

# EE290C: 28nm SoC for IoT

## Analog Layout and Matching

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- Please do not post these slides on the Internet. There are no “secrets” in these slides, but they contain some actual layouts from my consulting.

# Topics

## Part 1

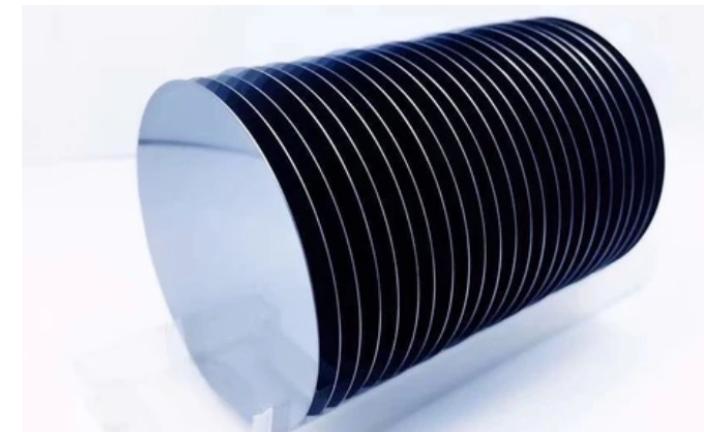
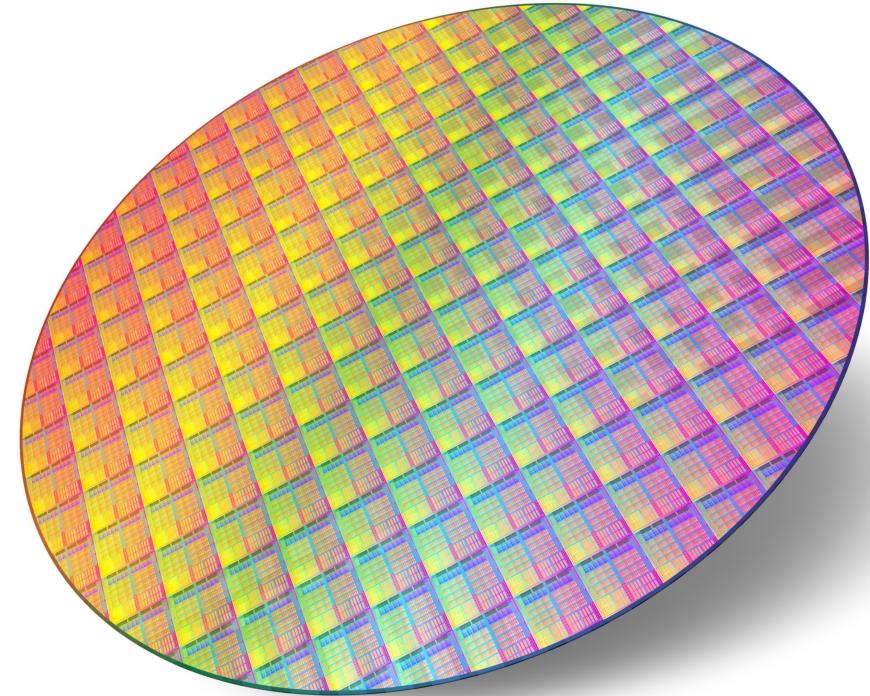
- Analog Layout
  - Matching
  - Common centroid
  - Symmetry
  - Unit Cell Concept
- Current Source
- Differential Pair

## Part 2

- Capacitors
  - Love your MOM
- Resistors
- More arrays
- Inductors
- Electromigration Rules
- Antenna Rules
- Examples:
  - DAC
  - Op-Amp
  - Tiles of blocks
  - Pin placement
- Layout Hierarchy vs Schematic

# Matching

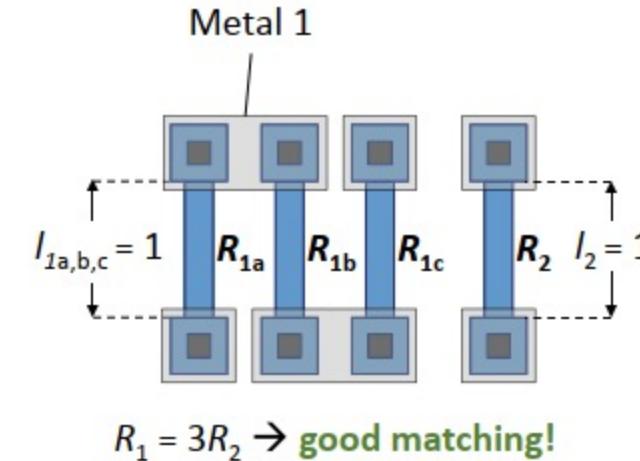
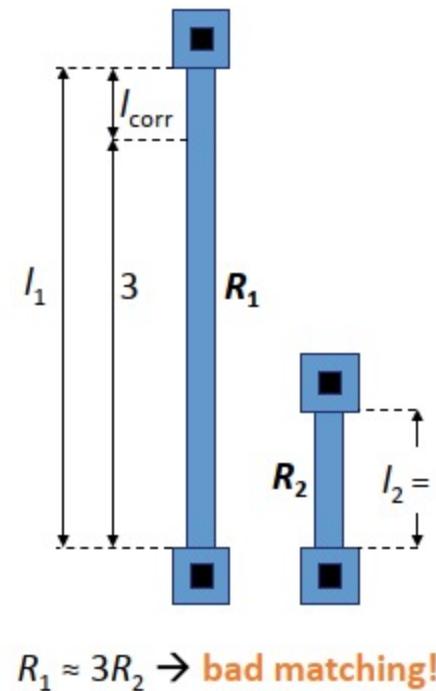
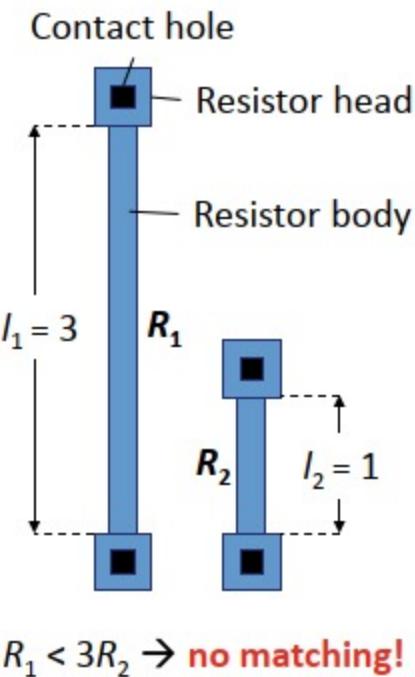
- Systematic Offsets – can usually fix with good layout
- Wafer-to-wafer variations, lot-to-lot, die-to-die, and transistor-to-transistor !
- Process Induced “Global” Mismatch – all transistors experience the same variations (say  $V_{TH}$ ), similar to “corner sims”
  - FF, FS, SF, SS, TT
- Process Induced “Local” Mismatch
  - Even two transistors on the same die sitting next to each other will be slightly different !
  - $V_{TH}$  mismatch



# Mismatch Between Transistors

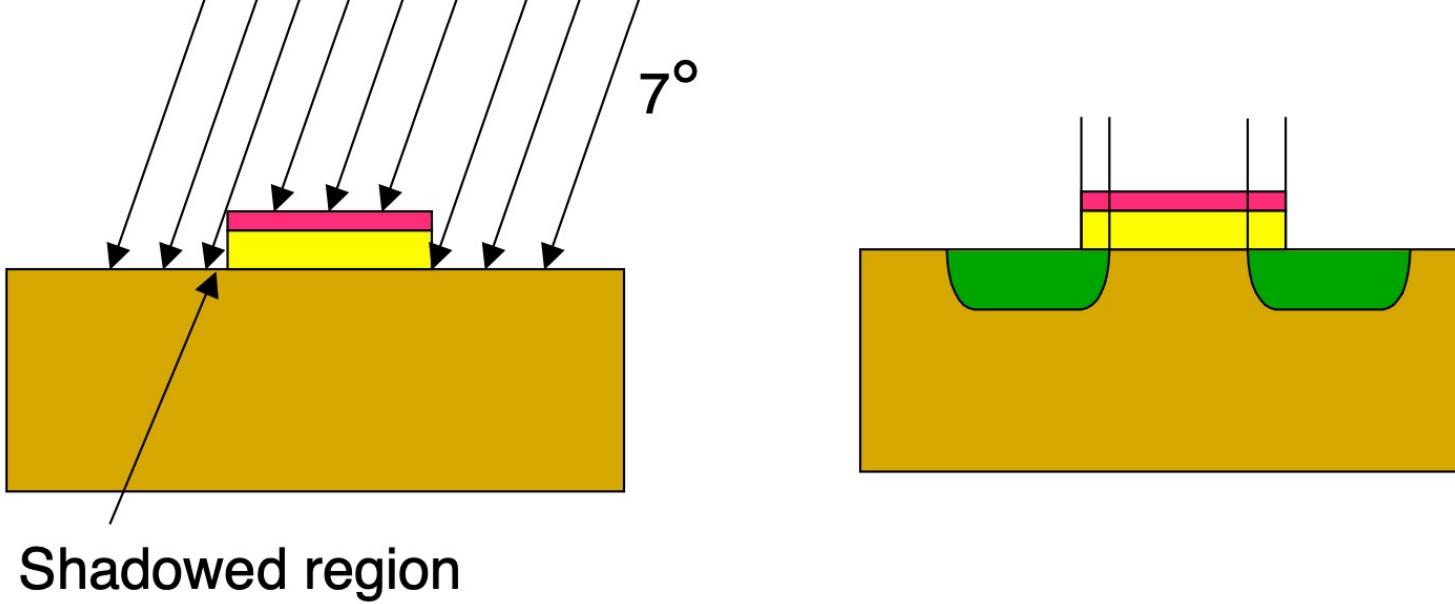
- Systematic offset (metal on top or different environment)
  - Eliminate this component as much as possible
- Geometry
  - Line edge roughness
  - Avoid minimum sized structures
- Doping
  - Threshold voltage implant
  - Avoid small transistors
  - A given W/L can be realized with a non-minimum sized device

# Matching: Resistors Example



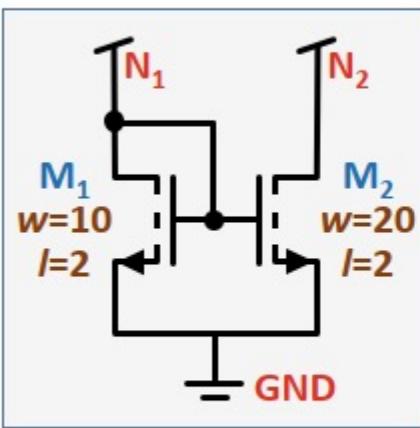
Lienig, Scheible, "Fundamentals of Layout Design for Electronic Circuits," Springer 2020.

# Asymmetry due to Fabrication

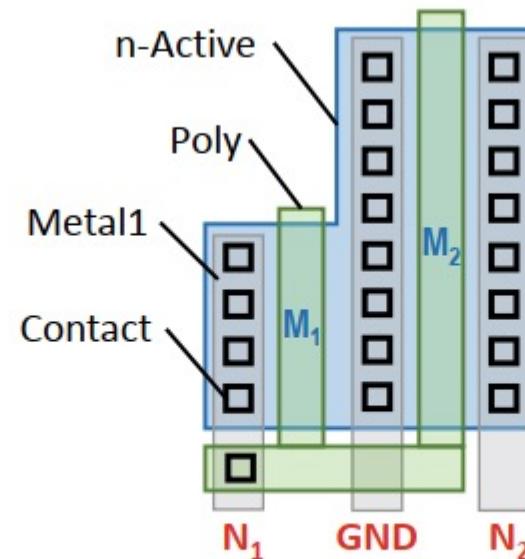


- Implants are tilted by 7 degrees
- Source and Drain are not Symmetric !

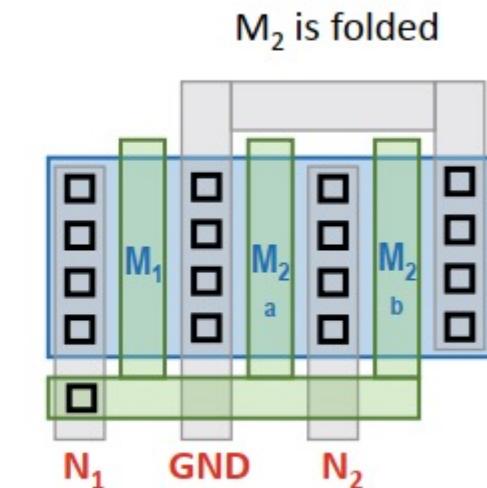
# Matching: MOS Current Mirror



Current mirror circuit



No splitting → **bad matching!**



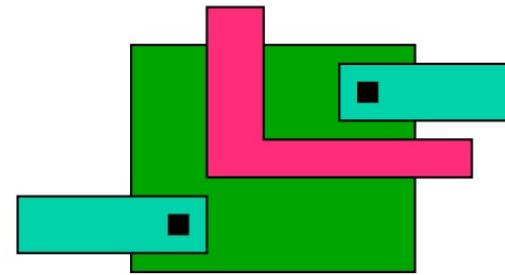
M<sub>2</sub> is folded

Splitting → **good matching!**  
Better matching

Lienig, Scheible, "Fundamentals of Layout Design for Electronic Circuits," Springer 2020.

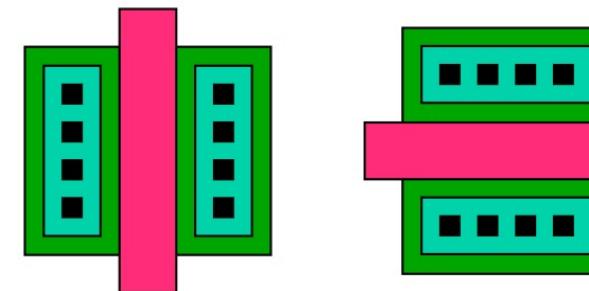
# Multiple Single Transistors

- Use Regular Rectangular Shapes

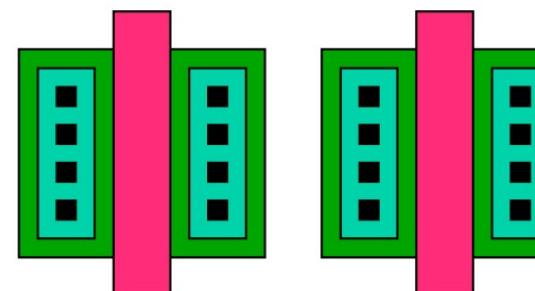


- Use Parallel Elements

- Direction matters with Silicon



- Try to have current flow in the same direction

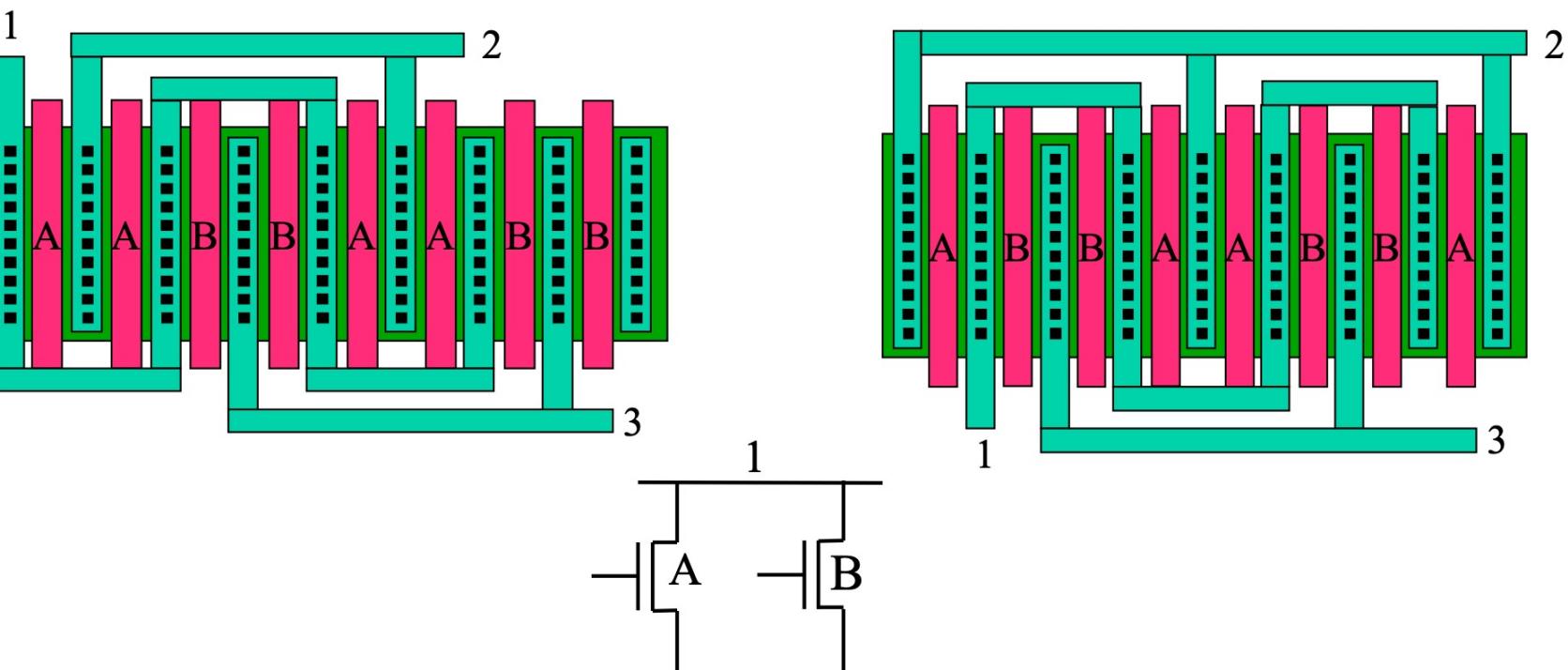


# Layout Patterns

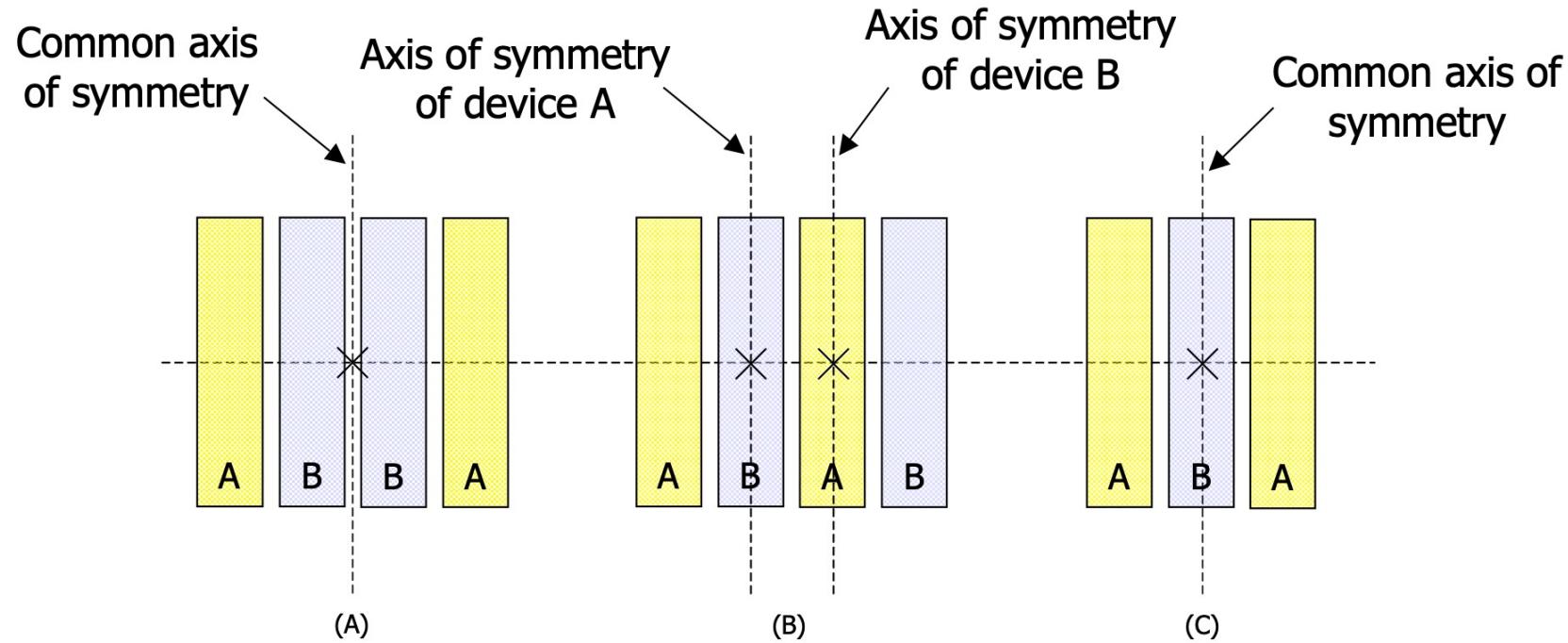
- Translational Symmetry versus Reflection
- Beautiful is not always better !
- When matching, keep everything the same as much as possible
- Assume up and down and left and right are asymmetric !

# Interdigitated Devices

- Two matched transistors are laid out together using one big interdigitated transistor
- Which pattern is better ? AABBAABB or ABBAABBA?

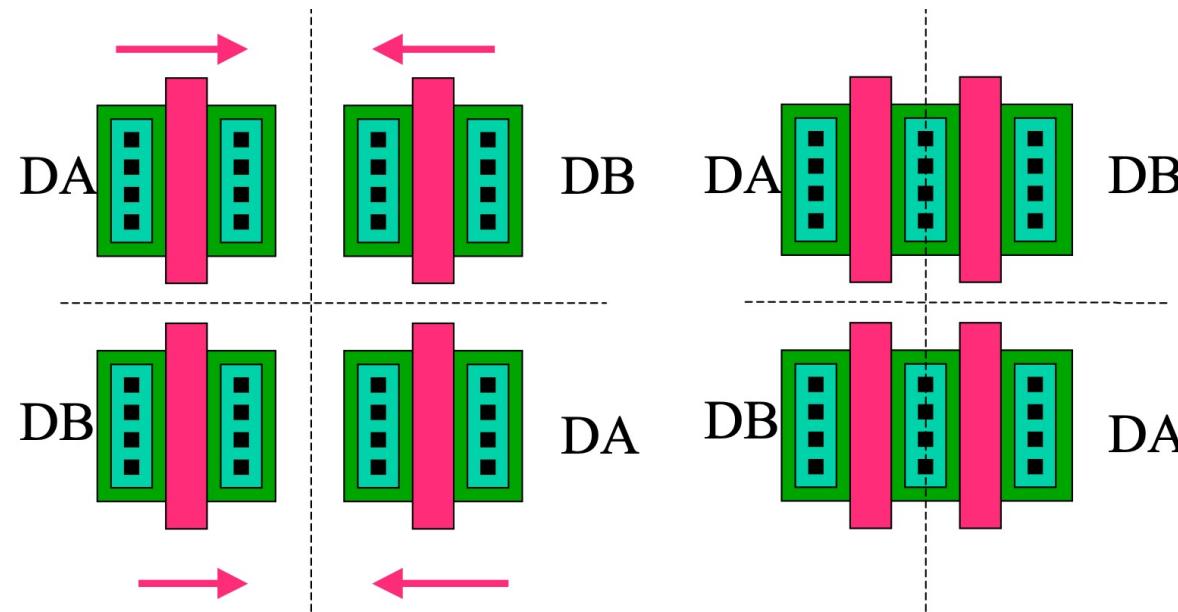


# Axis of Symmetry



## Common Centroid

- Using a common-centroid layout style, we can cancel gradients (to first order)

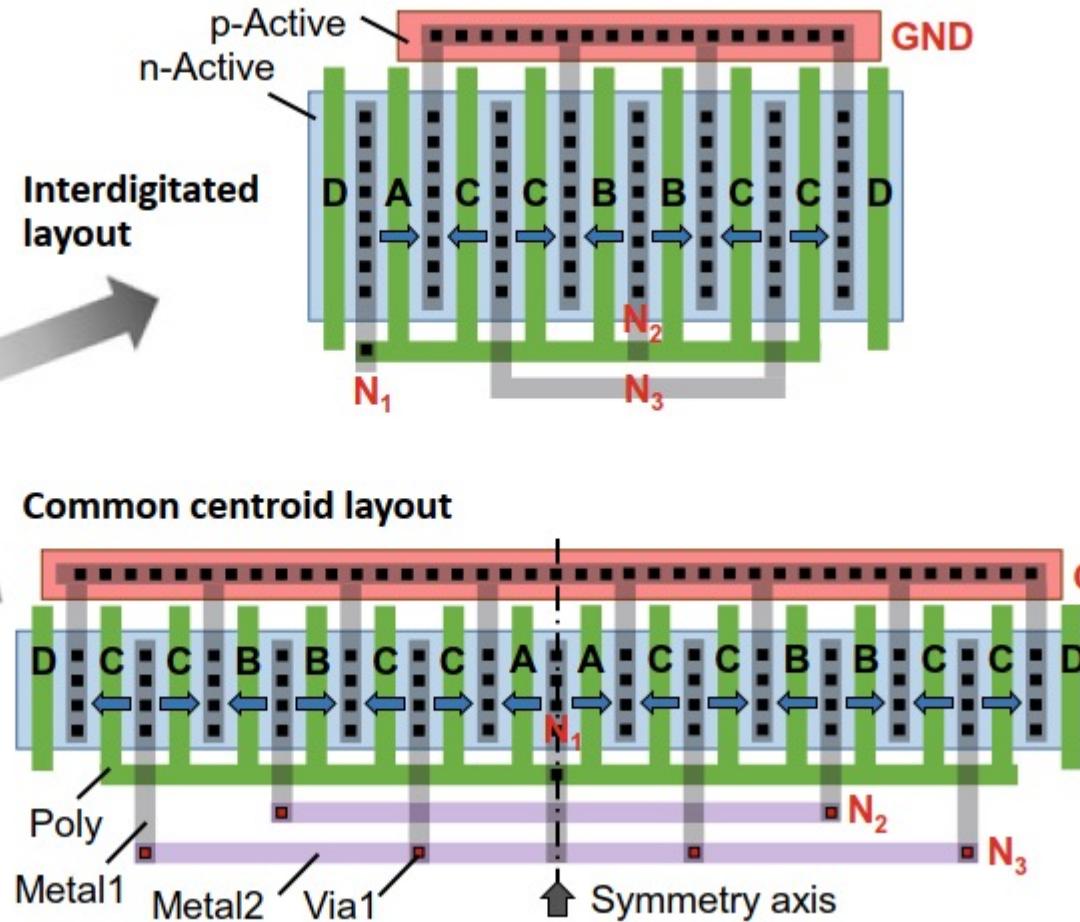
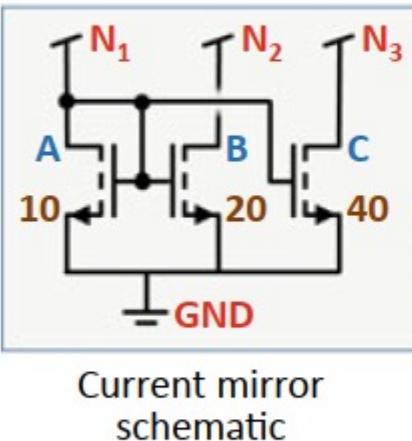


# Interdigitation Patterns

- Can also stack transistors in vertical direction

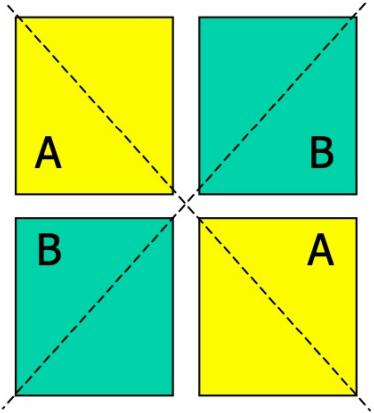
A	AA	AAA	AAAA
AB*	ABBA	ABBAAB*	ABABBABA
ABC*	ABCCBA	ABCBCACBCA*	ABCABCCBACBA
ABCD*	ABCDDCBA	ABCBCADBCDA*	ABCDDCBAABCDDCBA
ABA	ABAABA	ABAABAABA	ABAABAABAABA
ABABA	ABABAABABA	ABABAABABAABABA	ABABAABABAABABAABABA
AABA*	AABAABAA	AABAAABAAABA*	AABAABAAAABAABAA
AABAA	AABAAAABAA	AABAAAABAAAABAA	AABAAAABAAAABAAAABAA

# Common Centroid

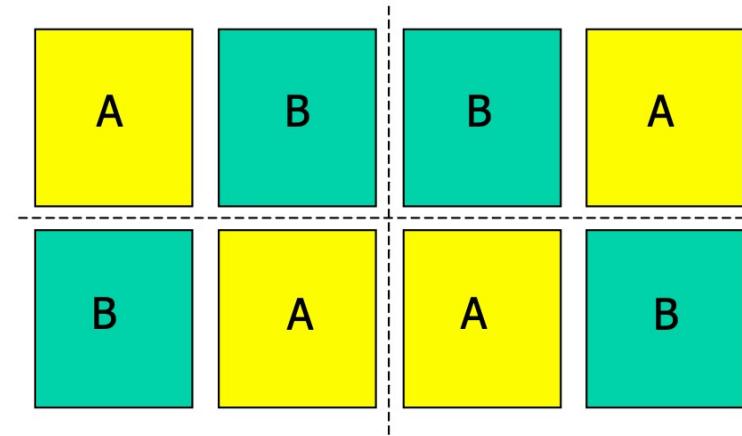


Lienig, Scheible, "Fundamentals of Layout Design for Electronic Circuits," Springer 2020.

# Common Centroid Arrays



Cross coupling

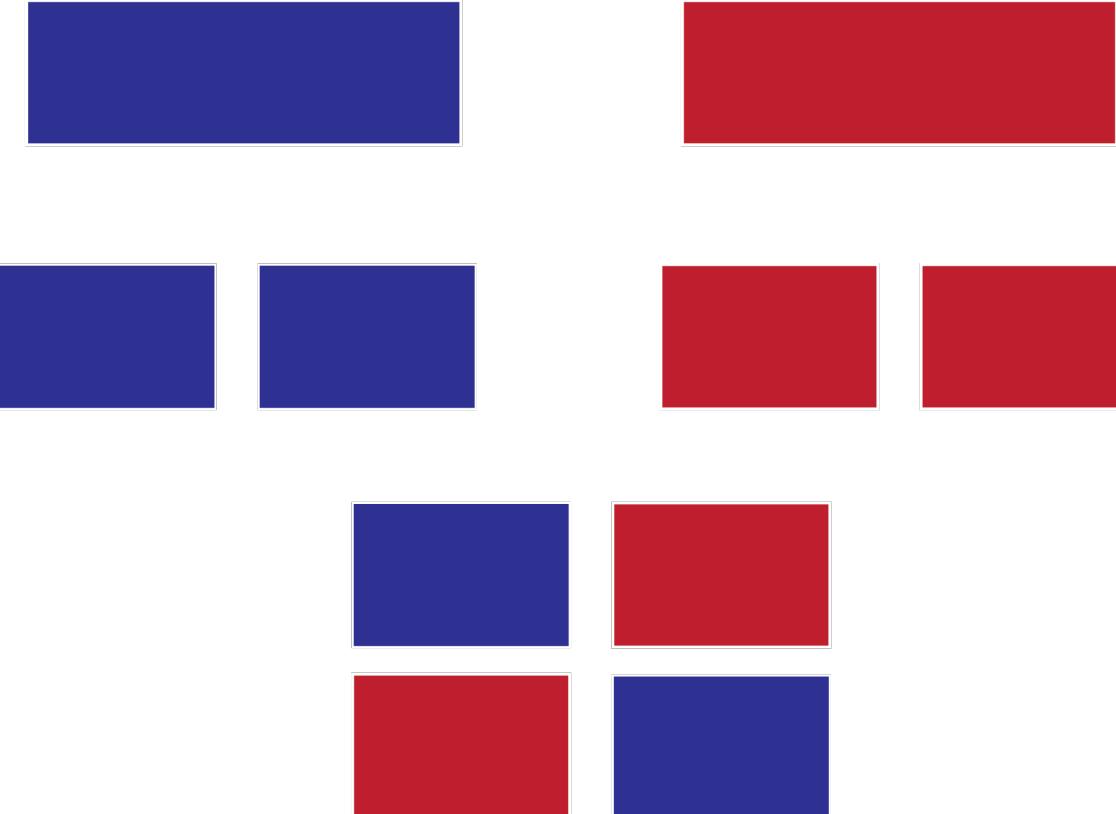


Tiling (more sensitive to high-order gradients)

# Common Centroid Patterns

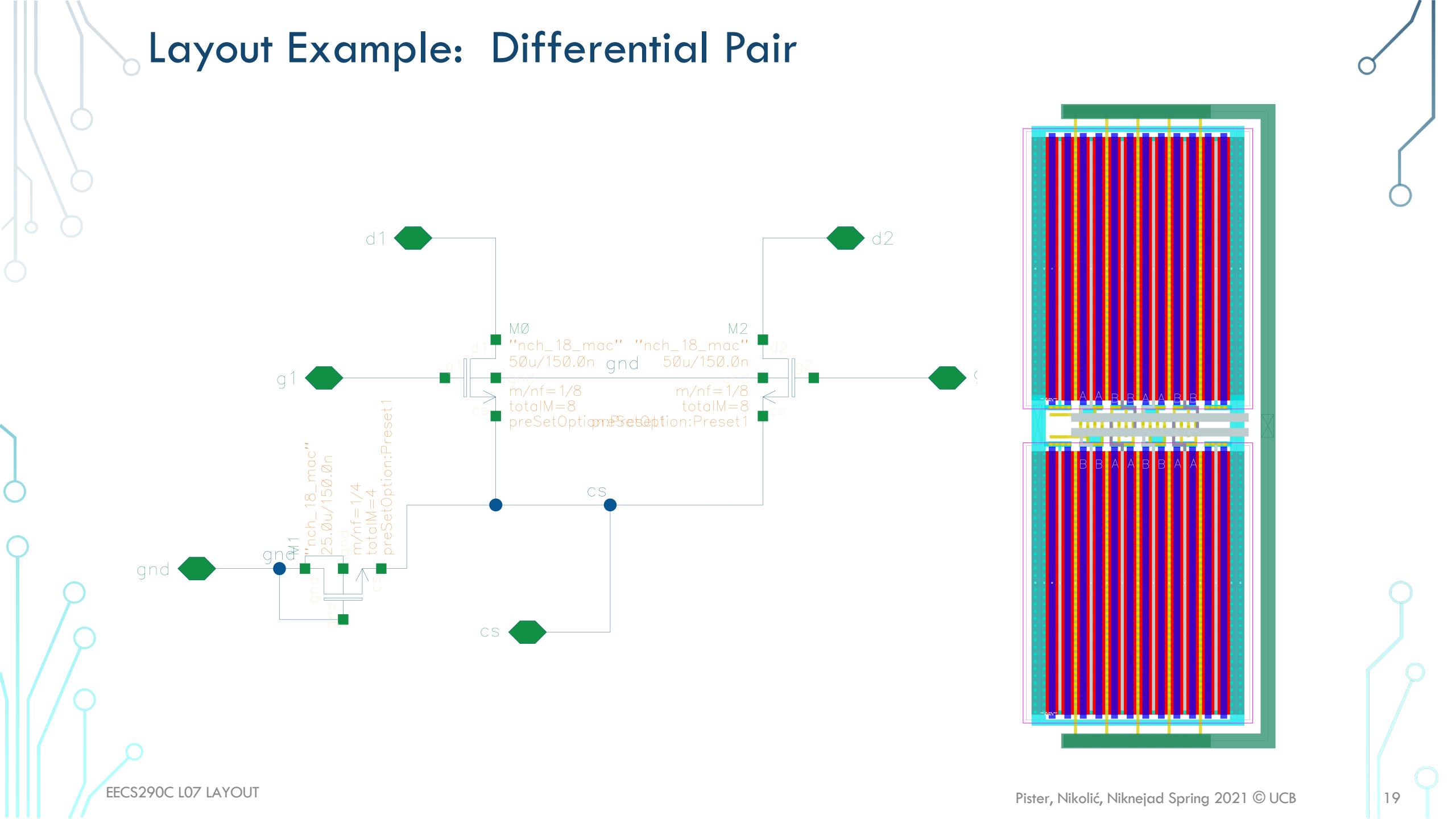
ABBA	ABBAABBA	ABBAABBA	ABBAABBA
BAAB	BAABBAAB	BAABBAAB	BAABBAAB
ABA	ABAABA	ABAABA	ABAABAABA
BAB	BABBAB	BABBAB	BABBABAB
ABCCBA	ABCCBAABC	ABCCBAABC	ABCCBAABC
CBAABC	CBAABCCBA	CBAABCCBA	CBAABCCBA
AAB	AABBAAA	AABBAAA	AABBAAA
BAA	BAAAAB	BAAAAB	BAAAAB
		AABBAA	AABBAA
		BAAAAB	BAAAAB
		AABBAA	AABBAA
		BAAAAB	BAAAAB
		AABBAA	AABBAA

## Example: Differential Pair



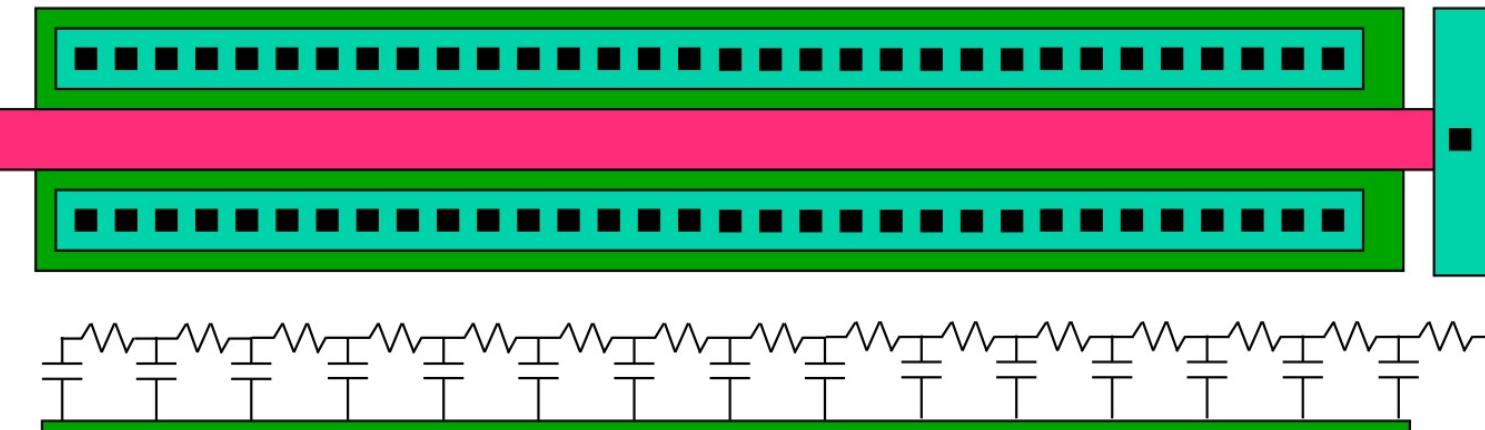
- Split transistors into smaller units
- Layout units in an array in a common centroid fashion
- Impossible to keep metal interconnections symmetric, so make sure IR drops are not an issue

# Layout Example: Differential Pair

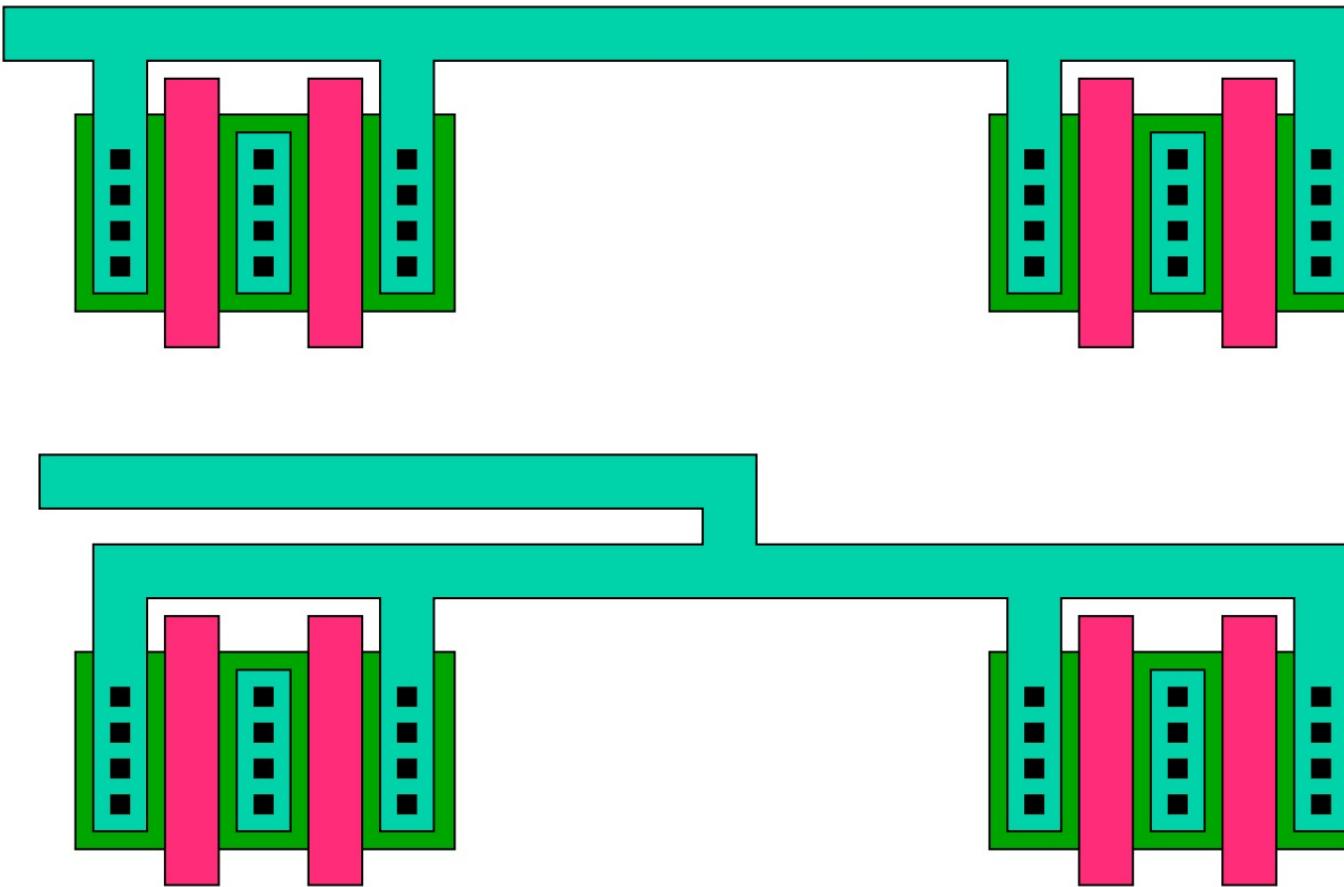


# Don't forget about “IR” Drops

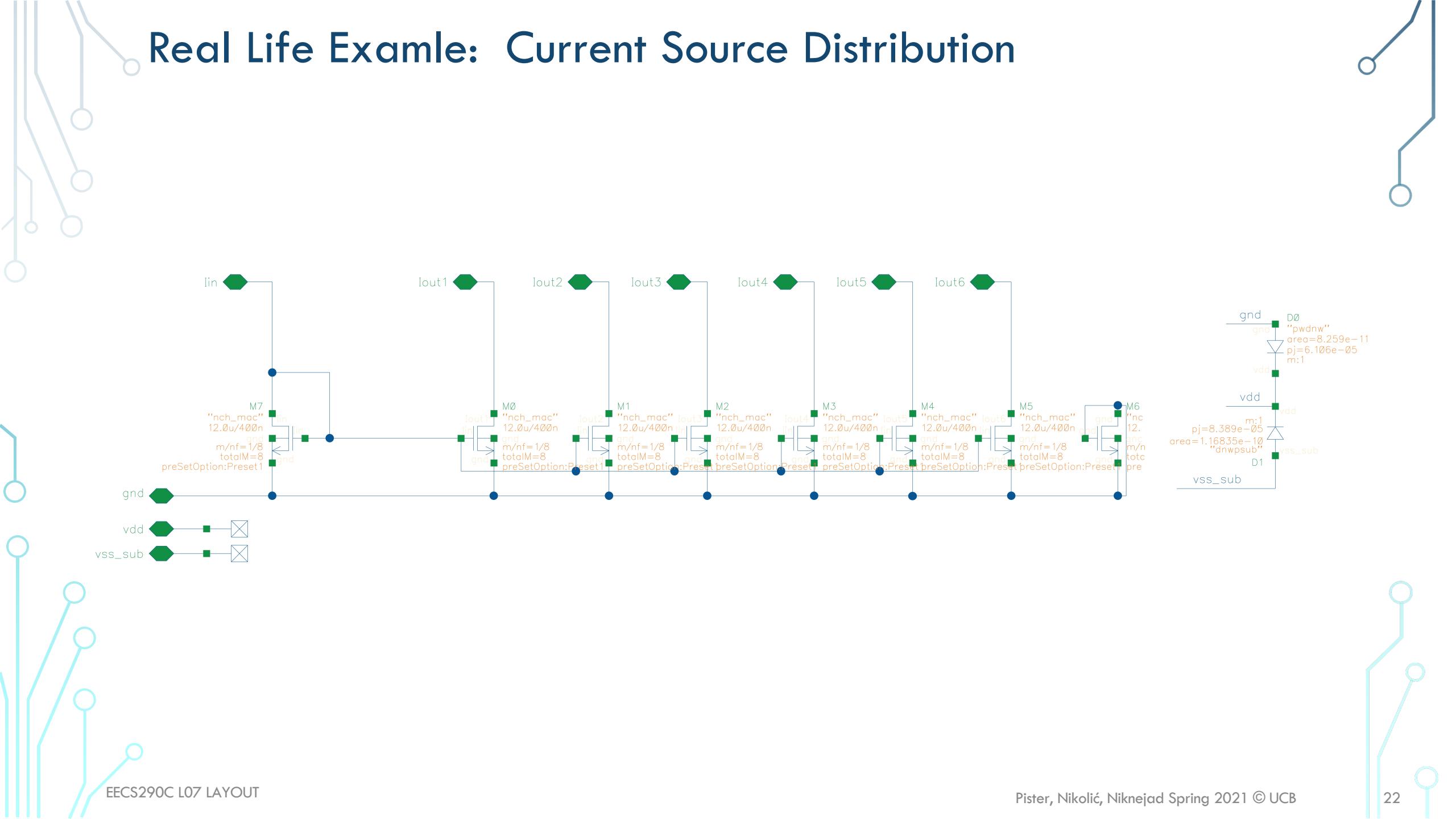
- Metals have considerable resistance since they are very narrow
- Poly is especially resistive
- Vias are tiny and introduce 10's of ohms of resistance ... use many in parallel
- Make sure  $I^*R$  drops are small and for differential circuits, make it symmetric so it's a common mode offset
- Use more fingers



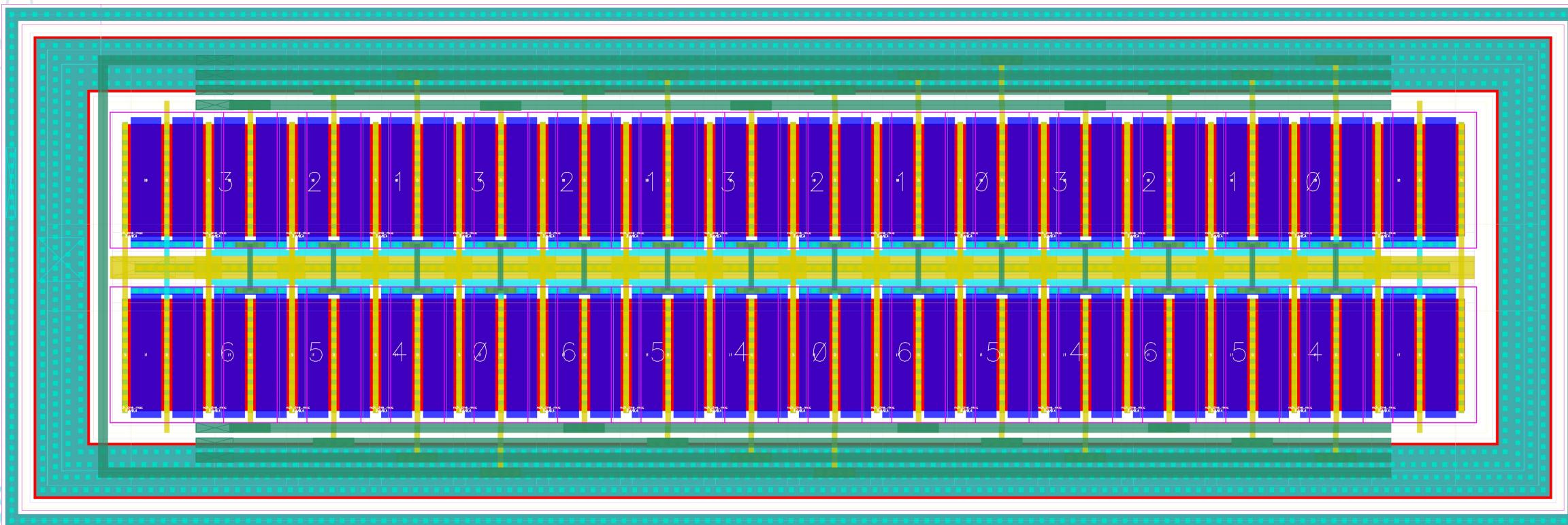
# Matched Interconnect



# Real Life Example: Current Source Distribution

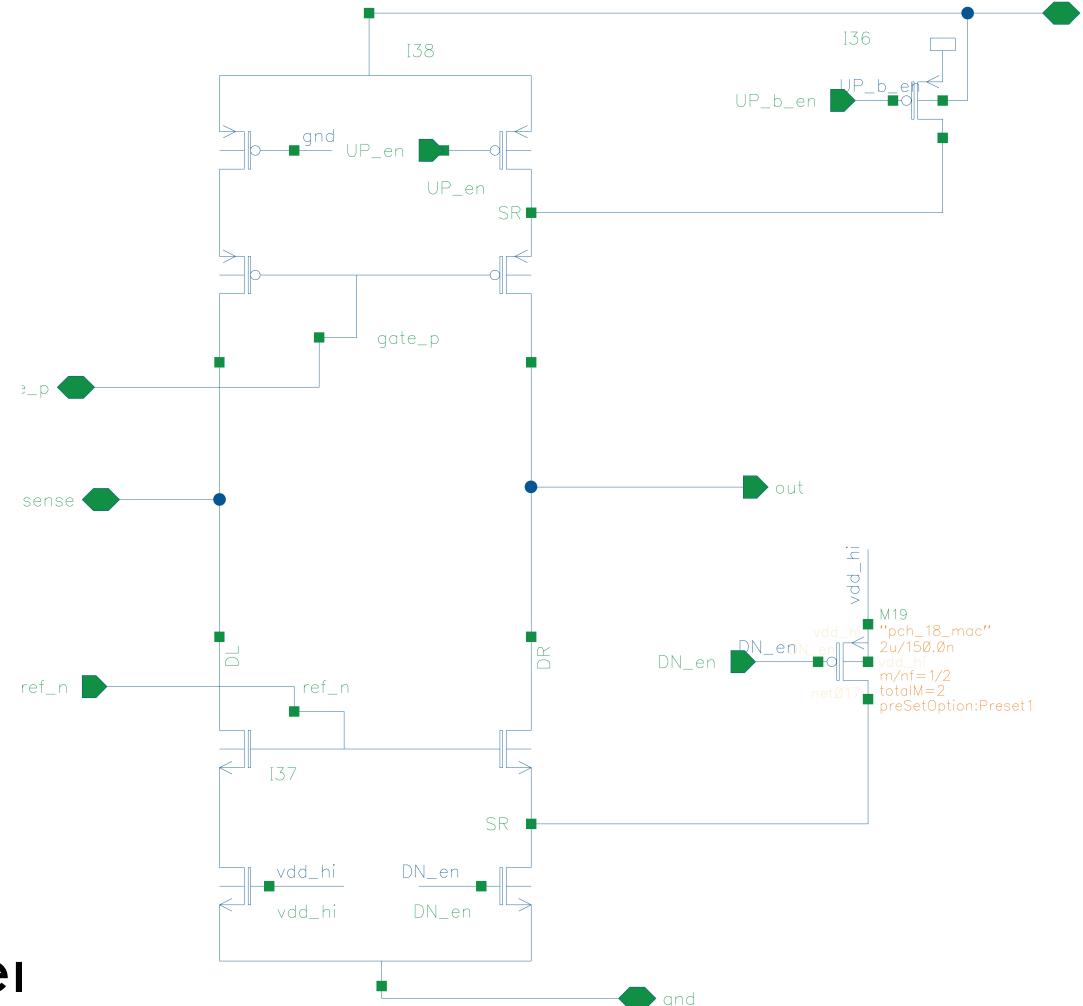


# Multi-Output Current Source Layout

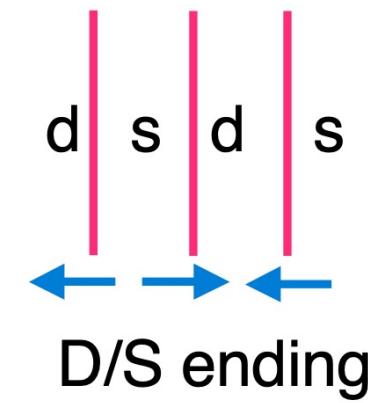
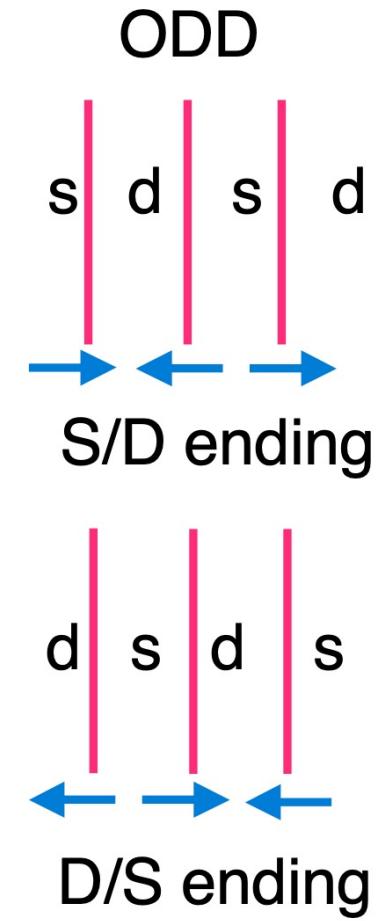
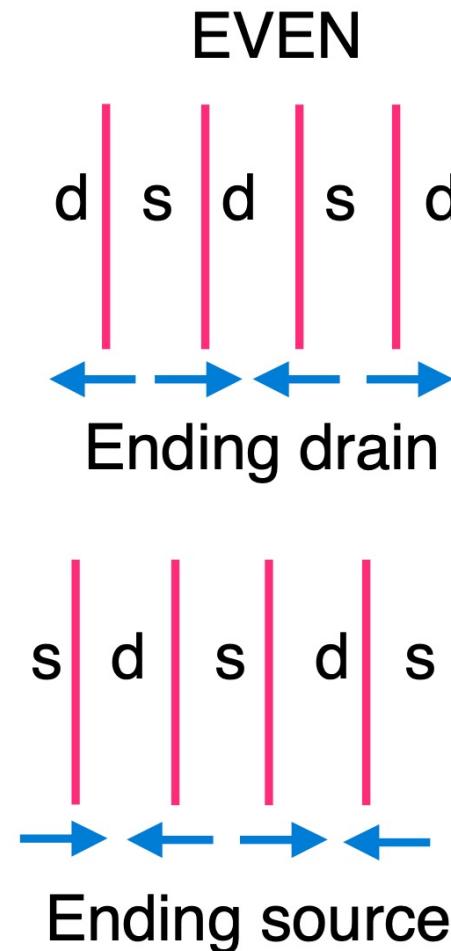
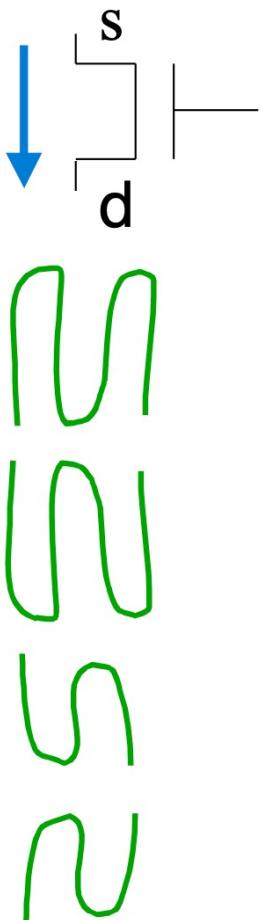


# Stacked and Parallel Layout

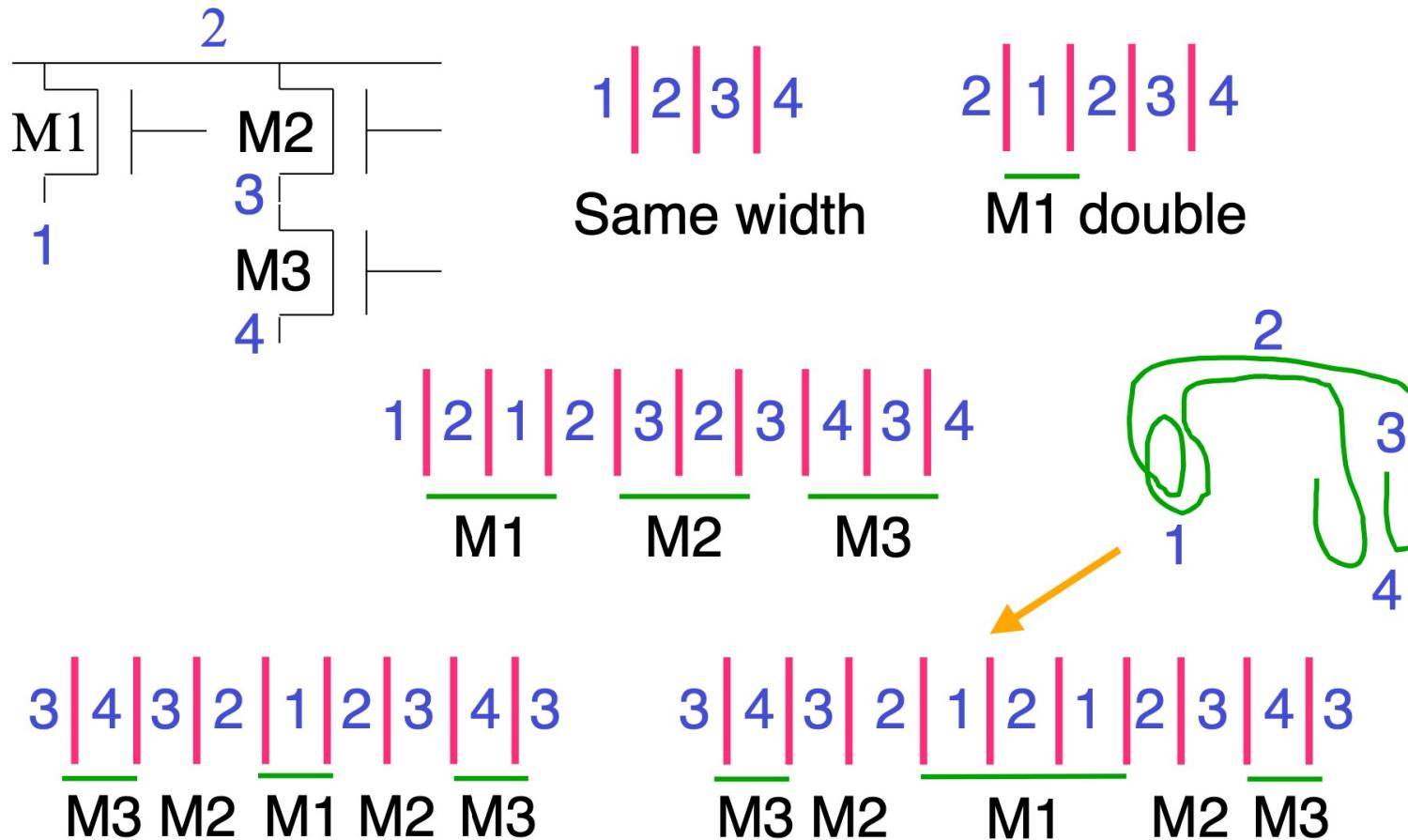
- When transistors are connected and share a junction (think of a current source, or a cascode connection), we can group them together
- To do this, the transistors need to have the same finger width
  - Modify your schematic so that all transistors are a multiple of a unit finger width, say 0.5um or 1um
- In the schematic, identify which transistors can be grouped together



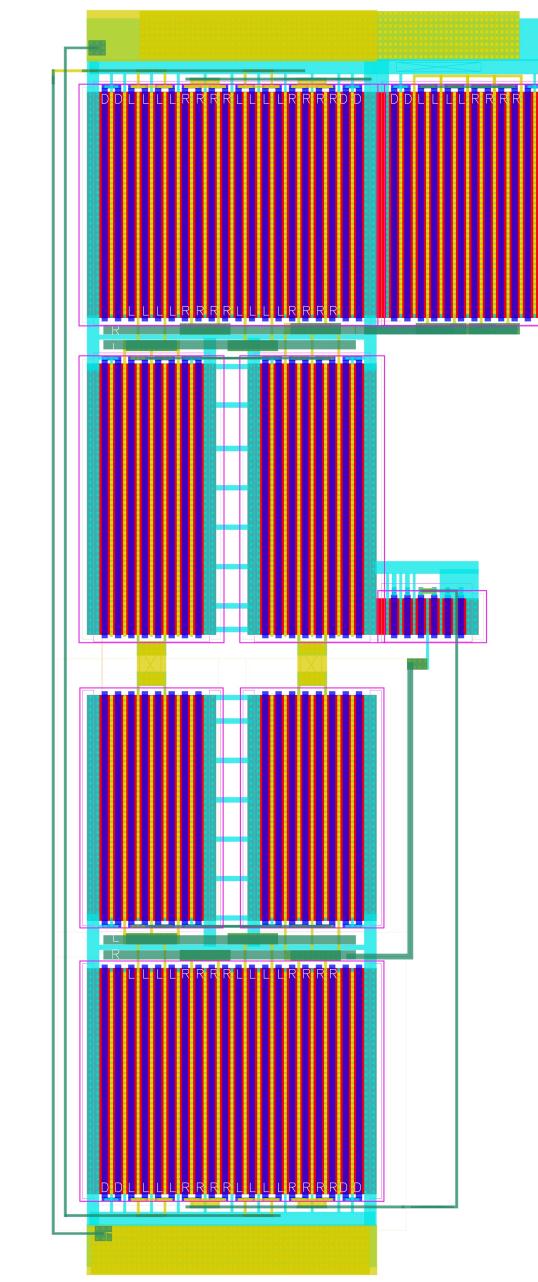
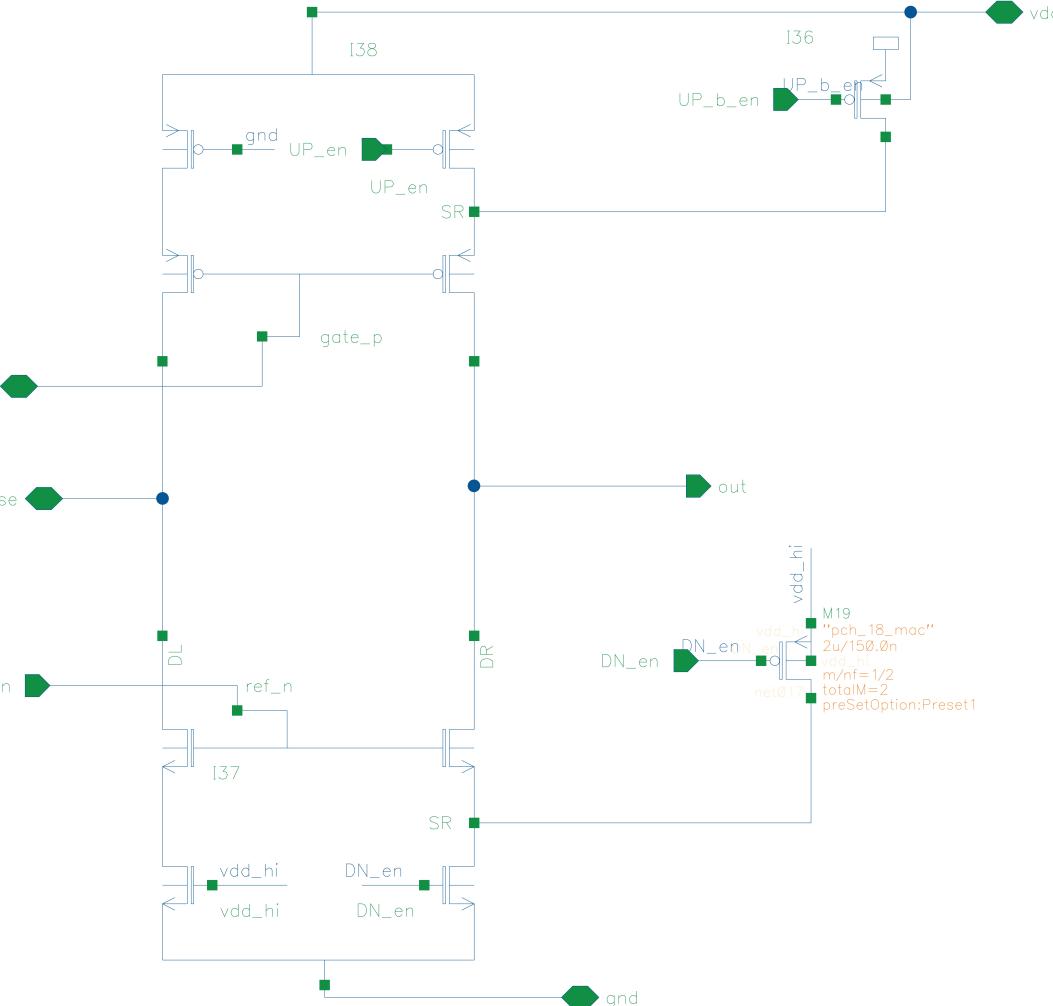
# Odd/Even Fingers and S/D Swaps



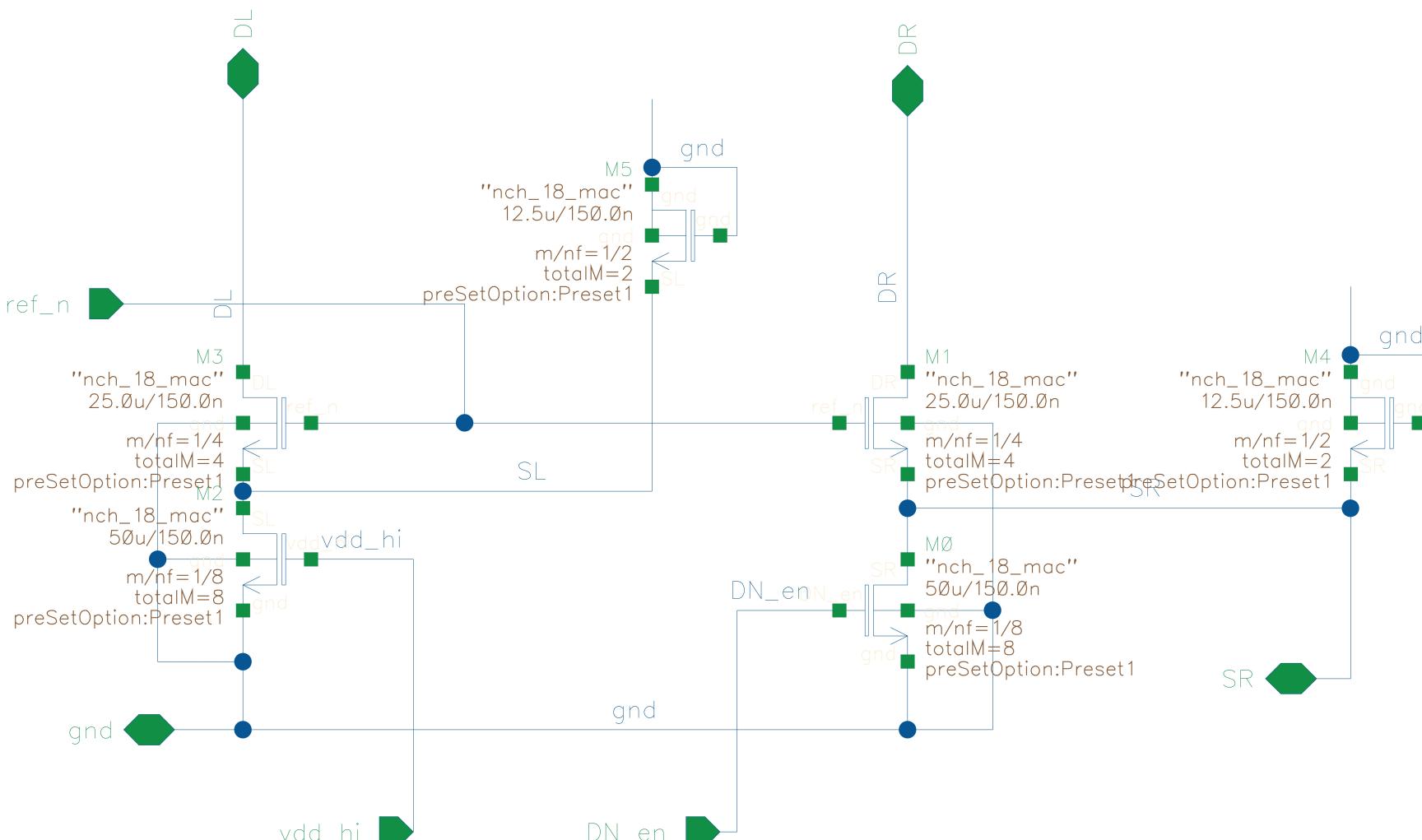
# Example of Stacking and Grouping



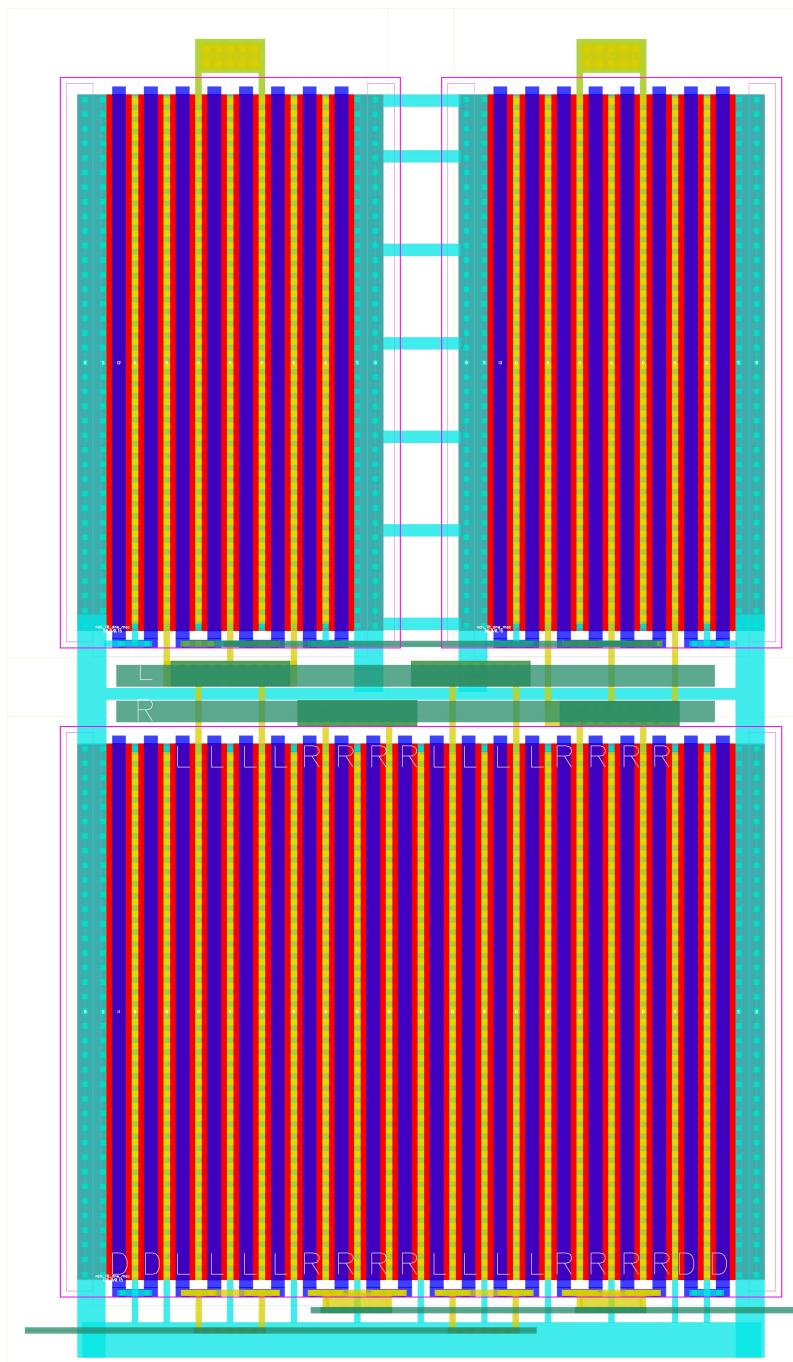
# Layout Example: Charge Pump



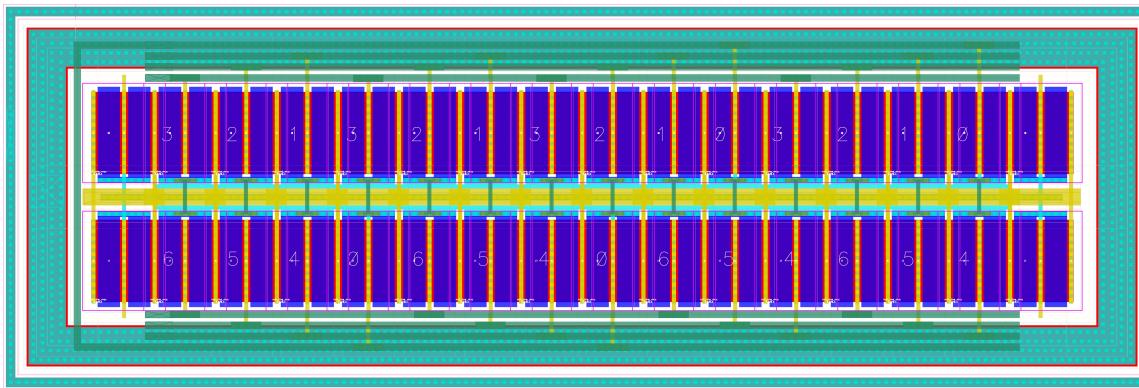
# Charge Pump: N-Mirror



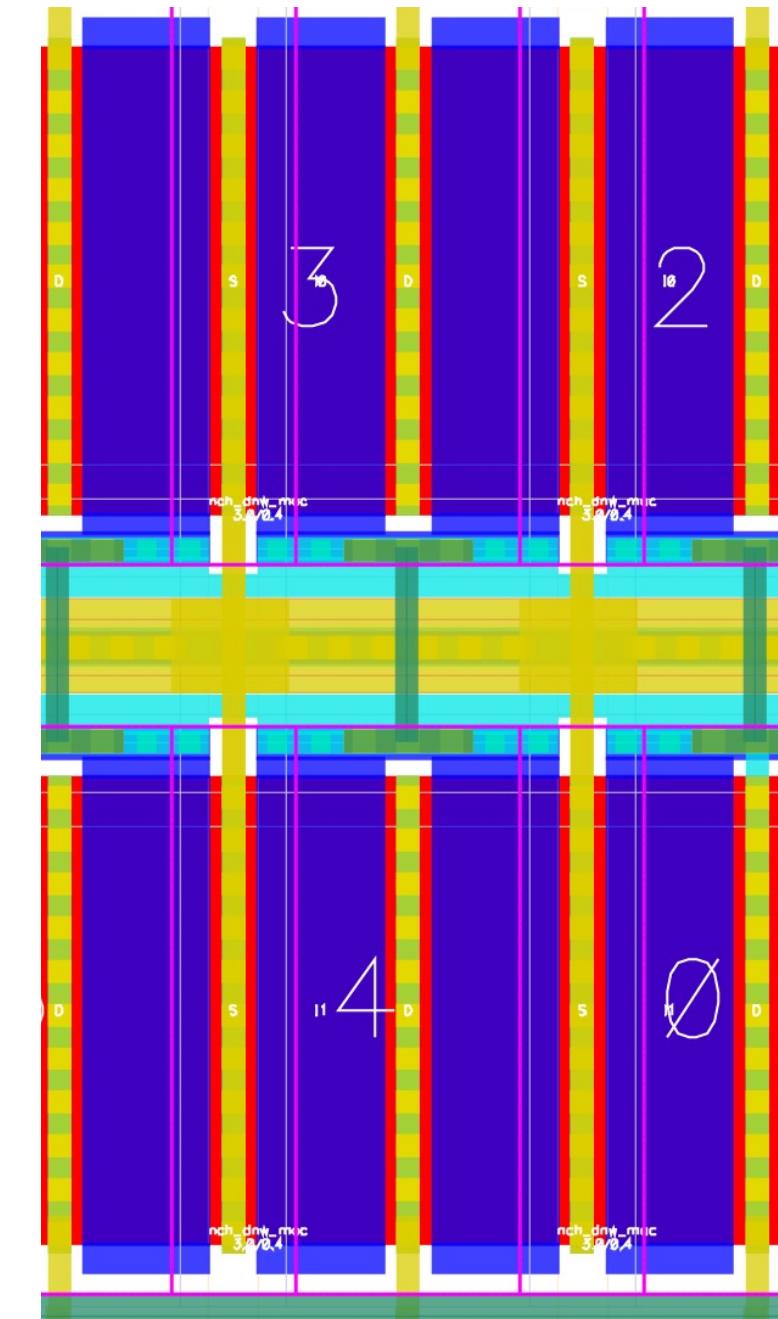
# N-Mirror Layout



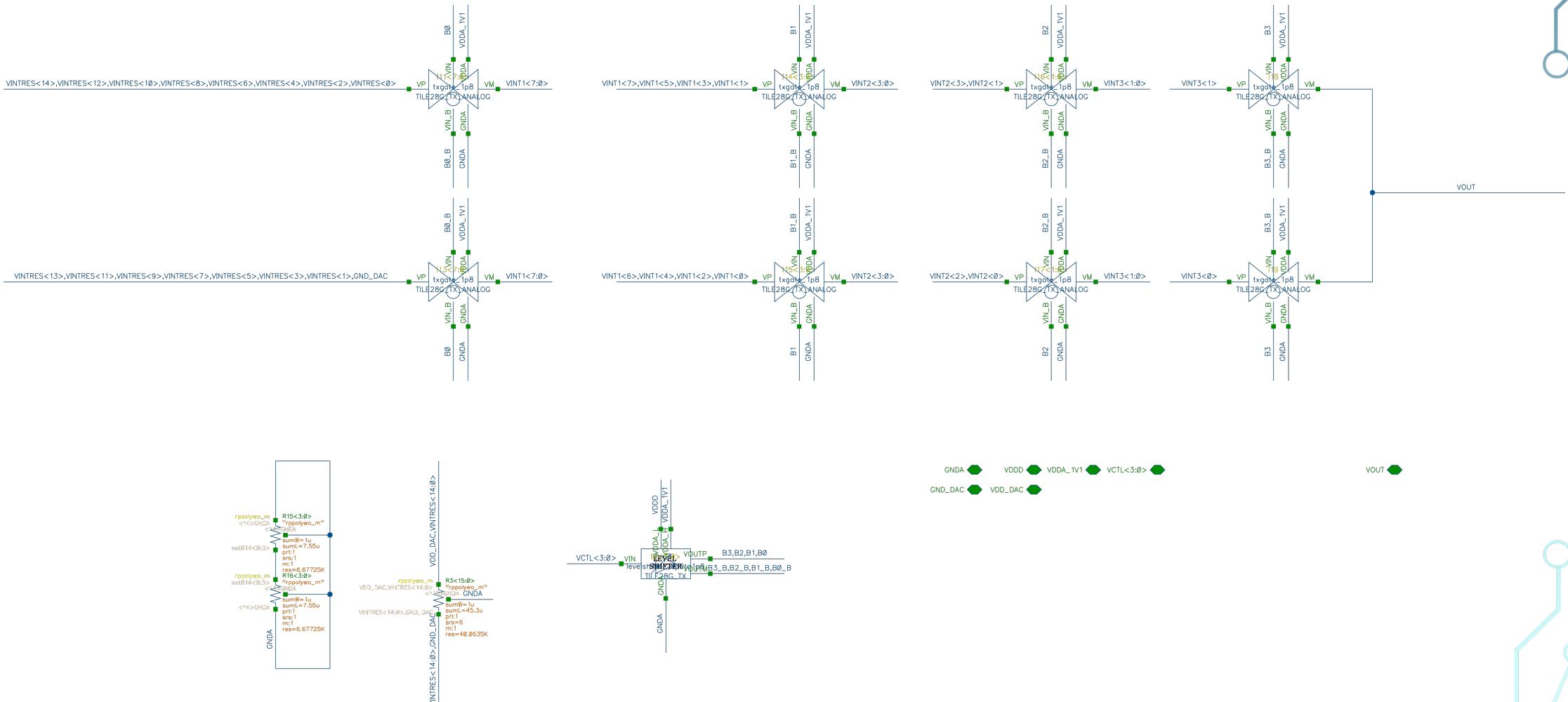
# Make a “Unit Cell”



- In analog layout, the best results (not the most area efficient) are obtained by using a “unit cell” and replicating the unit cell
- Make the “environment” around the cell as matched as possible

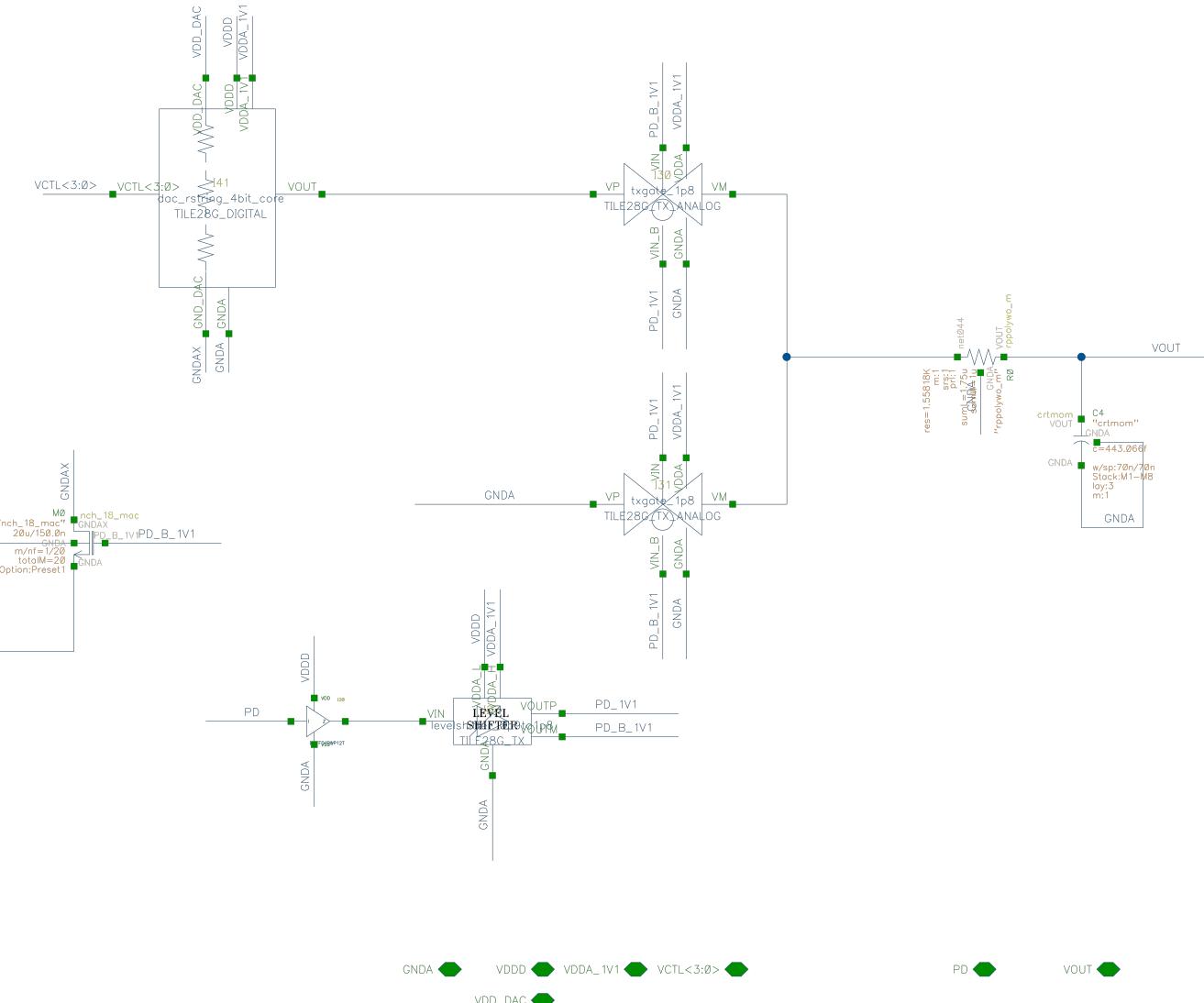


# Example: R-DAC Design

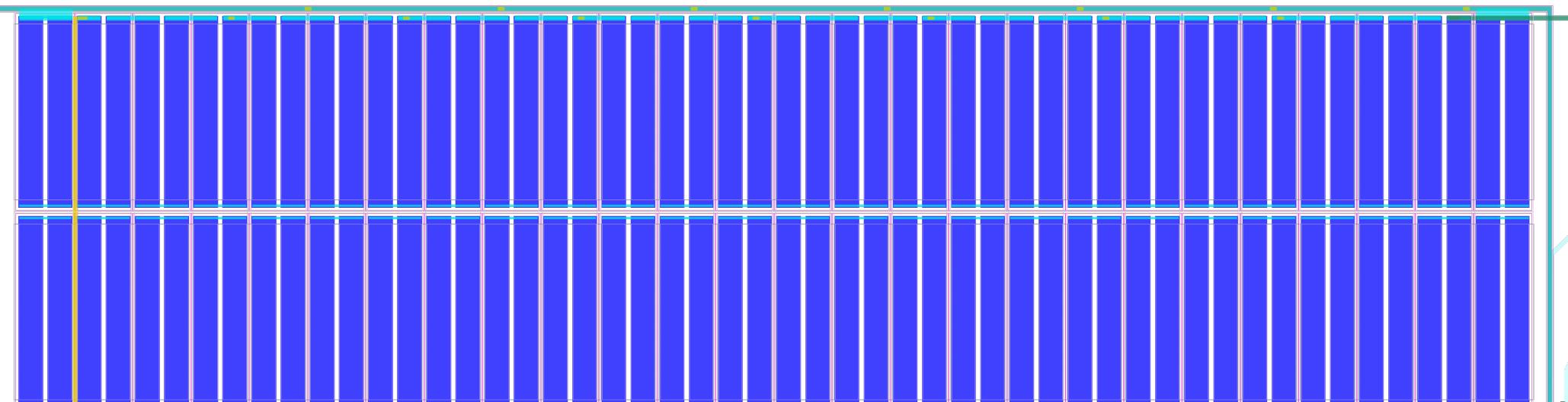
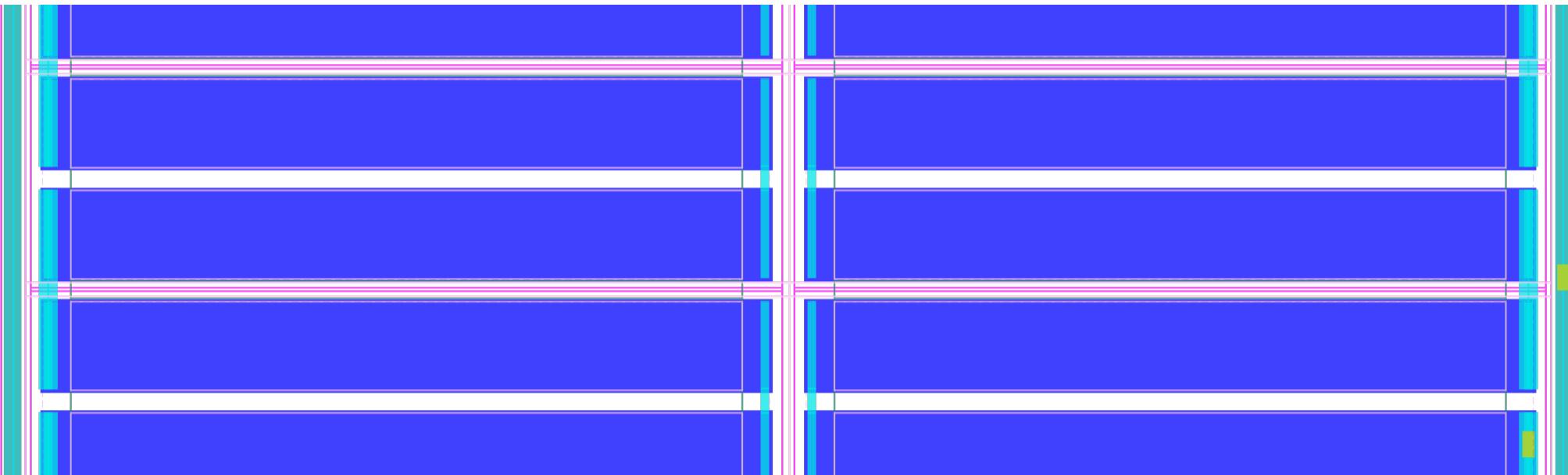


# Aside on Good Design Practices

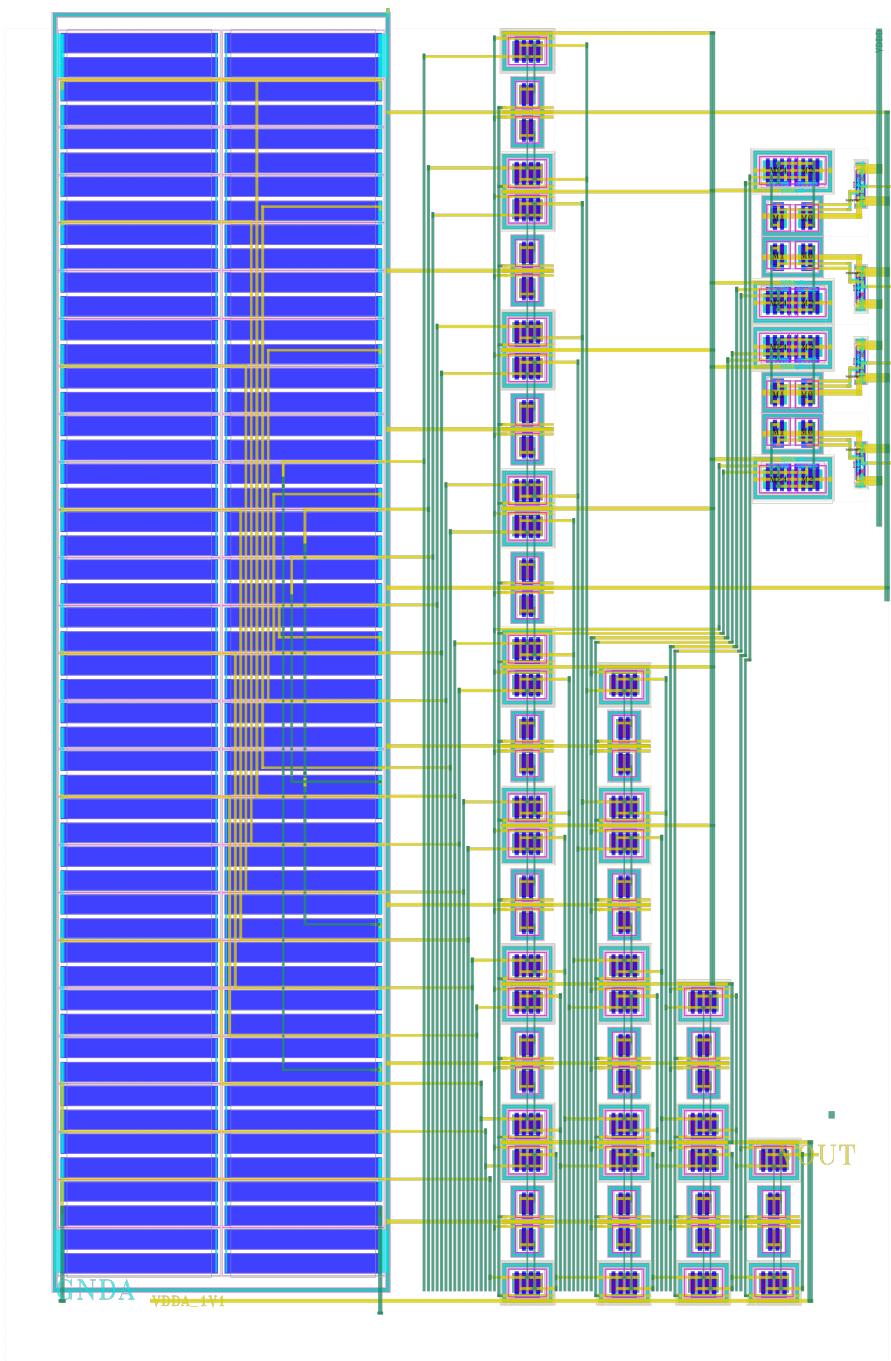
- Make an enable and make sure current is zero when “off”
- Make a default value of zero or vdd when off
- Put a filter on the output of this DAC (good practice since it’s a DC)
- Make sure you don’t drive any loads except gates since there’s no output buffer



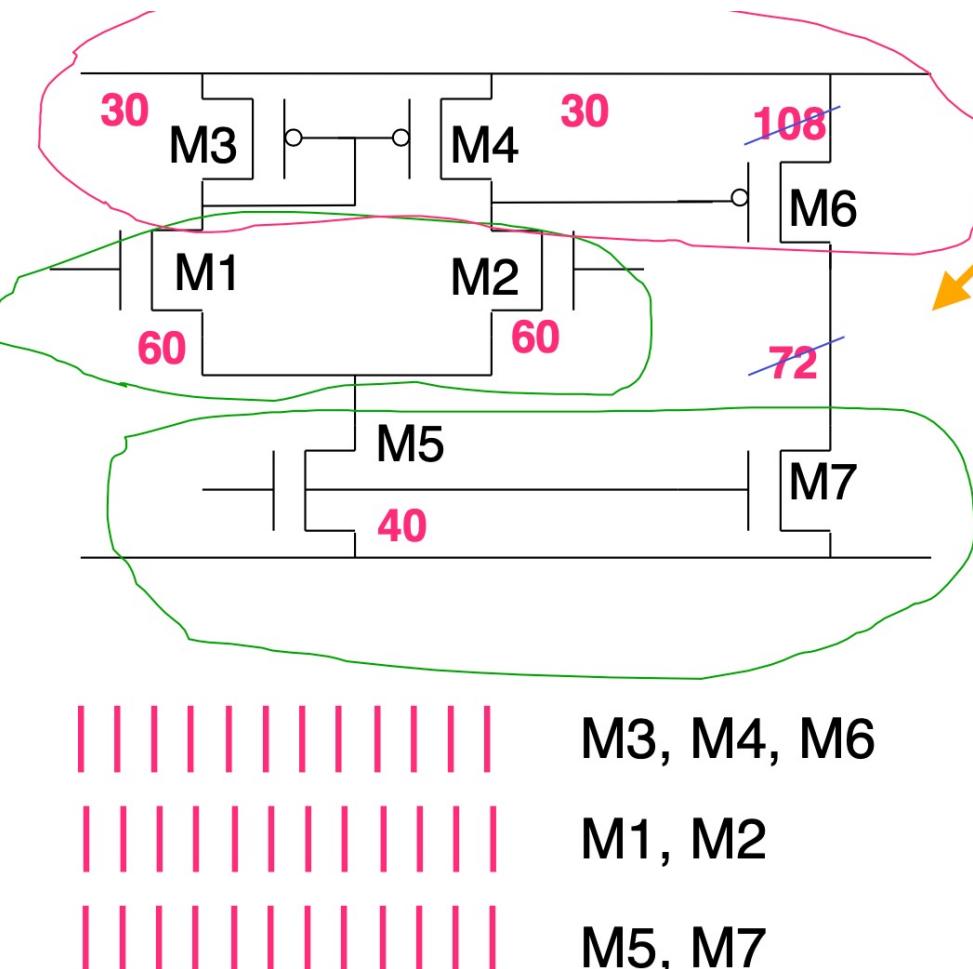
# Layout – Unit Cell Concept



# Top Level



## Example: OTA Layout



Only width matters

Possible stacks:  
1 p-channel, 2 n-channel

change the size of M6 and M7  
to 80 and 120 respectively

Width of each finger?  
We want the same number of  
fingers per stack (k).

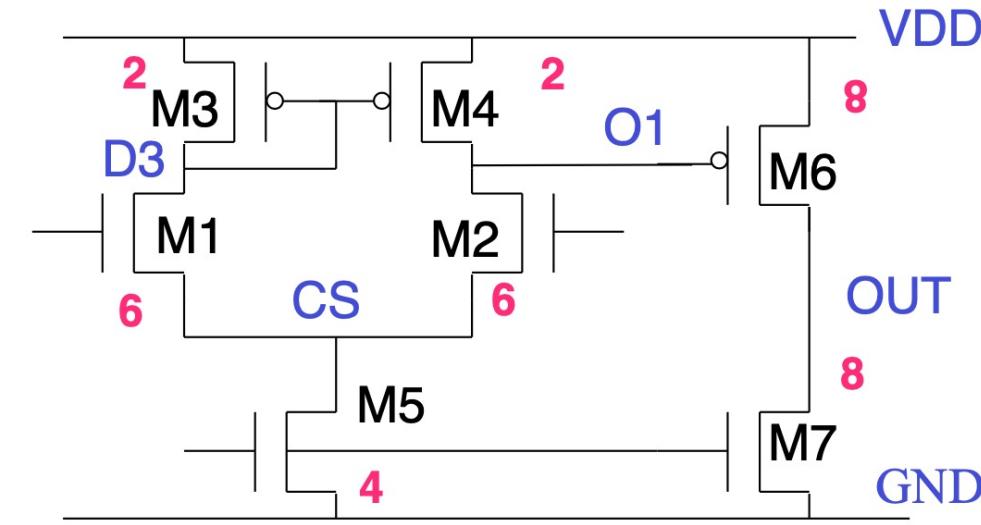
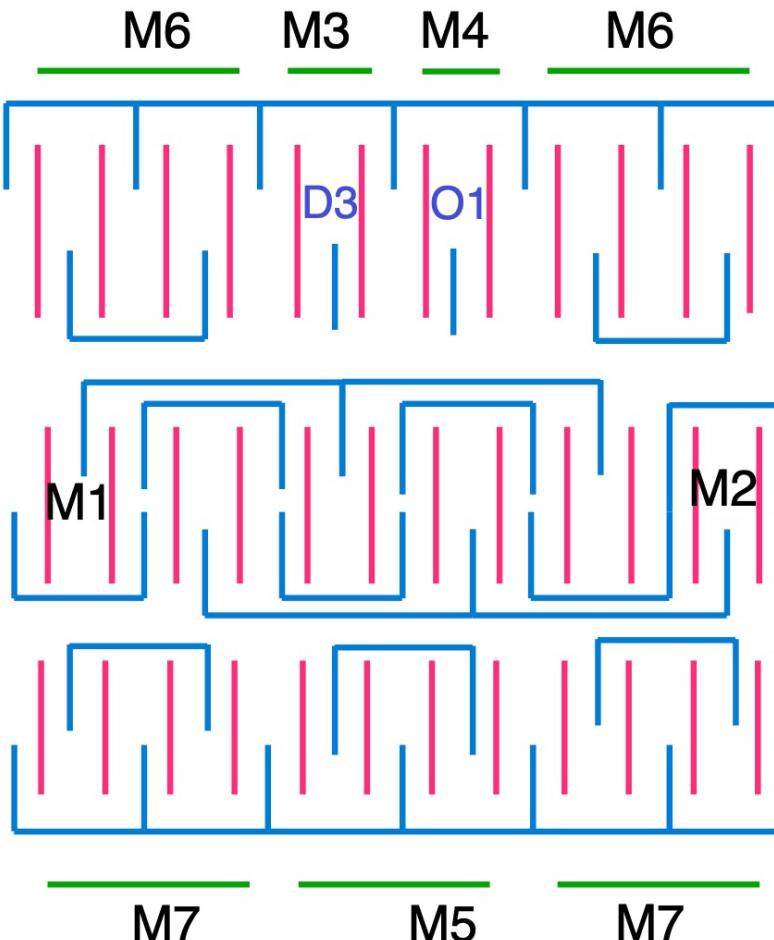
$$W_{p_1} = 180/k$$

$$W_{n_1} = 120/k$$

$$W_{n_2} = 120/k$$

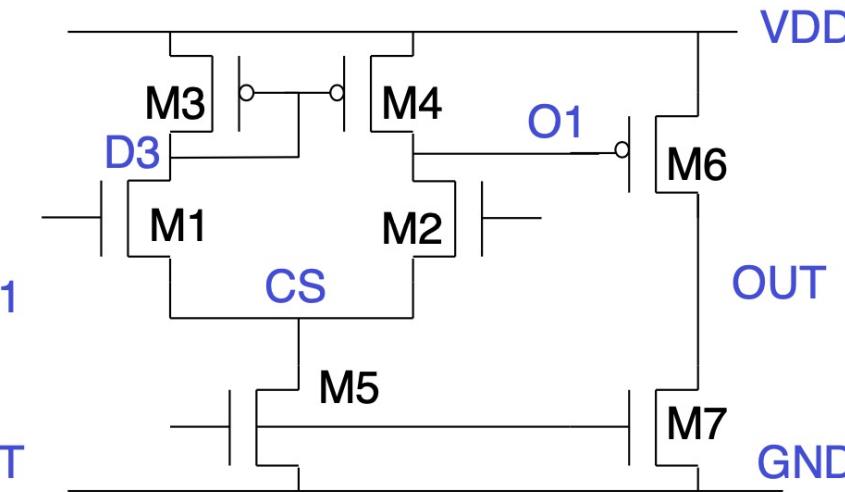
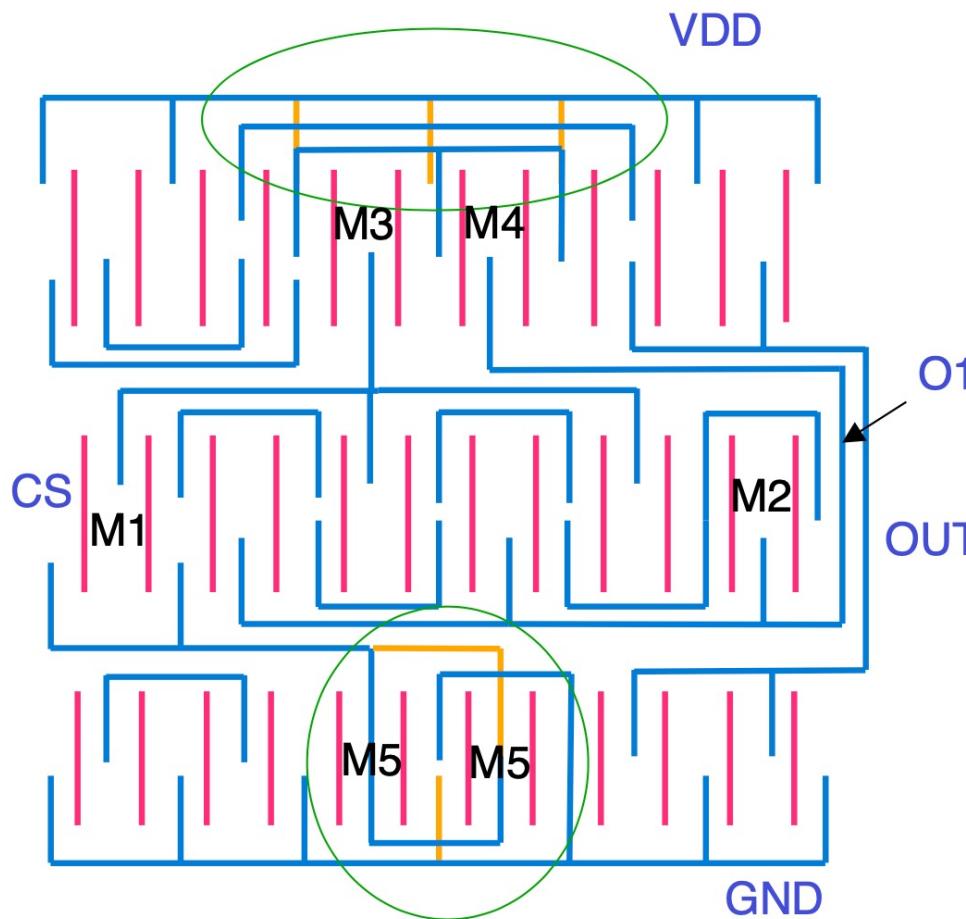
for M3 and M4 use 2 fingers

# OTA Layout Continued



First attempt of  
interconnections  
(not completed)

# Metal Interconnect



Use metal for carrying current!  
Poly connections are not a problem (usually)

# Layout of Passives



## Process Options

- Available for many processes
- Add features to “baseline process”
- E.g.
  - Capacitor option (MIM, 2 level poly, channel implant)
    - Note that “MOM” capacitors are free (no extra layers)
  - Low  $V_{TH}$  devices
  - “High voltage” devices (3.3V) [Always have thick oxide IO devices]
  - Flash
  - Silicide stop option
  - ...

# Passives: Resistors

- No provisions in standard CMOS
- Resistors are bad for digital circuits →
  - Minimized in standard CMOS
- Sheet resistance of available layers:

Layer	Sheet resistance
Aluminum	60 mΩ/□
Polysilicon	5 Ω/□
N+/P+ diffusion	5 Ω/□
N-well	1 kΩ/□

- Example: 100kΩ poly resistor  
→ 1μm wide by 20,000μm long

## Silicide Technology

- Implants used to lower resistance of source/drain and polysilicon.
- Titanium Silicide ( $TiSi_2$ ) is widely used salicide (self-aligned silicide), has low resistivity ( $13-17 \text{ m}\Omega\text{-cm}$ ) with melting point of  $1540^\circ\text{C}$ . Another widely used silicide is  $CoSi_2$ .
- Platinum Silicide ( $PtSi$ ) is a highly reliable contact metallization between the silicon substrate and the metal layers (Al).

# Silicide Block Option

Layer	R/ $\square$ [ $\Omega/\square$ ]	T <sub>C</sub> [ppm/ $^{\circ}$ C] @ T = 25 $^{\circ}$ C	V <sub>C</sub> [ppm/V]	B <sub>C</sub> [ppm/V]
N+ poly	100	-800	50	50
P+ poly	180	200	50	50
N+ diffusion	50	1500	500	-500
P+ diffusion	100	1600	500	-500
N-well	1000	-1500	20,000	30,000

- Non-silicided layers have significantly larger sheet resistance
- Resistor nonidealities:
  - Temperature coefficient:  $R = f(T)$
  - Voltage coefficient:  $R = f(V)$

# Resistor Example

Goal:  $R = 100 \text{ k}\Omega$ ,  $T_C = 1/R \times dR/dT = 0$

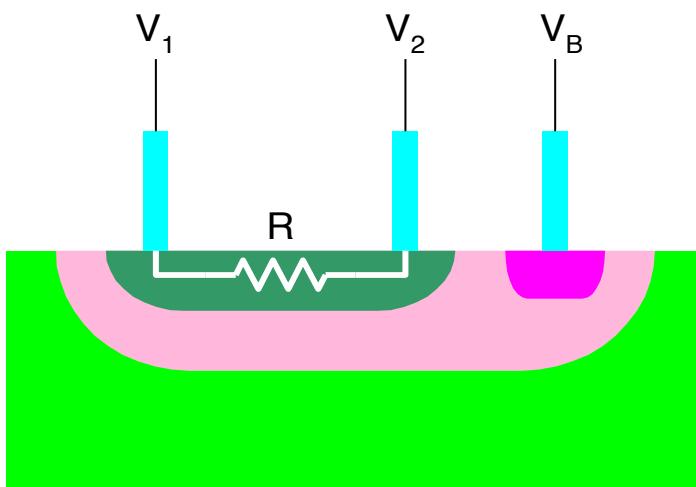
Solution: combination of N+ and P+ poly resistors in series

$$\begin{aligned} R &= R_N(1 + T_{CN}\Delta T) + R_P(1 + T_{CP}\Delta T) \\ &= \underbrace{R_N + R_P}_{R} + \underbrace{(R_N T_{CN} + R_P T_{CP})\Delta T}_{0} \Rightarrow \end{aligned}$$

$$R_N = R \frac{1}{1 - \frac{T_{CN}}{T_{CP}}} = 20 \text{ k}\Omega = 200 \text{ squares}$$

$$R_P = R \frac{1}{1 - \frac{T_{CP}}{T_{CN}}} = 80 \text{ k}\Omega = 444.4 \text{ squares}$$

# Voltage Coefficient



$$R = \frac{V_1 - V_2}{I}$$

$$\approx R_o \left[ 1 + T_C (T - 25^\circ) + V_C (V_1 - V_2) + B_C \left( \frac{V_1 + V_2}{2} - V_B \right) \right]$$

## Example:

Diffusion resistor

→ Applied voltage modulates depletion width  
(cross-section of conductive channel)

→ Well acts as a shield

# Resistor Matching

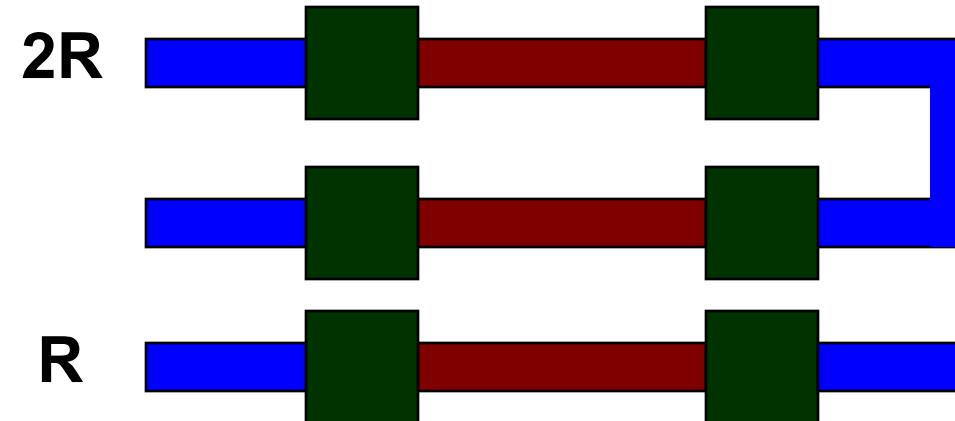
- Types of mismatch
  - Systematic (e.g. contacts)
  - Run-to-run variations
  - Random variations between devices
- Absolute resistor value
  - E.g. filter time constant, bias current (BG reference)
  - $\sim 15$  percent variations (or more)
- Resistor ratios
  - E.g. opamp feedback network
  - Insensitive to absolute resistor value
  - “unit-element” approach rejects systematic variations  
(large area for non-integer ratios)
  - Process gradients
  - 0.1 ... 1 percent matching possible with careful layout

# Systematic Variations from Layout

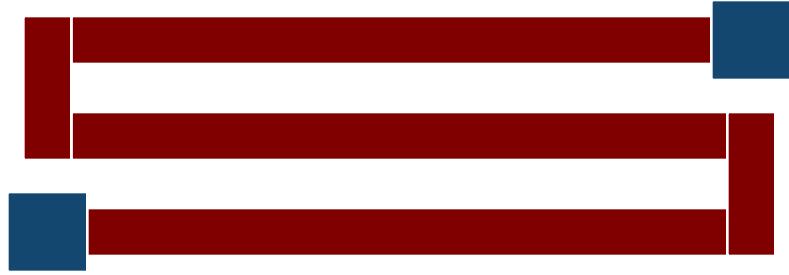
- Example:



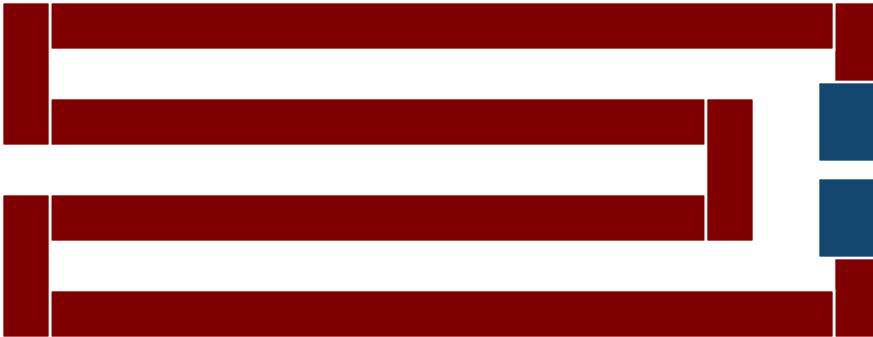
- Use unit element instead:



Serpentine layout for large values:

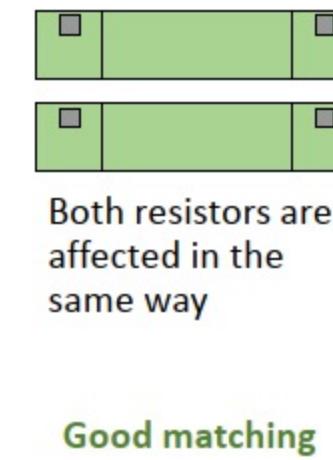
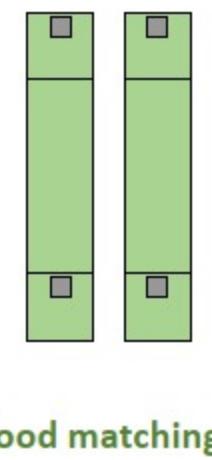
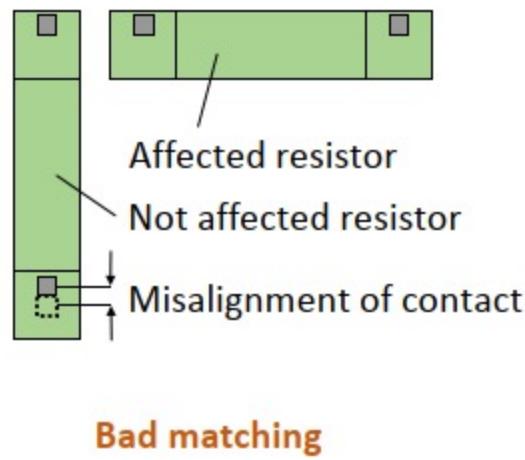


Better layout (mitigates offset due to thermoelectric effects):



See Hastings, "The art of analog layout," Prentice Hall, 2001.

# Misalignment



# Electromigration and Parasitics

- **Electromigration rules limit maximum current density**
  - Same value resistance must be physically larger to carry more current
  - Hence, more current → larger capacitive parasitics

# Capacitors

- Simplest capacitor:

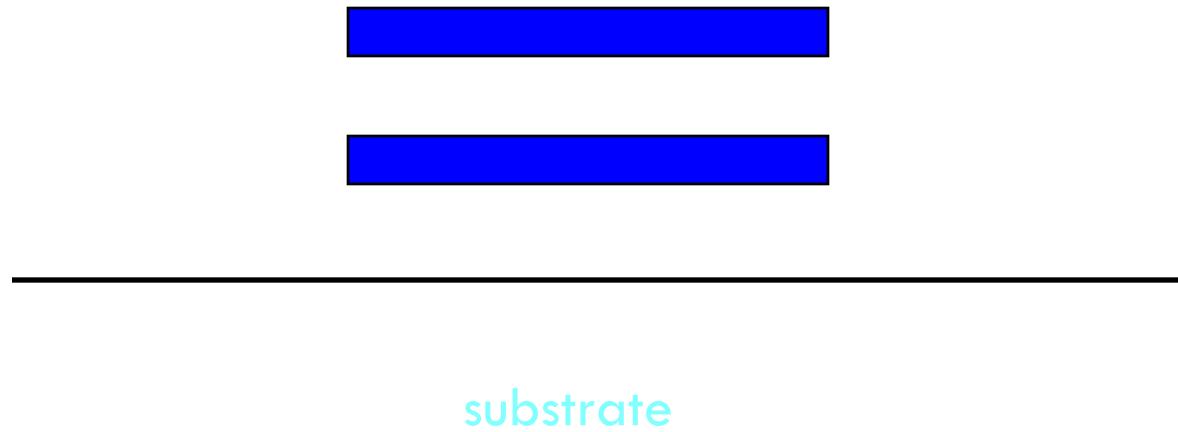


substrate

- What's the problem with this?

# Capacitors

- “Improved” capacitor:

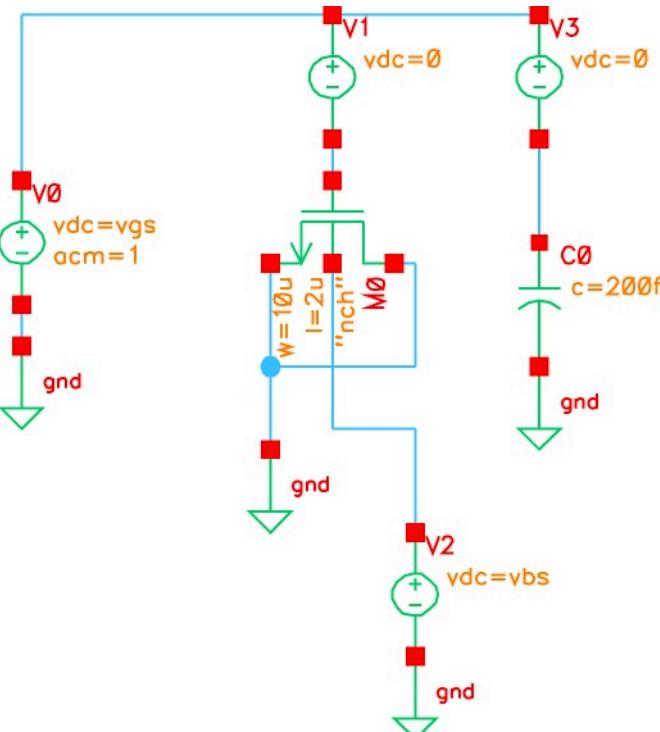


- Is this only 1 capacitor?

# Capacitors Options

Type	C [aF/ $\mu\text{m}^2$ ]	V <sub>C</sub> [ppm/V]	T <sub>C</sub> [ppm/ $^\circ\text{C}$ ]
Gate (thick oxide)	8000	Huge	Big
MIM (option)	4000		
MOM (> 5 metals)	~2000		
Poly-poly (option)	1000	10	25
Metal-metal	50	20	30
Metal-substrate	30		
Metal-poly	50		
Poly-substrate	120		
Junction capacitors	~ 1000	Big	Big

# MOS Capacitor



- High capacitance in inversion:
  - Linear region
  - Strong inversion
- SPICE:

$$C = \frac{I}{\omega V}$$

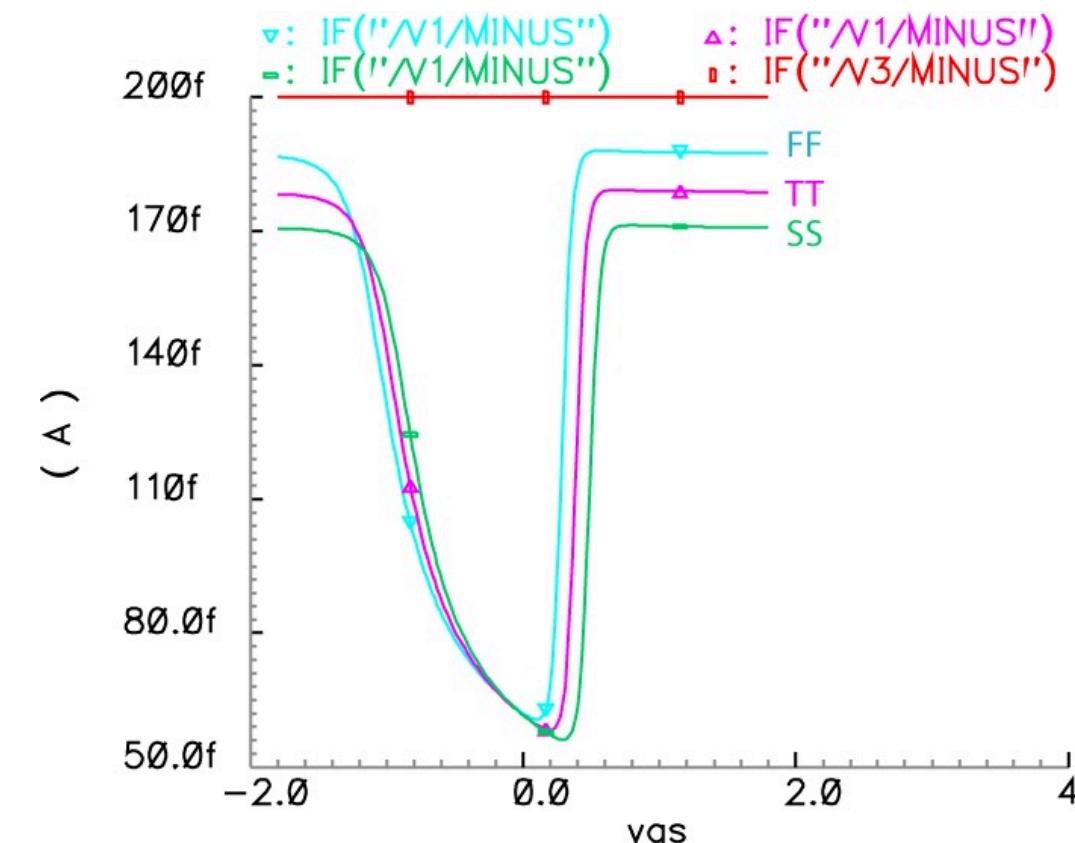
$$V = 1V$$

$$\omega = 1$$

$$\rightarrow C = I$$

# MOS Capacitor

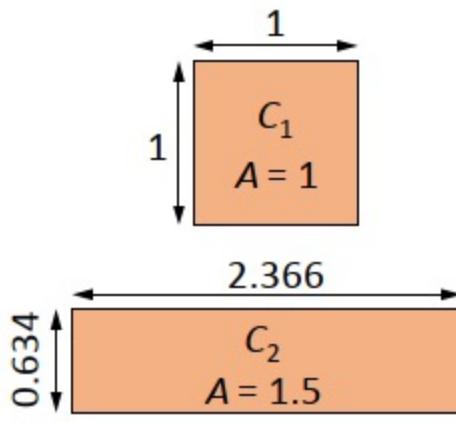
- High non-linearity, temperature coefficient
- But, still useful in many applications, e.g.:
  - (Miller) compensation capacitor
  - Bypass capacitor (supply, bias)



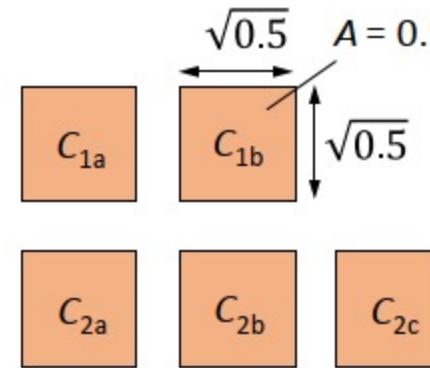
# Poly-Poly Capacitor

- Applications:
  - Feedback networks
  - Filters (SC and continuous time)
  - Charge redistribution DACs & ADCs
- Cross-section
- Bottom- and top-plate parasitics
- Shields

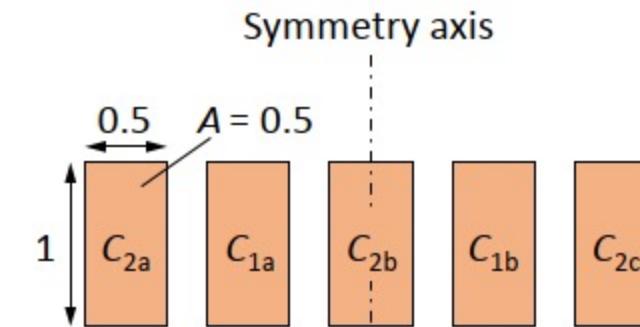
# Matching Capacitors



Minimum matching



Good matching



Very good matching

Excellent matching...

# Capacitor Layout

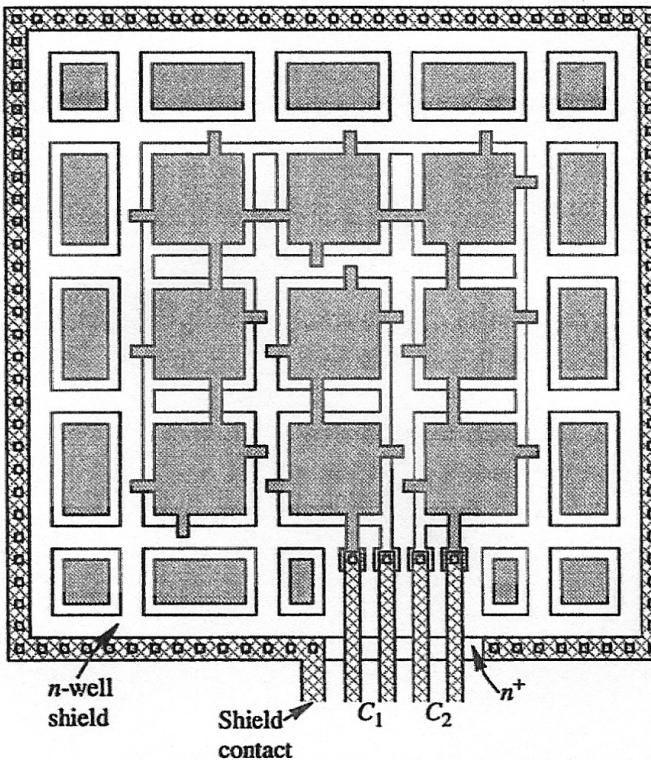


Figure 6.17 A layout of two capacitors with ratio  $C_2/C_1 = \frac{1}{2}$ , incorporating several of the techniques discussed in the text (adapted from Ref. 23).

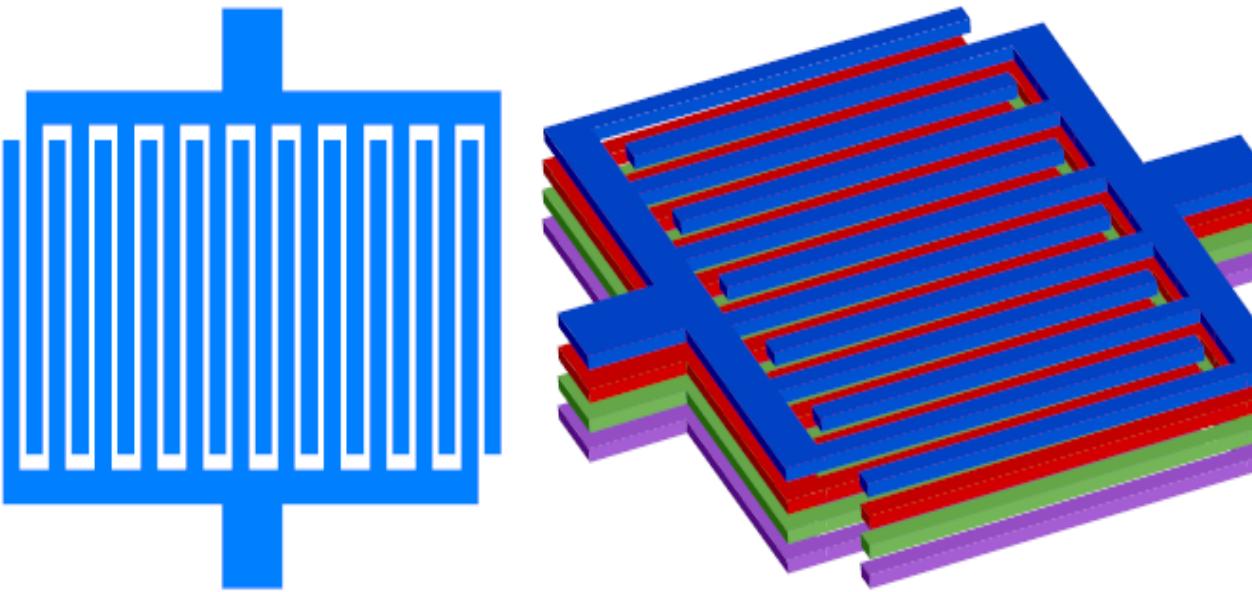
- Unit elements
- Shields:
  - Etching
  - Fringing fields
- “Common-centroid”
- Wiring and interconnect parasitics

Ref.: Y. Tsividis, “Mixed Analog-Digital VLSI Design and Technology,” McGraw-Hill, 1996.

## MIM Capacitors

- Many processes have add-on options such as a MIM capacitor.
- This is simply a metal-insulator-metal (MIM) structure situated in the oxide layers. The insulator is a very thin layer ( $\sim 25$  nm), resulting in high density and relatively low back-plate parasitics.

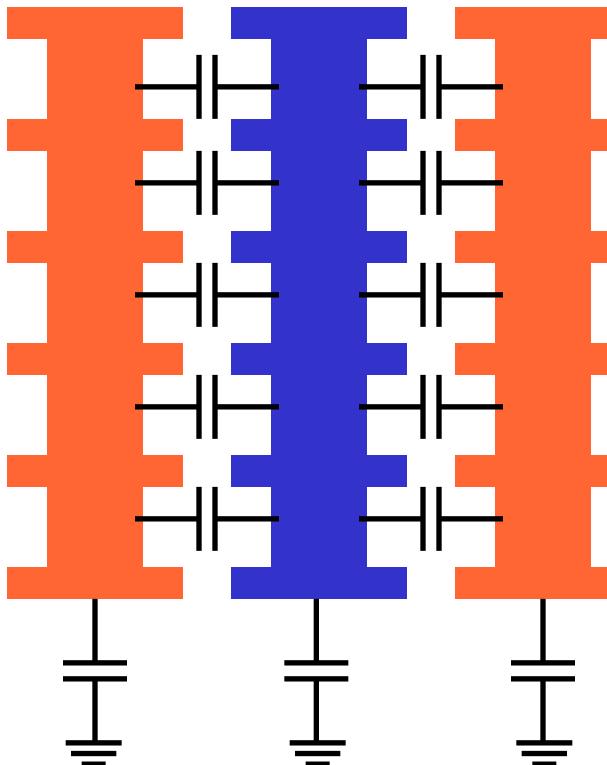
# “MOM” Capacitors



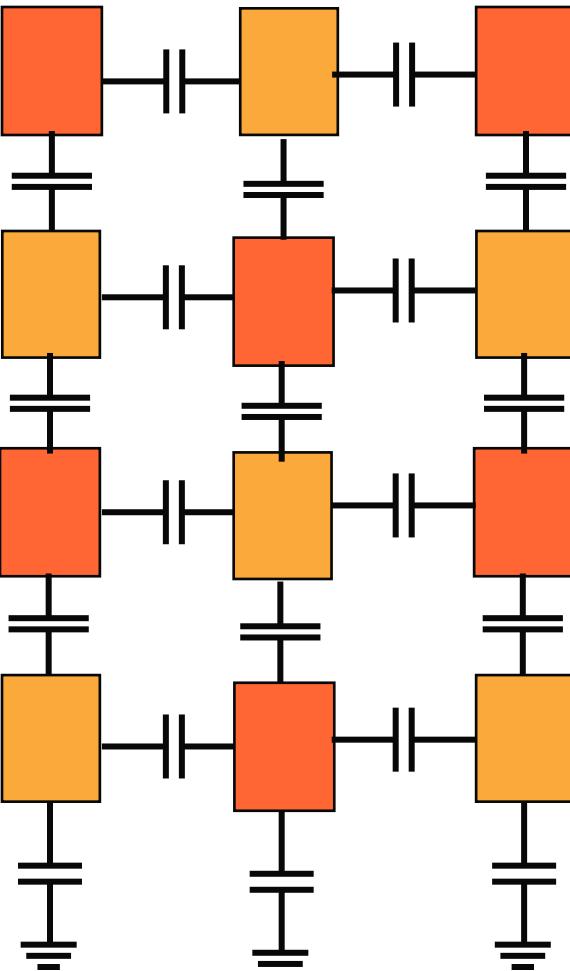
- Metal-Oxide-Metal capacitor. Free and most common in modern CMOS.
- Use lateral flux ( $\sim L_{\min}$ ) and multiple metal layers to realize high capacitance values

# MOM Capacitor Cross Section

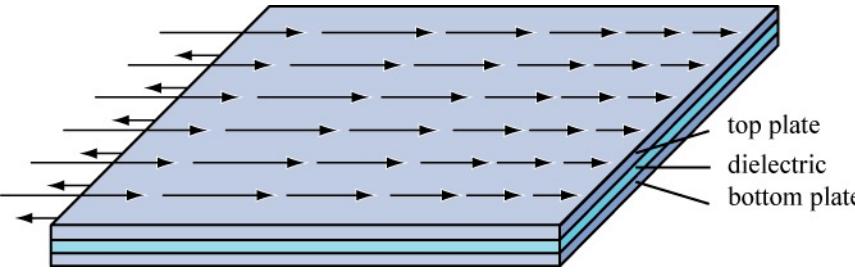
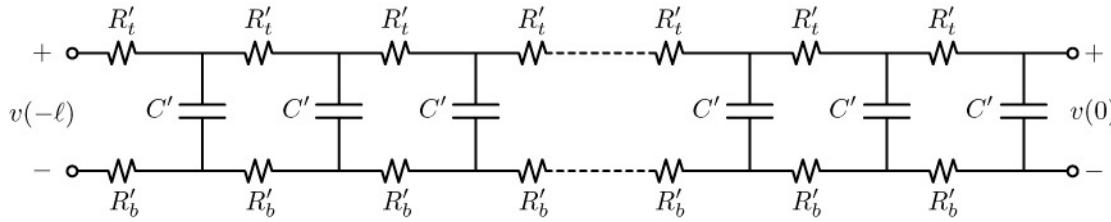
- Use a wall of metal and vias to realize high density
- More layers – higher density
  - May want to chop off lower layers to reduce  $C_{bot}$
- Reasonably good matching and accuracy



# Another MOM Option



# Capacitor Q



- Current density in capacitor plates drops due to vertical displacement current.
- Must analyze the distributed RC circuit...

## Capacitor Impedance

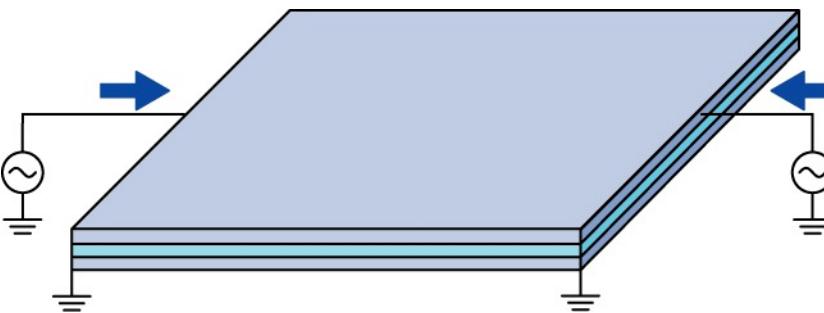
- For a capacitor contacted at one side, we can show that

$$Z_{in} = \frac{\gamma}{j\omega C'} \coth(\gamma\ell)$$

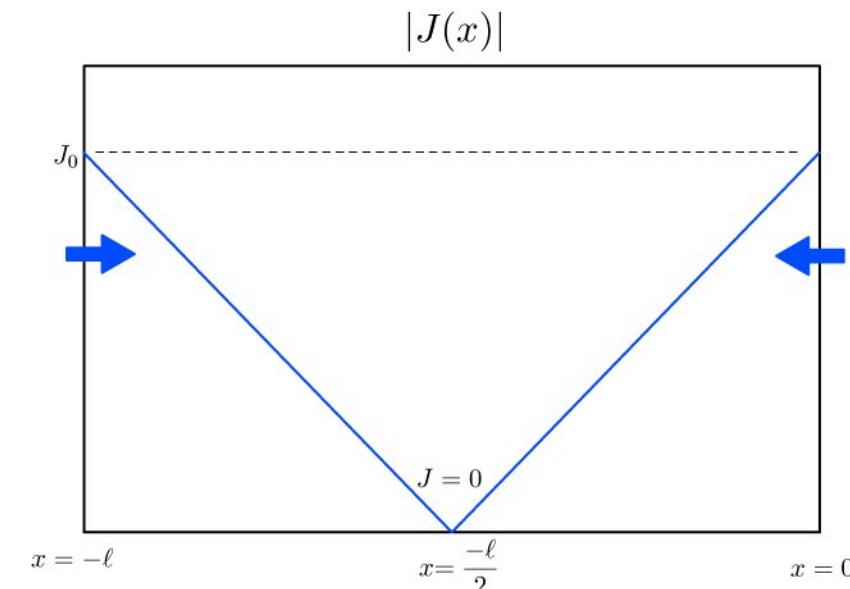
- We can simplify this for small structures

$$Z_{in} \approx \frac{R_T + R_B}{3} + \frac{1}{j\omega C}$$

# Double Contact Structure

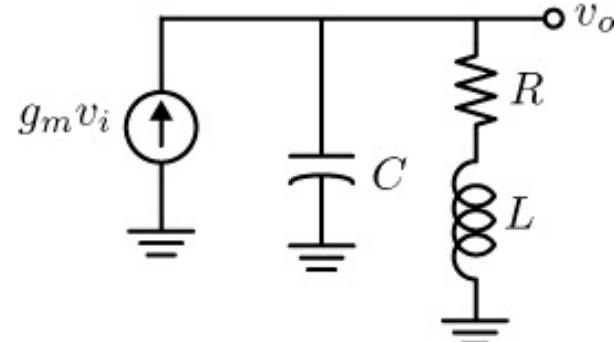
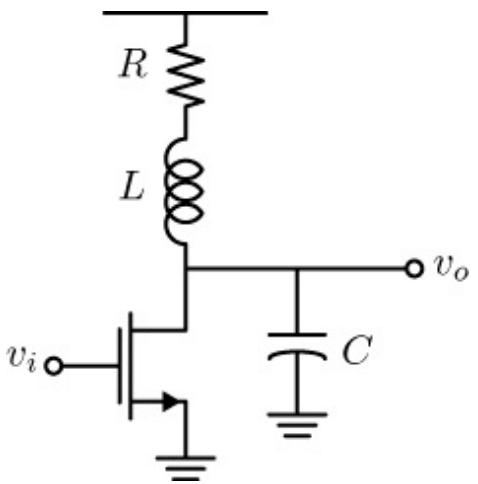


$$Z_{in} \approx \frac{R}{12} + \frac{1}{j\omega C}$$



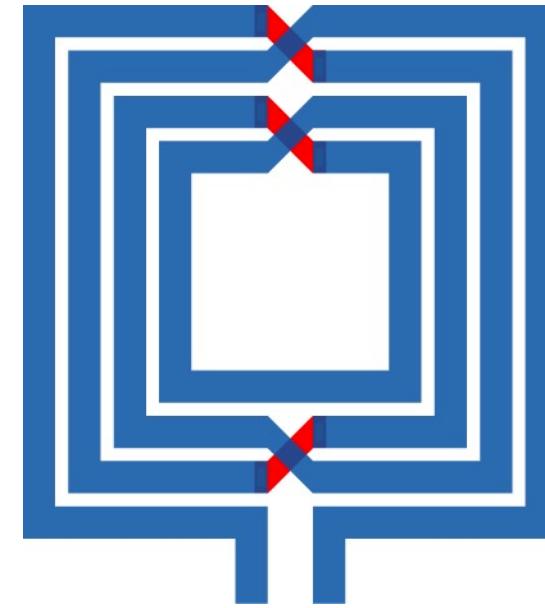
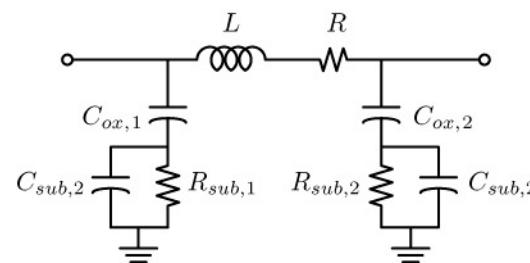
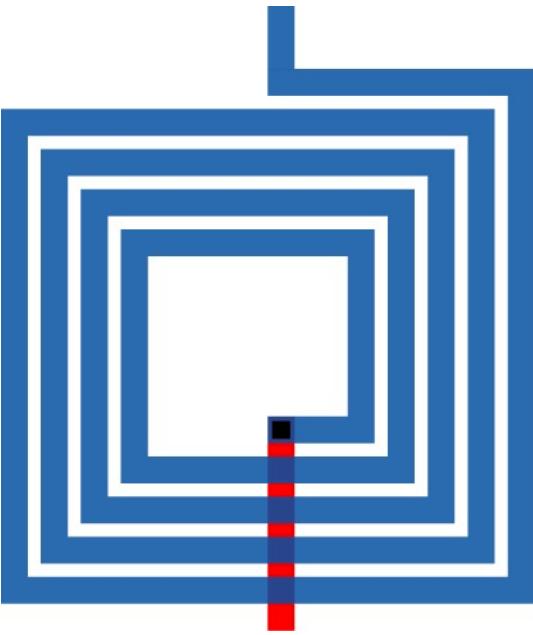
- Instead of doing distributed analysis, we can guess the current distribution as linear and quickly calculate the effective resistance.
- For a double contact case, the resistance drops by 4 times.

# What About Inductors?



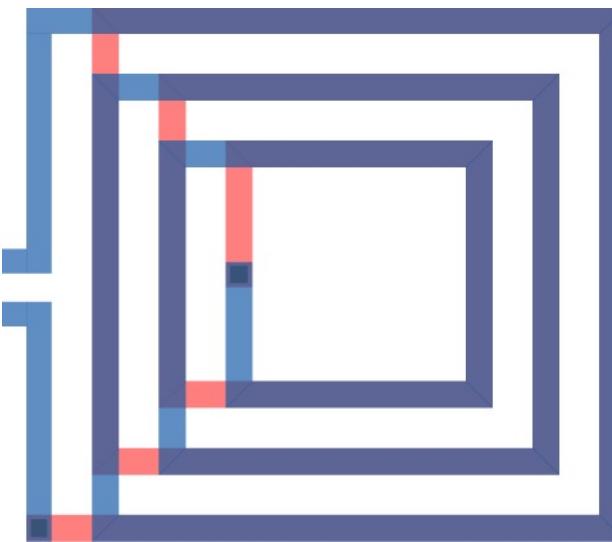
- Mostly not used in analog/mixed-signal design
  - Usually too big
  - More of a pain to model than R's and C's
  - But they do occasionally get used
- Example inductor app.: shunt peaking
  - Can boost bandwidth by up to 85%!
  - Q not that important (L in series with R)
  - But frequency response may not be flat

# Spiral Inductors

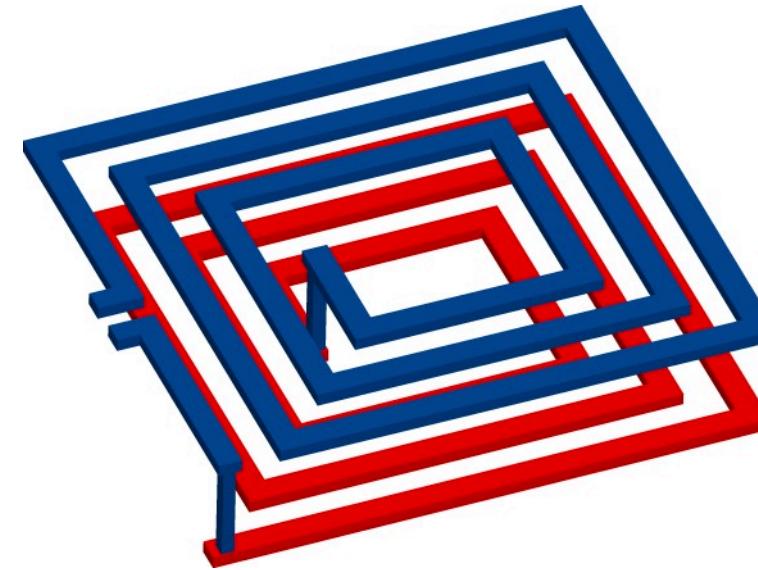


- Used widely in RF circuits for small L ( $\sim 1-10\text{nH}$ ).
- Use top metal for Q and high self resonance frequencies.
  - Very good matching and accuracy – if you model them right

# Multi-Layer Inductors



Top view



3D view (not to scale)

- By taking advantage of multiple metal layers and mutual coupling, we can realize very large inductors ( $\sim 100 \text{ nH}$ ).

## References

- Lienig, Scheible, “Fundamentals of Layout Design for Electronic Circuits,” Springer 2020.
- Maloberti, F, “Layout of Analog CMOS Integrated Circuits” (Part 2)
- Hastings, “The art of analog layout,” Prentice Hall, 2001.
- A. M. Niknejad, *Electromagnetics for High-Speed Analog and Digital Communication Circuits*, Cambridge University Press, 2007. ([pdf](#))

# Layout Editing Concepts

- Polygons
  - Merge
  - Chop
- F4 → options
- Array Copy vs Tile
- Vias
- Stretching / Edges
- Live DRC / DRD editing
- MPP's (Multi-Part Path)
- Grouping vs Hierarchy
  - Edit in place
- Bus layout
- Learn to Probe Nets / Mark Nets
- Learn how to do a hierarchy copy (and modify cells to point to new cells)