Problem 1: Laplace transform

(a)
$$f(t) = \sum_{k=0}^{\infty} a^k \delta(t-kT)$$
, $a>0$

$$\Rightarrow \mathcal{L}[f(t)] = \mathcal{L} \left[\sum_{k=0}^{\infty} a^k S(t-kT) \right]$$

Using the linearity property of laplace transform

NOWS

$$2[s(t)] = \int_{0}^{\infty} s(t) e^{-st} dt$$

$$= \int_{0}^{\infty} s(t) dt$$

$$= \int_{0}^{\infty} s(t) dt$$

By the t-shift property of laplace transform

$$2 \left[\frac{1}{k} \left(\frac{1}{k} - \frac{1}{k} \right) \right] = e^{-kTs}$$

$$2 \left[\frac{1}{k} \left(\frac{1}{k} - \frac{1}{k} \right) \right] = e^{-kTs}$$

$$2 \left[\frac{1}{k} \left(\frac{1}{k} - \frac{1}{k} \right) \right] = e^{-kTs}$$

$$2 \left[\frac{1}{k} - \frac{1}{k} - \frac{1}{k} - \frac{1}{k} \right]$$

$$2 \left[\frac{1}{k} - \frac{$$

Now the term inside the summation forms a geometric series with common ratio a geometric series with common ratio a e-st tence the infinite sum converges are. I tence the infinite sum converges to a finite value iff [ae-st]<1

$$f(s) = \frac{1}{1-\alpha e^{-sT}}$$

for ROC, we have

(b)
$$f(t) = \sin(\omega_0 t + b) e^{-qt} u(t)$$

$$\Rightarrow f(t) = \left[\sin(\omega_0 t) \cos(b) + \cos(\omega_0 t) \sin(b) \right] e^{-qt} u(t)$$

$$f(t) = \cos(b) \sin(\omega_0 t) e^{-qt} u(t) + \sin(b) \cos(\omega_0 t) e^{-qt} u(t)$$

Now using the LT table, we have
$$\sin(\omega_0 t) e^{-qt} u(t) \stackrel{\checkmark}{=} \frac{\omega_0}{(s+q)^2 + \omega_0^2}$$

$$\cos(\omega_0 t) e^{-qt} u(t) \stackrel{\checkmark}{=} \frac{3+q}{(s+q)^2 + \omega_0^2}$$

Hence using linearity of LT,
$$f(s) = \frac{\omega_0 \cos(b) + (s+q) \sin(b)}{(s+q)^2 + \omega_0^2}$$

Poles of FCS): (Sta) + w = 0

$$=$$
 $(sta)^{\gamma} = -\omega^{\gamma}$

Since f(t) is a right-sided signal, so ROC is all right-sidel and to the right of the rightmost pole.

Problem 2: Inverse Laplace transform

$$\Rightarrow \chi(s) = \frac{s+1}{(s+3)(s+v)}$$

Now using Partial Fraction expansion,

$$\frac{S+1}{(S+3)(S+2)} = \frac{r_1}{S+3} + \frac{r_2}{S+2}$$

using cover up procedure,

$$r_1 = \frac{2}{(5+3)} \times \frac{3}{(5)} = \frac{2}{-1} = 2$$

$$\int_{2-}^{2} (S_{+}) \times (S) \Big|_{S=-2} = \frac{-1}{1} = -1$$

$$\chi(s) = \frac{2}{5+3} - \frac{1}{5+2}$$

line the ROC is right-sided, so the time-domain signal is also right-sided.

Using lookup table,

$$x(t) = 2e^{-3t}u(t) - e^{-2t}u(t)$$

(b)
$$\chi(s) = \frac{s+1}{(s+1)^2 + 4}$$
, Roc: $\sigma > -1$

Then using lookup table,

Using the S-shift property of laplace transform,

$$x(t) = e^{-t}y(t)$$

 $x(t) = e^{-t}(e_{s}(2t)u(t))$

Problem 3: Stability and Causality of LTI systems

$$\ddot{y}(t) - \dot{y}(t) - 2y(t) = x(t)$$

(a) Taking the laplace transforms of both sigles

sigles

$$5^{2}Y(5) - 5J(0^{-}) - \dot{J}(0^{-}) - 5Y(5) + J(0^{-}) - 2Y(5) = X(5)$$

$$S^{1}Y(3) - Sy(6) - 3(6) - y(6) + y(6) = X(S)$$

 $Y(3)[s^{2}-S-2] - Sy(6) - y(6) + y(6) = X(S)$

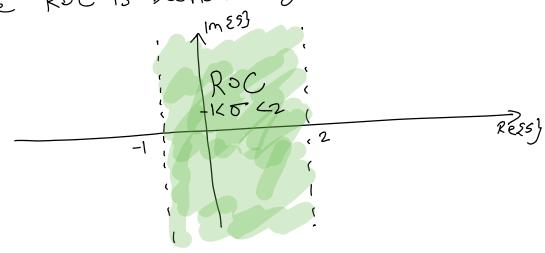
$$= \frac{Y(s)}{Y(s)} = H(s) = \frac{1}{s^2 - s - 2}$$

(b) i) If the system is stable, then the Roc of H(s) should include the imaginary axis. Now,

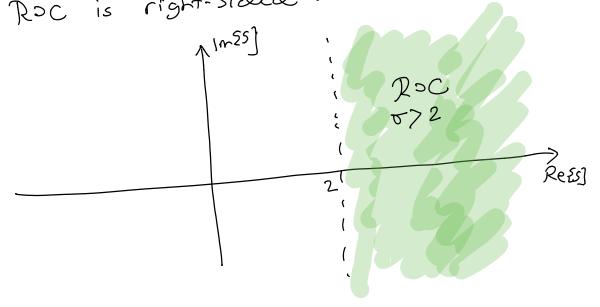
$$H(5) = \frac{1}{(5+1)(5-2)}$$

Hence H(s) has two poles: S=-1,2The only scenario in which the ROC
of H(s) can contain the imaginary
of H(s) is the imaginary
axis is if h(t) is two-sided and

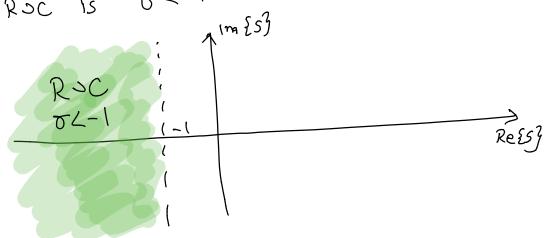
the Roc is bounded by the two poles.



ii) The system is causal iff h(t) = o for t <0. This can only Lappen if the ROC is right-sided.



(iii) If the system is relither stable nor cousal, then the only option for the ROC is TX-1.



@ If the system is causal, then the ROC is right-sided and the impulse response is also right-siled. Hence,

$$H(s) = \frac{1}{(s+1)(s-2)}$$

Using Partial Fraction expansion,

$$\frac{1}{(s+1)(s-2)} = \frac{r_1}{s+1} + \frac{r_2}{s-2}$$

By cover-up procedure,

$$r_1 = (s+1)H(s)|_{s=-1} = -1/3$$

$$(2-(5-2))(5))_{5=2} = 1/3$$

$$\Rightarrow H(S) = \frac{-1/3}{S+1} + \frac{1/3}{S-2}$$

Using lookup tables

using lookup tables
$$h(t) = -1/3 e^{-t} u(t) + 1/3 e^{2t} u(t)$$