

# All-Pass and Minimum Phase Systems

Digital Signal Processing

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# Review: Rational Transfer Functions

A **rational transfer function** looks like

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\sum_{k=0}^M b_k z^{-k}}{\sum_{k=0}^N a_k z^{-k}}$$

Applying the FTOA, we can factor the numerator and denominator:

$$H(z) = \left( \frac{b_0}{a_0} \right) \frac{\prod_{k=0}^M (1 - c_k z^{-1})}{\prod_{k=0}^N (1 - d_k z^{-1})}$$

The  $c_k$  are **zeros** of  $H(z)$  (zeros of the numerator).

The  $d_k$  are **poles** of  $H(z)$  (zeros of the denominator).

# All-Pass Systems

## Definition

An **all-pass system** is an LTI system whose frequency response has magnitude equal to one. In other words, its frequency response,  $H(e^{i\omega})$ , satisfies:

$$|H(e^{i\omega})| = 1, \quad \text{for all } \omega \in [-\pi, \pi).$$

# Simplest All-Pass System

Consider a transfer function  $H(z)$  of the form:

$$H(z) = \frac{z^{-1} - c}{1 - \bar{c}z^{-1}}.$$

The magnitude of its frequency response is:

$$\begin{aligned} |H(e^{i\omega})| &= \frac{|e^{-i\omega} - c|}{|1 - \bar{c}e^{-i\omega}|} && \text{plug in } z = e^{i\omega} \\ &= \frac{|e^{-i\omega}| |1 - ce^{i\omega}|}{|1 - \bar{c}e^{-i\omega}|} && \text{pull out } e^{-i\omega} \text{ factor} \\ &= \frac{|1 - ce^{i\omega}|}{|1 - \bar{c}e^{-i\omega}|} && |e^{-i\omega}| = 1 \\ &= \frac{|1 - \bar{c}e^{-i\omega}|}{|1 - \bar{c}e^{-i\omega}|} = 1 && \text{conjugates} \end{aligned}$$

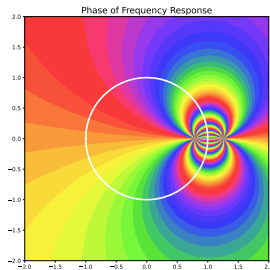
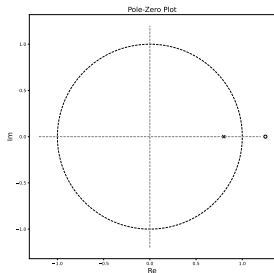
# Simplest All-Pass System

Rearranging the simple all-pass system:

$$H(z) = \frac{z^{-1} - c}{1 - \bar{c}z^{-1}} = -c \frac{1 - c^{-1}z^{-1}}{1 - \bar{c}z^{-1}}.$$

We get a zero at  $c^{-1}$  and a pole at  $\bar{c}$ .

Let  $\bar{c} = re^{-i\omega}$ . Then  $c^{-1} = \frac{1}{r}e^{-i\omega}$ . This is the pole-zero inverse pair relationship we saw last time:



# Real Coefficient All-Pass System

The transfer function  $H(z) = \frac{z^{-1}-c}{1-\bar{c}z^{-1}}$  represents the system:

$$y[n] = -cx[n] + x[n-1] + \bar{c}y[n-1].$$

It has complex coefficients.

If we want a real-valued system, we add a conjugate zero / pole

$$H(z) = \frac{(z^{-1} - c)(z^{-1} - \bar{c})}{(1 - \bar{c}z^{-1})(1 - cz^{-1})} = \frac{z^{-2} - 2\text{Re}(c)z^{-1} + |c|^2}{1 - 2\text{Re}(c)z^{-1} + |c|^2z^{-2}},$$

which now is a system with only real-valued coefficients.

# General Real-Valued All-Pass System

The general form of a real-valued all-pass system is

$$H(z) = \prod_{j=1}^J \frac{z^{-1} - d_j}{1 - d_j z^{-1}} \prod_{k=1}^K \frac{(z^{-1} - c_k)(z^{-1} - \bar{c}_k)}{(1 - \bar{c}_k z^{-1})(1 - c_k z^{-1})},$$

where  $c_j \in \mathbb{C}$  and  $d_j \in \mathbb{R}$ .

# Inverse Systems

Can we “undo” an LTI system? That is, given an output

$$y[n] = h[n] * x[n],$$

can we get back the input signal  $x[n]$ ?

This means we want an inverse system,  $h^{-1}[n]$ , such that:

$$h^{-1}[n] * (h[n] * x[n]) = x[n], \quad \text{for all signals } x[n].$$

This implies  $h^{-1}[n] * h[n] = \delta[n]$ .

Taking the  $z$ -transform, we have

$$H^{-1}(z)H(z) = 1 \quad \implies \quad H^{-1}(z) = \frac{1}{H(z)}$$



# Inverse of a Rational Transfer Function

Let

$$H(z) = \left( \frac{b_0}{a_0} \right) \frac{\prod_{k=0}^M (1 - c_k z^{-1})}{\prod_{k=0}^N (1 - d_k z^{-1})}$$

Then the inverse just flips the numerator and denominator:

$$H^{-1}(z) = \frac{1}{H(z)} = \left( \frac{a_0}{b_0} \right) \frac{\prod_{k=0}^N (1 - d_k z^{-1})}{\prod_{k=0}^M (1 - c_k z^{-1})}$$

The poles of  $H(z)$  become zeros of  $H^{-1}(z)$ , and zeros become poles.

# Exercise: Inverse of FBCF

What is the inverse system for the FBCF?

$$y[n] = x[n] + gy[n - k]$$

# Solution

Transfer function for FBCF:

$$H(z) = \frac{1}{1 - gz^{-k}}.$$

Inverse transfer function:

$$H^{-1}(z) = 1 - gz^{-k}.$$

This is the FFCF! (with negative gain)

# Minimum-Phase Systems

## Definition

A **minimum-phase system** is an LTI system that is stable, causal, and whose inverse is also stable and causal.

Because the poles and zeros flip roles in the inverse, a **minimum-phase system must have all of its poles and zeros inside the unit circle.**

# Decomposition of Stable, Causal Systems

## Theorem

*Let  $H(z)$  be the transfer function for a stable, causal LTI system. Then  $H(z)$  can be decomposed into a product*

$$H(z) = H_{\min}(z)H_{\text{ap}}(z),$$

*where  $H_{\min}(z)$  is a minimum-phase system, and  $H_{\text{ap}}(z)$  is an all-pass system.*