\omega$ .



# Double-Field Revolving Theory

- After time  $t$  seconds, the angle through which the flux vectors have rotated is  $\theta = \omega t$ . Resolving the flux vectors along-X-axis and Y-axis, we have,

$$\text{Total X-component} = \frac{\phi_m}{2} \cos \omega t + \frac{\phi_m}{2} \cos \omega t = \phi_m \cos \omega t$$

$$\text{Total Y-component} = \frac{\phi_m}{2} \sin \omega t - \frac{\phi_m}{2} \sin \omega t = 0$$

$$\text{Resultant flux, } \phi = \sqrt{(\phi_m \cos \omega t)^2 + 0^2} = \phi_m \cos \omega t$$

- When the rotating flux vectors are in phase the resultant vector is  $\phi = \phi_m$ ; when out of phase by  $180^\circ$ , the resultant vector  $\phi = 0$  [Fig.-2].
- At standstill, two torques produced by opposite revolving fluxes are equal and opposite and the net torque developed is zero. Therefore, single-phase induction motor is not self-starting [Fig.-3].
- If the motor is once started, it will develop net torque in the direction in which it has been started and will function as a motor [Fig.-3].

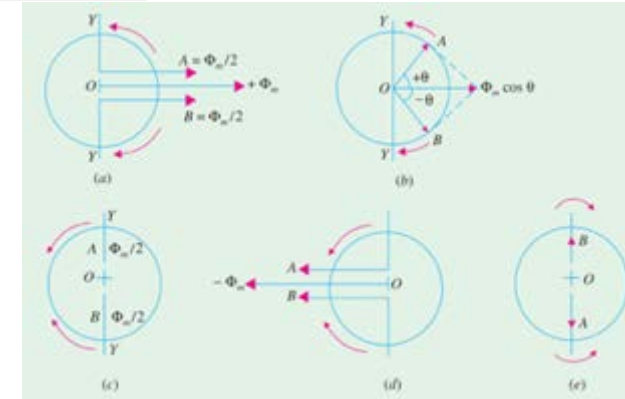


Fig. 1

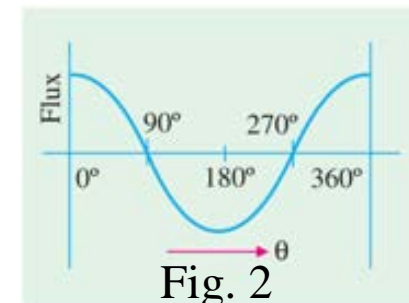


Fig. 2

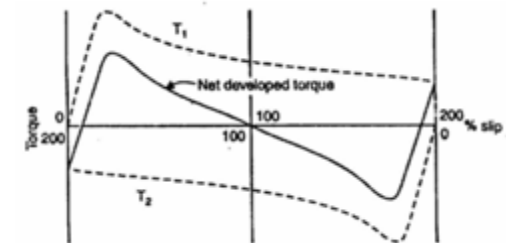


Fig. 3

# Cross-Field Theory

- The quadrature pulsating rotor field reacts against the pulsating main field to produce a resultant magnetic field. The resultant magnetic field is a fairly constant rotating magnetic field that rotates in the same direction as the direction of the rotation of the rotor.
- A squirrel-cage induction motor will continue to rotate, producing induction motor torque in a rotating magnetic field, once a rotational emf has been initiated.

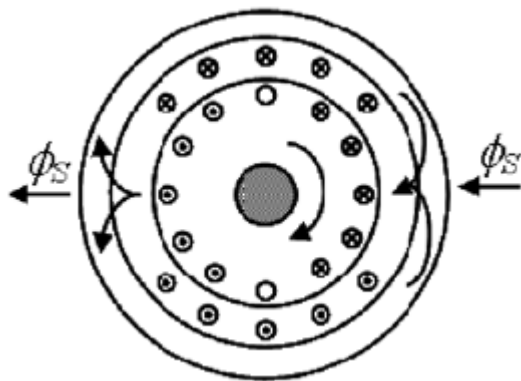


Fig: Stator field  $\Phi_s$  sets up flux along horizontal axis; rotor rotating in clockwise direction



Fig: Cross field when stator field is zero.

# Making Single-Phase Induction Motor Self-Starting

- To make a single-phase induction motor self-starting, phase splitting is done by temporarily converting a two-phase motor during the starting period.
- The stator is provided with an extra **starting (auxiliary)** winding, in addition to the **main (running)** winding.
- The two windings are spaced **90° electrically apart** and are connected in parallel across the single-phase supply.
- When the motor attains sufficient speed, the starting (auxiliary) winding may be removed depending upon the type of the motor.
- There are many methods by which the necessary phase-difference between the two currents can be created.
  - (i) **Split-phase motors** - started by two phase motor action through the use of an auxiliary or starting winding.
  - (ii) **Capacitor motors** - started by two-phase motor action through the use of an auxiliary winding and a capacitor.
  - (iii) **Shaded-pole motors** - started by the motion of the magnetic field produced by means of a shading coil around a portion of the pole structure.

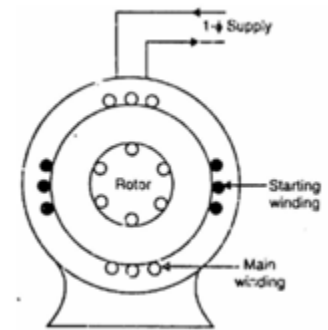


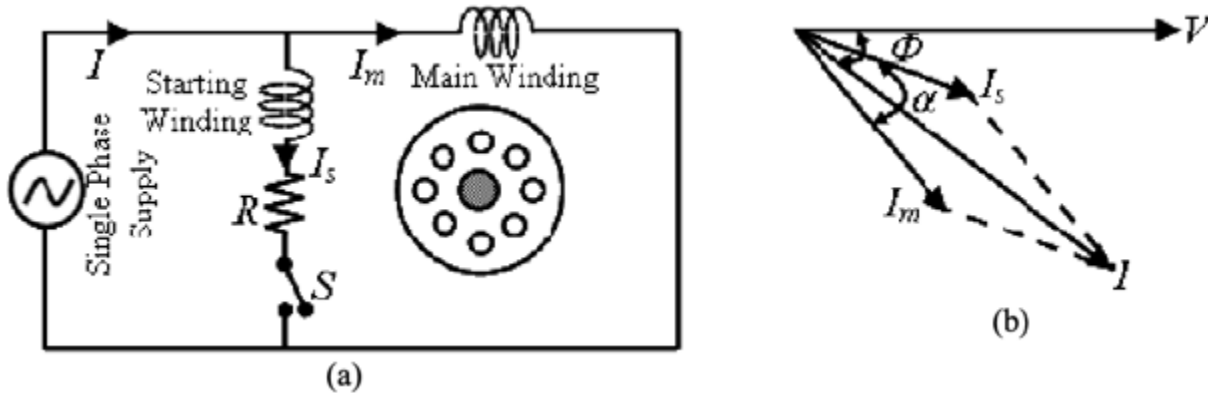
Fig: Auxiliary winding in stator to make 1- $\Phi$  motor self starting





# Split-Phase Induction Motors

The single-phase induction motor is equipped with an auxiliary winding in addition to the main winding. The starting winding is connected in parallel with the main running winding.

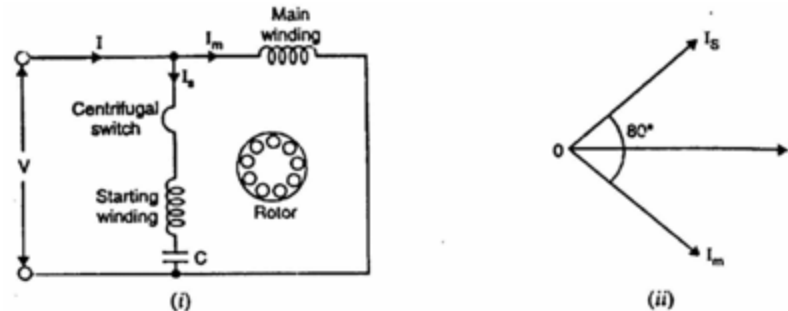


- Due to their low cost, split-phase induction motors are most popular single-phase motors in the market.
- An important characteristic of these motors is that they are essentially constant-speed motors. The speed variation is 2-5% from no-load to full-load.
- These motors are suitable where a **moderate starting torque** is required and where starting periods are infrequent e.g., to drive: (a) fans (b) washing machines (c) oil burners (d) small machine tools etc.
- The power rating of such motors generally lies between **60 W and 250 W**.



# Capacitor-Start Motor

- The capacitor-start motor is identical to a split-phase motor except that the starting winding has as many turns as the main winding.
- Moreover, a capacitor  $C$  is connected in series with the starting winding.



- Although starting characteristics of a capacitor-start motor are better than those of a split-phase motor, both machines possess the same running characteristics because the main windings are identical.
- The phase angle between the two currents is about  $80^\circ$  compared to about  $25^\circ$  in a split-phase motor.
- The starting winding of a capacitor start motor heats up less quickly and is well suited to applications involving either frequent or prolonged starting periods.
- Capacitor-start motors are used where **high starting torque** is required and where the starting period may be long e.g., to drive: (a) compressors (b) large fans (c) pumps (d) high inertia loads
- The power rating of such motors lies between **120 W and 7.5 kW**.



# Difference between capacitor start and capacitor run motor

- Capacitor-run motor is identical to a capacitor-start motor except that starting (auxiliary) winding is not opened after starting so that both the windings remain connected to the supply when running as well as at starting.
- In Capacitor-run motor, a single capacitor  $C$  is used for both starting and running. This design eliminates the need of a centrifugal switch and at the same time improves the power factor and efficiency of the motor.
- In the Capacitor-start motor, two capacitors  $C_1$  and  $C_2$  are used in the starting winding. The smaller capacitor  $C_1$  required for optimum running conditions is permanently connected in series with the starting winding. The much larger capacitor  $C_2$  is connected in parallel with  $C_1$  for optimum starting and remains in the circuit during starting. The starting capacitor  $C_2$  is disconnected when the motor approaches about 75% of synchronous speed. The motor then runs as a single-phase induction motor.
- Because of the capacitor-run motor produces a constant torque, it is vibration free and can be used in: (a) hospitals (b) studios and (c) other places where silence is important.

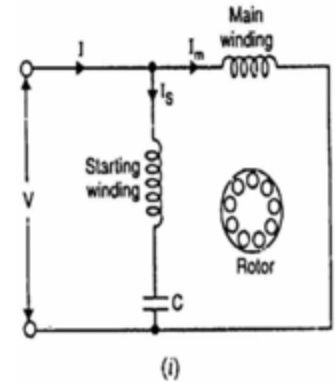


Fig: Capacitor Run Motor

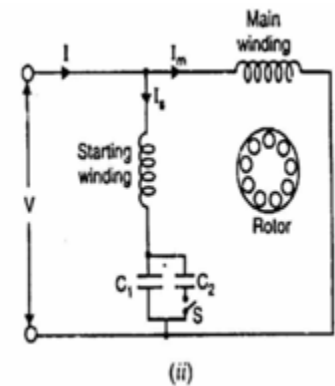


Fig: Capacitor Start Motor



# Example Problems

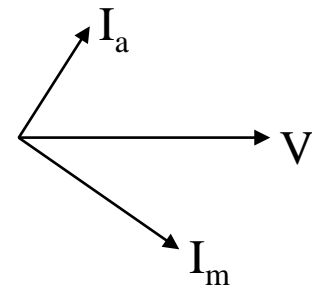
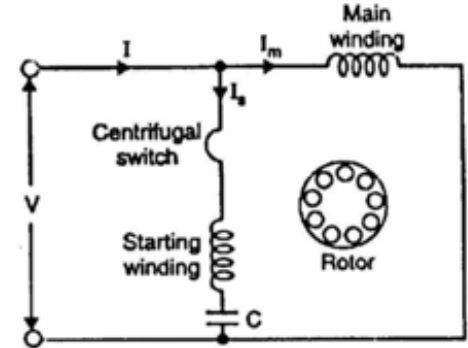
**Problem [6.1]:** The impedance of the main and auxiliary windings of a 50 Hz single-phase induction motor are  $Z_m = (3+j3) \Omega$  and  $Z_a = (6+j3) \Omega$  respectively. What will be the value of the capacitor to be connected in series with auxiliary winding to achieve a phase difference of  $90^\circ$  between the currents of the two windings?

$$I_m = \frac{V \angle 0}{3 + j3} = \frac{V \angle 0}{4.24 \angle 45^\circ} = \frac{V \angle -45^\circ}{4.24} \quad I_a = \frac{V \angle 0}{6 + j3} = \frac{V \angle 0}{6.7 \angle 26.5^\circ} = \frac{V \angle -26.5^\circ}{6.7}$$

The current flowing through the auxiliary winding after connecting a capacitor  $C$  in series should make an angle  $90^\circ$  with  $I_m$  or make an angle  $90^\circ - 45^\circ = 45^\circ$  with the applied voltage  $V$ . Since the new current of auxiliary winding should be leading the voltage  $V$  by an angle of  $45^\circ$ , the capacitive reactance of the auxiliary circuit is greater than the inductive reactance. Thus

$$\tan 45^\circ = \frac{X_c - X_L}{R} \quad 1 = \frac{(1/\omega C) - 3}{6} \quad (1/\omega C) - 3 = 6 \quad (1/\omega C) = 9$$

$$\omega C = 1/9; \quad C = 1/9\omega \quad C = 353.6 \mu F$$



# Example Problems

**Problem [6.2]:** A 50 Hz Capacitor-Start induction motor has a resistance  $5\ \Omega$  and an inductive reactance  $20\ \Omega$  in both main and auxiliary winding. Determine the value of resistance and capacitance to be added in series with auxiliary winding to send the same current in each winding with a phase difference of  $90^\circ$ .

$$Z = 5 + j20 = 20.6\angle 76^\circ$$

$$I_m = I_a = \frac{V\angle 0}{5 + j20} = \frac{V\angle 0}{20.6\angle 76^\circ} = \frac{V\angle -76^\circ}{20.6}$$

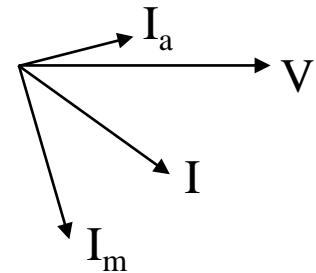
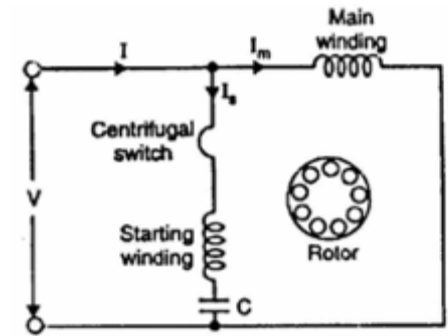
The current flowing through the auxiliary winding after connecting resistor  $R$  and a capacitor  $C$  in series should make an angle  $90^\circ$  with  $I_m$  or make an angle  $90^\circ - 76^\circ = 14^\circ$  with the applied voltage  $V$ . Since the new current of auxiliary winding should be leading the voltage  $V$  by an angle of  $14^\circ$ , the capacitive reactance of the auxiliary circuit is greater than the inductive reactance. Thus

$$\cos 14^\circ = \frac{5 + R}{Z}; \quad 5 + R = Z \cos 14^\circ; \quad R = 20.6 \cos 14^\circ - 5 = 19.99 - 5 = 14.99\ \Omega$$

$$\text{Again } \sin 14^\circ = \frac{X_C - X_L}{Z}; \quad X_C - X_L = Z \sin 14^\circ; \quad X_C = 20.6 \sin 14^\circ + X_L$$

$$X_C = 20.6 \sin 14^\circ + 20 = 24.98\ \Omega$$

$$(1/\omega C) = 24.98 \quad C = 127\ \mu F$$



# Example Problems

**Example 36.3.** A 250 W, 230 V, 50 Hz capacitor-start motor has the following constants for the main and auxiliary winding: Main winding,  $Z_m = (4.5 + j3.7)$  ohm, Auxiliary winding,  $Z_a = (9.5 + j3.5)$  ohm. Determine the value of the starting capacitor that will place the main and auxiliary winding currents in quadrature at starting.

**Solution:** Let  $X_C$  be the reactance of the capacitor connected in the auxiliary winding.

Then  $Z_a = 9.5 + j(3.5 - X_C) = (9.5 - jX)$  ohm

where,  $X$  is the net reactance

Now,  $Z_m = (4.5 + j3.7) = 5.82 \angle 39.4^\circ$  ohm

Obviously,  $I_m$  lags behind  $V$  by  $39.4^\circ$ .

Since time phase angle between  $I_m$  and  $I_a$  has to be  $90^\circ$ ,

$I_a$  must lead  $V$  by  $(39.4^\circ - 90^\circ) = -50.6^\circ$ .

For auxiliary winding,

$\tan \phi_a = (3.5 - X_C) / R$

or  $\tan(-50.6^\circ) = (3.5 - X_C) / 9.5 = -1.217$

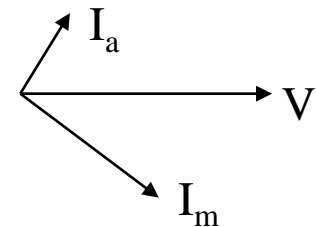
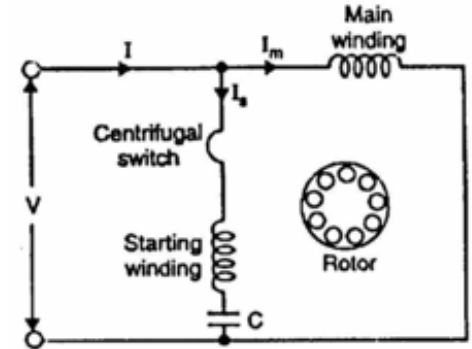
Or  $(3.5 - X_C) = -9.5 \times 1.217 = -11.56$  ohm

$\therefore X_C = 11.56 + 3.5 = 15.06$  ohm

Or  $1 / (2 \times \pi \times 50 \times C) = 15.06$

or  $C = 1 / (2 \times \pi \times 50 \times 15.06) = 211 \times 10^{-6} \text{ F}$

$\therefore C = 211 \mu\text{F}$ .



# Example Problems

**Example 36.1.** Find the mechanical power output at a slip of 0.05 of the 185-W, 4-pole, 110-V, 60-Hz single-phase induction motor, whose constants are given below:

Resistance of the stator main winding

$$R_1 = 1.86 \text{ ohm}$$

Reactance of the stator main winding

$$X_1 = 2.56 \text{ ohm}$$

Magnetizing reactance of the stator main winding

$$X_m = 53.5 \text{ ohm}$$

Rotor resistance at standstill

$$R_2 = 3.56 \text{ ohm}$$

Rotor reactance at standstill

$$X_2 = 2.56 \text{ ohm}$$

**Solution.** Here,  $X_m = 53.5 \Omega$ , hence  $x_m = 53.5/2 = 26.7 \Omega$

Similarly,  $r_2 = R_2/2 = 3.56/2 = 1.78 \Omega$  and  $x_2 = X_2/2 = 2.56/2 = 1.28 \Omega$

$$\therefore Z_f = \frac{j x_m \left( \frac{r_2}{s} + j x_2 \right)}{\frac{r_2}{s} + j (x_2 + x_m)} = x_m \frac{\frac{r_2}{s} \cdot x_m + j \left[ (r_2/s)^2 + x_2 x_0 \right]}{(r_2/s)^2 + x_0^2} \quad \text{where } x_0 = (x_m + x_2)$$

$$\therefore Z_f = 26.7 \frac{(1.78/0.05) \times 26.7 + j [(1.78/0.05)^2 + 1.28 \times 27.98]}{(1.78/0.05)^2 + (27.98)^2}$$

$$= 12.4 + j 17.15 = 21.15 \angle 54.2^\circ$$

$$\text{Similarly, } Z_b = \frac{j x_m \left( \frac{r_2}{2-s} + j x_2 \right)}{\frac{r_2}{2-s} + j (x_2 + x_m)} = x_m \frac{\left( \frac{r_2}{2-s} \right) x_m + j \left[ \left( \frac{r_2}{2-s} \right)^2 + x_0 x_2 \right]}{\left( \frac{r_2}{2-s} \right)^2 + x_0^2}$$

$$= 26.7 \frac{(1.78/1.95) \times 26.7 + j [(1.78/1.95)^2 + 1.28 \times 27.98]}{(1.78/1.95)^2 + (27.98)^2}$$

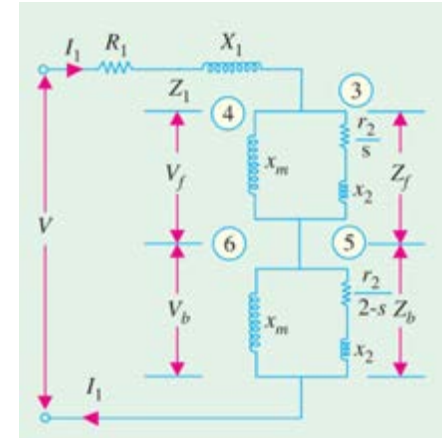
$$= 0.84 + j 1.26 = 1.51 \angle 56.3^\circ$$

$$Z_1 = R_1 + j X_1 = 1.86 + j 2.56 = 3.16 \angle 54^\circ$$

Total circuit impedance is

$$Z_{01} = Z_1 + Z_f + Z_b = (1.86 + j 2.56) + (12.4 + j 17.15) + (0.84 + j 1.26)$$

$$= 15.1 + j 20.97 = 25.85 \angle 54.3^\circ$$



$$\text{Motor current } I_1 = 110/25.85 = 4.27 \text{ A}$$

$$V_f = I_1 Z_f = 4.27 \times 21.15 = 90.4 \text{ V}; V_b = I_1 Z_b = 4.27 \times 1.51 = 6.44 \text{ V}$$

$$Z_3 = \sqrt{\left( \frac{r_2}{s} \right)^2 + x_2^2} = 35.7 \Omega, Z_5 = \sqrt{\left( \frac{r_2}{2-s} \right)^2 + x_2^2} = 1.57 \Omega$$

$$I_3 = V_f / Z_3 = 90.4/35.7 = 2.53 \text{ A}, I_5 = V_b / Z_5 = 6.44/1.57 = 4.1 \text{ A}$$

$$T_f = I_3^2 R_2 / s = 228 \text{ synch. watts}, T_5 = I_5^2 r_2 / (2-s) = 15.3 \text{ synch. watts.}$$

$$T = T_f - T_b = 228 - 15.3 = 212.7 \text{ synch. watts}$$

$$\text{Output} = \text{synch. watt} \times (1-s) = 212.7 \times 0.95 = 202 \text{ W}$$

Since friction and windage losses are not given, this also represents the net output.



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# Example Problems

**Example 36.2.** Find the mechanical power output of the 185-W, 4-pole, 110-V, 60-Hz single-phase induction motor, whose constants are given below at a slip of 0.05.

$R_1 = 1.86 \text{ ohm}$ ,  $X_1 = 2.56 \text{ ohm}$ ,  $X_m = 53.5 \text{ ohm}$ ,  $R_2 = 3.56 \text{ ohm}$ ,  $X_2 = 2.56 \text{ ohm}$

Core loss = 3.5 W, Friction and windage loss = 13.5 W.

**Solution.** It would be seen that major part of this problem has already been solved in Example 36.1. Let us, now, assume that  $V_f = 82.5\%$  of 110 V = 90.7 V. Then the core-loss current  $I_c = 35/90.7 = 0.386 \text{ A}$ ;  $r_c = 90.7/0.386 = 235 \text{ } \Omega$ .

## Motor I

conductance of core-loss branch =  $1/r_c = 1/235 = 0.00426 \text{ S}$

susceptance of magnetising branch =  $-j/x_m = -j/26.7 = -j0.0374 \text{ S}$

$$\text{admittance of branch 3} = \frac{(r_2/s) - jx_2}{(r_2/s)^2 + x_2^2} = 0.028 - j0.00101 \text{ S}$$

$$\begin{aligned} \text{admittance of 'motor' I is } Y_f &= 0.00426 - j0.0374 + 0.028 - j0.00101 \\ &= 0.03226 - j0.03841 \text{ S} \end{aligned}$$

$$\text{impedance } Z_f = 1/Y_f = 12.96 + j15.2 \text{ or } 19.9 \text{ } \Omega$$

## Motor II

$$\text{admittance of branch 5} = \frac{\frac{r_2}{2-s} - jx_2}{\left(\frac{r_2}{2-s}\right)^2 + x_2^2} = \frac{0.91 - j1.28}{2.469} = 0.369 - j0.517$$

$$\begin{aligned} \text{admittance of 'motor' II, } Y_b &= 0.00426 - j0.0374 + 0.369 - j0.517 \\ &= 0.3733 - j0.555 \text{ S} \end{aligned}$$

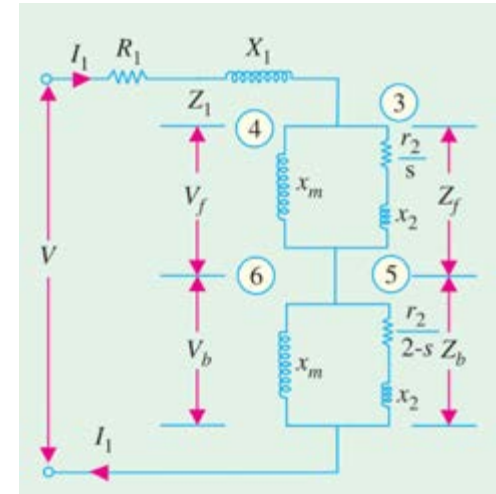
Impedance of 'motor' II,  $Z_b = 1/Y_b = 0.836 + j1.242 \text{ or } 1.5 \text{ } \Omega$

Impedance of entire motor (Fig. 36.16)  $Z_{01} = Z_1 + Z_f + Z_b = 15.66 + j19 \text{ or } 24.7 \text{ } \Omega$

$$I_1 = V/Z_{01} = 110/24.7 = 4.46 \text{ A}$$

$$V_f = I_1 Z_f = 4.46 \times 19.9 = 88.8 \text{ V}$$

$$V_b = 4.46 \times 1.5 = 6.69 \text{ V}$$



$$I_3 = 88.8/35.62 = 2.5 \text{ A}$$

$$I_5 = 6.69/1.57 = 4.25 \text{ A}$$

$$T_f = I_3^2 (r_2/s) = 222 \text{ synch. watt}$$

$$T_b = I_5^2 \left( \frac{r_2}{2-s} \right) = 16.5 \text{ synch. watt}$$

$$T = T_f - T_b = 205.5 \text{ synch. watt}$$

$$\text{Watts converted} = \text{synch. watt} (1 - s)$$

$$= 205.5 \times 0.95 = 195 \text{ W}$$

$$\text{Net output} = 195 - 13.5 = \mathbf{181.5 \text{ W.}}$$



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# Thank You

