

Filter synthesis

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TSIA201



Part I

Linear phase FIR filters

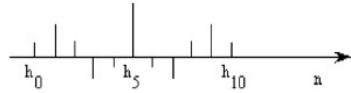
Linear phase FIR filters

- ▶ Impulse response : $h(n) \in \mathbb{R}, n = 0 \dots N-1$
- ▶ Frequency response : $H(e^{j2\pi v}) = e^{j2\pi(\beta - \alpha v)} H_R(v)$
- ▶ Constant and equal group and phase delays (if $\beta = 0$)
- ▶ Advantages : always **causal and stable**, preserve the waveform of a narrow-band signal (if $\beta = 0$)
- ▶ Drawback : high computational complexity

Characterization

- ▶ **1 - periodicity** of $H(e^{j2\pi v}) \Rightarrow \alpha = p/2, p \in \mathbb{Z}$ and H_R is 2-periodic
- ▶ **Hermitian symmetry** $\Rightarrow d = e^{2j\pi\beta} = 1$ or j and H_R is even or odd
- ▶ As $H_R(v)$ is 2-periodic, we can define $G(e^{j2\pi v}) = d H_R(2v)$ where $g(n)$ is real, even or odd
- ▶ We can thus write $H(z^2) = z^{-p} G(z)$ (filter of length $2N-1$)
- ▶ We choose $p = N-1$ for a causal filter
- ▶ **4 possibilities** depending on d and the parity of N :
 - ▶ $d = 1, N$ even or odd : $h(n) = h(N-1-n)$
 - ▶ $d = j, N$ even or odd : $h(n) = -h(N-1-n)$

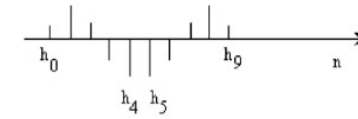
- ▶ Type 1 : N odd, symmetric ($d = 1$)



$$H(e^{j2\pi\nu}) = e^{-j2\pi\nu \frac{N-1}{2}} \sum_{n=0}^{\frac{N-1}{2}} a_n \cos(2\pi\nu n)$$

- ▶ Use : low-pass, high-pass, band-pass

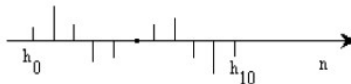
- ▶ Type 2 : N even, symmetric ($d = 1$)



$$H(e^{j2\pi\nu}) = e^{-j2\pi\nu \frac{N-1}{2}} \sum_{n=1}^{\frac{N-1}{2}} b_n \cos\left(2\pi\nu \left(n - \frac{1}{2}\right)\right)$$

- ▶ Property : $H(-1) = 0$ ($\nu = \frac{1}{2}$)
- ▶ Use : low-pass, band-pass

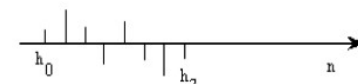
- ▶ Type 3 : N odd, antisymmetric ($d = i$)



$$H(e^{j2\pi\nu}) = i e^{-j2\pi\nu \frac{N-1}{2}} \sum_{n=1}^{\frac{N-1}{2}} c_n \sin(2\pi\nu n)$$

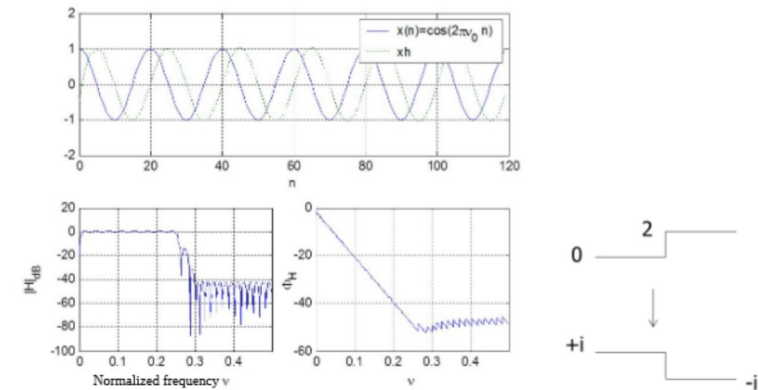
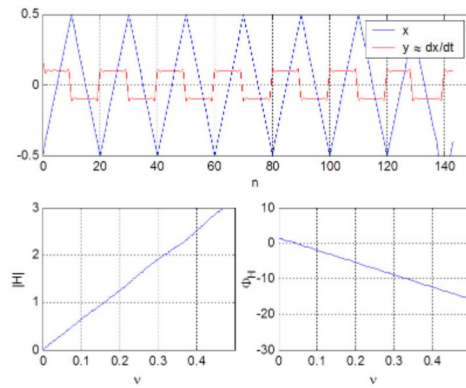
- ▶ Property : $H(1) = H(-1) = 0$ ($\nu = 0$ or $\frac{1}{2}$)
- ▶ Use : differentiator ($H(f) = i2\pi f$), Hilbert transform ($H(f) = -i \text{sign}(f)$) in band-pass form

- ▶ Type 4 : N even, antisymmetric ($d = i$)



$$H(e^{j2\pi\nu}) = i e^{-j2\pi\nu \frac{N-1}{2}} \sum_{n=1}^{\frac{N-1}{2}} d_n \sin\left(2\pi\nu \left(n - \frac{1}{2}\right)\right)$$

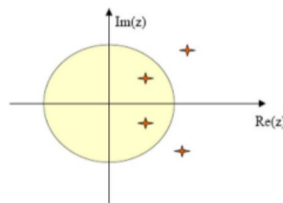
- ▶ Property : $H(1) = 0$ ($\nu = 0$)
- ▶ Use : differentiator ($H(f) = i2\pi f$), Hilbert transform ($H(f) = -i \text{sign}(f)$) in high-pass form



Position of the zeros

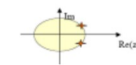
- Complex zero outside the unit circle :

$$(1 - \rho e^{i\theta} z^{-1})(1 - \rho e^{-i\theta} z^{-1})(1 - \frac{1}{\rho} e^{i\theta} z^{-1})(1 - \frac{1}{\rho} e^{-i\theta} z^{-1})$$

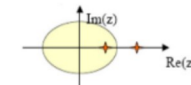


Position of the zeros

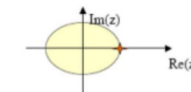
- Zero on the unit circle



- Real zero



- Real zero on the unit circle



Type I N odd symmetric		-	Low-pass High-pass Band-pass
Type II N even symmetric		$H(-1) = 0$	Low-pass, Band-pass
Type III N odd antisym.		$H(1) = 0$ $H(-1) = 0$	Differentiator, Hilbert Transform, Band-pass
Type IV N even antisym.		$H(1) = 0$	Differentiator, Hilbert Transform, High-pass

Part II

FIR filters: iterative methods

Iterative methods

- ▶ Advantages
 - ▶ Optimal design
 - ▶ Flexible method
 - ▶ Constant amplitude ripples
 - ▶ Minimum order for a given template
- ▶ Drawbacks
 - ▶ Computationally demanding synthesis
 - ▶ Not suitable for real-time processing
 - ▶ Not suitable for long filters (numerical stability issues)

Filter template

