

17/03/2021

Eqn for V_0 (pdf file of 26/Feb \rightarrow page-5)

From this eqn, Power corresponding to ω_1 is

$$P_{\omega_1} = \frac{1}{2} (a_1 V_0)^2 \quad \checkmark \quad \text{--- (1)}$$

Sim, power corresponding to third order IMD product $(2\omega_1 - \omega_2)$ or $(2\omega_2 - \omega_1)$

$$P_{2\omega_1 - \omega_2} = \frac{1}{2} \left(\frac{3}{4} a_3 V_0^3 \right)^2 \quad \checkmark \quad \text{--- (2)}$$

Recall

By definition of $11P_3$ (3rd order intercept pt), $P_{\omega_1} = P_{2\omega_1 - \omega_2}$



$$\therefore \textcircled{1} = \textcircled{2}$$

$$\frac{1}{2}(a_1 v_0)^2 = \frac{1}{2} \left(\frac{3}{4} a_3 v_0^3 \right)^2$$

$$\Rightarrow \frac{1}{2} a_1^2 v_0^2 = \frac{9}{32} a_3^2 v_0^6$$

$$\Rightarrow v_0^4 = \frac{a_1^2}{a_3^2} \cdot \frac{32}{2 \times 9}$$

$$\Rightarrow v_0^4 = \frac{a_1^2}{a_3^2} \cdot \frac{16}{9}$$

$$\Rightarrow v_0^2 = \frac{4}{3} \frac{a_1}{a_3}$$

$$\Rightarrow \boxed{v_0 = \sqrt{\frac{4}{3} \frac{a_1}{a_3}}}$$

$$v_0 = v_{IP}$$

$$\Rightarrow \boxed{v_{IP} = \sqrt{\frac{4}{3} \frac{a_1}{a_3}}}$$

$\textcircled{3}$

$$\underline{11 P_3 \text{ (or } (P_3 \text{ or } P_3))}$$

$$P_3 = P_{w,1} \Big|_{\underline{v_a = v_{cp}}} = \frac{1}{2} a_1^2 v_a^2 \quad (\text{from ①})$$

$$= \frac{1}{2} a_1^2 v_{cp}^2$$

$$= \frac{1}{2} a_1^2 \cdot \frac{4a_1}{3a_3} \quad [\text{from ②}]$$

this eqn is independent

of v_0
($= v_{cp}$)

$$\boxed{P_3 = \frac{2a_1^3}{3a_3}}$$

— ④

$$\underline{SFDR} = \frac{P_{\omega_1}}{P_{2\omega_1 - \omega_2}}$$

P_{ω_1}

$$P_{2\omega_1 - \omega_2} = \frac{9}{32} a_3^2 V_0^6$$

$$= \frac{\frac{1}{8} a_1^6 V_0^6}{\frac{4}{9} a_3^2}$$

(from ②)

(rewritten)

$$P_{2\omega_1 - \omega_2} = \frac{(P_{\omega_1})^3}{(P_3)^2}$$

→ from ①

→ from ④

⑤

From (5), 3rd order IMD power increases
as cube of input signal power

$$(5) : P_{2\omega_1 - \omega_2} = \frac{(P_{\omega_1})^3}{(P_3)^2}$$

$$\Rightarrow N_0 = \frac{(P_{\omega_1})^3}{(P_3)^2}$$

$$\Rightarrow P_{\omega_1} = (N_0)^{1/3} (P_3)^{2/3} \quad (6)$$

$$SFDR = \frac{P_{\omega_1}}{P_{2\omega_1 - \omega_2}} \quad \left| \quad P_{2\omega_1 - \omega_2} \approx N_0 \right.$$

$$\Rightarrow \text{SFDR} = \frac{(N_0)^{1/3} (P_3)^{2/3}}{N_0} \rightarrow \text{[from (6)]}$$

$$\text{SFDR} = \left(\frac{P_3}{N_0} \right)^{2/3}$$

c/p noise range (dB)

in dB,

$$\text{SFDR} = \frac{2}{3} (P_3 - N_0)$$

dBm

dBm

noise floor

Example (continued)

→ extended

A mobile RX has NF of 7 dB, 1 dB Compression (compression)

point of 25 dBm, a gain of 40 dB. If

this mobile RX is connected to an antenna having a noise temp T_A of 150 K,

find linear to spurious free dynamic range (LDR) (SFDR)

range

→ also, 3rd order intercept pt is 35 dBm

Assume $T_0 = 17^\circ\text{C} = 290\text{ K}$

$B = 100\text{ MHz} = 10^8\text{ Hz}$

Soln:

$$LDR =$$

$$P_{-1dB} - N_0$$

(dB)

$$SFDR =$$

$$\frac{2}{3} (11P_3 - N_0)$$

(dB)

Given, $P_{-1dB} = 25 \text{ dBm}$

$$11P_3 = 35 \text{ dBm}$$

$$N_0 =$$

?

(calculated earlier)

Recall

$$N_0 = \underbrace{KTABG}$$

(p noise f
antenna
is amplified)

$$+ \underbrace{KT_e BG}$$

noise added
because of down
to amplified

$$\begin{aligned}
 \Rightarrow N_o &= G k_B [T_A + T_e] \\
 &= G k_B [T_A + (F-1)T_o] \\
 &= 10^4 \times 1.38 \times 10^{-23} \times 10^8 \times \\
 &\quad [150 + (5-1)290]
 \end{aligned}$$

$$\begin{aligned}
 [\because G = 40 \text{ dB} &\Rightarrow 10^4 \\
 \& \text{ NF} = 7 \text{ dB} &\Rightarrow F = 5]
 \end{aligned}$$

$$\begin{aligned}
 \therefore N_o &= 1.8 \times 10^{-8} \text{ W} \\
 &= 1.8 \times 10^{-5} \text{ mW}
 \end{aligned}$$

$$N_0 = 1.8 \times 10^{-5} \text{ mW}$$

$$N_0 / \text{dBm} = -50 + 10 \log(1.8)$$

$$= -47.4 \text{ dBm}$$

$$\therefore \text{LDR} = P_{\text{1dB}} - N_0$$

$$= 25 \text{ dBm} - (-47.4 \text{ dBm})$$

$$\boxed{\text{LDR} = 72 \text{ dB}}$$

$$\therefore \text{SFDR} = \frac{2}{3} (11P_3 - N_0)$$

$$= \frac{2}{3} (35 + 47.4)$$

$$\boxed{\text{SFDR} = 54 \text{ dB}}$$

$$\underline{\underline{\text{SFDR} < \text{LDR}}}$$

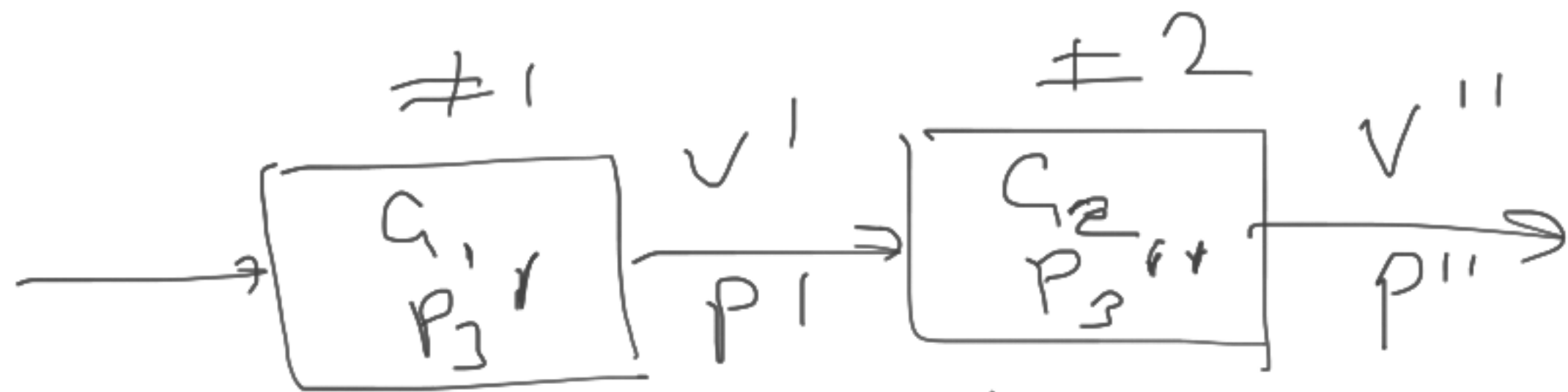
P_3 (or $1/P_3$) of Cascaded Systems

Recall

$$F = F_1 + \frac{F_2 - 1}{G_1} + \frac{F_3 - 1}{G_1 G_2} + \dots$$

$$T_e = T_{e1} + \frac{T_{e2}}{G_1} + \frac{T_{e3}}{G_1 G_2} + \dots$$

$$P_3 = ? ?$$



$G_1 =$ gain of system $\neq 1$
 $G_2 =$ " " " " $\neq 2$
 $P_3' = P_3$ of system $\neq 1$
 $P_3'' =$ " " " " $\neq 2$

3rd order distortion power of o/p of
first stage is

$$P'_{2\omega_1 - \omega_2} = \frac{(P'_{\omega_1})^3}{(P'_3)^2} \quad \text{[from (5)]}$$

where P'_{ω_1} is derived from ω_1 at o/p of
 first stage -

→ Voltage associated with this power is

$$V'_{2\omega_1 - \omega_2} = \sqrt{P'_{2\omega_1 - \omega_2} \cdot Z_0}, \quad \text{system } Z_0 = \text{impedance (}\Omega\text{)}$$

$$[\Rightarrow V = \sqrt{PR}]$$

$$\Rightarrow V'_{2\omega_1 - \omega_2} = \frac{\sqrt{(P'_{\omega_1})^3 Z_0}}{P'_3} \quad \text{— from (6)]}$$

Total 3rd order distortion voltage o/p of
Second stage is ^{sum} this voltage times voltage
 gain of 2nd stage +
 distortion voltage of 2nd stage & combined
 (i.e., total distortion voltage at o/p of 2nd stage is

$$V_{2\omega_1 - \omega_2}^{II} = \frac{\sqrt{G_2 (P_{\omega_1}')^3 Z_0}}{P_3'} + \frac{\sqrt{(P_{\omega_1}'')^2 Z_0}}{P_3''}$$

↑
 voltage gain
 ∴ sq root

↑
 first order

↑
 2nd order

Since $P_{\omega_1}'' = G_2 P_{\omega_1}' \Rightarrow P_{\omega_1}' = \frac{P_{\omega_1}''}{G_2}$

$$V''_{2\omega_1 - \omega_2} = \frac{\sqrt{G_2 \left(\frac{P''_{\omega_1}}{G_2} \right)^3 Z_0}}{P_3'} + \frac{\sqrt{(P''_{\omega_1})^3 Z_0}}{P_3''}$$

$$= \frac{\sqrt{(P''_{\omega_1})^3 Z_0}}{G_2 P_3'} + \frac{\sqrt{(P''_{\omega_1})^3 Z_0}}{P_3''}$$

$$V''_{2\omega_1 - \omega_2} = \left(\frac{1}{G_2 P_3'} + \frac{1}{P_3''} \right) \sqrt{(P''_{\omega_1})^3 Z_0}$$

\therefore o/p power =

$$P_{2\omega_1 - \omega_2}'' = \frac{(V_{2\omega_1 - \omega_2}'')^2}{Z_0}$$

$$[\because P = \frac{V^2}{R}]$$

Z_0

$$\Rightarrow P_{2\omega_1 - \omega_2}'' = \left(\frac{1}{G_2 P_3'} + \frac{1}{P_3''} \right) \cdot (P_{\omega_1}'')^3$$

$$= (P_{\omega_1}'')^3 / (P_3)''^2$$

where P_3 (3rd order intercept pt. of cascaded systems)

$$P_3 \equiv \left(\frac{1}{G_2 P_3'} + \frac{1}{P_3''} \right)^{-1}$$

$$\Rightarrow P_3 = \left(\frac{1}{G_2 P_3'} + \frac{1}{P_3''} \right)^{-1} \text{ for 2 stages}$$

So on, $P_3 = \left(\frac{1}{G_3 G_2 P_3'} + \frac{1}{G_2 P_3''} + \frac{1}{P_3'''} \right)^{-1}$ for 3 stages

$$\& P_3 = \left(\frac{1}{G_4 G_3 G_2 P_3'} + \frac{1}{G_3 G_2 P_3''} + \frac{1}{G_2 P_3'''} + \frac{1}{P_4''''} \right)^{-1}$$

for 4 stages

\Rightarrow Last system gain is important
 & first " " " " not important
 (In NF, it was otherwise)

Example



G_{gain}

$11 P_3 \text{ Pt}$

$$G_1 = 20 \text{ dB}$$

$$P_3' = 7 \text{ dBm} = 5$$

$$G_2 = -6 \text{ dB} = \frac{1}{4} = 0.25$$

$$P_3'' = 22 \text{ dBm} = 160 \text{ mW}$$

$$\begin{aligned} 20 \text{ dBm} &= 100 \text{ mW} \\ 23 \text{ dBm} &= 400 \text{ mW} \end{aligned}$$

$$P_3 = ?$$

$$P_3 = \left(\frac{1}{G_2 P_3'} + \frac{1}{P_3''} \right)^{-1} = \left(\frac{1}{0.25 \times 160} + \frac{1}{5} \right)^{-1}$$

all factors

$$P_3 = 4.4 \text{ mW}$$

$$P_3 = 6.4 \text{ dBm}$$