

Control Systems

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Abstract—This manual is an introduction to control systems in feedback circuits. Links to sample Python codes are available in the text.

Download python codes using

svn co <https://github.com/gadepall/school/trunk/control/feedback/codes>

1	FEEDBACK VOLTAGE AMPLIFIER: SERIES-SHUNT
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For the feedback transresistance amplifier in 5.1.1), use small-signal analysis to find the open-loop gain 'G', Feedback factor 'H' and Closed-loop

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gain 'T'. Let $R_F \gg R_L$ and $r_o \gg R_L$. Find the value of T for $R_L = 10K\Omega$, $R_F = 100K\Omega$ and the transistor current gain $\beta = 100$.

5.1. Draw the equivalent control system for the feedback Transresistance amplifier shown in 5.1.1

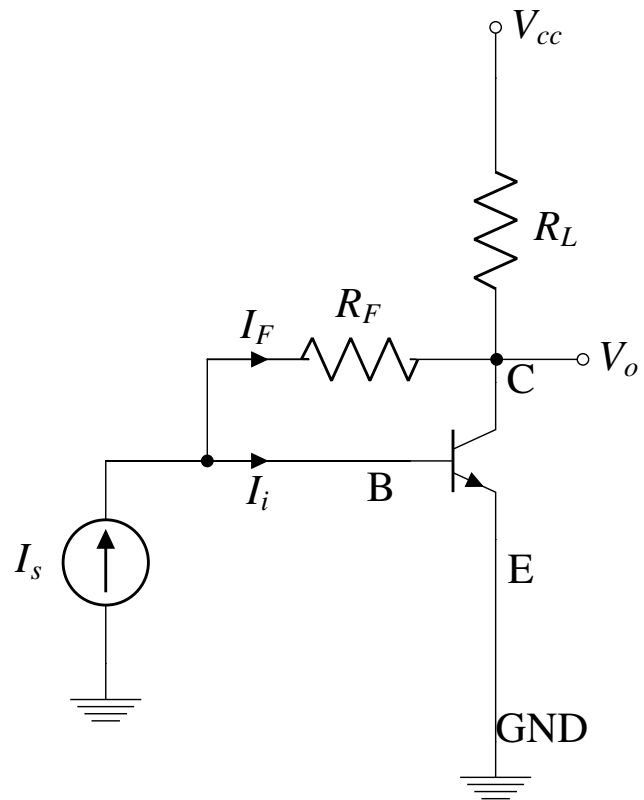


Fig. 5.1.1

Solution: see Fig. 5.1.2

5.2. For the feedback Transresistance amplifier shown in 5.1.1, Draw its small signal model. Early effect in Transistor is neglected.

Solution: see Fig. 5.2

While drawing a Small-Signal Model, we ground all constant voltage sources and open all constant current sources. All Small-Signal parameters are obtained from DC-Analysis

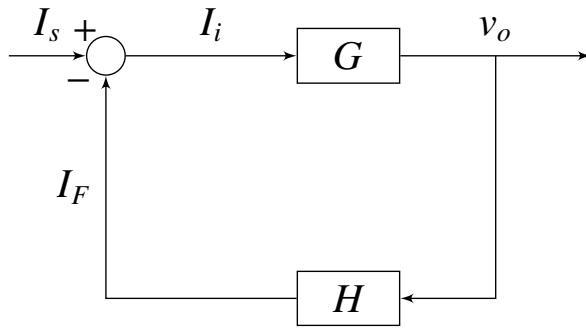


Fig. 5.1.2

of the circuit. Neglecting Early effect, in SmallSignal Analysis a npn-Transistor is modelled as a Current Source with value of current equal to $g_m V_{be}$ flowing from Collector to Emitter. Whereas a pnp-Transistor is modelled as a Current Source with value of current equal to $g_m V_{be}$ flowing from Emitter to Collector.

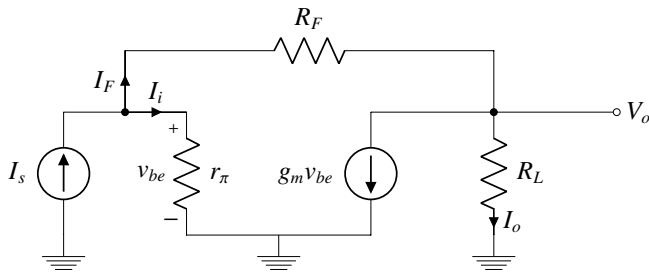


Fig. 5.2: Small Signal Model

5.3. Find small signal parameters g_m and v_{be} using DC analysis

Solution: small signal parameters of bjt are given in (5.3.1) and (5.3.2)

$$g_m = \frac{I_C}{V_T} \quad (5.3.1)$$

$$r_\pi = \frac{V_T}{I_B} \quad (5.3.2)$$

The Large signal model of circuit becomes as shown in figure 5.3

Where $V_T = 25\text{mvolts}$

$$V_{BE} = 0.7\text{volts} \implies V_B = 0.7\text{volts} \quad (5.3.3)$$

$$I_E = I_B + I_C \quad (5.3.4)$$

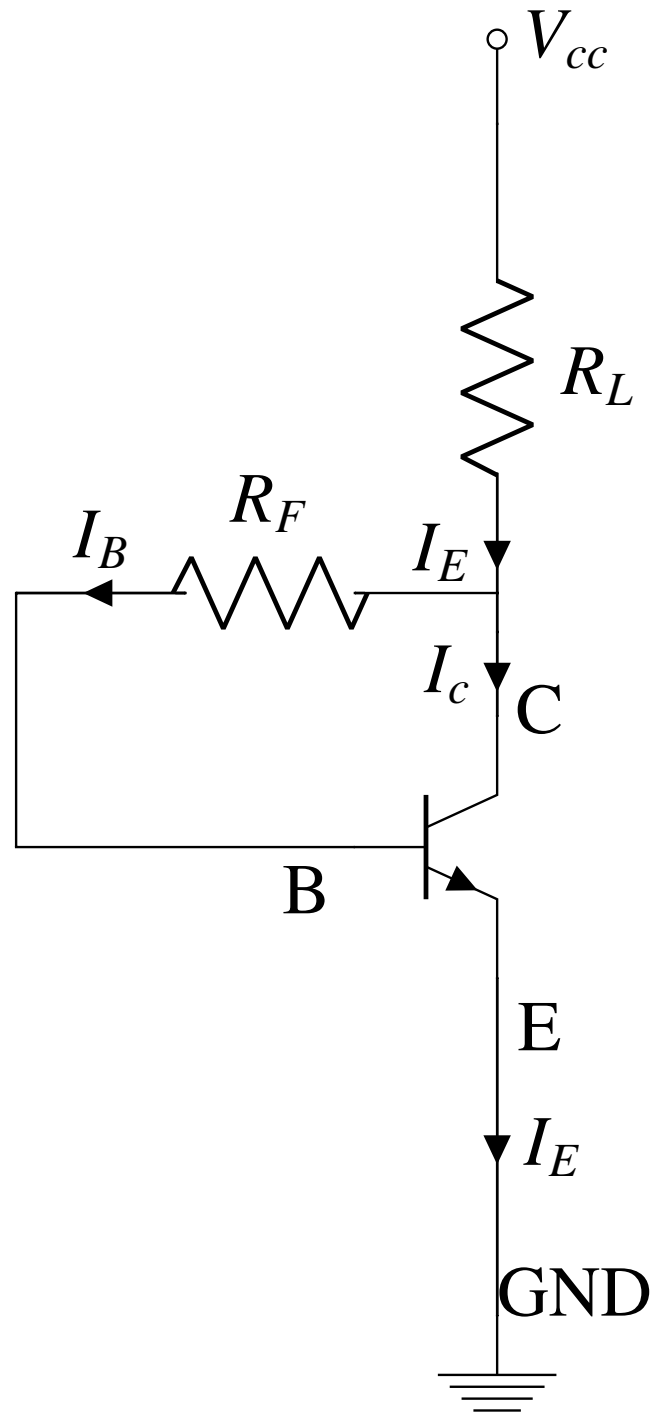


Fig. 5.3: Large signal model

$$I_C = \beta I_B \quad (5.3.5)$$

From applying KVL and KCL on Fig.

$$V_{cc} - I_E R_L - I_B R_F - 0.7 = 0 \quad (5.3.6)$$

$$\implies V_{cc} - (\beta + 1) I_B R_L - I_B R_F - 0.7 = 0 \quad (5.3.7)$$

$$I_B = \frac{V_{cc} - 0.7}{(\beta + 1)R_L + R_F} \quad (5.3.8)$$

$$I_C = \beta \frac{V_{cc} - 0.7}{(\beta + 1)R_L + R_F} \quad (5.3.9)$$

from (5.3.1), (5.3.2), I_B and I_C

$$g_m = \frac{\beta}{V_T} \frac{V_{cc} - 0.7}{(\beta + 1)R_L + R_F} \quad (5.3.10)$$

$$r_\pi = V_T \frac{(\beta + 1)R_L + R_F}{V_{cc} - 0.7} \quad (5.3.11)$$

5.4. Write all node/loop equations of Small-Signal model using KCL/KVL. Given that $R_F \gg R_L$

Solution:

$$v_{be} = I_i r_\pi \quad (5.4.1)$$

$$v_{be} - I_F R_F = V_o \quad (5.4.2)$$

$$V_o = (I_F - g_m v_{be}) R_L \quad (5.4.3)$$

5.5. Find the expression for feedback factor H.

Solution:

$$H = \frac{I_F}{V_o} \quad (5.5.1)$$

substituting (5.4.2) in (5.4.3)

$$V_o = (I_F - g_m V_o - g_m I_F R_F) R_L \quad (5.5.2)$$

$$\Rightarrow (1 + g_m R_L) V_o = I_F (R_L - g_m R_F R_L) \quad (5.5.3)$$

$$H = \frac{I_F}{V_o} = \frac{1 + g_m R_L}{R_L (1 - g_m R_F)} \quad (5.5.4)$$

$$\Rightarrow H \approx -\frac{1}{R_F} \quad (5.5.5)$$

5.6. Find the expression for Open loop Gain G.

Solution:

$$G = \frac{V_o}{I_i} \quad (5.6.1)$$

Substituting (5.4.1) in (5.4.2) and substituting I_F from (5.5.4)

$$I_i r_\pi - \left(\frac{1 + g_m R_L}{R_L (1 - g_m R_F)} \right) R_F V_o = V_o \quad (5.6.2)$$

$$\Rightarrow G = \frac{V_o}{I_i} = \frac{r_\pi R_L (1 - g_m R_F)}{R_F + R_L} \quad (5.6.3)$$

Upon approximating since $R_F \gg R_L$

$$G = -g_m r_\pi R_L \quad (5.6.4)$$

5.7. Find the expression for Closed Loop Gain $T = \frac{V_o}{I_s}$ We know that Closed Loop Gain

$$T = \frac{G}{1 + GH} \quad (5.7.1)$$

Substituting expressions from (5.5.5) and (5.6.3)

$$T = -\frac{g_m r_\pi R_L}{1 + \left(\frac{g_m r_\pi R_L}{R_F} \right)} \quad (5.7.2)$$

5.8. For the parameters given in table 5.8. Find G, H and T. **Solution:** Substituting the parameters in

Parameters	Value
V_{cc}	5V
I_s	1μ
R_F	$100K\Omega$
R_L	$10K\Omega$
β	100

TABLE 5.8

(5.3.10) and (5.3.11) gives,

$$r_\pi = 6.6667 \times 10^3 \Omega \quad (5.8.1)$$

$$g_m = 0.015S \quad (5.8.2)$$

Substituting g_m , r_π obtained in (5.5.5)

$$H = -10^{-5} \quad (5.8.3)$$

Substituting g_m , r_π obtained in (5.6.4)

$$G = -10^6 \quad (5.8.4)$$

Substituting g_m , r_π obtained in (5.7.2)

$$T = -90909.09 \quad (5.8.5)$$

5.9. Draw the block diagram and circuit diagram for H.

Solution: see figs 5.9.5 and 5.9.6

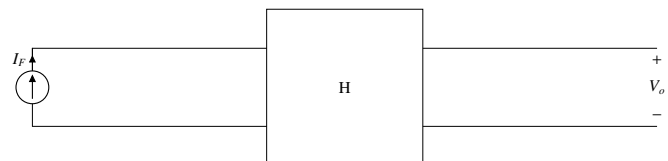


Fig. 5.9.5: Feedback block diagram

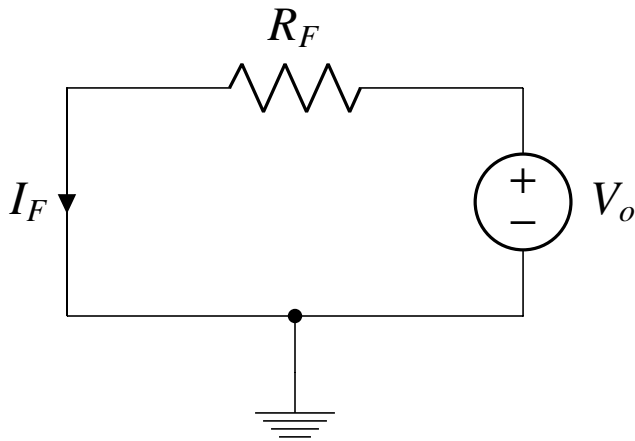


Fig. 5.9.6: Feedback circuit

From KVL on 5.9.6 we can see that

$$H = \frac{I_F}{V_o} = -\frac{1}{R_F} \quad (5.9.1)$$

5.10. Find the input and output resistances of the feedback network.

Solution: From the feedback amplifier circuit fig.5.9.6 To find the input resistance R_{11} short the output node V_o to ground.

$$R_{11} = R_F \quad (5.10.1)$$

To find the output resistance R_{22} remove the current source and short input terminals.

$$R_{22} = R_F \quad (5.10.2)$$

5.11. Draw the block diagram and circuit diagram for G.

Solution: see figs 5.11.7 and 5.11.8

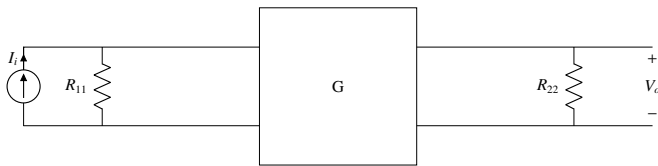


Fig. 5.11.7: Open loop block diagram

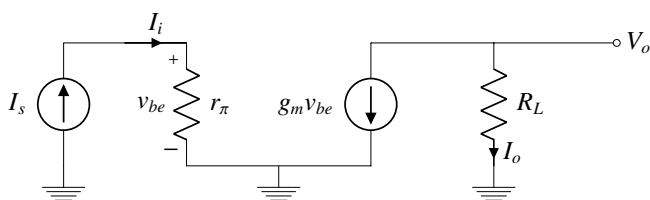


Fig. 5.11.8: Open loop block circuit diagram

5.12. Find G

Solution: From fig.5.11.8,

$$V_{be} = I_i r_\pi \quad (5.12.1)$$

From KCL at node V_o ,

$$I_o = -g_m I_i r_\pi \quad (5.12.2)$$

$$V_o = -g_m I_i r_\pi R_L \quad (5.12.3)$$

Therefore,

$$G = \frac{V_o}{I_i} = -g_m r_\pi R_L \quad (5.12.4)$$

5.13. Simulate the circuit using ngspice

Solution: The following file gives instructions on how to simulate the circuit.

codes/ee18btech11046/spice/README

The following netlist simulates the feedback amplifier using parameters in table 5.8.

codes/ee18btech11046/spice/
ee18btech11046_bjt.net

The Output Voltage obtained from spice is plotted in fig.5.13.9

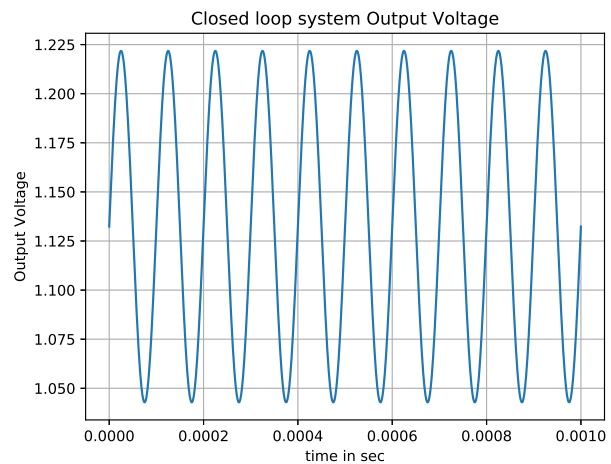


Fig. 5.13.9: Output Voltage

codes/ee18btech11046/spice/ee18btech11046.py

We can observe that V_o is sum of sine wave amplified by a factor of 89500 for small signal input and large signal output V_C which is close to the calculated values.