# Phase and Group Delay of Frequency Response

Digital Signal Processing

March 21, 2024



#### **Review: Transfer Function**

Given an LTI system with impulse response h[n]:

$$y[n] = x[n] * h[n].$$

The **transfer function** is the z-transform of h[n]. It is given by the ratio:

$$H(z) = \frac{Y(z)}{X(z)},$$

where  $x[n] \overset{\mathcal{Z}}{\longleftrightarrow} X(z)$  and  $y[n] \overset{\mathcal{Z}}{\longleftrightarrow} Y(z)$ .

#### Review: Transfer Function of LCCDE

A linear constant-coefficient difference equation (LCCDE) is an LTI system of the form:

$$\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k].$$

It's transfer function is

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\sum_{k=0}^{M} b_k z^{-k}}{\sum_{k=0}^{N} a_k z^{-k}}.$$

#### **Review: Frequency Response**

Looking at frequency response of an LTI:

$$Y(e^{i\omega}) = H(e^{i\omega})X(e^{i\omega})$$

Remember complex multiplication in Euler form:

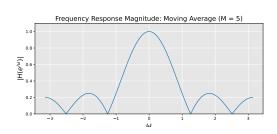
$$re^{i\theta} \cdot se^{i\phi} = (rs)e^{i(\theta+\phi)}$$

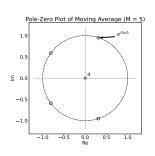
So, we have:

Magnitude:  $|Y(e^{i\omega})| = |H(e^{i\omega})| \cdot |X(e^{i\omega})|$ 

Phase:  $\operatorname{Arg}(Y(e^{i\omega})) = \operatorname{Arg}(H(e^{i\omega})) + \operatorname{Arg}(X(e^{i\omega}))$ 

## **Review: Moving Average (MA)**

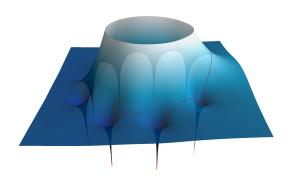




$$H(z) = \frac{1}{5} \sum_{k=0}^{4} z^{-k} = \frac{\prod_{k=1}^{4} (z - b_k)}{z^4},$$

where 
$$b_k = e^{\frac{i2\pi k}{5}}$$
.

#### **Moving Average Transfer Function**



Magnitude of full *z*-Transform, |H(z)|, as height function.

Frequency response is the red curve.

 $\textbf{From https://tttapa.github.io/Pages/Mathematics/Systems-and-Control-Theory/Digital-filters/Systems-and-Control-Theory/Systems-and-Control-Theory/Systems-and-Control-Theory/Systems-and-Control-Theory/Systems-and-Control-Theory/Systems-an$ 

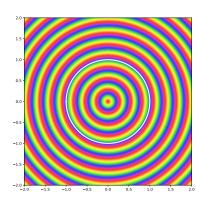
#### **Visualization of Complex Functions**

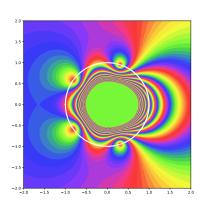
Can't directly plot  $H:\mathbb{C}\to\mathbb{C}$  (there are 4 dimensions!)

Let f(z)=(R(z),G(z),B(z)) be a map that assigns a color (red/green/blue) to every point  $z\in\mathbb{C}$ .

Then we can display f(H(z)) to visualize the complex function H(z).

#### **Another Visualization of Magnitude**

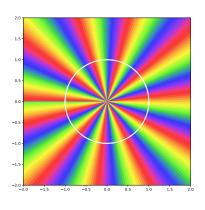




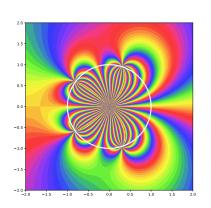
|z|

|H(z)| for MA

# What About Phase?

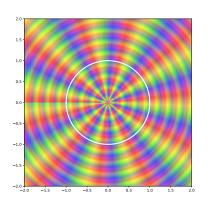


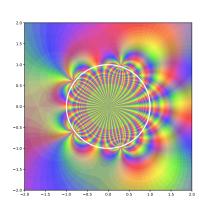
Arg(z)



Arg(H(z)) for MA

# **Visualizing Magnitude and Phase**

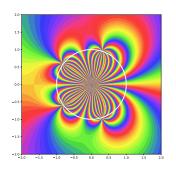


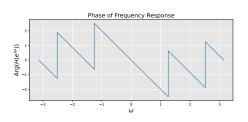


H(z) for MA

#### Phase Response of MA

Look at phase restricted to unit circle:  $Arg(H(e^{i\omega}))$ 

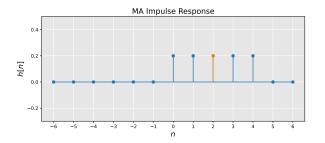




Note discontinuities (jumps by  $\pi$ ) when we hit a zero. Otherwise it is linear with slope = -2.

#### **Phase Response of MA**

What is the significance of linear phase with slope = -2?



It tells us the average delay!

#### Phase Response of Ideal Delay

Ideal Delay:

$$y[n] = x[n-k]$$

Its transfer function is:

$$H(z) = z^{-k}$$

Restricting to unit circle gives the frequency response:

$$H(e^{i\omega}) = e^{-i\omega k}$$

It has phase:

$$Arg(H(e^{i\omega})) = -\omega k,$$

which is linear in  $\omega$ , with slope = -k.

#### Phase Delay

This leads us to the following definition:

#### **Definition**

The **phase delay** of an LTI system at frequency  $\omega$  is

$$\tau_{\rm ph}(\omega) = -\frac{{\rm Arg}(H(e^{i\omega}))}{\omega}.$$

#### **Another Way to Get Slope**

If  $Arg(H(e^{i\omega})) = -\omega k$ , I can also recover the slope, -k, as the **derivative** w.r.t.  $\omega$ . This leads to what is called the group delay.

#### **Definition**

The **group delay** of an LTI system at frequency  $\omega$  is

$$\tau_{\rm gr}(\omega) = -\frac{d}{d\omega} \text{Arg}(H(e^{i\omega}))$$

# Why Two Types of Delays?

For an ideal delay, y[n] = x[n-k], we have seen both phase and group delay are the same:

$$\tau_{\rm ph}(\omega) = \tau_{\rm gr}(\omega) = k.$$

More generally, these are the same whenever we have a **linear phase response** system:

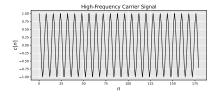
$$Arg(H(e^{i\omega})) = -\omega k.$$

But if the phase is not linear in  $\omega$ , these won't be the same:

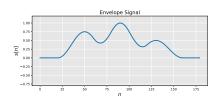
$$\tau_{\rm ph}(\omega) = -\frac{\operatorname{Arg}(H(e^{i\omega}))}{\omega} \neq -\frac{d}{d\omega}\operatorname{Arg}(H(e^{i\omega})) = \tau_{\rm gr}(\omega).$$

## Sinusoid Signal with Envelope

Consider a high-frequency sinusoid:  $c[n] = \cos(\omega_0 n)$ , which we'll call the **high-frequency carrier**:



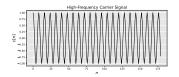
And some lower-frequency **envelope**, s[n]:



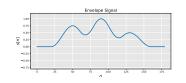
#### Sinusoid Signal with Envelope

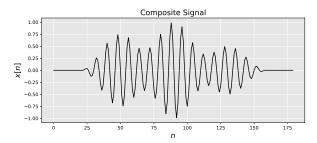
Now multiply them together to get the signal:

$$x[n] = s[n]\cos(\omega_0 n).$$





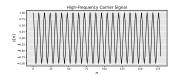




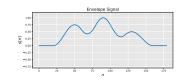
#### Sinusoid Signal with Envelope

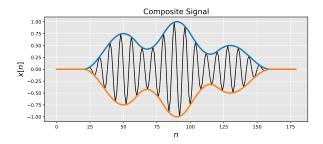
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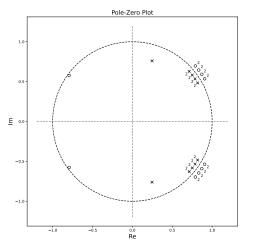
#### **Phase and Group Delay**

Now when applying a general LTI system, H(z), the phase delay will act on the high-frequency carrier, while the group delay will act on the envelope:

$$y[n] \approx |H(e^{i\omega_0})| s[n - \tau_{\rm gr}(\omega_0)] \cos(\omega_0 n + \tau_{\rm ph}(\omega_0)).$$

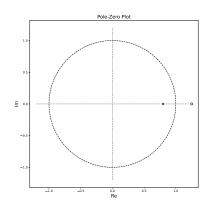
# Example<sup>1</sup>

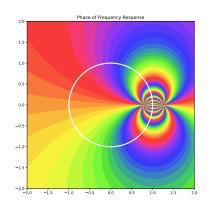
#### Consider an LTI system with the following pole-zero plot:



<sup>&</sup>lt;sup>1</sup>From Oppenheim & Schafer, pg. 278

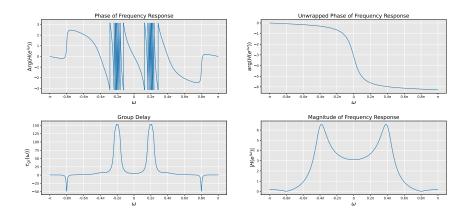
#### **Pole-Zero Inverse Pairs**





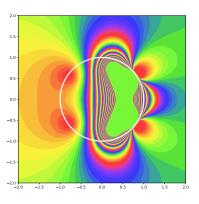
$$H(z) = \left(\frac{4}{5}\right) \frac{1 - \frac{5}{4}z^{-1}}{1 - \frac{4}{5}z^{-1}}$$

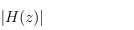
#### **Pole-Zero Inverse Pairs**

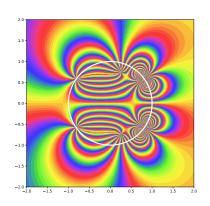


Pole-zero inverse pairs cause a narrowband group delay, while leaving the magnitude fixed.

## **Example: Transfer Function**

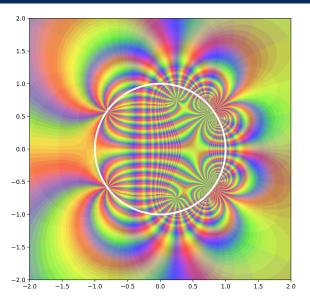






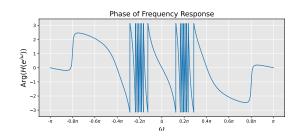
Arg(H(z))

## **Example: Transfer Function**

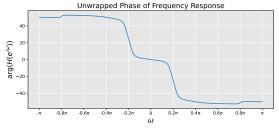


#### **Example: Phase of Frequency Response**

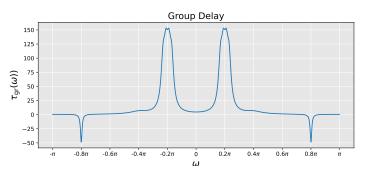
Wrapped:



Unwrapped:

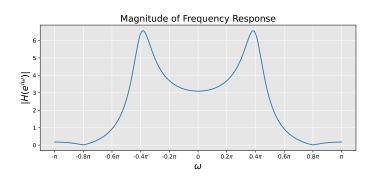


#### **Example: Group Delay**



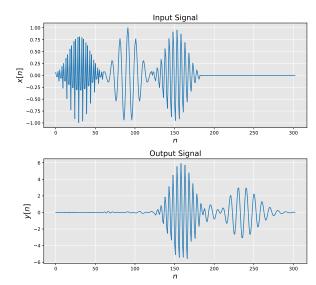
$$\tau_{\rm gr}(\omega) = -\frac{d}{d\omega} Arg(H(e^{i\omega}))$$

# **Example: Magnitude of Frequency Response**



#### **Example: Input and Output**

Input signal has three frequencies:  $0.8\pi, 0.2\pi, 0.4\pi$ 



Output:

Input: