

12.1 Consider the following open-loop TF, where $G(s) = Y(s)/U(s)$

$$\begin{aligned} \text{i. } G(s) &= \frac{(s+3)}{(s+4)^2} & \text{ii. } G(s) &= \frac{s}{(s+5)(s+7)} \\ \text{iii. } G(s) &= \frac{20s(s+7)}{(s+3)(s+7)(s+9)} & \text{iv. } G(s) &= \frac{30(s+2)(s+3)}{(s+4)(s+5)(s+6)} \\ \text{v. } G(s) &= \frac{s^2+8s+15}{(s^2+4s+10)(s^2+3s+12)} \end{aligned}$$

For each of these TF, do the following

- Draw the signal-flow graph in **phase-variable form**
- Add state-variable feedback to the signal-flow graph
- For each closed-loop signal-flow graph, write the state equations
- Write, by inspection, the closed-loop TF $T(s)$ for your closed-loop signal-flow graphs
- Verify your answers for $T(s)$ by finding the closed-loop TF from the state equations and Eq.3.73

12.2 The following open-loop TF can be represented by signal-flow graphs in **cascade form**

$$\text{i. } G(s) = \frac{30(s+2)(s+7)}{s(s+3)(s+5)} \quad \text{ii. } G(s) = \frac{5(s^2+3s+7)}{(s+2)(s^2+2s+10)}$$

For each, do the following

- Draw the signal-flow graph and show the state variable feedback
- Find the closed-loop TF with state variable feedback

Solution

12.3 The following open-loop TF can be represented by signal-flow graphs in **parallel form**

$$\text{a. } G(s) = \frac{50(s^2+7s+25)}{s(s+10)(s+20)} \quad \text{b. } G(s) = \frac{50(s+3)(s+4)}{(s+5)(s+6)(s+7)}$$

For each, do the following

- Draw the signal-flow graph and show the state variable feedback
- Find the closed-loop TF with state variable feedback

12.4 Given the following open-loop plant

$$G(s) = \frac{20}{(s+2)(s+4)(s+8)}$$

design a controller to yield a $\%OS = 4.32\%$ and $T_s = 4s$. Place the third pole 10 times as far from the imaginary axis as the dominant pole pair. Use the phase variables for state-variable feedback

12.5 Section 12.2 showed that controller design is easier to implement if the uncompensated system is represented in phase-variable form with its typical **lower companion matrix**. We alluded to the fact that the design can just as easily progress using the controller canonical form with its **upper companion matrix**

- Redo the general controller design covered in Section 12.2, assuming that the plant is represented in controller canonical form rather than phase-variable form
- Apply your derivation to the system

$$G(s) = \frac{20(s+5)}{s(s+1)(s+4)}, \quad \%OS = 9.5\%, \quad T_s = 0.74s$$

if the uncompensated plant is represented in controller canonical form

12.6 Given the following open-loop plant

$$G(s) = \frac{100(s+2)(s+20)}{(s+1)(s+3)(s+4)}$$

design a controller to yield 15% overshoot with a peak time of 0.5s. Use the **controller canonical form** for state-variable feedback

12.7 Given the following open-loop plant

$$G(s) = \frac{20(s+2)}{s(s+5)(s+7)}$$

design a controller to yield $\%OS = 10\%$ and $T_s = 2s$. Place the third pole 10 times as far from the imaginary axis as the dominant pole pair. Use the [phase variables](#) for state-variable feedback

12.8 Given the following open-loop plant

$$G(s) = \frac{20}{(s+2)(s+4)(s+8)}$$

design a controller to yield a $\%OS = 4.32\%$ and $T_s = 4s$. Place the third pole 10 times as far from the imaginary axis as the dominant pole pair. Use the [cascade form](#) for state-variable feedback

12.18 If an open-loop plant

$$G(s) = \frac{100}{s(s+5)(s+9)}$$

is represented in parallel form, design a controller to yield a closed-loop response of 15% overshoot and a peak time of 0.2s. Design the controller by first transforming the plant to controller canonical form

12.20 Consider the plant

$$G(s) = \frac{1}{s(s+3)(s+7)}$$

whose state variables are not available. Design an observer for the observer canonical variables to yield a transient response described by $\zeta = 0.4$ and $\omega_n = 75$. Place the third pole 10 times farther from the imaginary axis than the dominant poles