

# Assignment 1

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Download all python and C codes from

[https://github.com/narasimha-123/EE3250/Assignment 1-C/Codes](https://github.com/narasimha-123/EE3250/Assignment%201-C/Codes)

Download all sound files and data files from

[https://github.com/narasimha-123/EE3250/Assignment 1-C/Data](https://github.com/narasimha-123/EE3250/Assignment%201-C/Data)

and latex-tikz codes from

[https://github.com/narasimha-123/EE3250/Assignment 1-C](https://github.com/narasimha-123/EE3250/Assignment%201-C)

Run the following commands to compile and then run the C files to generate .dat files

```
gcc FILENAME.c -o FILENAME -lm
./FILENAME
```

The .dat files generated using C codes are plotted and generated as Sound files by executing this python file

Codes/ee18btech11046.py

```
#order of the filter
order=4
```

```
#cutoff frequency 4kHz
cutoff_freq=4000.0
```

```
#digital frequency
Wn=2*cutoff_freq/sampl_freq
```

```
# b and a are numerator and denominator
polynomials respectively
b, a = signal.butter(order,Wn, 'low')
```

```
#filter the input signal with butterworth filter
output_signal = signal.filtfilt(b, a,
    input_signal)
#output_signal = signal.lfilter(b, a,
    input_signal)
```

```
#write the output signal into .wav file
sf.write('Sound_With_ReducedNoise.wav',
    output_signal, fs)
```

## 1 DIGITAL FILTER

### 1.1 Download the sound file from

[wget https://raw.githubusercontent.com/gadepall/EE1310/master/filter/codes/Sound\\_Noise.wav](https://raw.githubusercontent.com/gadepall/EE1310/master/filter/codes/Sound_Noise.wav)

### 1.2 Write the python code for removal of out of band noise and execute the code.

**Solution:**

```
import soundfile as sf
from scipy import signal

#read .wav file
input_signal,fs = sf.read('Sound_Noise.wav')

#sampling frequency of Input signal
sampl_freq=fs
```

## 2 DIFFERENCE EQUATION

### 2.1 Write the difference equation of above Digital filter obtained in problem 1.2.

**Solution:**

$$\sum_{m=0}^M a(m) y(n-m) = \sum_{k=0}^N b(k) x(n-k) \quad (2.0.1)$$

$$y(n) - 2.52y(n-1) + 2.56y(n-2) - 1.206y(n-3) + 0.22013y(n-4) = 0.00345x(n) + 0.0138x(n-1) + 0.020725x(n-2) + 0.0138x(n-3) + 0.00345x(n-4) \quad (2.0.2)$$

### 2.2 Sketch x(n) and y(n).

**Solution:** The following code yields Fig. 2.2

Codes/xnyn.c

The filtered sound signal obtained through difference equation is found in

Data/Sound\_de.wav

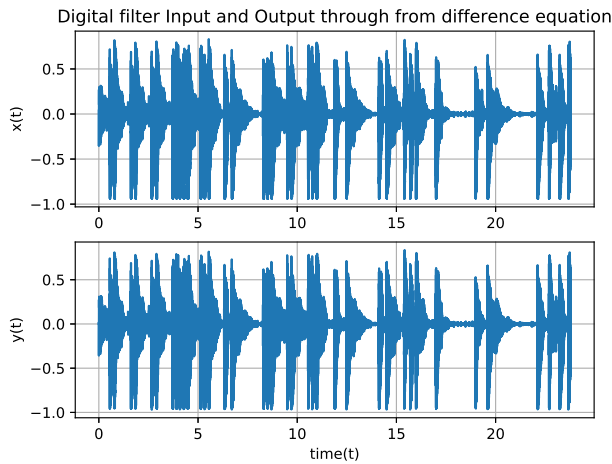


Fig. 2.2

### 3 Z-TRANSFORM

3.1

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n} \quad (3.0.1)$$

Show that

$$\mathcal{Z}\{x(n-1)\} = z^{-1}X(z) \quad (3.0.2)$$

and find

$$\mathcal{Z}\{x(n-k)\} \quad (3.0.3)$$

**Solution:** From (3.0.1),

$$\mathcal{Z}\{x(n-k)\} = \sum_{n=-\infty}^{\infty} x(n-k)z^{-n} \quad (3.0.4)$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-1} = z^{-1} \sum_{n=-\infty}^{\infty} x(n)z^{-n} \quad (3.0.5)$$

resulting in (3.0.2). Similarly, it can be shown that

$$\mathcal{Z}\{x(n-k)\} = z^{-k}X(z) \quad (3.0.6)$$

3.2 Find

$$H(z) = \frac{Y(z)}{X(z)} \quad (3.0.7)$$

from (2.0.2) assuming that the Z-transform is a linear operation.

**Solution:** Applying (3.0.6) in (2.0.2) we get,

$$\begin{aligned} H(z) &= \frac{Y(z)}{X(z)} \\ &= \frac{b[0] + b[1]z^{-1} + b[2]z^{-2} + b[3]z^{-3} + b[4]z^{-4}}{a[0] + a[1]z^{-1} + a[2]z^{-2} + a[3]z^{-3} + a[4]z^{-4}} \end{aligned} \quad (3.0.8)$$

3.3 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}). \quad (3.0.9)$$

Plot  $|H(e^{j\omega})|$ .

**Solution:** The following code plots Fig. 3.3.

Codes/dtft.py

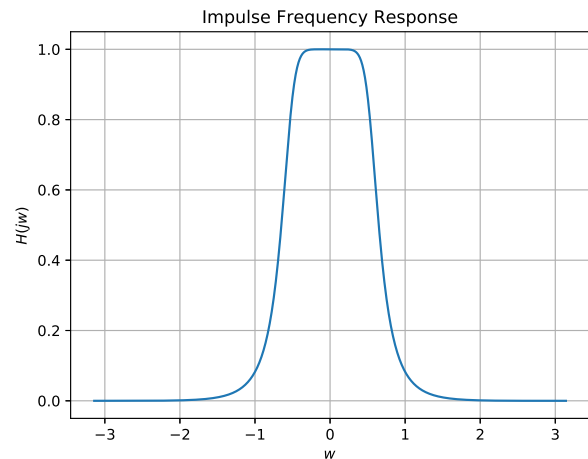


Fig. 3.3:  $|H(e^{j\omega})|$

### 4 IMPULSE RESPONSE

4.1 From the difference equation eq. 2.0.2. Sketch  $h(n)$

**Solution:** We know that when an impulse is given as input to a system we get the Impulse response  $h(n)$  of the system as output.

From eq.2.0.1,

By substituting  $x(n-k) = \delta(n-k)$ , then  $y(n-k)$  becomes  $h(n-k)$  for all  $k=0,1,2,3,4$ .

Now, the following code plots Fig.4.1

Codes/hndef.c

4.2 Check whether  $h(n)$  obtained is stable.

**Solution:** The system is defined by the equation 2.0.2 For a system to be stable, output should be bounded when the input is bounded. This is known as BIBO stability.

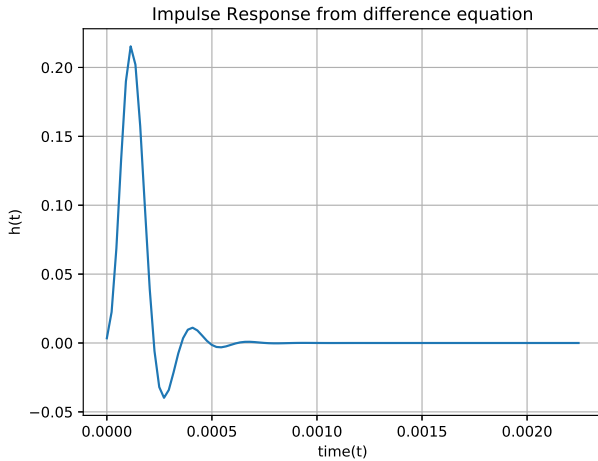


Fig. 4.1:  $h(n)$

Since  $x(n)$  is bounded, let  $B_x$  be some finite value

$$|y(n)| \leq B_x \sum_{-\infty}^{\infty} |x(n-k)| \quad (4.0.1)$$

From convolution formula,

$$|y(n)| = \left| \sum_{-\infty}^{\infty} h(k)x(n-k) \right| \quad (4.0.2)$$

$$|y(n)| \leq \sum_{-\infty}^{\infty} |h(k)| |x(n-k)| \quad (4.0.3)$$

Let  $B_x$  be the maximum value  $x(n-k)$  can take, then

$$|y(n)| \leq B_x \sum_{-\infty}^{\infty} |h(k)| \quad (4.0.4)$$

If

$$\sum_{-\infty}^{\infty} |h(k)| < \infty \quad (4.0.5)$$

Then

$$|y(n)| \leq B_y < \infty \quad (4.0.6)$$

Therefore we can say that  $y(n)$  is bounded if  $x(n)$  and  $h(n)$  are bounded.

Since the audio input is bounded, the system is said to be stable if  $h(n)$  is also bounded

$$\sum_{n=-\infty}^{\infty} |h(n)| < \infty \quad (4.0.7)$$

The above equation can be re written as,

$$\sum_{n=-\infty}^{\infty} |h(n)z^{-n}|_{|z|=1} < \infty \quad (4.0.8)$$

$$\sum_{n=-\infty}^{\infty} |h(n)| |z^{-n}|_{|z|=1} < \infty \quad (4.0.9)$$

From Triangle inequality,

$$\left| \sum_{n=-\infty}^{\infty} h(n)z^{-n} \right|_{|z|=1} < \infty \quad (4.0.10)$$

$$\Rightarrow |H(n)|_{|z|=1} < \infty \quad (4.0.11)$$

Therefore, the Region of Convergence(ROC) should include the unit circle for the system to be stable.

Since,  $h(n)$  is right sided the ROC is outside the outer most pole. From the equation (3.0.8) Poles of the given transfer equation is:

$$z(\text{approx}) = 0.69382 \pm 0.41i, \quad 0.56617835 \pm 0.134423 \quad (4.0.12)$$

From the above poles, we can see that that the ROC of the system is  $|z| > \sqrt{0.69382^2 + 0.41^2}$ . From the figure we can observe that ROC of

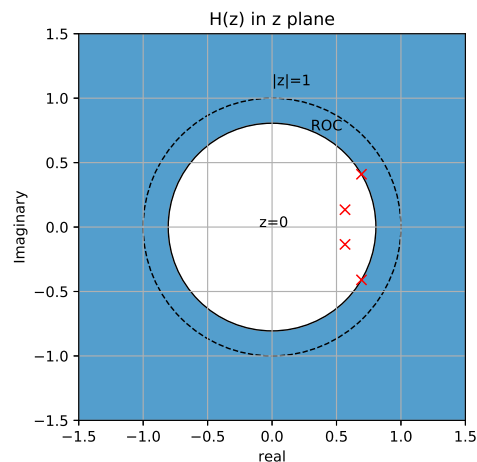


Fig. 4.2:  $X(k)$  and  $H(k)$

the system includes unit circle  $|z| = 1$ .

The code for plotting the figure is:

Codes/ROC.py

Which implies that the given IIR filter is stable, because  $h(n)$  is absolutely summable.

**Verification:-**

Given bounded input  $x(n)$  (audio sample) and system difference equation 2.0.2

From 2.2 we can see that the maximum value of  $x(n)$  is 0.839 and minimum value is greater than -0.94171.

Similarly 2.2 we can also see that the maximum value of  $y(n)$  is 0.822256 and minimum value is -0.953761 and it tends to zero after the length of signal.

We can see that the bounded input  $x(n)$  gives bounded output  $y(n)$ . Therefore we can say that the system is BIBO stable.

4.3 Compute Filtered output using convolution formula using  $h(n)$  obtained in 4.1

$$y(n) = x(n) * h(n) = \sum_{n=-\infty}^{\infty} x(k)h(n-k) \quad (4.0.13)$$

**Solution:** The following code plots Fig. 4.3

Codes/ynconv.py

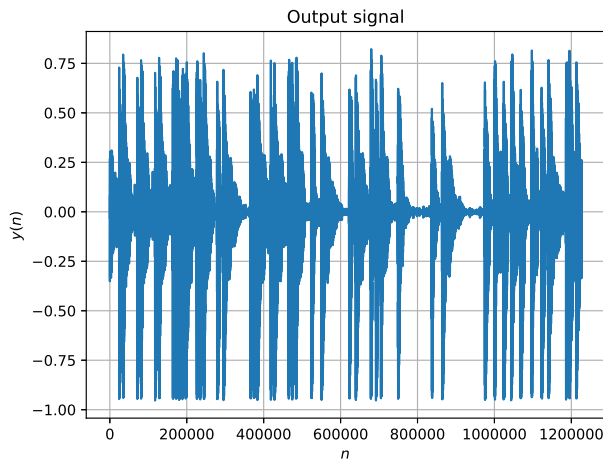


Fig. 4.3:  $y(n)$  from the definition of convolution

The filtered sound signal through convolution from this method is found in

Data/Sound\_conv.wav

We can observe that the output obtained is same as  $y(n)$  obtained in Fig. 2.2

## 5 FFT AND IFFT

5.1 Compute

$$X(k) \triangleq \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \quad (5.0.1)$$

and  $H(k)$  using  $h(n)$ .

**Solution:** For this given IIR system with audio sample as  $x(n)$  and  $h(n)$  as impulse response  $h(n)$  obtained in 4.1

DFT of a Input Signal  $x(n)$  is

$$X(k) \triangleq \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \quad (5.0.2)$$

DFT of a Impulse Response  $h(n)$  is

$$H(k) \triangleq \sum_{n=0}^{N-1} h(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \quad (5.0.3)$$

The following code plots FFT of  $x(n)$  and  $h(n)$ .

Codes/xnhnfft.c

Magnitude and Phase plots obtained through above code is

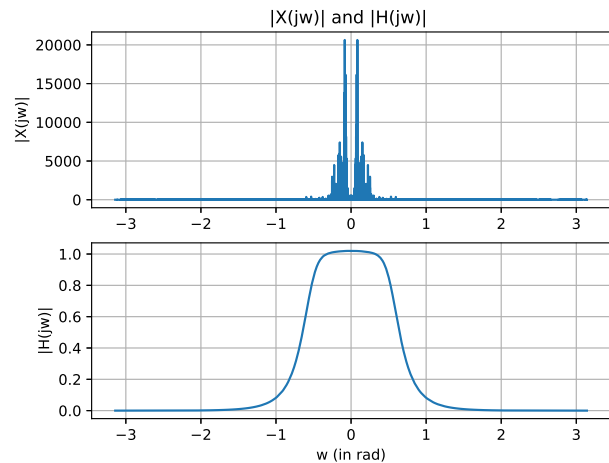


Fig. 5.1:  $X(k)$  and  $H(k)$

5.2 From

$$Y(k) = X(k)H(k) \quad (5.0.4)$$

Compute

$$y(n) \triangleq \sum_{k=0}^{N-1} Y(k)e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1 \quad (5.0.5)$$

**Solution:** The following code plots Fig.5.2

Codes/ynfft.c

The filtered sound signal from this method is found in

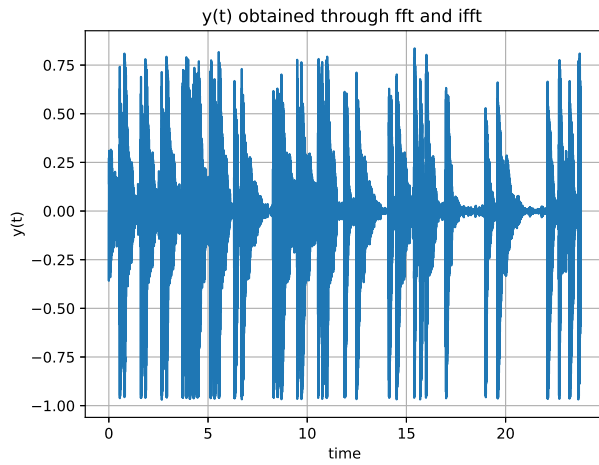


Fig. 5.2:  $y(n)$

Data/Sound\_fft.wav

We can observe from the above plot that it is same as the  $y(n)$  observed in Fig.2.2