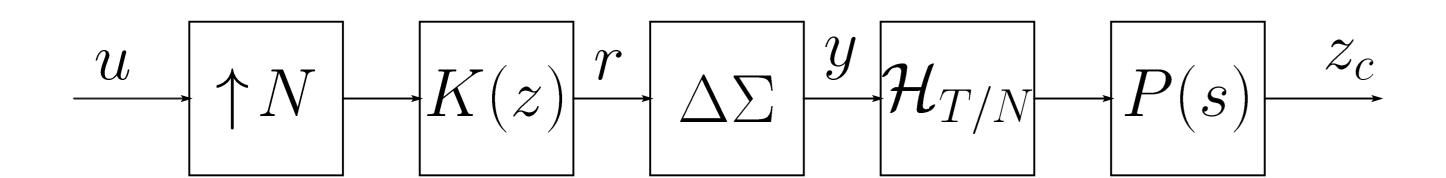
# Oversampling $\Delta\Sigma$ DA Converters



u: digital signal (1/T [Hz] and b [bits])

 $\uparrow N$ : upsampler

K(z): interpolation filter to be designed

 $\Delta\Sigma$ :  $\Delta\Sigma$  modulator

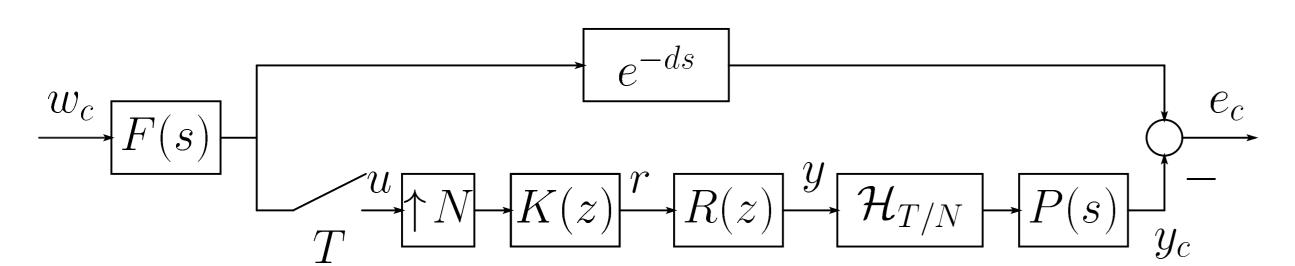
 $\mathcal{H}_{T/N}$ : hold device with sampling time T/N [sec]

P(s): analog lowpass filter

 $z_c$ : reconstructed analog signal

Our objective here is to design the interpolation filter K(z) to interpolate samples taking account of the analog performance.

### Interpolator Design



 $w_c$ : analog signal in  $L^2$ 

F(s): frequency characteristic of analog input signal

#### Design Problem

Given a stable, strictly causal F(s), stable, causal P(s), stable, strictly causal R(z), upsampling factor N, delay d, sampling time T, find K(z) which minimizes

$$\|\mathcal{E}\|_{\infty} := \sup_{w_c \in L^2} \frac{\|e_c\|_{L^2}}{\|w_c\|_{L^2}}.$$

The optimal filter K(z) can be obtained via sampled-data  $H^{\infty}$  optimization.

# Design Example

#### Design Parameters

sampling time: T = 1

upsampling ratio: N = 8

reconstruction delay: d = 8

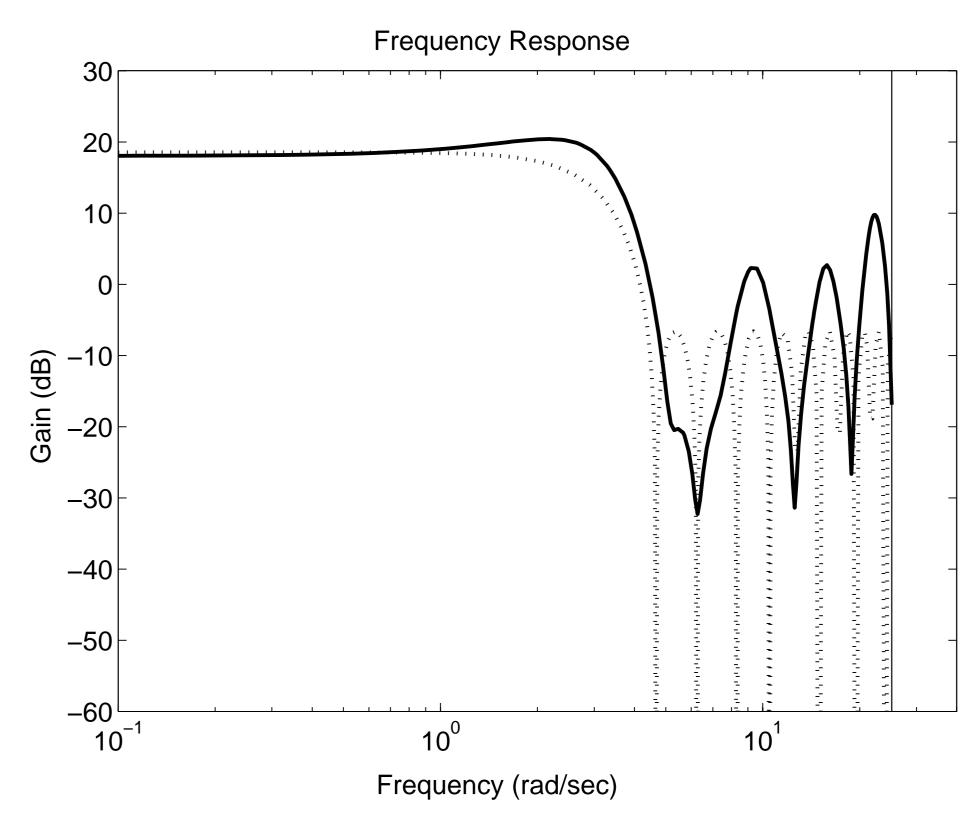
analog filters:

$$P(s) = \frac{1}{\{(T/\pi)s + 1\}^2}$$

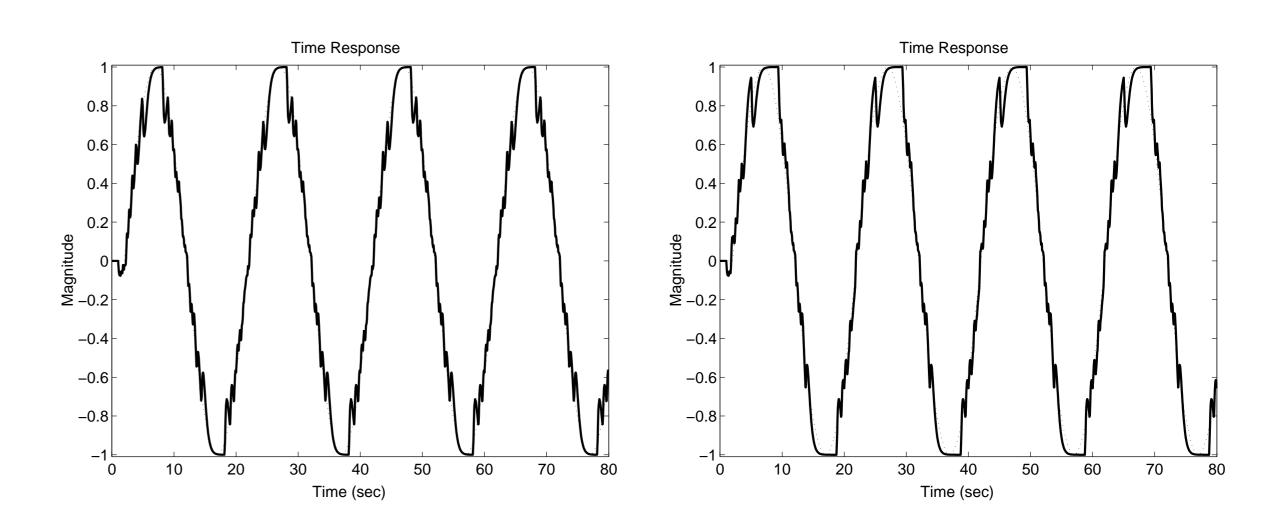
$$F(s) = \frac{1}{\{(T_F/\pi)s + 1\}\{(0.1T_F/\pi)s + 1\}}, \quad T_F = 22.05$$

quantizer:

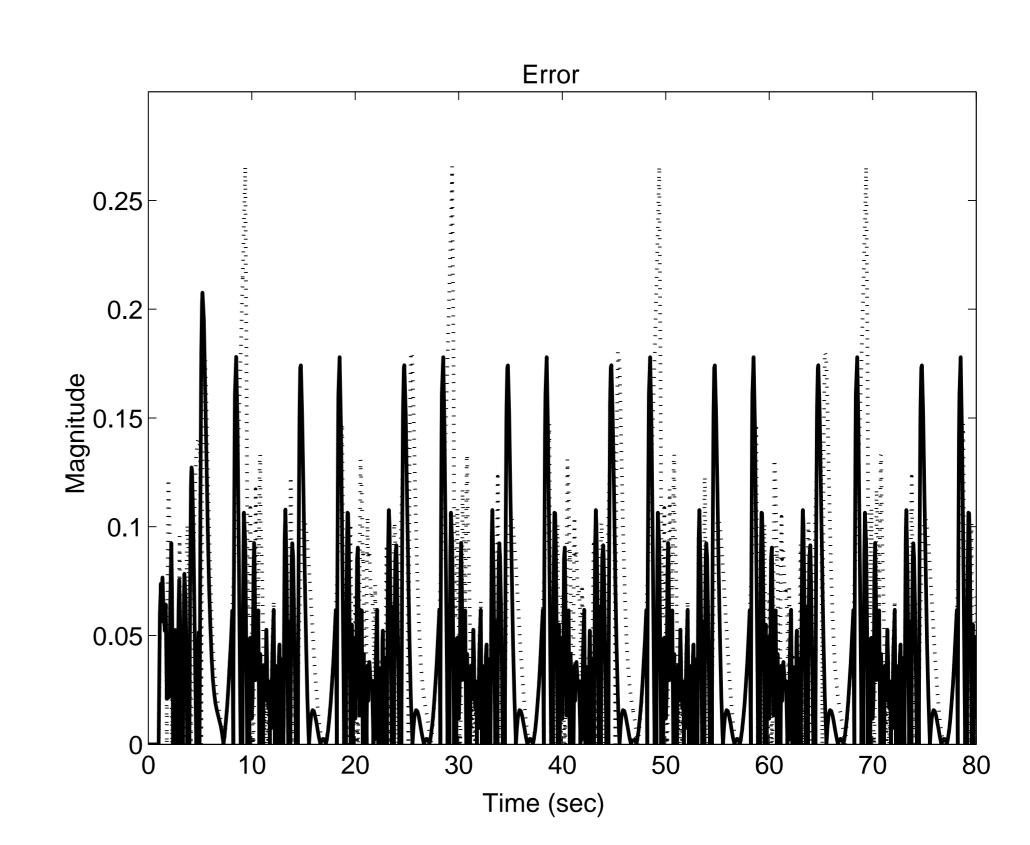
$$Q(\psi) = \operatorname{sgn}(\psi) = \begin{cases} 1, & \psi \ge 0, \\ -1 & \psi < 0. \end{cases}$$



Interpolation filter K(z) (solid: sampled-data  $H^{\infty}$  optimal, dotted: equiripple design)



Time response against a sinusoidal wave  $u[k] = \sin(0.1\pi k)$  (left: proposed, right: conventional)



Absolute error (solid: proposed, dotted: conventional)

Conparison of error

Comparison of circl			
		Proposed design	Conventional design
	$  e_c  _{\infty}$	$2.08 \times 10^{-1}$	$2.67 \times 10^{-1}$
	$  e_c  _2$	$5.68 \times 10^{-1}$	$7.21 \times 10^{-1}$
	$\overline{\mathrm{RMS}(e_c)}$	$6.34 \times 10^{-2}$	$8.06 \times 10^{-2}$

In the table, RMS is the root-mean-square values defined as follows: For fixed  $T_f>0$ ,  $\mathrm{RMS}(e_c):=\left\{\tfrac{1}{T_f}\int_0^{T_f}|e_c(t)|^2dt\right\}^{\frac{1}{2}}.$ 

## Conclusion

We have proposed a new design method for  $\Delta\Sigma$  modulators and oversampling  $\Delta\Sigma$  DA converters via  $H^\infty$  optimization. We have presented design examples and shown the advantages of the present method.