

# Plotting Bode plot with complex roots and zeros with hand

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## Introduction

Given second order transfer function can be generalized following form;

$$H(s) = \frac{\omega_n^2}{s^2 + 2\xi\omega_n s + \omega_n^2}$$

where  $\xi$  is **damping ratio** and  $\omega_n$  is natural frequency.

When  $\xi > 1$  system response said over-damped, and it can easily separate two first order system cascaded.

Then transfer function rewritten in following form;

$$H(s) = \frac{K}{(s + \alpha)(s + \beta)}$$

Then body plot of that transfer function is straight-forward.

-Regulate the transfer function

$$H(s) = \frac{\frac{K}{\alpha\beta}}{(1 + \frac{s}{\alpha})(1 + \frac{s}{\beta})}$$

-Find the dc-gain (gain at  $\omega = 0$ )

$$G_{dc} = \frac{K}{\alpha\beta}$$

Since there is no zero (no  $s$  in the numerator) in the following function also no pole at zero (no zero in the denominator), plot came with straight line before first pole hit.

-Plot the asymptotes and bode

When  $\xi = 1$  we said system response critically-damped. Denominator can rewrite as  $(s + \omega_n)^2$ . That means now we have double pole in a particular location. Bode plot algorithm is still same just make sure know we have  $40 \frac{dB}{decade}$

## Complex poles and zeros

### Complex poles

All points in second order term with  $\xi < 1$  is we plot asymptotes like  $\xi = 1$  case. With closing  $\xi$  to one (from zero), transfer function start behave;

$$H(s) = \frac{\omega_n}{(s + \omega_n)^2}$$

The can plot above function bode plot easily. So after that we just should edit a little by looking  $\xi$ , damping ratio.

$$H(s) = \frac{1}{s^2 + 2\xi s + 1}$$

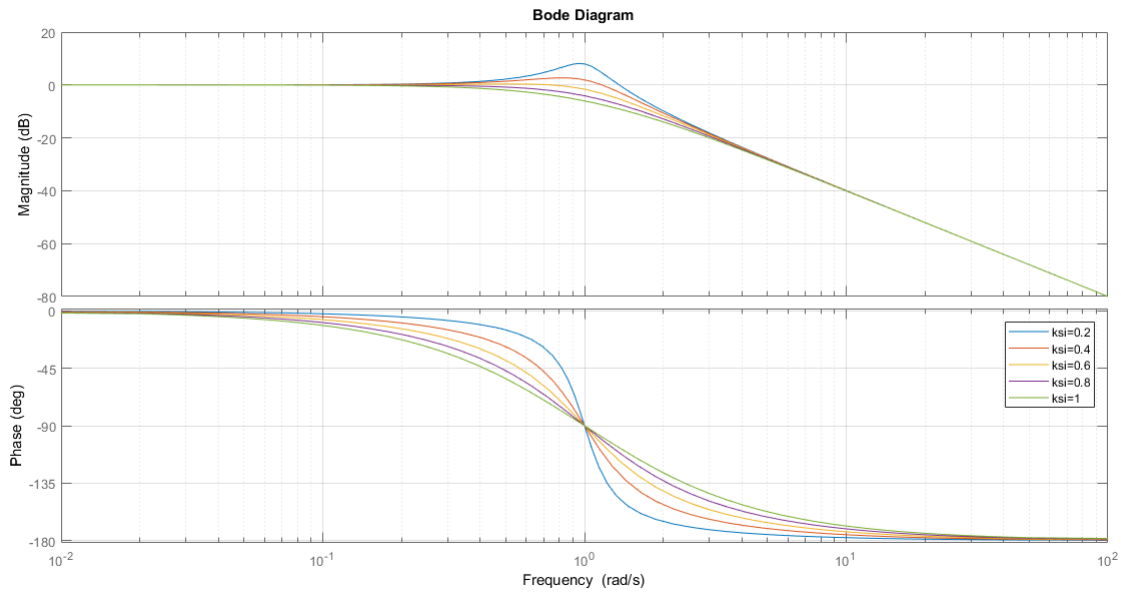


Figure 1: Bode plot of  $H(s)$  with different  $\xi$

$\xi = 1$  , we draw asymptotes and we did -3 dB approximation (for poles). However with  $\xi < \frac{1}{\sqrt{2}}$  actual gain underline the other side of the asymptotes (positive over-shoot characteristic for poles). Cut-off frequency also changes slightly with changing  $\xi < \frac{1}{\sqrt{2}}$

$$\omega_{n'} = \omega_n \sqrt{1 - 2\xi^2}$$

## Complex zeros

$$H(s) = s^2 + 2\xi s + 1$$

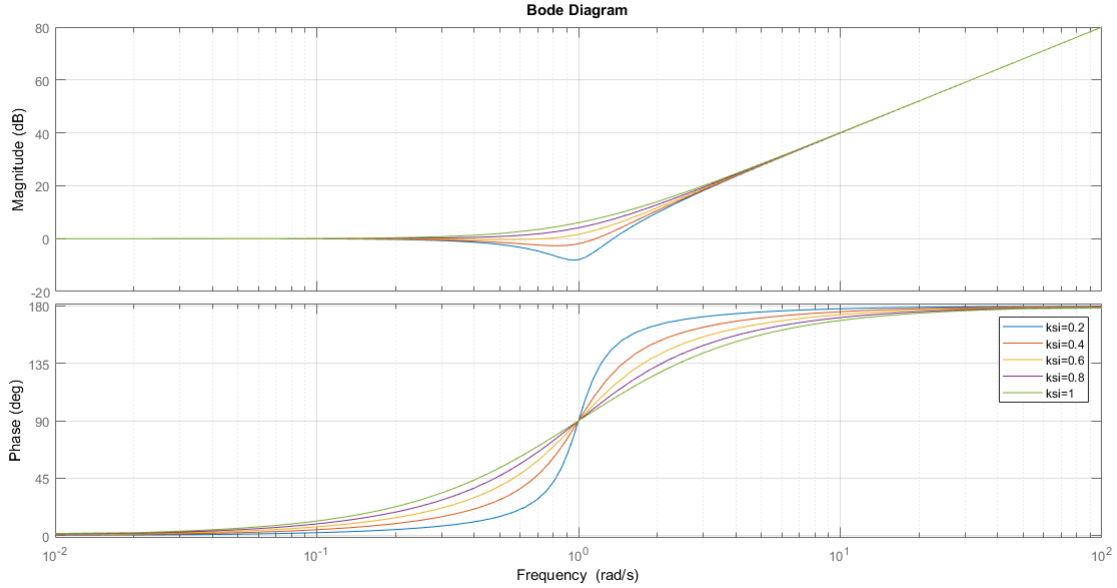


Figure 2: Bode plot of  $H(s)$  with different  $\xi$

Similarly,  $\xi = 1$  , we draw asymptotes and we did +3 dB approximation (for poles). Now the actual gain again is at other side of the asymptote (negative over-shoot characteristic for zeros)

## Example

$$H(s) = 128 \frac{s^2 + 4}{(s + 32)(s^2 + 8s + 64)}$$

### Behaviour at DC

System is type-0, no pole at  $\omega = 0$  also no zero at  $\omega = 0$  therefore straight line comes from the left side of the graph

### DC gain

Normalizing the transfer function;

$$H(s) = \frac{1}{4} \frac{1 + \frac{s^2}{4}}{(1 + \frac{s}{32})(1 + \frac{s}{8} + \frac{s^2}{64})}$$

$$G_{dc} = 20 \log\left(\frac{1}{4}\right) = -12 \text{ dB}$$

### Second order Behaviour

#### Second order pole

$$s^2 + 2\xi\omega_n + \omega_n^2 = s^2 + 8s + 64$$

$$\omega_n = 8$$

$$\xi = 0.5$$

Double pole at  $\omega = 8$

#### Second order zero

$$s^2 + 2\xi\omega_n + \omega_n^2 = s^2 + 8$$

$$\omega_n = 2$$

$$\xi = 0$$

Double zero at  $\omega = 2$

Matlab plot;

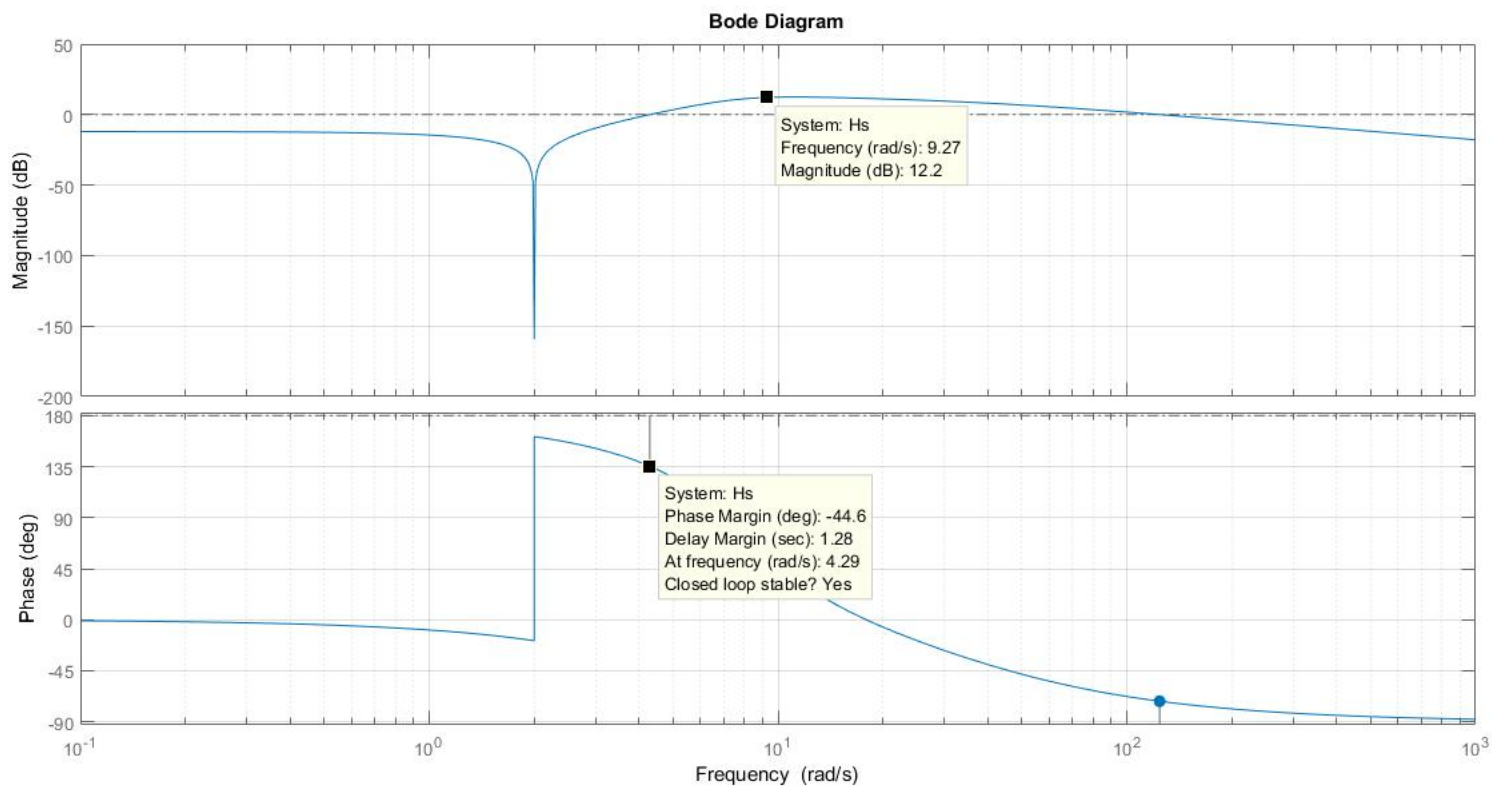


Figure 3: Bode plot of  $H(s)$  with different  $\xi$

By hand;