

# Linear Time-Invariant (LTI) Systems

Digital Signal Processing

January 31, 2023



# Linear Systems

## Definition

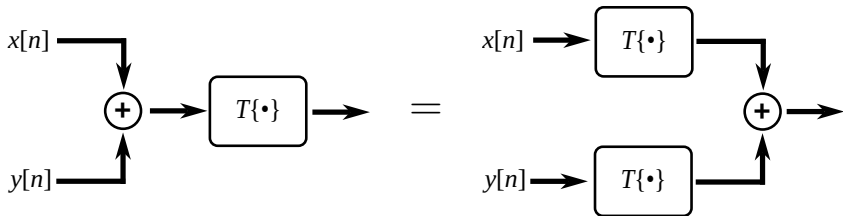
A **linear system** is a system  $T$  that satisfies:

- ① Additivity:  $T\{x[n] + y[n]\} = T\{x[n]\} + T\{y[n]\},$
- ② Scaling:  $T\{ax[n]\} = aT\{x[n]\},$

for all signals  $x[n], y[n]$ , and all scalar constants,  $a$ .

# Linearity Property in Diagrams

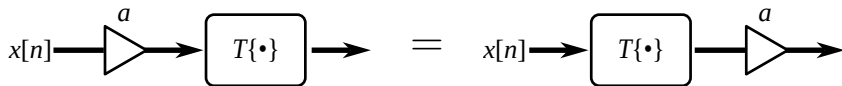
## Additivity:



$$T\{x[n] + y[n]\} = T\{x[n]\} + T\{y[n]\}$$

# Linearity Property in Diagrams

## Scaling:



$$T\{ax[n]\} = aT\{x[n]\}$$

# Linear Systems (again)

An *equivalent* definition of linearity combines additivity and scaling into one rule:

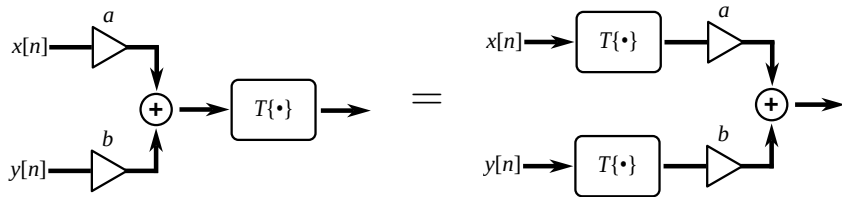
## Definition

A **linear system** is a system  $T$  that satisfies:

$$T\{ax[n] + by[n]\} = aT\{x[n]\} + bT\{y[n]\},$$

for all signals  $x[n]$ ,  $y[n]$ , and all scalar constants,  $a$ ,  $b$ .

# Linearity Property in Diagrams (again)



$$T\{ax[n] + by[n]\} = aT\{x[n]\} + bT\{y[n]\}$$

# Examples

Are the following linear systems or non-linear systems?

- $T\{x[n]\} = 2x[n]$  Linear
- $T\{x[n]\} = x[n - 1]$  Linear
- $T\{x[n]\} = x[n]^2$  Non-linear
- $T\{x[n]\} = nx[n]$  Linear
- $T\{x[n]\} = x[2n]$  Linear
- $T\{x[n]\} = x[n] + 1$  Non-linear

# Time-Invariant Systems

## Definition

A system,  $T$ , is called **time-invariant**, or **shift-invariant**, if it satisfies

$$y[n] = T\{x[n]\} \Rightarrow y[n - N] = T\{x[n - N]\},$$

for all signals  $x[n]$  and all shifts  $N \in \mathbb{Z}$ .



# Examples

Are the following time-invariant or time-variant systems?

- $T\{x[n]\} = 2x[n]$  Time-invariant
- $T\{x[n]\} = x[n - 1]$  Time-invariant
- $T\{x[n]\} = x[n]^2$  Time-invariant
- $T\{x[n]\} = nx[n]$  Time-variant
- $T\{x[n]\} = x[2n]$  Time-variant
- $T\{x[n]\} = x[n] + 1$  Time-invariant

# Linear Time-Invariant (LTI) Systems

## Definition

A **linear time-invariant (LTI) system** is one that is both linear and time-invariant.

# Examples

Are the following LTI or not LTI systems?

- $T\{x[n]\} = 2x[n]$  LTI
- $T\{x[n]\} = x[n - 1]$  LTI
- $T\{x[n]\} = x[n]^2$  not LTI
- $T\{x[n]\} = nx[n]$  not LTI
- $T\{x[n]\} = x[2n]$  not LTI
- $T\{x[n]\} = x[n] + 1$  not LTI

# LTI Fun Fact

The **only** way to get an LTI system is by composing time shifts and scalings by constants.

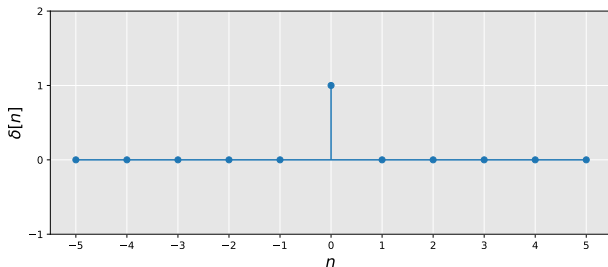
In other words, any LTI system,  $T$ , can be written as

$$T\{x[n]\} = \sum_{m=-\infty}^{\infty} a_m x[n - m],$$

for some scalar constants,  $a_m$ .

# Impulse Response

Recall our **unit sample function** or **impulse**:



$$\delta[n] = \begin{cases} 1, & n = 0, \\ 0, & n \neq 0. \end{cases}$$

# Impulse Response

## Definition

The **impulse response** of a system,  $T$ , is the output it produces when given the unit impulse function as input. This is denoted:

$$h[n] = T\{\delta[n]\}.$$

# Impulse Response

Recall any sequence,  $x[n]$ , can be written as a sum of scaled, shifted impulses:

$$x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n - k].$$

This is the principle of **superposition**.

# Impulse Response for an LTI System

Given an LTI,  $T$ :

$$\begin{aligned} T\{x[n]\} &= T\left\{\sum_{k=-\infty}^{\infty} x[k]\delta[n-k]\right\} && \text{superposition for } x[n] \\ &= \sum_{k=-\infty}^{\infty} T\{x[k]\delta[n-k]\} && \text{additivity property} \\ &= \sum_{k=-\infty}^{\infty} x[k]T\{\delta[n-k]\} && \text{scaling property} \\ &= \sum_{k=-\infty}^{\infty} x[k]h[n-k] && \text{definition of impulse response} \end{aligned}$$



# Convolution

## Definition

The convolution of two sequence,  $x[n]$ ,  $h[n]$ , is given by

$$(x * h)[n] = \sum_{k=-\infty}^{\infty} x[k]h[n - k].$$

With this notation, any LTI system,  $T$ , with impulse response,  $h$ , can be computed as

$$T\{x[n]\} = x[n] * h[n].$$

# Properties of Convolution

## Commutativity:

$$x[n] * h[n] = h[n] * x[n]$$

# Properties of Convolution

## Associativity:

$$(x[n] * h_1[n]) * h_2[n] = x[n] * (h_1[n] * h_2[n])$$

This means that we can apply  $h_1[n]$  to  $x[n]$  followed by  $h_2[n]$ , or we can convolve the impulse responses  $h_2[n] * h_1[n]$  and then apply the resulting system to  $x[n]$ .

# Properties of Convolution

## Linearity:

$$(ax[n]) * h[n] = a(x[n] * h[n])$$

$$(x[n] + y[n]) * h[n] = (x[n] * h[n]) + (y[n] * h[n])$$

# Properties of Convolution

## Time-Invariance / Shift-Invariance:

Let  $D\{x[n]\} = x[n - N]$  be an ideal delay by  $N$ . Then

$$D\{x[n] * h[n]\} = D\{x[n]\} * h[n]$$

This means that we can convolve  $x[n]$  and  $h[n]$  and then shift the result, or we can shift  $x[n]$  and then convolve it with  $h[n]$ .

# Equivalence of LTI Systems and Convolutions

## Theorem

*A system,  $T\{\}$ , is linear and time-invariant if and only if it can be written as a convolution,*

$$T\{x[n]\} = (x * h)[n],$$

*for some signal,  $h$ .*

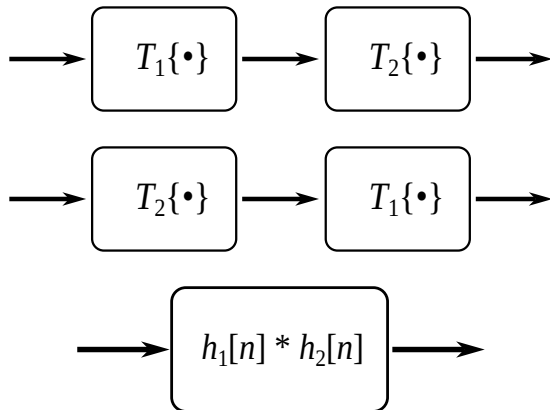
# Commutativity of LTI Systems

Let  $T_1$  and  $T_2$  be LTI systems, with impulse responses  $h_1, h_2$ , respectively.

$$\begin{aligned}T_2\{T_1\{x[n]\}\} &= (x[n] * h_1[n]) * h_2[n] \\&= x[n] * (h_1[n] * h_2[n]) && \text{associativity of } * \\&= x[n] * (h_2[n] * h_1[n]) && \text{commutativity of } * \\&= (x[n] * h_2[n]) * h_1[n] && \text{associativity again} \\&= T_1\{T_2\{x[n]\}\}\end{aligned}$$

# Commutativity of LTI Systems

If  $T_1$  and  $T_2$  are LTI systems, the following are equivalent:





## Definition

A signal,  $x[n]$ , is **bounded** if  $|x[n]| \leq B$  for some  $B < \infty$  and for all  $n \in \mathbb{Z}$

## Definition

A system,  $T\{\cdot\}$ , is said to be **bounded-input, bounded-output (BIBO) stable** if for every bounded input  $x[n]$ , the resulting output  $T\{x[n]\}$  is also bounded.

# BIBO Stability of LTI Systems

## Theorem

*An LTI system is BIBO stable if and only if its impulse response,  $h[n]$ , is absolutely summable:*

$$\sum_{k=-\infty}^{\infty} |h[n]| < \infty.$$

# Causality

## Definition

A system is said to be **causal** if, for any  $n_0 \in \mathbb{Z}$ ,  $T\{x[n_0]\}$  depends only on previous values of  $x[n]$ , for  $n \leq n_0$

A causal system cannot “look into the future.”

If  $x[n] = y[n]$  for all  $n < n_0$ , then  $T\{x[n]\} = T\{y[n]\}$  for all  $n < n_0$ .

# Causality of LTI Systems

## Theorem

*An LTI system is causal if and only if its impulse response function,  $h[n]$ , satisfies  $h[n] = 0$  for all  $n < 0$ .*

*Sketchy proof.*

Our LTI system output evaluated for some  $n_0$  is:

$$(h * x)[n_0] = \sum_{k=-\infty}^{\infty} h[k]x[n_0 - k]$$

This will avoid using  $x[n]$  for  $n > n_0$  if and only if  $h[k] = 0$  when  $n_0 - k > n_0$ . That is, when  $k < 0$ .

# Working with Finite-Length Signals

The convolution equation deals with signals,  $x[n]$ ,  $h[n]$ , that are defined for **infinite time**:  $-\infty < n < \infty$ :

$$x[n] * h[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k], \quad \text{for } n \in \mathbb{Z}.$$

Of course, on a computer we can only store signals that are **finite sequences**, that is, arrays with index  $n \in [0, L-1]$ .

# Padding

For a finite-length signal,  $x[n]$ , defined for  $n \in [0, L - 1]$ , we can extend it to all  $n \in \mathbb{Z}$  by **padding**.

Multiple ways to pad:

- Pad with zeros:  $x[n] = 0$  for  $n < 0$  and  $n \geq L$
- Periodic padding:  $x[n] = x[n \bmod L]$  for  $n \in \mathbb{Z}$
- many more ...

# Convolution with Zero Padding

Zero padding means we can truncate the  $k$  and  $n$  indices in our convolution equation to be between  $[0, L - 1]$ :

$$x[n] * h[n] = \sum_{k=0}^{L-1} x[k]h[n - k], \quad \text{for } n \in [0, L - 1].$$

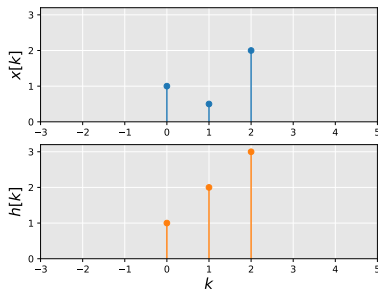
Does this work? **No!  $n - k$  can be negative.**

Instead, truncate  $k$  at  $n$ :

$$x[n] * h[n] = \sum_{k=0}^n x[k]h[n - k], \quad \text{for } n \in [0, L - 1].$$

# Convolution Example

Computing  $y[n] = x[n] * h[n] = \sum_{k=0}^n x[k]h[n-k]$



$$x[n] = (0.5, 1.5, 2.5)$$

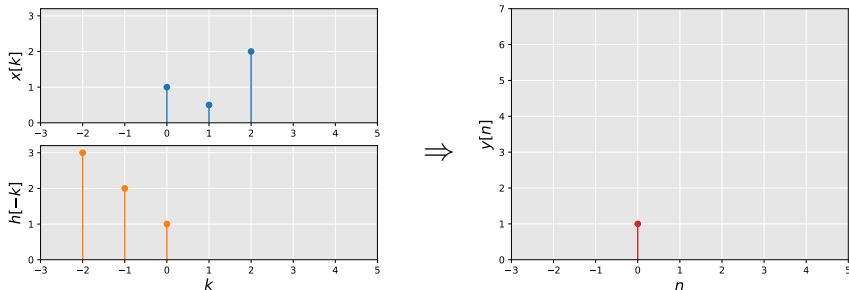
$$h[n] = (1.0, 2.0, 3.0)$$



# Convolution Example

Computing  $y[n] = x[n] * h[n] = \sum_{k=0}^n x[k]h[n-k]$

For  $n = 0$ , flip  $h$  about 0 to get  $h[-k]$ .

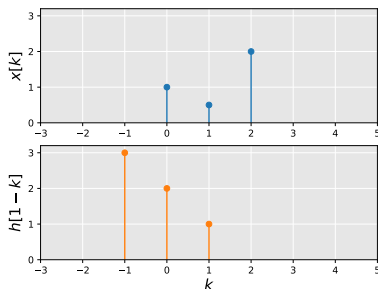


$$\begin{aligned} y[0] &= x[0] \times h[0] \\ &= 1.0 \times 1.0 = 1.0 \end{aligned}$$

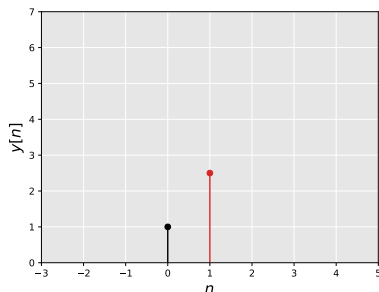
# Convolution Example

Computing  $y[n] = x[n] * h[n] = \sum_{k=0}^n x[k]h[n-k]$

For  $n = 1$ , shift  $h$  right by one to get  $h[1-k]$ .



$\Rightarrow$

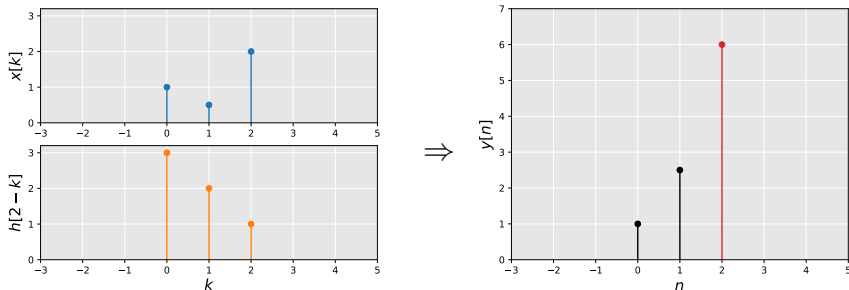


$$\begin{aligned} y[1] &= x[0]h[1] + x[1]h[0] \\ &= 1.0 \times 2.0 + 0.5 \times 1.0 = 2.5 \end{aligned}$$

# Convolution Example

Computing  $y[n] = x[n] * h[n] = \sum_{k=0}^n x[k]h[n-k]$

For  $n = 2$ , shift  $h$  right again to get  $h[2-k]$ .

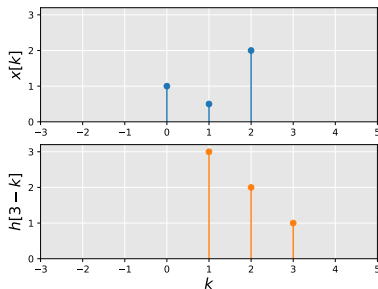


$$\begin{aligned} y[2] &= x[0]h[2] + x[1]h[1] + x[2]h[0] \\ &= 1.0 \times 3.0 + 0.5 \times 2.0 + 2.0 \times 1.0 = 6.0 \end{aligned}$$

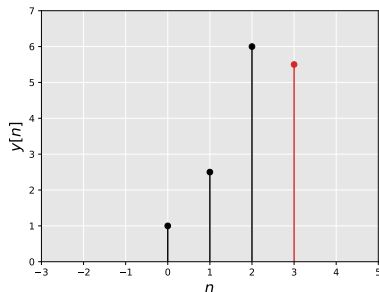
# Convolution Example

Computing  $y[n] = x[n] * h[n] = \sum_{k=0}^n x[k]h[n-k]$

For  $n = 3$ , shift  $h$  right again to get  $h[3-k]$ .



$\Rightarrow$

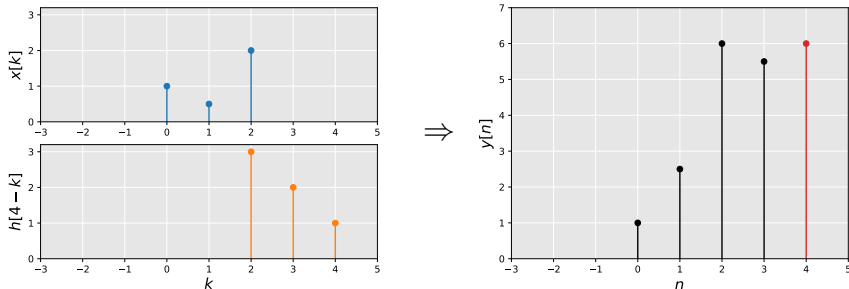


$$\begin{aligned} y[3] &= x[1]h[2] + x[2]h[1] \\ &= 0.5 \times 3.0 + 2.0 \times 2.0 = 5.5 \end{aligned}$$

# Convolution Example

Computing  $y[n] = x[n] * h[n] = \sum_{k=0}^n x[k]h[n-k]$

For  $n = 4$ , shift  $h$  right again to get  $h[4-k]$ .



$$\begin{aligned} y[4] &= x[2]h[2] \\ &= 2.0 \times 3.0 = 6.0 \end{aligned}$$

# Output Length

## Fact

*The convolution of two  $L$ -length signals will have length  $2L - 1$ .*

$$x[n] * h[n] = \sum_{k=0}^n x[k]h[n-k] \quad \text{for } n \in [0, 2L-2]$$

So, we need to pad  $h[n]$  with zeros on the right, from  $n = [L, 2L - 2]$ .

# Differing Length Inputs

## Fact

*If  $x[n]$  has length  $L_x$  and  $h[n]$  has length  $L_h$ , then  $x[n] * h[n]$  has length  $L_x + L_h - 1$ .*

Need to pad  $h[n]$  with zeros to the right, for  $n = [L_h, L_x + L_h - 2]$ .

**Note:** It's cheaper to have the longer length signal on the right! (less padding) Because of commutativity, we can always swap  $x[n] * h[n] = h[n] * x[n]$ .