AUTOMATIC CONTROL THEORY

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Chapter 3 Linear system in frequency domain

- 3.1. SISO system
- 3.2. Modeling a nonlinear system by linearization
- 3.3. Typical linear dynamic systems
- 3.4. System stability and stability criteria
- 3.5. Transient and steady- state responses
- 3.6. Overshoot, settling time, steady-state error.
- 3.7. Open-loop controller design
- 3.8. PID controller
- 3.9. PID controller design methods

•
$$T: \boldsymbol{u}(t) \to \boldsymbol{y}(t)$$

•
$$u(t) = \begin{bmatrix} u_1(t) \\ u_2(t) \\ \vdots \\ u_r(t) \end{bmatrix}; y(t) = \begin{bmatrix} y_1(t) \\ y_2(t) \\ \vdots \\ y_s(t) \end{bmatrix}$$

• System is linear if the superposition holds:

$$T\left\{\sum_{i}^{n} a_{i} \boldsymbol{u}_{i}(t)\right\} = \sum_{i}^{n} a_{i} T\{\boldsymbol{u}_{i}(t)\}$$

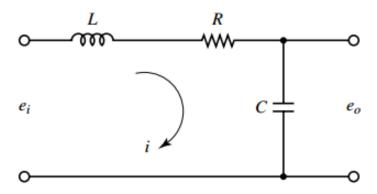
- SISO system: r = s = 1
- MIMO system: r > 1; s > 1
- MISO system : r > 1; s = 1
- SISO system can be represented as:
 - □Differential equations
 - \square Transfer function G(s)
 - ☐State-space model

• Differential equation

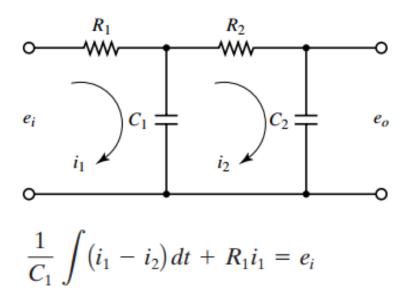
$$a_0y + a_1\frac{dy}{dt} + \dots + a_n\frac{d^ny}{dt^n} = b_0u + b_1\frac{du}{dt} + \dots + b_m\frac{d^mu}{dt^m}$$

- a_i, b_i : constant
- n > m integer number
- *u* input
- *y* output

- Electrical system: Kirchoff Law 1 & 2
 - □Sum of currents is zero for a node.
 - □Sum of voltages is zero for a loop.
- Resistor: u = Ri
- Capacitor: $i_C(t) = C \frac{du_C(t)}{dt}$;
- Inductor: $u_L(t) = L \frac{di_L(t)}{dt}$
- *u* voltage, *i* current.



$$L\frac{di}{dt} + Ri + \frac{1}{C} \int i \, dt = e_i$$
$$\frac{1}{C} \int i \, dt = e_o$$

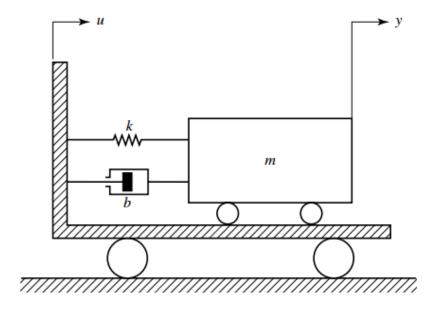


$$\frac{1}{C_1} \int (i_2 - i_1) dt + R_2 i_2 + \frac{1}{C_2} \int i_2 dt = 0$$

$$\frac{1}{C_2} \int i_2 dt = e_o$$

• Mechanical system:

- Newton's second law:
- $\sum F_i = ma$
- F_i force
- *m* mass
- a acceleration



$$m\frac{d^2y}{dt^2} = -b\left(\frac{dy}{dt} - \frac{du}{dt}\right) - k(y - u)$$

$$m\frac{d^2y}{dt^2} + b\frac{dy}{dt} + ky = b\frac{du}{dt} + ku$$

3.2. Modeling a nonlinear system by linearization

$$\dot{x} = f(x, u)$$
$$y = g(x)$$

Working point:

$$u = u_0$$

$$f(x_0, u_0) = 0$$

$$\dot{z} = Az + Bv$$

$$A = \frac{df}{dx} \left| x = x_0, u = u_0 \right| B = \frac{df}{du} \left| x = x_0, u = u_0 \right|$$

$$v = u - u_0$$

$$z = x - x_0$$

3.2. Modeling a nonlinear system by linearization

$$\dot{x} = -x^3 - x + u$$
$$y = x$$

Working point:

$$u = 0$$

 $f(x_0, u_0) = 0$
 $x_0 = 0$
 $A = -1; B = 1$

$$\dot{z} = -z + v$$

$$v = u - u_0$$

$$z = x - x_0$$

- Unit Impulse:
- Impulse Response:

$$g(t) = L^{-1}\{G(s)\}$$

- Unit step function:
- Unit Step Response:

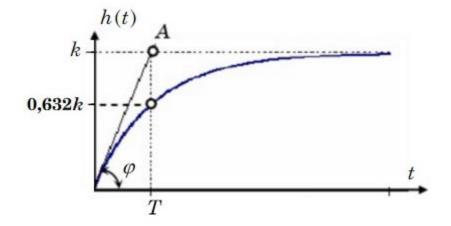
$$h(t) = L^{-1} \left\{ \frac{G(s)}{s} \right\}$$

• First-order system:

$$G(s) = \frac{k}{1 + Ts}$$

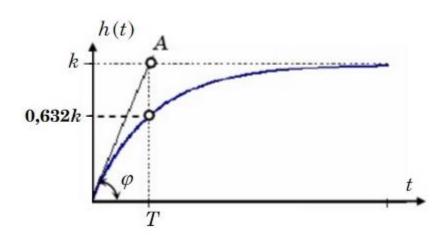
• Step response:

$$h(t) = L^{-1} \left\{ \frac{G(s)}{s} \right\} = k(1 - e^{-\frac{t}{T}})$$



$$\frac{dh(0)}{dt} = \lim_{s \to \infty} s[sH(s) - h(0)] = \frac{k}{T} = \tan(\varphi)$$

- Determine k và T:
 - \square Apply 1(*t*) to the system.
 - $\square k$ steady state output
 - ☐ T t-coordinate of A

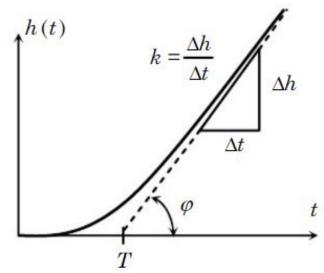


• First-order system with integrator

$$G(s) = \frac{k}{s(1+Ts)}$$

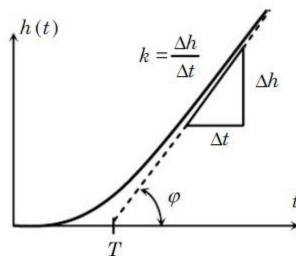
$$H(s) = \frac{G(s)}{s} = \frac{k}{s^2(1+Ts)}$$

$$h(t) = k[t - T(1 - e^{-\frac{t}{T}})]$$



$$h(t) = k[t - T(1 - e^{-\frac{t}{T}})]$$

- Steady-state, $h(t) = k(t T) = h_{tc}(t)$
- $k = tan(\varphi)$
- Asymptote at ∞ .
- T: t-coordinate of the intersection between the asymptote and the time axis.
- $k = tan(\varphi)$.

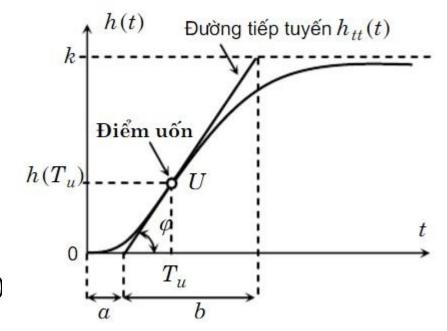


Second-order system

$$G(s) = \frac{k}{(1+sT_1)(1+sT_2)}, T_2 < T_1, x = \frac{T_2}{T_1}$$

•
$$H(s) = \frac{k}{s(1+sT_1)(1+sT_2)}$$

•
$$h(t) = k \left[1 - \frac{T_1 e^{-\frac{t}{T_1}} - T_2 e^{-\frac{t}{T_2}}}{T_1 - T_2} \right]$$



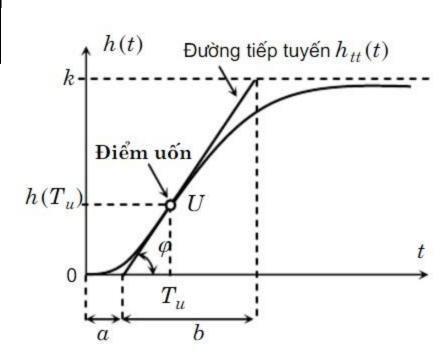
- Điểm uốn (inflection point)
- Đường tiếp tuyến (tangent line)

•
$$\frac{dh(t)}{dt} = k \frac{e^{-\frac{t}{T_1}} - e^{-\frac{t}{T_2}}}{T_1 - T_2}$$

•
$$\frac{d^2h(t)}{dt^2} = k \left[\frac{e^{-\frac{t}{T_1}}}{T_1(T_2 - T_1)} + \frac{e^{-\frac{t}{T_2}}}{T_2(T_1 - T_2)} \right]$$
• $h(t)$

•
$$T_u = \frac{T_1 T_2}{T_2 - T_1} ln \frac{T_2}{T_1}$$

•
$$v^* = \frac{dh(T_u)}{dt} = \frac{k}{T_1} \left(\frac{T_2}{T_1}\right)^{\frac{T_2}{T_1 - T_2}}$$



• Let
$$x = \frac{T_2}{T_1}$$
, $v^* = \frac{k}{T_1} x^{\frac{x}{1-x}}$

• But
$$v^* = \frac{k}{b}$$
,

• So
$$\frac{b}{T_1} = \chi \frac{x}{x-1}$$

•
$$a = T_u - \frac{h(T_u)}{v^*}$$

$$\frac{a}{b} = x^{\frac{x}{1-x}} \frac{x \ln x + x^2 - 1}{x - 1} - 1 = g(x)$$

$$x = g^{-1} \left(\frac{a}{b}\right)$$

$$0 < \frac{a}{b} < 0.103$$

Second-order system with oscillation

$$G(s) = \frac{k}{1 + 2DTs + T^{2}s^{2}}, \quad 0 < D < 1$$

$$H(s) = \frac{G(s)}{s}$$

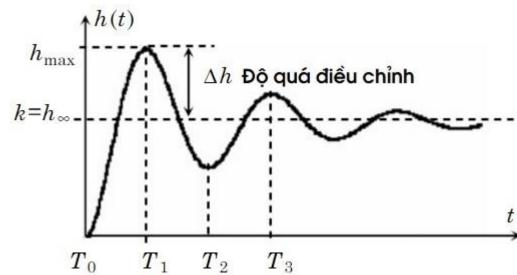
$$h(t) = k \left(1 - \frac{e^{-\frac{D}{T}t}}{\sqrt{1 - D^{2}}} \sin\left(\frac{\sqrt{1 - D^{2}}}{T}t + arcos(D)\right) \right)$$

$$\frac{dh(t)}{dt} = k \frac{e^{-\frac{D}{T}t}}{T\sqrt{1-D^2}} \sin\left(\frac{\sqrt{1-D^2}}{T}t\right); T_i = \frac{i\pi T}{\sqrt{1-D^2}}$$

$$h_{max} = h(T_1) = k[1 + e^{-\frac{\pi D}{\sqrt{1-D^2}}}]$$

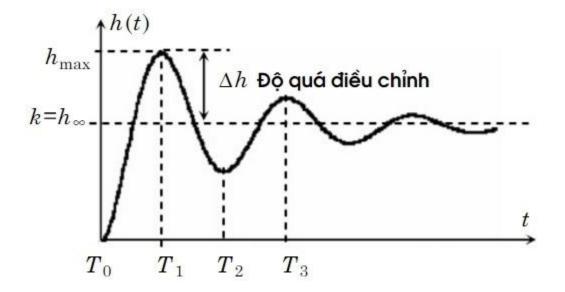
•
$$\Delta h = h_{max} - k = ke^{-\frac{\pi D}{\sqrt{1 - D^2}}}$$

$$D = \frac{1}{\sqrt{1 + \frac{\pi^2}{\ln^2 \frac{\Delta h}{k}}}}$$



$$T_{1} = \frac{\pi T}{\sqrt{1 - D^{2}}}$$

$$T = \frac{T_{1}\sqrt{1 - D^{2}}}{\pi}$$



Frequency characteristics

$$G(s) = \frac{Y(s)}{X(s)} = \frac{b_0 + b_1 s + \dots + b_m s^m}{a_0 + a_1 s + \dots + s^n} \qquad (m \le n)$$

$$G(j\omega) = G(s) \Big|_{s = j\omega}$$

$$G(s) = \frac{1}{1+s}$$

$$G(j\omega) = \frac{1}{1+j\omega} = \frac{1}{1+\omega^2} - \frac{j\omega}{1+\omega^2}$$

$$Re\{G(j\omega)\} = \frac{1}{1 + \omega^2} \to Re \ge 0$$

$$Im\{G(j\omega)\} = \frac{1}{1 + \omega^2} \to Im \le 0$$

$$(Re - \frac{1}{2})^2 + Im^2 = \frac{1}{4}$$

Bode plot

$$L(\omega) = 20 \lg |G(j\omega)| - \text{biên tần}$$

$$\varphi(\omega) = \arctan\left(\frac{Im\{G_f(j\omega)\}}{Re\{G_f(j\omega)\}}\right)$$

- Gain: G(s) = k, k > 0
- First-order system: $G(s) = \frac{k}{1+Ts}$

• PD:
$$G(s) = 1 + Ts$$

• I:
$$G(s) = \frac{1}{T_I s}$$

- D: $G(s) = T_D s$
- Second-order system:

$$G(s) = \frac{k}{1 + 2DTs + T^2s^2}, \qquad 0 < D < 1$$

• Conversion of diagram:

□ Series

□Parallel

□ Feedback

