

Signal Processing

ENGN2228

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Second Semester

Lecture 18

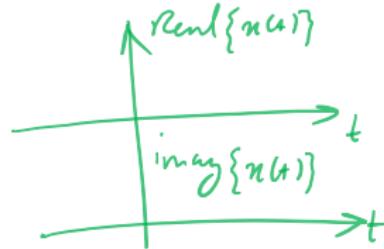


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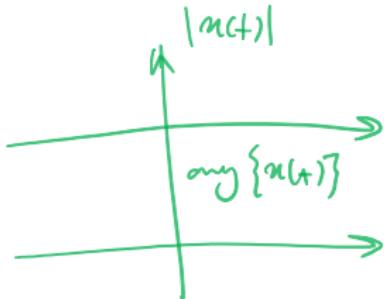
Complex numbers

$$x(t) = \underbrace{\alpha(t)}_{\text{Real part}} e^{j\beta(t)} = \underbrace{\alpha(t) \cos(\beta(t))}_{\text{Real part}} + j \underbrace{\alpha(t) \sin(\beta(t))}_{\text{Imaginary part}}$$

$$\left\{ \begin{array}{l} \text{Real}\{x(t)\} = \alpha(t) \cos(\beta(t)) \\ \text{Imag}\{x(t)\} = \alpha(t) \sin(\beta(t)) \end{array} \right.$$



$$\left\{ \begin{array}{l} |x(t)| = \sqrt{\alpha(t)^2 + \sin^2(\beta(t))} \\ \arg\{x(t)\} = \beta(t) \end{array} \right.$$



Example:

$$n(t) + n^*(t) = 2 \left[\operatorname{Re}\{n(t)\} \right]$$

Let $x(t) = \sqrt{2}(1+j)e^{j\pi/4}e^{(-1+j2\pi)t}$. Sketch and label the following:

(a) $\operatorname{Re}\{x(t)\} = -e^{-t} \sin(2\pi t)$

(b) $\operatorname{Im}\{x(t)\} = e^{-t} \cos(2\pi t)$

(c) $x(t+2) + x^*(t+2) = 2 \left[\operatorname{Re}\{n(t+2)\} \right] = 2 \left[-e^{-t-2} \sin(2\pi(t+2)) \right]$

(d) $|n(t)| = e^{-t}$

(e) $\arg\{n(t)\} = 2\pi t + \frac{\pi}{2}$

$$n(t) = \sqrt{2} \left(\frac{1}{\sqrt{2}} e^{j\pi/4} \right) e^{j\pi/4} e^{-t} e^{j2\pi t} = \underline{e^{-t}} \underline{e^{j(2\pi t + \pi/2)}}$$

$$= \bar{e}^{-t} \left[\underbrace{\cos(2\pi t + \frac{\pi}{2})}_{-\sin(2\pi t)} + j \underbrace{\sin(2\pi t + \frac{\pi}{2})}_{\cos(2\pi t)} \right]$$



Example: From the last lecture

$$T=2\pi, \omega_0 = \frac{2\pi}{T} = \pi \text{ rad/s}$$

$$a_k = \frac{1}{T} \int_T^T n(t) e^{-jk\omega_0 t} dt$$

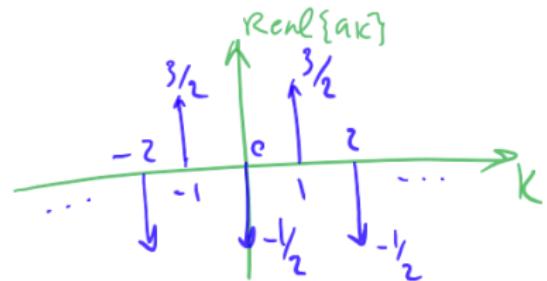
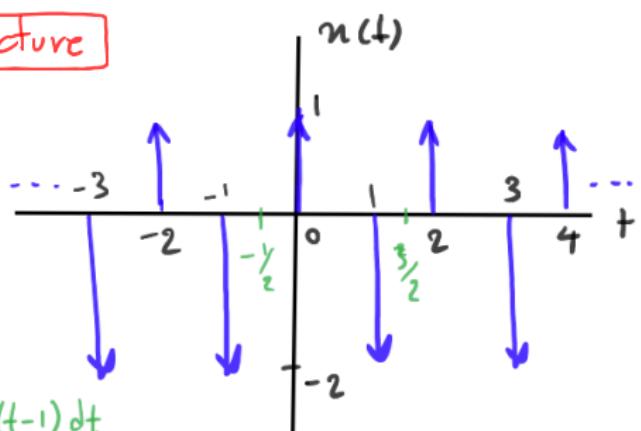
$$a_0 = \frac{1}{2} \int_{-\frac{1}{2}}^{\frac{3}{2}} n(t) dt - \frac{1}{2} \int_{-\frac{1}{2}}^{\frac{3}{2}} \delta(t) - 2\delta(t-1) dt$$

$$= \frac{1}{2}(1-2) = -\frac{1}{2}$$

$$a_K = \frac{1}{2} \int_{-\frac{1}{2}}^{\frac{3}{2}} [\delta(t) - 2\delta(t-1)] e^{-jk\omega_0 t} dt$$

$$= \frac{1}{2} \int_{-\frac{1}{2}}^{\frac{3}{2}} \delta(t) - 2\delta(t-1) e^{-jk\omega_0 t} dt$$

$$= \frac{1}{2} - e^{j k \omega_0} = \frac{1}{2} - \left(e^{-j\pi}\right)^k = \frac{1}{2} - (-1)^k$$



$$\begin{aligned} \frac{1}{2} - (-1)^k &= \frac{1}{2} \\ \frac{1}{2} - (-1)^0 &= \frac{1}{2} \\ \frac{1}{2} - (-1)^1 &= \frac{1}{2} \\ \frac{1}{2} - (-1)^2 &= 0 \end{aligned}$$

Signals Classifications:

$u(t)$

$u[n]$

- Continuous or discrete

$$u(t) = \text{even}\{u(t)\} + \text{odd}\{u(t)\}$$

$$\text{even}\{u(t)\} = \frac{u(t) + u(-t)}{2}$$

$$\text{odd}\{u(t)\} = \frac{u(t) - u(-t)}{2}$$

- Even, odd or none

- Periodic or non-periodic

- Energy or power

$$E\{u(t)\} = \int_{-\infty}^{+\infty} |u(t)|^2 dt < \infty$$

$$T \rightarrow \infty$$

$$\frac{E_u}{\omega_0} = T$$

$$\cos(\omega_0 t) + \sin(\beta \omega_0 t)$$

$$\frac{E_u}{\omega_0} m$$

$$\cos(\omega_0 n)$$



Example:

Let $x[n] = \cos(\Omega_x(n + P_x) + \theta_x)$.

(a) Determine the period of $x[n]$ for each of the following three cases:

	Ω_x	P_x	θ_x	(i) $n[n] = \cos\left(\frac{\pi}{3}(n+0) + 2\pi\right) = \cos\left(\frac{\pi}{3}n\right)$
(i)	$\pi/3$	0	2π	$\frac{2\pi}{\omega_0}m = \frac{2\pi}{\pi/3}m = 6m \xrightarrow{m=1} N_0=6$
(ii)	$3\pi/4$	2	$\pi/4$	$\frac{2\pi}{\omega_0}m = \frac{2\pi}{3\pi/4}m = \frac{8}{3}m \xrightarrow{m=3} N_0=8$
(iii)	$3/4$	1	$1/4$	

$$(ii) n[n] = \cos\left(\frac{3\pi}{4}(n+2) + \frac{\pi}{4}\right) = \cos\left(\frac{3\pi}{4}n + \frac{7\pi}{4}\right)$$

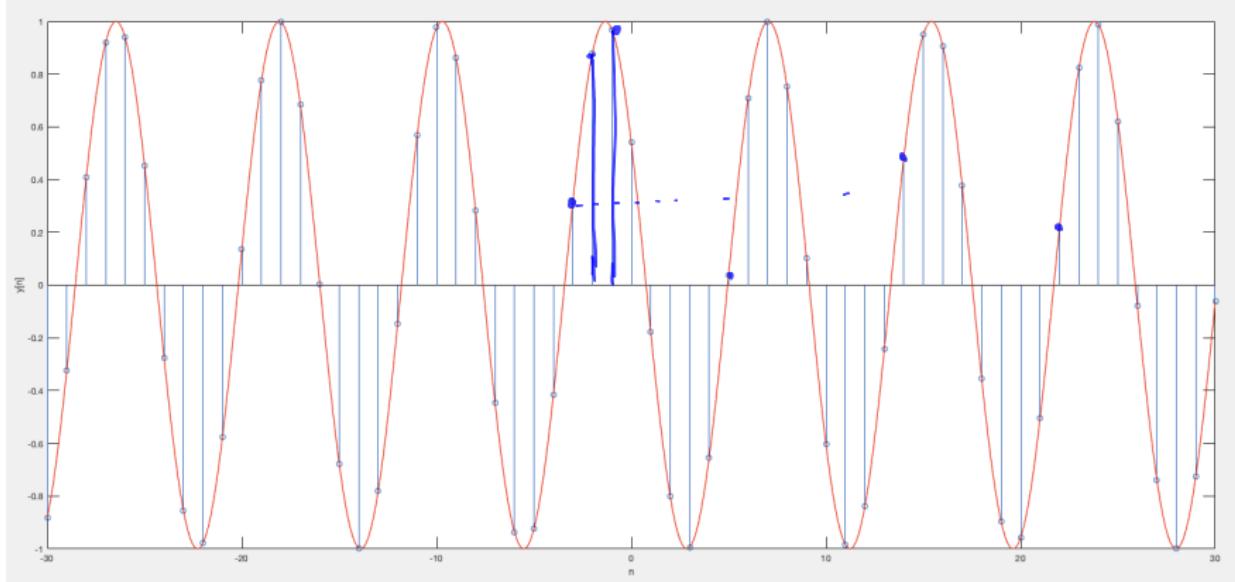
$$\frac{2\pi}{\frac{3\pi}{4}}m = \frac{8}{3}m \xrightarrow{m=3} N_0=8$$

$$(iii) n[n] = \cos\left(\frac{3}{4}(n+1) + \frac{1}{4}\right) = \cos\left(\frac{3}{4}n + 1\right)$$

$$\frac{2\pi}{\frac{3}{4}}m = \frac{8}{3}\pi m \longrightarrow \text{This signal is not periodic}$$



Example (con'd):



System Properties :

- Causal or non-causal
- With memory or memory-less

* • Time-invariant or time-variant

- Linear or non-linear
- Stable or unstable

$$\downarrow h(t)$$

$$\int_{-\infty}^{\infty} |h(t)| dt < \infty$$

$$y(t) = f_t(u(t))$$

$$y(t-t_0) = f_{t-t_0}(u(t-t_0)) \quad \textcircled{I}$$

$$u_i(t) = u(t-t_0)$$

$$y_i(t) = f_t(u_i(t)) = f_t(u(t-t_0)) \quad \textcircled{II}$$

$$y(t) = f(u(t))$$

$$f(a u_1(t) + b u_2(t)) = a f(u_1(t)) + b f(u_2(t))$$



Example:



$y(t), y[n]$	Memoryless	Linear	Time-Invariant	Causal
(a) $(2 + \sin t)x(t)$	✓	✓	✗	---
(b) $x(2t)$	---	✓	✗	✗
(c) $\sum_{k=-\infty}^{\infty} x[k]$	---	✓	✓	---
(d) $\sum_{k=-\infty}^n x[k]$	✗	✓	✓	✓
(e) $\frac{dx(t)}{dt}$	---	✓	---	---
(f) $\max\{x[n], x[n-1], \dots, x[-\infty]\}$	✗	✗	---	---

$$y[n] = f[n[n]]$$

$$f[a n_1[n] + b n_2[n]] = \max\{a n_1[n] + b n_2[n], \dots, a n_1[-\infty] + b n_2[-\infty]\}$$

$$\neq a \max\{n_1[n], \dots, n_1[-\infty]\} + b \max\{n_2[n], \dots, n_2[-\infty]\}$$



Solutions:

Memoryless:

- (a) $y(t) = (2 + \sin t)x(t)$ is memoryless because $y(t)$ depends only on $x(t)$ and not on prior values of $x(t)$.
- (d) $y[n] = \sum_{k=-\infty}^n x[k]$ is not memoryless because $y[n]$ does depend on values of $x[\cdot]$ before the time instant n .
- (f) $y[n] = \max\{x[n], x[n - 1], \dots, x[-\infty]\}$ is clearly not memoryless.

Linear:

(a)
$$\begin{aligned}y(t) &= (2 + \sin t)x(t) = T[x(t)], \\T[ax_1(t) + bx_2(t)] &= (2 + \sin t)[ax_1(t) + bx_2(t)] \\&= a(2 + \sin t)x_1(t) + b(2 + \sin t)x_2(t) \\&= aT[x_1(t)] + bT[x_2(t)]\end{aligned}$$

Therefore, $y(t) = (2 + \sin t)x(t)$ is linear.

(b)
$$\begin{aligned}y(t) &= x(2t) = T[x(t)], \\T[ax_1(t) + bx_2(t)] &= ax_1(2t) + bx_2(2t) \\&= aT[x_1(t)] + bT[x_2(t)]\end{aligned}$$

Therefore, $y(t) = x(2t)$ is linear.



Solutions:

(c) $y[n] = \sum_{k=-\infty}^{\infty} x[k] = T[x[n]],$

$$\begin{aligned}T[ax_1[n] + bx_2[n]] &= a \sum_{k=-\infty}^{\infty} x_1[k] + b \sum_{k=-\infty}^{\infty} x_2[k] \\&= aT[x_1[n]] + bT[x_2[n]]\end{aligned}$$

Therefore, $y[n] = \sum_{k=-\infty}^{\infty} x[k]$ is linear.

(d) $y[n] = \sum_{k=-\infty}^n x[k]$ is linear (see part c).

(e) $y(t) = \frac{dx(t)}{dt} = T[x(t)],$

$$\begin{aligned}T[ax_1(t) + bx_2(t)] &= \frac{d}{dt}[ax_1(t) + bx_2(t)] \\&= a \frac{dx_1(t)}{dt} + b \frac{dx_2(t)}{dt} = aT[x_1(t)] + bT[x_2(t)]\end{aligned}$$

Therefore, $y(t) = dx(t)/dt$ is linear.

(f) $y[n] = \max\{x[n], \dots, x[-\infty]\} = T[x[n]],$

$$\begin{aligned}T[ax_1[n] + bx_2[n]] &= \max\{ax_1[n] + bx_2[n], \dots, ax_1[-\infty] + bx_2[-\infty]\} \\&\neq a \max\{x_1[n], \dots, x_1[-\infty]\} + b \max\{x_2[n], \dots, x_2[-\infty]\}\end{aligned}$$

Therefore, $y[n] = \max\{x[n], \dots, x[-\infty]\}$ is not linear.



Solutions:

Time-invariant:

(a) $y(t) = (2 + \sin t)x(t) = T[x(t)],$
 $T[x(t - T_0)] = (2 + \sin t)x(t - T_0)$
 $\neq y(t - T_0) = (2 + \sin(t - T_0))x(t - T_0)$

Therefore, $y(t) = (2 + \sin t)x(t)$ is not time-invariant.

(b) $y(t) = x(2t) = T[x(t)],$
 $T[x(t - T_0)] = x(2t - 2T_0) \neq x(2t - T_0) = y(t - T_0)$

Therefore, $y(t) = x(2t)$ is not time-invariant.

(c) $y[n] = \sum_{k=-\infty}^{\infty} x[k] = T[x[n]],$
 $T[x[n - N_0]] = \sum_{k=-\infty}^{\infty} x[k - N_0] = y[n - N_0]$

Therefore, $y[n] = \sum_{k=-\infty}^{\infty} x[k]$ is time-invariant.

(d) $y[n] = \sum_{k=-\infty}^n x[k] = T[x[n]],$
 $T[x[n - N_0]] = \sum_{k=-\infty}^n x[k - N_0] = \sum_{l=-\infty}^{n - N_0} x[l] = y[n - N_0]$

Therefore, $y[n] = \sum_{k=-\infty}^n x[k]$ is time-invariant.



Solutions:

(e) $y(t) = \frac{dx(t)}{dt} = T[x(t)],$

$$T[x(t - T_0)] = \frac{d}{dt}x(t - T_0) = y(t - T_0)$$

Therefore, $y(t) = dx(t)/dt$ is time-invariant.

Causal:

(b) $y(t) = x(2t),$
 $y(1) = x(2)$

The value of $y(\cdot)$ at time = 1 depends on $x(\cdot)$ at a future time = 2. Therefore, $y(t) = x(2t)$ is not causal.

(d) $y[n] = \sum_{k=-\infty}^n x[k]$

Yes, $y[n] = \sum_{k=-\infty}^n x[k]$ is causal because the value of $y[\cdot]$ at any instant n depends only on the previous (past) values of $x[\cdot]$.

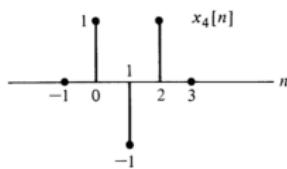
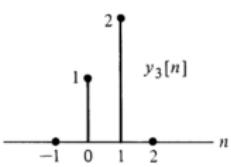
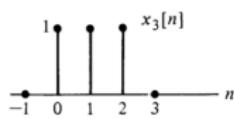
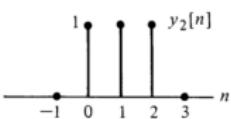
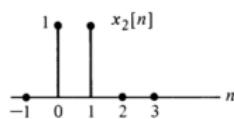
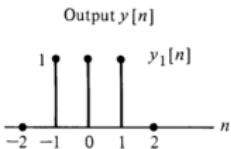
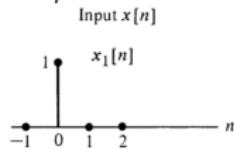


LTI Systems:



LTI Systems Example:

The system is linear:



?

$y_4[n]$

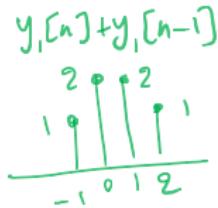
Is this system time-invariant?

$$n_4[n] = 2n_1[n] - 2n_2[n] + n_3[n]$$

$$y_4[n] = 2y_1[n] - 2y_2[n] + y_3[n]$$

$$\underbrace{n_1[n] + n_2[n-1]}_{y_1[n] + y_2[n-1]} = n_2[n]$$

$$y_1[n] + y_2[n-1] \Rightarrow$$



$$y_2[n] \Rightarrow$$

Since $y_4[n] \neq y_1[n] + y_2[n-1]$

the system is not time-invariant



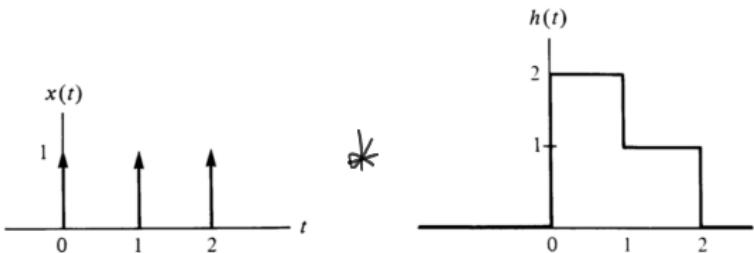
Convolution:

$$CT \quad y(t) = h(t) * g(t) = \int_{-\infty}^{\infty} n(\tau) h(t-\tau) d\tau$$

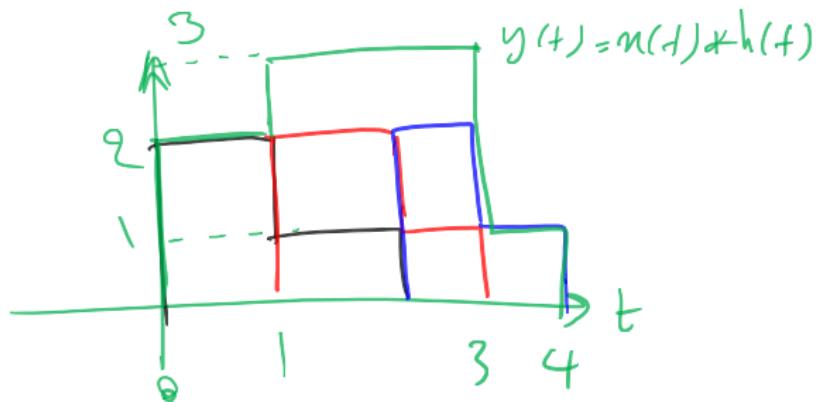
$$DT \quad y[n] = h[n] * n[n] = \sum_{k=-\infty}^{\infty} n[k] h[n-k]$$



Example:



$$n(t) * h(t) = ?$$



Definition (Fourier Analysis and Synthesis)

For $x(t) = x(t + T)$ periodic with period T and $\omega_0 = 2\pi/T$

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}, \quad t \in \mathbb{R} \quad (\text{Synthesis Equation})$$

$$a_k = \frac{1}{T} \int_T x(t) e^{-jk\omega_0 t} dt, \quad k \in \mathbb{Z} \quad (\text{Analysis Equation})$$



Motivation – Motivation (cont'd)

TABLE 3.1 PROPERTIES OF CONTINUOUS-TIME FOURIER SERIES

Property	Section	Periodic Signal	Fourier Series Coefficients
		$x(t) \left\{ \begin{array}{l} \text{Periodic with period } T \text{ and} \\ y(t) \text{ fundamental frequency } \omega_0 = 2\pi/T \end{array} \right.$	a_k b_k
Linearity	3.5.1	$Ax(t) + By(t)$	$Aa_k + Bb_k$
Time Shifting	3.5.2	$x(t - t_0)$	$a_k e^{-jk\omega_0 t_0} = a_k e^{-jk(2\pi/T)t_0}$
Frequency Shifting		$e^{jM\omega_0 t} x(t) = e^{jM(2\pi/T)t} X(t)$	a_{k-M}
Conjugation	3.5.6	$x^*(t)$	a_{-k}^*
Time Reversal	3.5.3	$x(-t)$	a_{-k}
Time Scaling	3.5.4	$x(\alpha t), \alpha > 0$ (periodic with period T/α)	a_k
Periodic Convolution		$\int_T x(\tau)y(t - \tau)d\tau$	$T a_k b_k$
Multiplication	3.5.5	$x(t)y(t)$	$\sum_{l=-\infty}^{+\infty} a_l b_{k-l}$
Differentiation		$\frac{dx(t)}{dt}$	$jk\omega_0 a_k = jk \frac{2\pi}{T} a_k$
Integration		$\int_{-\infty}^t x(t) dt$ (finite valued and periodic only if $a_0 = 0$)	$\left(\frac{1}{jk\omega_0}\right) a_k = \left(\frac{1}{jk(2\pi/T)}\right) a_k$
Conjugate Symmetry for Real Signals	3.5.6	$x(t)$ real	$\begin{cases} a_k = a_{-k}^* \\ \Re\{a_k\} = \Re\{a_{-k}\} \\ \Im\{a_k\} = -\Im\{a_{-k}\} \\ a_k = a_{-k} \\ \Im\{a_k\} = -\Im\{a_{-k}\} \end{cases}$
Real and Even Signals	3.5.6	$x(t)$ real and even	a_k real and even
Real and Odd Signals	3.5.6	$x(t)$ real and odd	a_k purely imaginary and odd
Even-Odd Decomposition of Real Signals		$\begin{cases} x_e(t) = \Re\{x(t)\} & [x(t) \text{ real}] \\ x_o(t) = \Im\{x(t)\} & [x(t) \text{ real}] \end{cases}$	$\Re\{a_k\}$ $j\Im\{a_k\}$

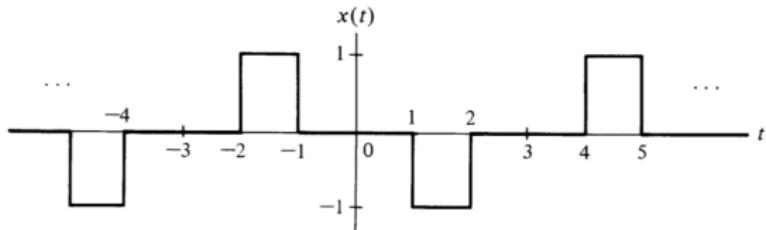
Parseval's Relation for Periodic Signals

$$\frac{1}{T} \int_T |x(t)|^2 dt = \sum_{k=-\infty}^{+\infty} |a_k|^2$$



Example:

$$a_K = ?$$



$$T = 6 \quad \frac{2\pi}{T} = \omega_0 \rightarrow \omega_0 = \frac{\pi}{3}$$

$$a_0 = \frac{1}{T} \int_T u(t) = 0$$

$$\begin{aligned} a_K &= \frac{1}{T} \int_T u(t) e^{-j K \omega_0 t} dt = \frac{1}{6} \int_{-3}^3 u(t) e^{-j K \omega_0 t} dt = +\frac{1}{6} \frac{1}{j K \omega_0} \left[e^{j K \omega_0 t} \right]_{-3}^3 - \frac{1}{6 j K \omega_0} \left[e^{-j K \omega_0 t} \right]_{-3}^3 \\ &= \frac{1}{j 2\pi K} \left[e^{+j \frac{\pi}{3} K} - e^{-j \frac{\pi}{3} K} + e^{+j \frac{9\pi}{3} K} - e^{-j \frac{9\pi}{3} K} \right] = \frac{1}{j \pi K} \left[\cos\left(\frac{\pi}{3} K\right) - \cos\left(\frac{9\pi}{3} K\right) \right] \end{aligned}$$



Time vs. Frequency domain:

		Periodic	Non-periodic	
		Fourier Series (FS)	Fourier Transform (FT)	Non-periodic
Continuous	Fourier Series (FS)	Fourier Transform (FT)	Periodic	
	Discrete-Time Fourier Series (DTFS)	Discrete-Time Fourier Transform (DTFT)		
		Discrete	Continuous	Frequency-domain Properties

