An  $(n \times n)$  square matrix **A** is said to be *positive semi-definite* if

$$\boldsymbol{x}^T \boldsymbol{A} \boldsymbol{x} \ge 0 \qquad \forall \boldsymbol{x}. \tag{A.56}$$

This definition implies that  $\varrho(\mathbf{A}) = r < n$ , and thus r eigenvalues of  $\mathbf{A}$  are positive and n - r are null. Therefore, a positive semi-definite matrix  $\mathbf{A}$  has a null space of finite dimension, and specifically the form vanishes when  $\mathbf{x} \in \mathcal{N}(\mathbf{A})$ . A typical example of a positive semi-definite matrix is the matrix  $\mathbf{A} = \mathbf{H}^T \mathbf{H}$  where  $\mathbf{H}$  is an  $(m \times n)$  matrix with m < n. In an analogous way, a negative semi-definite matrix can be defined.

Given the bilinear form in (A.51), the gradient of the form with respect to  $\boldsymbol{x}$  is given by

$$\nabla_{\boldsymbol{x}} B(\boldsymbol{x}, \boldsymbol{y}) = \left(\frac{\partial B(\boldsymbol{x}, \boldsymbol{y})}{\partial \boldsymbol{x}}\right)^{T} = \boldsymbol{A} \boldsymbol{y}, \tag{A.57}$$

whereas the gradient of B with respect to y is given by

$$\nabla_{\boldsymbol{y}} B(\boldsymbol{x}, \boldsymbol{y}) = \left(\frac{\partial B(\boldsymbol{x}, \boldsymbol{y})}{\partial \boldsymbol{y}}\right)^T = \boldsymbol{A}^T \boldsymbol{x}. \tag{A.58}$$

Given the quadratic form in (A.52) with  $\boldsymbol{A}$  symmetric, the gradient of the form with respect to  $\boldsymbol{x}$  is given by

$$\nabla_{\boldsymbol{x}} Q(\boldsymbol{x}) = \left(\frac{\partial Q(\boldsymbol{x})}{\partial \boldsymbol{x}}\right)^T = 2\boldsymbol{A}\boldsymbol{x}.$$
 (A.59)

Further, if x and A are differentiable functions of t, then

$$\dot{Q}(x) = \frac{d}{dt}Q(\boldsymbol{x}(t)) = 2\boldsymbol{x}^T \boldsymbol{A} \dot{\boldsymbol{x}} + \boldsymbol{x}^T \dot{\boldsymbol{A}} \boldsymbol{x}; \tag{A.60}$$

if A is constant, then the second term obviously vanishes.

#### A.7 Pseudo-inverse

The inverse of a matrix can be defined only when the matrix is square and nonsingular. The inverse operation can be extended to the case of non-square matrices. Consider a matrix  $\mathbf{A}$  of dimensions  $(m \times n)$  with  $\varrho(\mathbf{A}) = \min\{m, n\}$ 

If m < n, a right inverse of A can be defined as the matrix  $A_r$  of dimensions  $(n \times m)$  so that

$$AA_r = I_m$$
.

If instead m > n, a left inverse of A can be defined as the matrix  $A_l$  of dimensions  $(n \times m)$  so that

$$A_l A = I_n$$

If A has more columns than rows (m < n) and has rank m, a special right inverse is the matrix

$$\boldsymbol{A}_r^{\dagger} = \boldsymbol{A}^T (\boldsymbol{A} \boldsymbol{A}^T)^{-1} \tag{A.61}$$

which is termed right pseudo-inverse, since  $AA_r^{\dagger} = I_m$ . If  $W_r$  is an  $(n \times n)$  positive definite matrix, a weighted right pseudo-inverse is given by

$$\mathbf{A}_{r}^{\dagger} = \mathbf{W}_{r}^{-1} \mathbf{A}^{T} (\mathbf{A} \mathbf{W}_{r}^{-1} \mathbf{A}^{T})^{-1}. \tag{A.62}$$

If **A** has more rows than columns (m > n) and has rank n, a special left inverse is the matrix

$$\boldsymbol{A}_{l}^{\dagger} = (\boldsymbol{A}^{T} \boldsymbol{A})^{-1} \boldsymbol{A}^{T} \tag{A.63}$$

which is termed *left pseudo-inverse*, since  $\mathbf{A}_l^{\dagger} \mathbf{A} = \mathbf{I}_n$ .<sup>3</sup> If  $\mathbf{W}_l$  is an  $(m \times m)$  positive definite matrix, a weighted left pseudo-inverse is given by

$$\boldsymbol{A}_{l}^{\dagger} = (\boldsymbol{A}^{T} \boldsymbol{W}_{l} \boldsymbol{A})^{-1} \boldsymbol{A}^{T} \boldsymbol{W}_{l}. \tag{A.64}$$

The pseudo-inverse is very useful to invert a linear transformation y = Ax with A a full-rank matrix. If A is a square nonsingular matrix, then obviously  $x = A^{-1}y$  and then  $A_l^{\dagger} = A_r^{\dagger} = A^{-1}$ .

If A has more columns than rows (m < n) and has rank m, then the solution x for a given y is not unique; it can be shown that the expression

$$x = A^{\dagger} y + (I - A^{\dagger} A) k, \tag{A.65}$$

with k an arbitrary  $(n \times 1)$  vector and  $\mathbf{A}^{\dagger}$  as in (A.61), is a solution to the system of linear equations established by (A.35). The term  $\mathbf{A}^{\dagger}\mathbf{y} \in \mathcal{N}^{\perp}(\mathbf{A}) \equiv \mathcal{R}(\mathbf{A}^T)$  minimizes the norm of the solution  $\|\mathbf{x}\|$ . The term  $(\mathbf{I} - \mathbf{A}^{\dagger}\mathbf{A})\mathbf{k}$  is the projection of  $\mathbf{k}$  in  $\mathcal{N}(\mathbf{A})$  and is termed homogeneous solution; as  $\mathbf{k}$  varies, all the solutions to the homogeneous equation system  $\mathbf{A}\mathbf{x} = \mathbf{0}$  associated with (A.35) are generated.

On the other hand, if A has more rows than columns (m > n), the equation in (A.35) has no solution; it can be shown that an *approximate* solution is given by

$$\boldsymbol{x} = \boldsymbol{A}^{\dagger} \boldsymbol{y} \tag{A.66}$$

where  $\mathbf{A}^{\dagger}$  as in (A.63) minimizes  $\|\mathbf{y} - \mathbf{A}\mathbf{x}\|$ . If instead  $\mathbf{y} \in \mathcal{R}(\mathbf{A})$ , then (A.66) is a real solution.

Notice that the use of the weighted (left or right) pseudo-inverses in the solution to the linear equation systems leads to analogous results where the minimized norms are weighted according to the metrics defined by matrices  $W_r$  and  $W_l$ , respectively.

The results of this section can be easily extended to the case of (square or nonsquare) matrices  $\boldsymbol{A}$  not having full-rank. In particular, the expression (A.66) (with the pseudo-inverse computed by means of the singular value decomposition of  $\boldsymbol{A}$ ) gives the minimum-norm vector among all those minimizing  $\|\boldsymbol{y}-\boldsymbol{A}\boldsymbol{x}\|$ .

<sup>&</sup>lt;sup>3</sup> Subscripts l and r are usually omitted whenever the use of a left or right pseudo-inverse is clear from the context.

#### A.8 Singular Value Decomposition

For a nonsquare matrix it is not possible to define eigenvalues. An extension of the eigenvalue concept can be obtained by singular values. Given a matrix  $\mathbf{A}$  of dimensions  $(m \times n)$ , the matrix  $\mathbf{A}^T \mathbf{A}$  has n nonnegative eigenvalues  $\lambda_1 \geq \lambda_2 \geq \ldots \geq \lambda_n \geq 0$  (ordered from the largest to the smallest) which can be expressed in the form

$$\lambda_i = \sigma_i^2 \qquad \sigma_i \ge 0.$$

The scalars  $\sigma_1 \geq \sigma_2 \geq \ldots \geq \sigma_n \geq 0$  are said to be the *singular values* of matrix A. The *singular value decomposition* (SVD) of matrix A is given by

$$\mathbf{A} = \mathbf{U}\boldsymbol{\Sigma}\mathbf{V}^T \tag{A.67}$$

where U is an  $(m \times m)$  orthogonal matrix

$$\boldsymbol{U} = [\boldsymbol{u}_1 \quad \boldsymbol{u}_2 \quad \dots \quad \boldsymbol{u}_m], \tag{A.68}$$

V is an  $(n \times n)$  orthogonal matrix

$$\boldsymbol{V} = [\boldsymbol{v}_1 \quad \boldsymbol{v}_2 \quad \dots \quad \boldsymbol{v}_n] \tag{A.69}$$

and  $\Sigma$  is an  $(m \times n)$  matrix

$$\Sigma = \begin{bmatrix} D & O \\ O & O \end{bmatrix} \qquad D = \operatorname{diag}\{\sigma_1, \sigma_2, \dots, \sigma_r\}$$
 (A.70)

where  $\sigma_1 \geq \sigma_2 \geq \ldots \geq \sigma_r > 0$ . The number of non-null singular values is equal to the rank r of matrix A.

The columns of U are the eigenvectors of the matrix  $AA^T$ , whereas the columns of V are the eigenvectors of the matrix  $A^TA$ . In view of the partitions of U and V in (A.68), (A.69), it is  $Av_i = \sigma_i u_i$ , for i = 1, ..., r and  $Av_i = 0$ , for i = r + 1, ..., n.

Singular value decomposition is useful for analysis of the linear transformation  $\mathbf{y} = A\mathbf{x}$  established in (A.35). According to a geometric interpretation, the matrix  $\mathbf{A}$  transforms the unit sphere in  $\mathbb{R}^n$  defined by  $\|\mathbf{x}\| = 1$  into the set of vectors  $\mathbf{y} = A\mathbf{x}$  which define an *ellipsoid* of dimension r in  $\mathbb{R}^m$ . The singular values are the lengths of the various axes of the ellipsoid. The *condition number* of the matrix

$$\kappa = \frac{\sigma_1}{\sigma_r}$$

is related to the eccentricity of the ellipsoid and provides a measure of ill-conditioning ( $\kappa \gg 1$ ) for numerical solution of the system established by (A.35).

It is worth noticing that the numerical procedure of singular value decomposition is commonly adopted to compute the (right or left) pseudo-inverse  $A^{\dagger}$ , even in the case of a matrix A not having full rank. In fact, from (A.67), (A.70) it is

$$\boldsymbol{A}^{\dagger} = \boldsymbol{V} \boldsymbol{\Sigma}^{\dagger} \boldsymbol{U}^{T} \tag{A.71}$$

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with

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$$\Sigma^{\dagger} = \begin{bmatrix} D^{\dagger} & O \\ O & O \end{bmatrix} \qquad D^{\dagger} = \operatorname{diag} \left\{ \frac{1}{\sigma_1}, \frac{1}{\sigma_2}, \dots, \frac{1}{\sigma_r} \right\}.$$
 (A.72)

## **Bibliography**

A reference text on linear algebra is [169]. For matrix computation see [88]. The properties of pseudo-inverse matrices are discussed in [24].

# Rigid-body Mechanics

The goal of this appendix is to recall some fundamental concepts of *rigid body* mechanics which are preliminary to the study of manipulator kinematics, statics and dynamics.

#### **B.1 Kinematics**

A rigid body is a system characterized by the constraint that the distance between any two points is always constant.

Consider a rigid body  $\mathcal{B}$  moving with respect to an orthonormal reference frame O-xyz of unit vectors x, y, z, called *fixed frame*. The rigidity assumption allows the introduction of an orthonormal frame O'-x'y'z' attached to the body, called *moving frame*, with respect to which the position of any point of  $\mathcal{B}$  is independent of time. Let x'(t), y'(t), z'(t) be the unit vectors of the moving frame expressed in the fixed frame at time t.

The orientation of the moving frame O'-x'y'z' at time t with respect to the fixed frame O-xyz can be expressed by means of the *orthogonal*  $(3 \times 3)$  matrix

$$\boldsymbol{R}(t) = \begin{bmatrix} \boldsymbol{x}'^{T}(t)\boldsymbol{x} & \boldsymbol{y}'^{T}(t)\boldsymbol{x} & \boldsymbol{z}'^{T}(t)\boldsymbol{x} \\ \boldsymbol{x}'^{T}(t)\boldsymbol{y} & \boldsymbol{y}'^{T}(t)\boldsymbol{y} & \boldsymbol{z}'^{T}(t)\boldsymbol{y} \\ \boldsymbol{x}'^{T}(t)\boldsymbol{z} & \boldsymbol{y}'^{T}(t)\boldsymbol{z} & \boldsymbol{z}'^{T}(t)\boldsymbol{z} \end{bmatrix},$$
(B.1)

which is termed rotation matrix defined in the orthonormal special group SO(3) of the  $(3 \times 3)$  matrices with orthonormal columns and determinant equal to 1. The columns of the matrix in (B.1) represent the components of the unit vectors of the moving frame when expressed in the fixed frame, whereas the rows represent the components of the unit vectors of the fixed frame when expressed in the moving frame.

Let p' be the *constant* position vector of a generic point P of  $\mathcal{B}$  in the moving frame O'-x'y'z'. The motion of P with respect to the fixed frame O-xyz is described by the equation

$$\mathbf{p}(t) = \mathbf{p}_{O'}(t) + \mathbf{R}(t)\mathbf{p}', \tag{B.2}$$

where  $p_{O'}(t)$  is the position vector of origin O' of the moving frame with respect to the fixed frame.

Notice that a position vector is a *bound vector* since its line of application and point of application are both prescribed, in addition to its direction; the point of application typically coincides with the origin of a reference frame. Therefore, to transform a bound vector from a frame to another, both translation and rotation between the two frames must be taken into account.

If the positions of the points of  $\mathcal{B}$  in the moving frame are known, it follows from (B.2) that the motion of each point of  $\mathcal{B}$  with respect to the fixed frame is uniquely determined once the position of the origin and the orientation of the moving frame with respect to the fixed frame are specified in time. The origin of the moving frame is determined by three scalar functions of time. Since the orthonormality conditions impose six constraints on the nine elements of matrix  $\mathbf{R}(t)$ , the orientation of the moving frame depends only on three independent scalar functions, three being the minimum number of parameters to represent SO(3).

Therefore, a rigid body motion is described by arbitrarily specifying six scalar functions of time, which describe the body pose (position + orientation). The resulting rigid motions belong to the *special Euclidean group*  $SE(3) = \mathbb{R}^3 \times SO(3)$ .

The expression in (B.2) continues to hold if the position vector  $\mathbf{p}_{O'}(t)$  of the origin of the moving frame is replaced with the position vector of any other point of  $\mathcal{B}$ , i.e.,

$$\boldsymbol{p}(t) = \boldsymbol{p}_O(t) + \boldsymbol{R}(t)(\boldsymbol{p}' - \boldsymbol{p}_O') \tag{B.3}$$

where  $p_Q(t)$  and  $p'_Q$  are the position vectors of a point Q of  $\mathcal{B}$  in the fixed and moving frames, respectively.

In the following, for simplicity of notation, the dependence on the time variable t will be dropped.

Differentiating (B.3) with respect to time gives the known velocity composition rule

$$\dot{\boldsymbol{p}} = \dot{\boldsymbol{p}}_Q + \boldsymbol{\omega} \times (\boldsymbol{p} - \boldsymbol{p}_Q), \tag{B.4}$$

where  $\omega$  is the angular velocity of rigid body  $\mathcal{B}$ . Notice that  $\omega$  is a free vector since its point of application is not prescribed. To transform a free vector from a frame to another, only rotation between the two frames must be taken into account.

By recalling the definition of the skew-symmetric operator  $S(\cdot)$  in (A.32), the expression in (B.4) can be rewritten as

$$\begin{split} \dot{\boldsymbol{p}} &= \dot{\boldsymbol{p}}_Q + \boldsymbol{S}(\boldsymbol{\omega})(\boldsymbol{p} - \boldsymbol{p}_Q) \\ &= \dot{\boldsymbol{p}}_Q + \boldsymbol{S}(\boldsymbol{\omega})\boldsymbol{R}(\boldsymbol{p}' - \boldsymbol{p}_Q'). \end{split}$$

<sup>&</sup>lt;sup>1</sup> The minimum number of parameters represent a special orthonormal group SO(m) is equal to m(m-1)/2.

Comparing this equation with the formal time derivative of (B.3) leads to the result

$$\dot{\mathbf{R}} = \mathbf{S}(\boldsymbol{\omega})\mathbf{R}.\tag{B.5}$$

In view of (B.4), the *elementary displacement* of a point P of the rigid body  $\mathcal{B}$  in the time interval (t, t + dt) is

$$d\mathbf{p} = \dot{\mathbf{p}}dt = (\dot{\mathbf{p}}_Q + \boldsymbol{\omega} \times (\mathbf{p} - \mathbf{p}_Q))dt$$

$$= d\mathbf{p}_Q + \boldsymbol{\omega}dt \times (\mathbf{p} - \mathbf{p}_Q).$$
(B.6)

Differentiating (B.4) with respect to time yields the following expression for acceleration:

$$\ddot{\boldsymbol{p}} = \ddot{\boldsymbol{p}}_O + \dot{\boldsymbol{\omega}} \times (\boldsymbol{p} - \boldsymbol{p}_O) + \boldsymbol{\omega} \times (\boldsymbol{\omega} \times (\boldsymbol{p} - \boldsymbol{p}_O)). \tag{B.7}$$

#### **B.2 Dynamics**

Let  $\rho dV$  be the mass of an elementary particle of a rigid body  $\mathcal{B}$ , where  $\rho$  denotes the density of the particle of volume dV. Also let  $V_{\mathcal{B}}$  be the body volume and  $m = \int_{V_{\mathcal{B}}} \rho dV$  its total mass assumed to be constant. If  $\boldsymbol{p}$  denotes the position vector of the particle of mass  $\rho dV$  in the frame O-xyz, the centre of mass of  $\mathcal{B}$  is defined as the point C whose position vector is

$$\mathbf{p}_C = \frac{1}{m} \int_{V_{\mathcal{B}}} \mathbf{p} \rho dV. \tag{B.8}$$

In the case when  $\mathcal{B}$  is the union of n distinct parts of mass  $m_1, \ldots, m_n$  and centres of mass  $p_{C_1} \ldots p_{C_n}$ , the centre of mass of  $\mathcal{B}$  can be computed as

$$\boldsymbol{p}_C = \frac{1}{m} \sum_{i=1}^n m_i \boldsymbol{p}_{Ci}$$

with  $m = \sum_{i=1}^{n} m_i$ .

Let r be a line passing by O and  $d(\mathbf{p})$  the distance from r of the particle of  $\mathcal{B}$  of mass  $\rho dV$  and position vector  $\mathbf{p}$ . The moment of inertia of body  $\mathcal{B}$  with respect to line r is defined as the positive scalar

$$I_r = \int_{V_{\mathcal{B}}} d^2(\boldsymbol{p}) \rho dV.$$

Let r denote the unit vector of line r; then, the moment of inertia of  $\mathcal{B}$  with respect to line r can be expressed as

$$I_r = \mathbf{r}^T \left( \int_{V_{\mathbf{p}}} \mathbf{S}^T(\mathbf{p}) \mathbf{S}(\mathbf{p}) \rho dV \right) \mathbf{r} = \mathbf{r}^T \mathbf{I}_O \mathbf{r}, \tag{B.9}$$

where  $S(\cdot)$  is the skew-symmetric operator in (A.31), and the *symmetric*, positive definite matrix

$$I_{O} = \begin{bmatrix} \int_{V_{\mathcal{B}}} (p_{y}^{2} + p_{z}^{2}) \rho dV & -\int_{V_{\mathcal{B}}} p_{x} p_{y} \rho dV & -\int_{V_{\mathcal{B}}} p_{x} p_{z} \rho dV \\ * & \int_{V_{\mathcal{B}}} (p_{x}^{2} + p_{z}^{2}) \rho dV & -\int_{V_{\mathcal{B}}} p_{y} p_{z} \rho dV \\ * & * & \int_{V_{\mathcal{B}}} (p_{x}^{2} + p_{y}^{2}) \rho dV \end{bmatrix}$$

$$= \begin{bmatrix} I_{Oxx} & -I_{Oxy} & -I_{Oxz} \\ * & I_{Oyy} & -I_{Oyz} \\ * & * & I_{Ozz} \end{bmatrix}$$
(B.10)

is termed inertia tensor of body  $\mathcal{B}$  relative to pole O.<sup>2</sup> The (positive) elements  $I_{Oxx}$ ,  $I_{Oyy}$ ,  $I_{Ozz}$  are the inertia moments with respect to three coordinate axes of the reference frame, whereas the elements  $I_{Oxy}$ ,  $I_{Oxz}$ ,  $I_{Oyz}$  (of any sign) are said to be products of inertia.

The expression of the inertia tensor of a rigid body  $\mathcal{B}$  depends both on the pole and the reference frame. If orientation of the reference frame with origin at O is changed according to a rotation matrix  $\mathbf{R}$ , the inertia tensor  $\mathbf{I}'_O$  in the new frame is related to  $\mathbf{I}_O$  by the relationship

$$I_O = RI_O'R^T. (B.11)$$

The way an inertia tensor is transformed when the pole is changed can be inferred by the following equation, also known as *Steiner theorem* or parallel axis theorem:

$$I_O = I_C + mS^T(\mathbf{p}_C)S(\mathbf{p}_C), \tag{B.12}$$

where  $I_C$  is the inertia tensor relative to the centre of mass of  $\mathcal{B}$ , when expressed in a frame parallel to the frame with origin at O and with origin at the centre of mass C.

Since the inertia tensor is a symmetric positive definite matrix, there always exists a reference frame in which the inertia tensor attains a diagonal form; such a frame is said to be a *principal frame* (relative to pole O) and its coordinate axes are said to be *principal axes*. In the case when pole O coincides with the centre of mass, the frame is said to be a *central frame* and its axes are said to be *central axes*.

Notice that if the rigid body is moving with respect to the reference frame with origin at O, then the elements of the inertia tensor  $I_O$  become a function of time. With respect to a pole and a reference frame attached to the body (moving frame), instead, the elements of the inertia tensor represent six structural constants of the body which are known once the pole and reference frame have been specified.

<sup>&</sup>lt;sup>2</sup> The symbol '\*' has been used to avoid rewriting the symmetric elements.

Let  $\dot{\boldsymbol{p}}$  be the velocity of a particle of  $\mathcal{B}$  of elementary mass  $\rho dV$  in frame O-xyz. The linear momentum of body  $\mathcal{B}$  is defined as the vector

$$\boldsymbol{l} = \int_{V_{\mathcal{B}}} \dot{\boldsymbol{p}} \rho dV = m \dot{\boldsymbol{p}}_{C}. \tag{B.13}$$

Let  $\Omega$  be any point in space and  $p_{\Omega}$  its position vector in frame O-xyz; then, the angular momentum of body  $\mathcal{B}$  relative to pole  $\Omega$  is defined as the vector

$$oldsymbol{k}_{arOmega} = \int_{V_{oldsymbol{k}}} \dot{oldsymbol{p}} imes (oldsymbol{p}_{arOmega} - oldsymbol{p}) 
ho dV.$$

The pole can be either fixed or moving with respect to the reference frame. The angular momentum of a rigid body has the following notable expression:

$$\boldsymbol{k}_{\Omega} = \boldsymbol{I}_{C}\boldsymbol{\omega} + m\dot{\boldsymbol{p}}_{C} \times (\boldsymbol{p}_{C} - \boldsymbol{p}_{C}), \tag{B.14}$$

where  $I_C$  is the inertia tensor relative to the centre of mass, when expressed in a frame parallel to the reference frame with origin at the centre of mass.

The *forces* acting on a generic system of material particles can be distinguished into *internal* forces and *external* forces.

The internal forces, exerted by one part of the system on another, have null linear and angular momentum and thus they do not influence rigid body motion.

The external forces, exerted on the system by an agency outside the system, in the case of a rigid body  $\mathcal{B}$  are distinguished into *active* forces and reaction forces.

The active forces can be either concentrated forces or body forces. The former are applied to specific points of  $\mathcal{B}$ , whereas the latter act on all elementary particles of the body. An example of body force is the gravitational force which, for any elementary particle of mass  $\rho dV$ , is equal to  $\mathbf{g}_0 \rho dV$  where  $\mathbf{g}_0$  is the gravity acceleration vector.

The reaction forces are those exerted because of surface contact between two or more bodies. Such forces can be distributed on the contact surfaces or they can be assumed to be concentrated.

For a rigid body  $\mathcal{B}$  subject to gravitational force, as well as to active and or reaction forces  $f_1 \dots f_n$  concentrated at points  $p_1 \dots p_n$ , the resultant of the external forces f and the resultant moment  $\mu_{\Omega}$  with respect to a pole  $\Omega$  are respectively

$$\begin{aligned} \boldsymbol{f} &= \int_{V_{\mathcal{B}}} \boldsymbol{g}_{0} \rho dV + \sum_{i=1}^{n} \boldsymbol{f}_{i} = m \boldsymbol{g}_{0} + \sum_{i=1}^{n} \boldsymbol{f}_{i} \\ \boldsymbol{\mu}_{\Omega} &= \int_{V_{\mathcal{B}}} \boldsymbol{g}_{0} \times (\boldsymbol{p}_{\Omega} - \boldsymbol{p}) \rho dV + \sum_{i=1}^{n} \boldsymbol{f}_{i} \times (\boldsymbol{p}_{\Omega} - \boldsymbol{p}_{i}) \\ &= m \boldsymbol{g}_{0} \times (\boldsymbol{p}_{\Omega} - \boldsymbol{p}_{C}) + \sum_{i=1}^{n} \boldsymbol{f}_{i} \times (\boldsymbol{p}_{\Omega} - \boldsymbol{p}_{i}). \end{aligned} \tag{B.15}$$

In the case when f and  $\mu_{\Omega}$  are known and it is desired to compute the resultant moment with respect to a point  $\Omega'$  other than  $\Omega$ , the following relation holds:

$$\boldsymbol{\mu}_{\Omega'} = \boldsymbol{\mu}_{\Omega} + \boldsymbol{f} \times (\boldsymbol{p}_{\Omega'} - \boldsymbol{p}_{\Omega}). \tag{B.17}$$

Consider now a generic system of material particles subject to external forces of resultant f and resultant moment  $\mu_{\Omega}$ . The motion of the system in a frame O-xyz is established by the following fundamental principles of dynamics (Newton laws of motion):

$$f = \dot{l} \tag{B.18}$$

$$\boldsymbol{\mu}_{\Omega} = \dot{\boldsymbol{k}}_{\Omega} \tag{B.19}$$

where  $\Omega$  is a pole fixed or coincident with the centre of mass C of the system. These equations hold for any mechanical system and can be used even in the case of variable mass. For a system with constant mass, computing the time derivative of the momentum in (B.18) gives Newton equations of motion in the form

$$\boldsymbol{f} = m\ddot{\boldsymbol{p}}_C,\tag{B.20}$$

where the quantity on the right-hand side represents the resultant of inertia forces.

If, besides the assumption of constant mass, the assumption of rigid system holds too, the expression in (B.14) of the angular momentum with (B.19) yield Euler equations of motion in the form

$$\boldsymbol{\mu}_{\Omega} = \boldsymbol{I}_{\Omega} \dot{\boldsymbol{\omega}} + \boldsymbol{\omega} \times (\boldsymbol{I}_{\Omega} \boldsymbol{\omega}), \tag{B.21}$$

where the quantity on the right-hand side represents the resultant moment of inertia forces.

For a system constituted by a set of rigid bodies, the external forces obviously do not include the reaction forces exerted between the bodies belonging to the same system.

# B.3 Work and Energy

Given a force  $f_i$  applied at a point of position  $p_i$  with respect to frame O-xyz, the elementary work of the force  $f_i$  on the displacement  $dp_i = \dot{p}_i dt$  is defined as the scalar

$$dW_i = \boldsymbol{f}_i^T d\boldsymbol{p}_i.$$

For a rigid body  $\mathcal{B}$  subject to a system of forces of resultant f and resultant moment  $\mu_Q$  with respect to any point Q of  $\mathcal{B}$ , the elementary work on the rigid displacement (B.6) is given by

$$dW = (\mathbf{f}^T \dot{\mathbf{p}}_O + \boldsymbol{\mu}_O^T \boldsymbol{\omega}) dt = \mathbf{f}^T d\mathbf{p}_O + \boldsymbol{\mu}_O^T \boldsymbol{\omega} dt.$$
 (B.22)

The kinetic energy of a body  $\mathcal{B}$  is defined as the scalar quantity

$$\mathcal{T} = \frac{1}{2} \int_{V_{\mathcal{B}}} \dot{\boldsymbol{p}}^T \dot{\boldsymbol{p}} \rho dV$$

which, for a rigid body, takes on the notable expression

$$\mathcal{T} = \frac{1}{2} m \dot{\boldsymbol{p}}_C^T \dot{\boldsymbol{p}}_C + \frac{1}{2} \boldsymbol{\omega}^T \boldsymbol{I}_C \boldsymbol{\omega}$$
 (B.23)

where  $I_C$  is the inertia tensor relative to the centre of mass expressed in a frame parallel to the reference frame with origin at the centre of mass.

A system of position forces, i.e., the forces depending only on the positions of the points of application, is said to be *conservative* if the work done by each force is independent of the trajectory described by the point of application of the force but it depends only on the initial and final positions of the point of application. In this case, the elementary work of the system of forces is equal to minus the total differential of a scalar function termed *potential energy*, i.e.,

$$dW = -d\mathcal{U}. (B.24)$$

An example of a conservative system of forces on a rigid body is the gravitational force, with which is associated the potential energy

$$\mathcal{U} = -\int_{V_{\mathcal{R}}} \boldsymbol{g}_0^T \boldsymbol{p} \rho dV = -m \boldsymbol{g}_0^T \boldsymbol{p}_C.$$
 (B.25)

## **B.4 Constrained Systems**

Consider a system  $\mathcal{B}_r$  of r rigid bodies and assume that all the elements of  $\mathcal{B}_r$  can reach any position in space. In order to find uniquely the position of all the points of the system, it is necessary to assign a vector  $\mathbf{x} = [x_1 \dots x_p]^T$  of 6r = p parameters, termed configuration. These parameters are termed Lagrange or generalized coordinates of the unconstrained system  $\mathcal{B}_r$ , and p determines the number of degrees of freedom (DOFs).

Any limitation on the mobility of the system  $\mathcal{B}_r$  is termed *constraint*. A constraint acting on  $\mathcal{B}_r$  is said to be *holonomic* if it is expressed by a system of equations

$$h(x,t) = 0, (B.26)$$

where h is a vector of dimensions  $(s \times 1)$ , with s < m. On the other hand, a constraint in the form  $h(x, \dot{x}, t) = 0$  which is nonintegrable is said to be *nonholonomic*. For simplicity, only equality (or *bilateral*) constraints are considered. If the equations in (B.26) do not explicitly depend on time, the constraint is said to be *scleronomic*.

On the assumption that h has continuous and continuously differentiable components, and its Jacobian  $\partial h/\partial x$  has full rank, the equations in (B.26)

allow the elimination of s out of m coordinates of the system  $\mathcal{B}_r$ . With the remaining n=m-s coordinates it is possible to determine uniquely the configurations of  $\mathcal{B}_r$  satisfying the constraints (B.26). Such coordinates are the Lagrange or generalized coordinates and n is the number of degrees of freedom of the unconstrained system  $\mathcal{B}_r$ .

The motion of a system  $\mathcal{B}_r$  with n DOFs and holonomic equality constraints can be described by equations of the form

$$\boldsymbol{x} = \boldsymbol{x}(\boldsymbol{q}(t), t), \tag{B.27}$$

where  $q(t) = [q_1(t) \dots q_n(t)]^T$  is a vector of Lagrange coordinates.

The elementary displacement of system (B.27) relative to the interval (t, t+dt) is defined as

$$d\mathbf{x} = \frac{\partial \mathbf{x}(\mathbf{q}, t)}{\partial \mathbf{q}} \dot{\mathbf{q}} dt + \frac{\partial \mathbf{x}(\mathbf{q}, t)}{\partial t} dt.$$
 (B.28)

The *virtual displacement* of system (B.27) at time t, relative to an increment  $\delta \lambda$ , is defined as the quantity

$$\delta \boldsymbol{x} = \frac{\partial \boldsymbol{x}(\boldsymbol{q}, t)}{\partial \boldsymbol{q}} \delta \boldsymbol{q}. \tag{B.29}$$

The difference between the elementary displacement and the virtual displacement is that the former is relative to an actual motion of the system in an interval (t, t + dt) which is consistent with the constraints, while the latter is relative to an imaginary motion of the system when the constraints are made invariant and equal to those at time t.

For a system with time-invariant constraints, the equations of motion (B.27) become

$$\boldsymbol{x} = \boldsymbol{x}(\boldsymbol{q}(t)), \tag{B.30}$$

and then, by setting  $\delta \lambda = d\lambda = \dot{\lambda} dt$ , the virtual displacements (B.29) coincide with the elementary displacements (B.28).

To the concept of virtual displacement can be associated that of *virtual* work of a system of forces, by considering a virtual displacement instead of an elementary displacement.

If external forces are distinguished into active forces and reaction forces, a direct consequence of the principles of dynamics (B.18), (B.19) applied to the system of rigid bodies  $\mathcal{B}_r$  is that, for each virtual displacement, the following relation holds:

$$\delta W_m + \delta W_a + \delta W_h = 0, \tag{B.31}$$

where  $\delta W_m$ ,  $\delta W_a$ ,  $\delta W_h$  are the total virtual works done by the inertia, active, reaction forces, respectively.

<sup>&</sup>lt;sup>3</sup> In general, the Lagrange coordinates of a constrained system have a local validity; in certain cases, such as the joint variables of a manipulator, they can have a global validity.

In the case of *frictionless* equality constraints, reaction forces are exerted orthogonally to the contact surfaces and the virtual work is always null. Hence, (B.31) reduces to

$$\delta W_m + \delta W_a = 0. ag{B.32}$$

For a steady system, inertia forces are identically null. Then the condition for the equilibrium of system  $\mathcal{B}_r$  is that the virtual work of the active forces is identically null on any virtual displacement, which gives the fundamental equation of statics of a constrained system

$$\delta W_a = 0 \tag{B.33}$$

known as principle of virtual work. Expressing (B.33) in terms of the increment  $\delta \lambda$  of generalized coordinates leads to

$$\delta W_a = \boldsymbol{\zeta}^T \delta \boldsymbol{q} = 0 \tag{B.34}$$

where  $\zeta$  denotes the  $(n \times 1)$  vector of active generalized forces.

In the dynamic case, it is worth distinguishing active forces into *conserva*tive (that can be derived from a potential) and *nonconservative*. The virtual work of conservative forces is given by

$$\delta W_c = -\frac{\partial \mathcal{U}}{\partial \boldsymbol{q}} \delta \boldsymbol{q}, \tag{B.35}$$

where  $\mathcal{U}(\lambda)$  is the total potential energy of the system. The work of nonconservative forces can be expressed in the form

$$\delta W_{nc} = \boldsymbol{\xi}^T \delta \boldsymbol{q},\tag{B.36}$$

where  $\xi$  denotes the vector of nonconservative generalized forces. It follows that the vector of active generalized forces is

$$\zeta = \xi - \left(\frac{\partial \mathcal{U}}{\partial q}\right)^{T}.$$
 (B.37)

Moreover, the work of inertia forces can be computed from the total kinetic energy of system  $\mathcal{T}$  as

$$\delta W_m = \left(\frac{\partial \mathcal{T}}{\partial \mathbf{q}} - \frac{d}{dt} \frac{\partial \mathcal{T}}{\partial \dot{\mathbf{q}}}\right) \delta \mathbf{q}. \tag{B.38}$$

Substituting (B.35), (B.36), (B.38) into (B.32) and observing that (B.32) holds true for any increment  $\delta \lambda$  leads to Lagrange equations

$$\frac{d}{dt} \left( \frac{\partial \mathcal{L}}{\partial \dot{q}} \right)^{T} - \left( \frac{\partial \mathcal{L}}{\partial q} \right)^{T} = \xi, \tag{B.39}$$

where

$$\mathcal{L} = \mathcal{T} - \mathcal{U} \tag{B.40}$$

is the Lagrangian function of the system. The equations in (B.39) completely describe the dynamic behaviour of an n-DOF system with holonomic equality constraints.

The sum of kinetic and potential energy of a system with time-invariant constraints is termed *Hamiltonian* function

$$\mathcal{H} = \mathcal{T} + \mathcal{U}. \tag{B.41}$$

Conservation of energy dictates that the time derivative of the Hamiltonian must balance the power generated by the nonconservative forces acting on the system, i.e.,

$$\frac{d\mathcal{H}}{dt} = \boldsymbol{\xi}^T \dot{\boldsymbol{q}}.\tag{B.42}$$

In view of (B.37), (B.41), the equation in (B.42) becomes

$$\frac{d\mathcal{T}}{dt} = \boldsymbol{\zeta}^T \dot{\boldsymbol{q}}.\tag{B.43}$$

#### **Bibliography**

The fundamental concepts of rigid-body mechanics and constrained systems can be found in classical texts such as [87, 154, 224]. An authoritative reference on rigid-body system dynamics is [187].

#### Feedback Control

As a premise to the study of manipulator decentralized control and centralized control, the fundamental principles of  $feedback\ control$  of  $linear\ systems$  are recalled, and an approach to the determination of control laws for  $nonlinear\ systems$  based on the use of  $Lyapunov\ functions$  is presented.

# C.1 Control of Single-input/Single-output Linear Systems

According to classical automatic control theory of linear time-invariant single-input/single-output systems, in order to serve the output y(t) of a system to a reference r(t), it is worth adopting a negative feedback control structure. This structure indeed allows the use of approximate mathematical models to describe the input/output relationship of the system to control, since negative feedback has a potential for reducing the effects of system parameter variations and nonmeasurable disturbance inputs d(t) on the output.

This structure can be represented in the domain of complex variable s as in the block scheme of Fig. C.1, where G(s), H(s) and C(s) are the transfer functions of the system to control, the transducer and the controller, respectively. From this scheme it is easy to derive

$$Y(s) = W(s)R(s) + W_D(s)D(s), \tag{C.1}$$

where

$$W(s) = \frac{C(s)G(s)}{1 + C(s)G(s)H(s)}$$
 (C.2)

is the closed-loop input/output transfer function and

$$W_D(s) = \frac{G(s)}{1 + C(s)G(s)H(s)}$$
 (C.3)

is the disturbance/output transfer function.

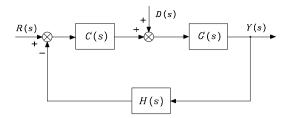


Fig. C.1. Feedback control structure

The goal of the controller design is to find a control structure C(s) ensuring that the output variable Y(s) tracks a reference input R(s). Further, the controller should guarantee that the effects of the disturbance input D(s) on the output variable are suitably reduced. The goal is then twofold, namely, reference tracking and disturbance rejection.

The basic problem for controller design consists of the determination of an action C(s) which can make the system asymptotically stable. In the absence of positive or null real part pole/zero and zero/pole cancellation in the open-loop function F(s) = C(s)G(s)H(s), a necessary and sufficient condition for asymptotic stability is that the poles of W(s) and  $W_D(s)$  have all negative real parts; such poles coincide with the zeros of the rational transfer function 1 + F(s). Testing for this condition can be performed by resorting to stability criteria, thus avoiding computation of the function zeros.

Routh criterion allows the determination of the sign of the real parts of the zeros of the function 1+F(s) by constructing a table with the coefficients of the polynomial at the numerator of 1+F(s) (characteristic polynomial).

Routh criterion is easy to apply for testing stability of a feedback system, but it does not provide a direct relationship between the open-loop function and stability of the closed-loop system. It is then worth resorting to Nyquist criterion which is based on the representation, in the complex plane, of the open-loop transfer function F(s) evaluated in the domain of real angular frequency  $(s = j\omega, -\infty < \omega < +\infty)$ .

Drawing of Nyquist plot and computation of the number of circles made by the vector representing the complex number  $1 + F(j\omega)$  when  $\omega$  continuously varies from  $-\infty$  to  $+\infty$  allows a test on whether or not the *closed-loop* system is asymptotically stable. It is also possible to determine the number of positive, null and negative real part roots of the characteristic polynomial, similarly to application of Routh criterion. Nonetheless, Nyquist criterion is based on the plot of the open-loop transfer function, and thus it allows the determination of a direct relationship between this function and closed-loop system stability. It is then possible from an examination of the Nyquist plot to draw suggestions on the controller structure C(s) which ensures closed-loop system asymptotic stability.

If the closed-loop system is asymptotically stable, the *steady-state response* to a sinusoidal input r(t), with d(t)=0, is sinusoidal, too. In this case, the function W(s), evaluated for  $s=j\omega$ , is termed *frequency response function*; the frequency response function of a feedback system can be assimilated to that of a low-pass filter with the possible occurrence of a *resonance peak* inside its *bandwidth*.

As regards the transducer, this should be chosen so that its bandwidth is much greater than the feedback system bandwidth, in order to ensure a nearly instantaneous response for any value of  $\omega$  inside the bandwidth of  $W(j\omega)$ . Therefore, setting  $H(j\omega) \approx H_0$  and assuming that the loop gain  $|C(j\omega)G(j\omega)H_0| \gg 1$  in the same bandwidth, the expression in (C.1) for  $s = j\omega$  can be approximated as

$$Y(j\omega) \approx \frac{R(j\omega)}{H_0} + \frac{D(j\omega)}{C(j\omega)H_0}.$$

Assuming  $R(j\omega) = H_0 Y_d(j\omega)$  leads to

$$Y(j\omega) \approx Y_d(j\omega) + \frac{D(j\omega)}{C(j\omega)H_0};$$
 (C.4)

i.e., the output tracks the desired output  $Y_d(j\omega)$  and the frequency components of the disturbance in the bandwidth of  $W(j\omega)$  produce an effect on the output which can be reduced by increasing  $|C(j\omega)H_0|$ . Furthermore, if the disturbance input is a constant, the steady-state output is not influenced by the disturbance as long as C(s) has at least a pole at the origin.

Therefore, a feedback control system is capable of establishing a proportional relationship between the desired output and the actual output, as evidenced by (C.4). This equation, however, requires that the frequency content of the input (desired output) be inside the frequency range for which the loop gain is much greater than unity.

The previous considerations show the advantage of including a proportional action and an integral action in the controller C(s), leading to the transfer function

$$C(s) = K_I \frac{1 + sT_I}{s} \tag{C.5}$$

of a proportional-integral controller (PI);  $T_I$  is the time constant of the integral action and the quantity  $K_I T_I$  is called proportional sensitivity.

The adoption of a PI controller is effective for low-frequency response of the system, but it may involve a reduction of *stability margins* and/or a reduction of closed-loop system bandwidth. To avoid these drawbacks, a *derivative action* can be added to the proportional and integral actions, leading to the transfer function

$$C(s) = K_I \frac{1 + sT_I + s^2 T_D T_I}{s}$$
 (C.6)

of a proportional-integral-derivative controller (PID);  $T_D$  denotes the time constant of the derivative action. Notice that physical realizability of (C.6)

demands the introduction of a high-frequency pole which little influences the input/output relationship in the system bandwidth. The transfer function in (C.6) is characterized by the presence of two zeros which provide a stabilizing action and an enlargement of the closed-loop system bandwidth. Bandwidth enlargement implies shorter response time of the system, in terms of both variations of the reference signal and recovery action of the feedback system to output variations induced by the disturbance input.

The parameters of the adopted control structure should be chosen so as to satisfy requirements on the system behaviour at steady state and during the transient. Classical tools to determine such parameters are the root locus in the domain of the complex variable s or the Nichols chart in the domain of the real angular frequency  $\omega$ . The two tools are conceptually equivalent. Their potential is different in that root locus allows a control law to be found which assigns the exact parameters of the closed-loop system time response, whereas Nichols chart allows a controller to be specified which confers good transient and steady-state behaviour to the system response.

A feedback system with strict requirements on the steady-state and transient behaviour, typically, has a response that can be assimilated to that of a second-order system. In fact, even for closed-loop functions of greater order, it is possible to identify a pair of complex conjugate poles whose real part absolute value is smaller than the real part absolute values of the other poles. Such a pair of poles is dominant in that its contribution to the transient response prevails over that of the other poles. It is then possible to approximate the input/output relationship with the transfer function

$$W(s) = \frac{k_W}{1 + \frac{2\zeta s}{\omega_n} + \frac{s^2}{\omega_n^2}}$$
 (C.7)

which has to be realized by a proper choice of the controller. Regarding the values to assign to the parameters characterizing the transfer function in (C.7), the following remarks are in order. The constant  $k_W$  represents the input/output steady-state gain, which is equal to  $1/H_0$  if  $C(s)G(s)H_0$  has at least a pole at the origin. The natural frequency  $\omega_n$  is the modulus of the complex conjugate poles, whose real part is given by  $-\zeta\omega_n$  where  $\zeta$  is the damping ratio of the pair of poles.

The influence of parameters  $\zeta$  and  $\omega_n$  on the closed-loop frequency response can be evaluated in terms of the resonance peak magnitude

$$M_r = \frac{1}{2\zeta\sqrt{1-\zeta^2}},$$

occurring at the resonant frequency

$$\omega_r = \omega_n \sqrt{1 - 2\zeta^2},$$

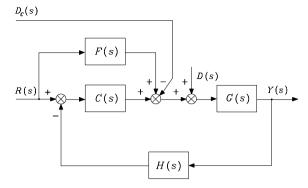


Fig. C.2. Feedback control structure with feedforward compensation

and of the 3dB bandwidth

$$\omega_3 = \omega_n \sqrt{1 - 2\zeta^2 + \sqrt{2 - 4\zeta^2 + 4\zeta^4}}.$$

A step input is typically used to characterize the transient response in the time domain. The influence of parameters  $\zeta$  and  $\omega_n$  on the *step response* can be evaluated in terms of the percentage of *overshoot* 

$$s\% = 100 \exp(-\pi \zeta / \sqrt{1 - \zeta^2}),$$

of the rise time

$$t_r \approx \frac{1.8}{\omega_n}$$

and of the settling time within 1%

$$t_s = \frac{4.6}{\zeta \omega_n}.$$

The adoption of a feedforward compensation action represents a feasible solution both for tracking a time-varying reference input and for enhancing rejection of the effects of a disturbance on the output. Consider the general scheme in Fig. C.2. Let R(s) denote a given input reference and  $D_c(s)$  denote a computed estimate of the disturbance D(s); the introduction of the feedforward action yields the input/output relationship

$$Y(s) = \left(\frac{C(s)G(s)}{1 + C(s)G(s)H(s)} + \frac{F(s)G(s)}{1 + C(s)G(s)H(s)}\right)R(s)$$

$$+ \frac{G(s)}{1 + C(s)G(s)H(s)} (D(s) - D_c(s)).$$
(C.8)

By assuming that the desired output is related to the reference input by a constant factor  $K_d$  and regarding the transducer as an instantaneous system  $(H(s) \approx H_0 = 1/K_d)$  for the current operating conditions, the choice

$$F(s) = \frac{K_d}{G(s)} \tag{C.9}$$

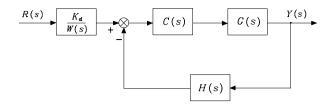


Fig. C.3. Feedback control structure with inverse model technique

yields the input/output relationship

$$Y(s) = Y_d(s) + \frac{G(s)}{1 + C(s)G(s)H_0} (D(s) - D_c(s)).$$
 (C.10)

If  $|C(j\omega)G(j\omega)H_0| \gg 1$ , the effect of the disturbance on the output is further reduced by means of an accurate estimate of the disturbance.

Feedforward compensation technique may lead to a solution, termed *inverse model control*, illustrated in the scheme of Fig. C.3. It should be remarked, however, that such a solution is based on dynamics cancellation, and thus it can be employed only for a minimum-phase system, i.e., a system whose poles and zeros have all strictly negative real parts. Further, one should consider physical realizability issues as well as effects of parameter variations which prevent perfect cancellation.

#### C.2 Control of Nonlinear Mechanical Systems

If the system to control does not satisfy the linearity property, the control design problem becomes more complex. The fact that a *system* is qualified as *nonlinear*, whenever linearity does not hold, leads to understanding how it is not possible to resort to general techniques for control design, but it is necessary to face the problem for each class of nonlinear systems which can be defined through imposition of special properties.

On the above premise, the control design problem of nonlinear systems described by the dynamic model

$$H(x)\ddot{x} + h(x,\dot{x}) = u \tag{C.11}$$

is considered, where  $[\boldsymbol{x}^T \ \dot{\boldsymbol{x}}^T]^T$  denotes the  $(2n \times 1)$  state vector of the system,  $\boldsymbol{u}$  is the  $(n \times 1)$  input vector,  $\boldsymbol{H}(\boldsymbol{x})$  is an  $(n \times n)$  positive definite (and thus invertible) matrix depending on  $\boldsymbol{x}$ , and  $\boldsymbol{h}(\boldsymbol{x}, \dot{\boldsymbol{x}})$  is an  $(n \times 1)$  vector depending on state. Several mechanical systems can be reduced to this class, including manipulators with rigid links and joints.

The *control* law can be found through a nonlinear compensating action obtained by choosing the following *nonlinear state feedback* law (*inverse dynamics* control):

$$u = \widehat{H}(x)v + \widehat{h}(x,\dot{x})$$
 (C.12)

where  $\widehat{\boldsymbol{H}}(\boldsymbol{x})$  and  $\widehat{\boldsymbol{h}}(\boldsymbol{x})$  respectively denote the *estimates* of the terms  $\boldsymbol{H}(\boldsymbol{x})$  and  $\boldsymbol{h}(\boldsymbol{x})$ , computed on the basis of measures on the system state, and  $\boldsymbol{v}$  is a new control input to be defined later. In general, it is

$$\widehat{\boldsymbol{H}}(\boldsymbol{x}) = \boldsymbol{H}(\boldsymbol{x}) + \Delta \boldsymbol{H}(\boldsymbol{x}) \tag{C.13}$$

$$\widehat{\boldsymbol{h}}(\boldsymbol{x}, \dot{\boldsymbol{x}}) = \boldsymbol{h}(\boldsymbol{x}, \dot{\boldsymbol{x}}) + \Delta \boldsymbol{h}(\boldsymbol{x}, \dot{\boldsymbol{x}})$$
 (C.14)

because of the unavoidable modelling approximations or as a consequence of an intentional simplification in the compensating action. Substituting (C.12) into (C.11) and accounting for (C.13), (C.14) yields

$$\ddot{\boldsymbol{x}} = \boldsymbol{v} + \boldsymbol{z}(\boldsymbol{x}, \dot{\boldsymbol{x}}, \boldsymbol{v}) \tag{C.15}$$

where

$$z(x, \dot{x}, v) = H^{-1}(x) (\Delta H(x)v + \Delta h(x, \dot{x})).$$

If tracking of a trajectory  $(\boldsymbol{x}_d(t), \dot{\boldsymbol{x}}_d(t), \ddot{\boldsymbol{x}}_d(t))$  is desired, the tracking error can be defined as

$$e = \begin{bmatrix} x_d - x \\ \dot{x}_d - \dot{x} \end{bmatrix} \tag{C.16}$$

and it is necessary to derive the error dynamics equation to study convergence of the actual state to the desired one. To this end, the choice

$$\boldsymbol{v} = \ddot{\boldsymbol{x}}_d + \boldsymbol{w}(\boldsymbol{e}), \tag{C.17}$$

substituted into (C.15), leads to the error equation

$$\dot{e} = Fe - Gw(e) - Gz(e, x_d, \dot{x}_d, \ddot{x}_d), \tag{C.18}$$

where the  $(2n \times 2n)$  and  $(2n \times n)$  matrices, respectively,

$$F = \left[egin{array}{cc} O & I \ O & O \end{array}
ight] \qquad G = \left[egin{array}{c} O \ I \end{array}
ight]$$

follow from the error definition in (C.16). Control law design consists of finding the error function w(e) which makes (C.18) globally asymptotically stable, i.e.,

$$\lim_{t\to\infty} \boldsymbol{e}(t) = \mathbf{0}.$$

In the case of *perfect* nonlinear compensation  $(z(\cdot) = 0)$ , the simplest choice of the control action is the *linear* one

$$w(e) = -K_P(x_d - x) - K_D(\dot{x}_d - \dot{x})$$

$$= [-K_P - K_D] e,$$
(C.19)

<sup>&</sup>lt;sup>1</sup> Global asymptotic stability is invoked to remark that the equilibrium state is asymptotically stable for any perturbation.

where asymptotic stability of the error equation is ensured by choosing *positive* definite matrices  $K_P$  and  $K_D$ . The error transient behaviour is determined by the eigenvalues of the matrix

$$\mathbf{A} = \begin{bmatrix} \mathbf{O} & \mathbf{I} \\ -\mathbf{K}_P & -\mathbf{K}_D \end{bmatrix} \tag{C.20}$$

characterizing the error dynamics

$$\dot{e} = Ae. \tag{C.21}$$

If compensation is *imperfect*, then  $z(\cdot)$  cannot be neglected and the error equation in (C.18) takes on the general form

$$\dot{\boldsymbol{e}} = \boldsymbol{f}(\boldsymbol{e}). \tag{C.22}$$

It may be worth choosing the control law w(e) as the sum of a nonlinear term and a linear term of the kind in (C.19); in this case, the error equation can be written as

$$\dot{\boldsymbol{e}} = \boldsymbol{A}\boldsymbol{e} + \boldsymbol{k}(\boldsymbol{e}),\tag{C.23}$$

where A is given by (C.20) and k(e) is available to make the system globally asymptotically stable. The equations in (C.22), (C.23) express nonlinear differential equations of the error. To test for stability and obtain advise on the choice of suitable control actions, one may resort to Lyapunov direct method illustrated below.

# C.3 Lyapunov Direct Method

The philosophy of the *Lyapunov direct method* is the same as that of most methods used in control engineering to study stability, namely, testing for stability without solving the differential equations describing the dynamic system.

This method can be presented in short on the basis of the following reasoning. If it is possible to associate an energy-based description with a (linear or nonlinear) autonomous dynamic system and, for each system state with the exception of the equilibrium state, the time rate of such energy is negative, then energy decreases along any system trajectory until it attains its minimum at the equilibrium state; this argument justifies an intuitive concept of stability.

With reference to (C.22), by setting  $f(\mathbf{0}) = \mathbf{0}$ , the *equilibrium state* is  $e = \mathbf{0}$ . A scalar function V(e) of the system state, continuous together with its first derivative, is defined a *Lyapunov function* if the following properties hold:

$$V(e) > 0 \qquad \forall e \neq 0$$

$$V(e) = 0$$
  $e = 0$   
 $\dot{V}(e) < 0$   $\forall e \neq 0$   
 $V(e) \to \infty$   $\|e\| \to \infty$ .

The existence of such a function ensures global asymptotic stability of the equilibrium e = 0. In practice, the equilibrium e = 0 is globally asymptotically stable if a positive definite, radially unbounded function V(e) is found so that its time derivative along the system trajectories is negative definite.

If positive definiteness of V(e) is realized by the adoption of a quadratic form, i.e.,

$$V(e) = e^T Q e \tag{C.24}$$

with Q a symmetric positive definite matrix, then in view of (C.22) it follows

$$\dot{V}(\mathbf{e}) = 2\mathbf{e}^T \mathbf{Q} \mathbf{f}(\mathbf{e}). \tag{C.25}$$

If f(e) is so as to render the function  $\dot{V}(e)$  negative definite, the function V(e) is a Lyapunov function, since the choice (C.24) allows system global asymptotic stability to be proved. If  $\dot{V}(e)$  in (C.25) is not negative definite for the given V(e), nothing can be inferred on the stability of the system, since the Lyapunov method gives only a sufficient condition. In such cases one should resort to different choices of V(e) in order to find, if possible, a negative definite  $\dot{V}(e)$ .

In the case when the property of negative definiteness does not hold, but  $\dot{V}(e)$  is only negative semi-definite

$$\dot{V}(\boldsymbol{e}) \leq 0,$$

global asymptotic stability of the equilibrium state is ensured if the only system trajectory for which  $\dot{V}(e)$  is *identically* null  $(\dot{V}(e) \equiv 0)$  is the equilibrium trajectory  $e \equiv 0$  (a consequence of La Salle theorem).

Finally, consider the stability problem of the nonlinear system in the form (C.23); under the assumption that k(0) = 0, it is easy to verify that e = 0 is an equilibrium state for the system. The choice of a Lyapunov function candidate as in (C.24) leads to the following expression for its derivative:

$$\dot{V}(e) = e^{T} (A^{T}Q + QA)e + 2e^{T}Qk(e). \tag{C.26}$$

By setting

$$\boldsymbol{A}^T \boldsymbol{Q} + \boldsymbol{Q} \boldsymbol{A} = -\boldsymbol{P},\tag{C.27}$$

the expression in (C.26) becomes

$$\dot{V}(\mathbf{e}) = -\mathbf{e}^T \mathbf{P} \mathbf{e} + 2\mathbf{e}^T \mathbf{Q} \mathbf{k}(\mathbf{e}). \tag{C.28}$$

The matrix equation in (C.27) is said to be a *Lyapunov equation*; for any choice of a symmetric positive definite matrix P, the solution matrix Q exists

and is symmetric positive definite if and only if the eigenvalues of  $\boldsymbol{A}$  have all negative real parts. Since matrix  $\boldsymbol{A}$  in (C.20) verifies such condition, it is always possible to assign a positive definite matrix  $\boldsymbol{P}$  and find a positive definite matrix solution  $\boldsymbol{Q}$  to (C.27). It follows that the first term on the right-hand side of (C.28) is negative definite and the stability problem is reduced to searching a control law so that  $\boldsymbol{k}(\boldsymbol{e})$  renders the total  $\dot{V}(\boldsymbol{e})$  negative (semi-)definite.

It should be underlined that La Salle theorem does not hold for *time-varying* systems (also termed *non-autonomous*) in the form

$$\dot{\boldsymbol{e}} = f(\boldsymbol{e}, t).$$

In this case, a conceptually analogous result which might be useful is the following, typically referred to as  $Barbalat\ lemma$  — of which it is indeed a consequence. Given a scalar function V(e,t) so that

- 1. V(e,t) is lower bounded
- 2.  $\dot{V}(e,t) < 0$
- 3. V(e,t) is uniformly continuous

then it is  $\lim_{t\to\infty} \dot{V}(\boldsymbol{e},t) = 0$ . Conditions 1 and 2 imply that  $V(\boldsymbol{e},t)$  has a bounded limit for  $t\to\infty$ . Since it is not easy to verify the property of uniform continuity from the definition, Condition 3 is usually replaced by

3'.  $\ddot{V}(\boldsymbol{e},t)$  is bounded

which is sufficient to guarantee validity of Condition 3. Barbalat lemma can obviously be used for time-invariant (autonomous) dynamic systems as an alternative to La Salle theorem, with respect to which some conditions are relaxed; in particular, V(e) needs not necessarily be positive definite.

## Bibliography

Linear systems analysis can be found in classical texts such as [61]. For the control of these systems see [82, 171]. For the analysis of nonlinear systems see [109]. Control of nonlinear mechanical systems is dealt with in [215].