# Design of Oversampling $\Delta\Sigma$ DA Converters via $H^\infty$ Optimization

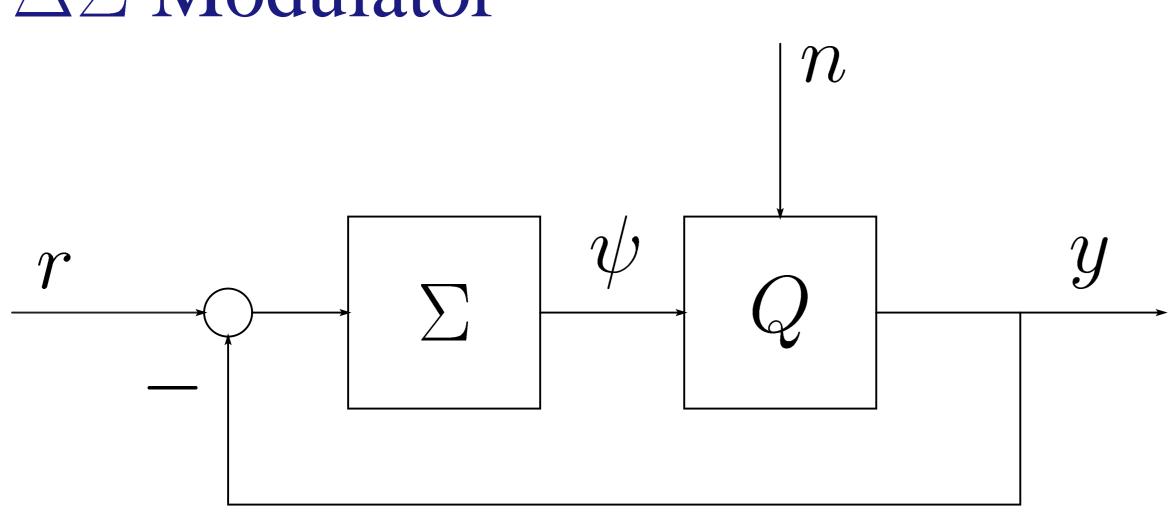
## Masaaki Nagahara\*, Toshihiro Wada, and Yutaka Yamamoto

Department of Applied Analysis and Complex Dynamical Systems, Graduate School of Informatics, Kyoto University, Kyoto 606-8501, Japan

## Abstract

In this paper, we propose a new method for designing oversampling  $\Delta\Sigma$  DA converters via  $H^\infty$  optimization. The design consists of two steps. One is that for  $\Delta\Sigma$  modulators. In  $\Delta\Sigma$  modulators, the accumulator 1/(z-1) is conventionally used in a feedback loop to shape quantization noise. In contrast, we give all stabilizing loop filters for the modulator, and propose an  $H^\infty$  design to shape the frequency response of the noise transfer function (NTF). The other is a design for interpolation filters in oversampling DA converters. While conventional designs are executed in the discrete-time domain, we take account of the characteristic of the original analog signal by using sampled-data  $H^\infty$  optimization. A design example is presented to show that our design is superior to conventional ones.

## $\Delta\Sigma$ Modulator



r: Input signal to be modulated

y: Modulated signal

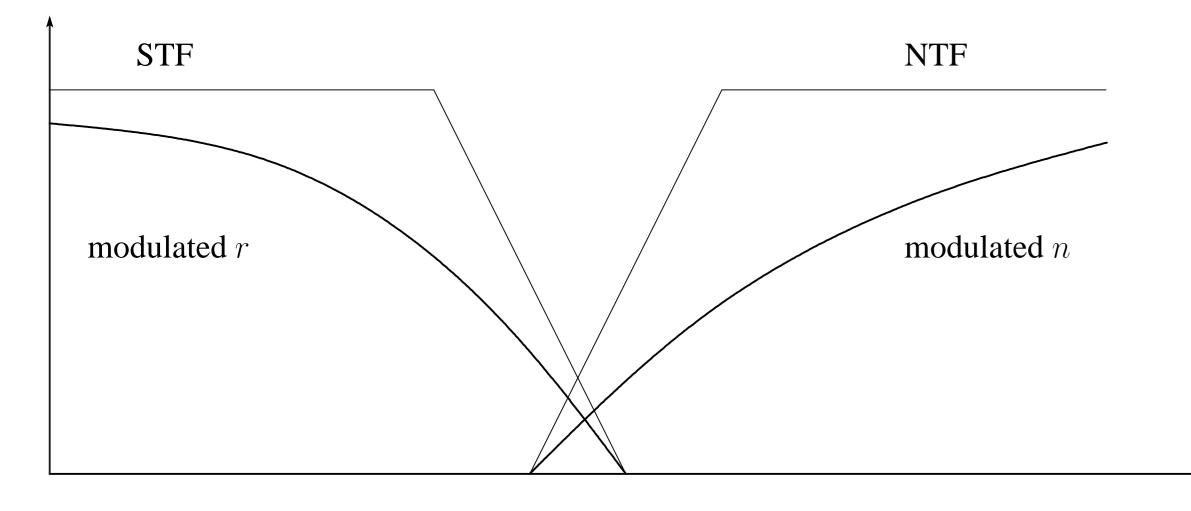
n: Quantization noise

 $R(z): r \rightarrow y$ : Signal Transfer Function (STF)

 $H(z): n \rightarrow y$ : Noise Transfer Function (NTF)

## Quantization Noise Shaping

If r contains few high frequency components, we can separate the noise n from the output signal y by a allpass or lowpass STF and a highpass NTF.



## Example

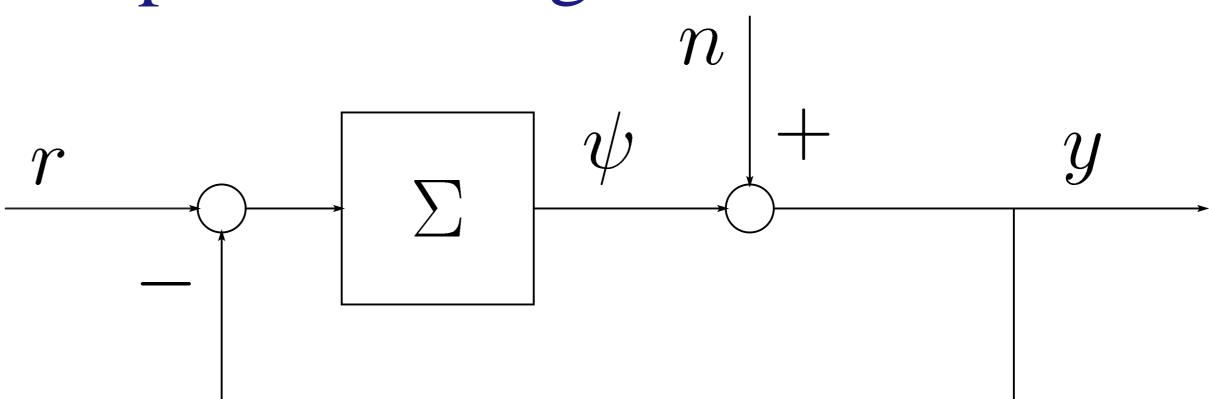
Conventional first order  $\Delta\Sigma$  Modulator.

$$\begin{split} \Sigma(z) &= \frac{1}{z-1} \\ y &= \frac{\Sigma(z)}{1+\Sigma(z)} r + \frac{1}{1+\Sigma(z)} n = z^{-1} r + (1-z^{-1}) n \end{split}$$

STF:  $R(z) = z^{-1}$ : Allpass

NTF:  $H(z) = 1 - z^{-1}$ : Highpass

## Loop Filter Design



#### Lemma 1

The above system is well-posed and internally stable if and only if

$$\Sigma(z) \in \left\{ \frac{R(z)}{1 - R(z)} : R(z) \text{ is stable and strictly causal} \right\}$$

## Design Problem

Given a stable transfer function  $H_{des}(z)$  (desired NTF) and a stable weighting function W(z), find H(z) with  $H(\infty) = 1$  (required for well-posedness) which minimizes

$$J(H) = ||(H - H_{\text{des}})W||_{\infty}.$$

#### Design by LMI

Assume H(z) is FIR, that is,

$$H(z) = \sum_{k=1}^{N} a_k z^{-k}.$$

Then, by using the *bounded real lemma*, the optimization is reducible to a linear matrix inequality (LMI) with respect to a matrix variable and the coefficients  $a_1, \ldots, a_N$ .

## Zeros of NTF

Define  $n_H(z):=z^N-\Sigma_{k=1}^N\,a_kz^{N-k}$  (the numerator of H(z)). Then, H(z) has at least M zeros at  $z=z_0$  if and only if

$$\frac{d^k n_H(z)}{dz^k}\bigg|_{z=z_0} = 0, k = 0, 1, \dots, M-1$$
 (Linear constraints).

#### **Stability Constraints**

We can add a stability condition considering nonlinear behaviors of  $\Delta\Sigma$  modulators by

$$||H||_{\infty} < C$$

where C = 1.5 for binary quantizer (*Lee Criterion*), or

$$C = \frac{1}{(2\nu + 1)} (M + 2 - ||r||_{\infty})$$

for M-step quantizer with  $\nu$ -th order H(z). This is also reducible to an LMI.