

EE204 : Analog Circuits
Dept of Electrical Engineering
IIT Bombay
Autumn Semester 2023

Tutorial-2

1. Figure 1 shows an op amp that is ideal except for having a finite open-loop gain and is used to realize an inverting amplifier whose gain has a magnitude $G = \frac{R_2}{R_1}$ (ideal closed loop gain). To compensate for the gain reduction due to the finite A_V , a resistor R_c is shunted across R_1 . Find the value of R_c in terms of R_2, R_1 and G such that obtained closed loop gain is equal to ideal closed loop gain.

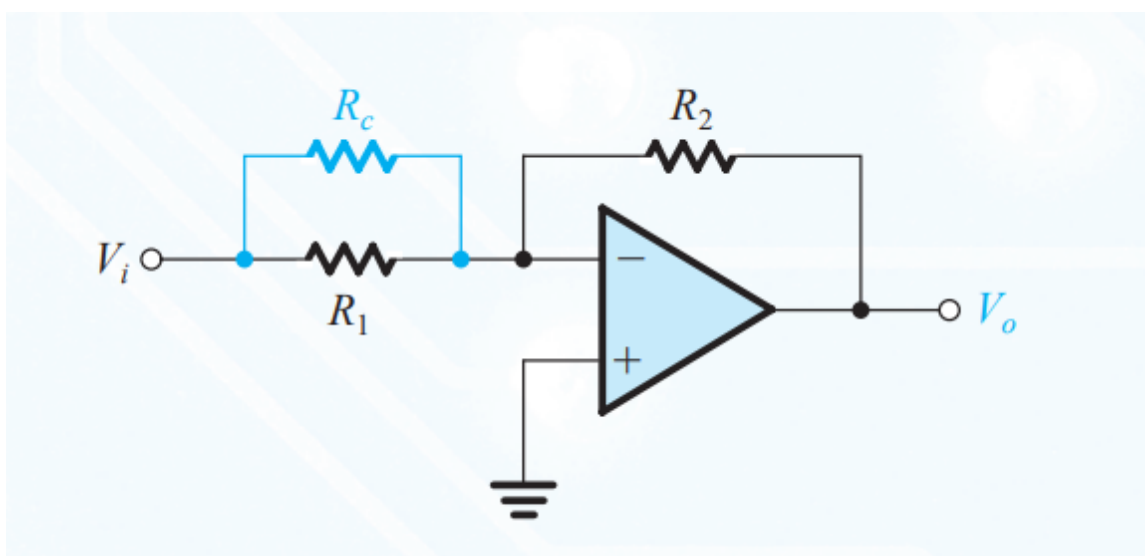
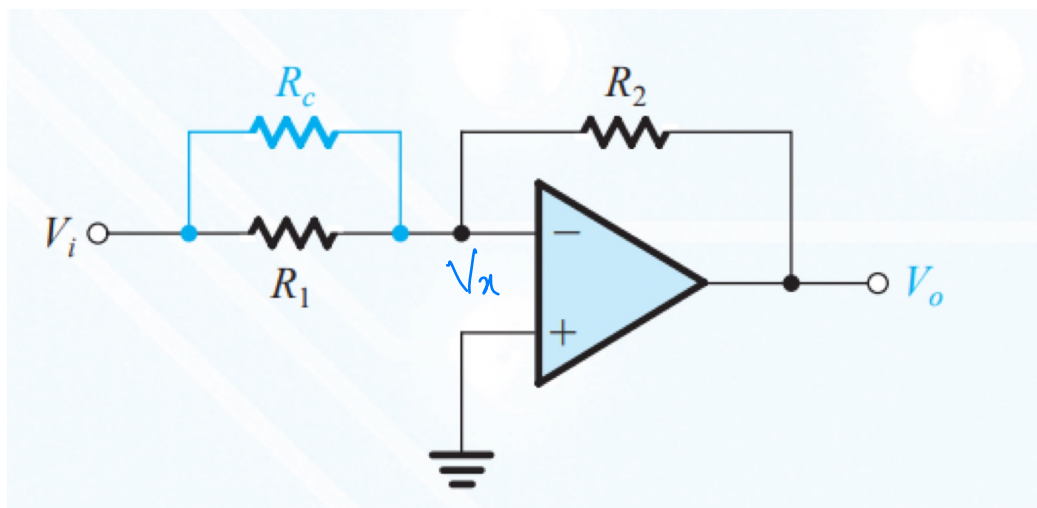


Figure 1: Circuit for Q1

Solu



$$\frac{V_i - V_n}{R_1 \parallel R_c} = \frac{V_n - V_o}{R_2}$$

$$V_o = A_v V_n$$

$$\frac{V_i + \frac{V_o}{A_v}}{R_1 \parallel R_c} = \frac{-\left(\frac{V_o}{A_v} + V_o\right)}{R_2}$$

$$R_2 V_i + \frac{R_2}{A_v} V_o = -(R_1 \parallel R_c) \left(\frac{V_o}{A_v} + V_o \right)$$

$$R_2 V_i = - \left(\frac{R_1 \parallel R_c}{A_v} V_o + R_1 \parallel R_c V_o + \frac{R_2}{A_v} V_o \right)$$

$$\frac{V_o}{V_i} = \frac{-R_2}{\frac{R_1 \parallel R_c}{A_v} + R_1 \parallel R_c + \frac{R_2}{A_v}}$$

$$\frac{V_o}{V_i} = \frac{-R_2}{\frac{R_1 R_c}{R_1 + R_c} \left(\frac{1}{A_v} + 1 \right) + \frac{R_2}{A_v}}$$

$$= \frac{-R_2}{R_1} \frac{1}{\frac{R_c}{R_1 + R_c} \left(\frac{1}{A_v} + 1 \right) + \frac{G}{A_v}}$$

$$\frac{R_c}{R_1 + R_c} \frac{(1 + A_v)}{A_v} + \frac{G}{A_v} = 1$$

$$A_v = \frac{R_c}{R_1 + R_c} (1 + A_v) + G$$

$$A_v R_1 + A_v R_c = R_c + R_c A_v + G R_1 + G R_c$$

$$(A_v - G) R_1 = R_c (1 + G)$$

$$R_1 = \frac{R_c (G + 1)}{A_v - G}$$

2. The MOSFET in Figure 2 has $V_t = 0.5 \text{ V}$, $\mu_n C_{ox} = 400 \frac{\mu\text{A}}{\text{V}^2}$ and $\lambda=0$. Find the required values of W/L and of R so that when $V_i = V_{DD} = +1.8 \text{ V}$, effective channel resistance(r_{DS}) = 50Ω , and $V_o = 50 \text{ mV}$.

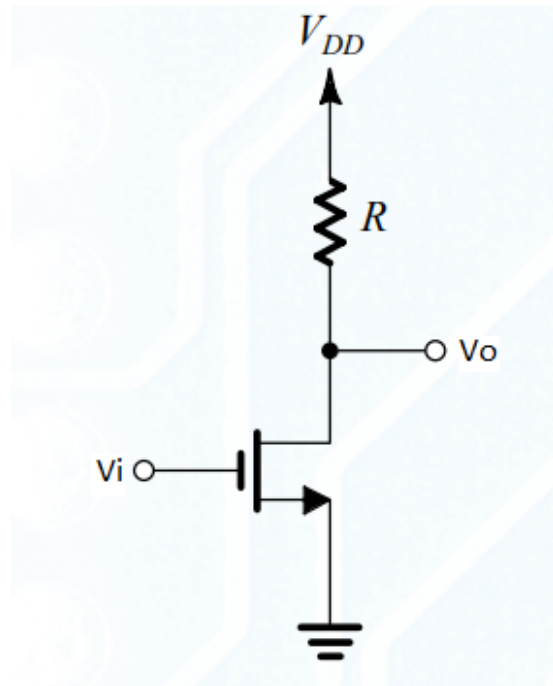


Figure 2: Circuit for Q2

Soln

Given $V_o = 50 \text{ mV}$; $r_{DS} = 50\Omega$; $V_{DS} = 1.8 \text{ V}$

as V_{DS} is very low

$$I_D = \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_t) V_{DS}$$

$$\frac{1}{r_{DS}} = g_{ds} = \frac{\delta I_D}{\delta V_{DS}} = \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_t)$$

$$I_D = \frac{V_o}{r_{DS}} = 1 \text{ mA}$$

$$1 \text{ mA} = 400 \frac{\mu\text{A}}{\text{V}^2} \frac{W}{L} (1.8 - 0.5) 50 \text{ mV}$$

$$\boxed{\frac{W}{L} = 38.46}$$

$$V_o = V_{DD} - I_D R$$

$$50\text{mV} = 1.8 - (1\text{mA})R$$

$$R = 1.75\text{k}\Omega$$

3. For the devices in the circuits of Fig, $|V_t| = 0.5$ V, $\lambda = 0$, $\mu_n C_{ox} = 40 \frac{\mu A}{V^2}$, $L = 1 \mu m$, and $W = 10 \mu m$. Find V_2 and I_2 . How do these values change if Q3 and Q4 are made to have $W = 100 \mu m$?

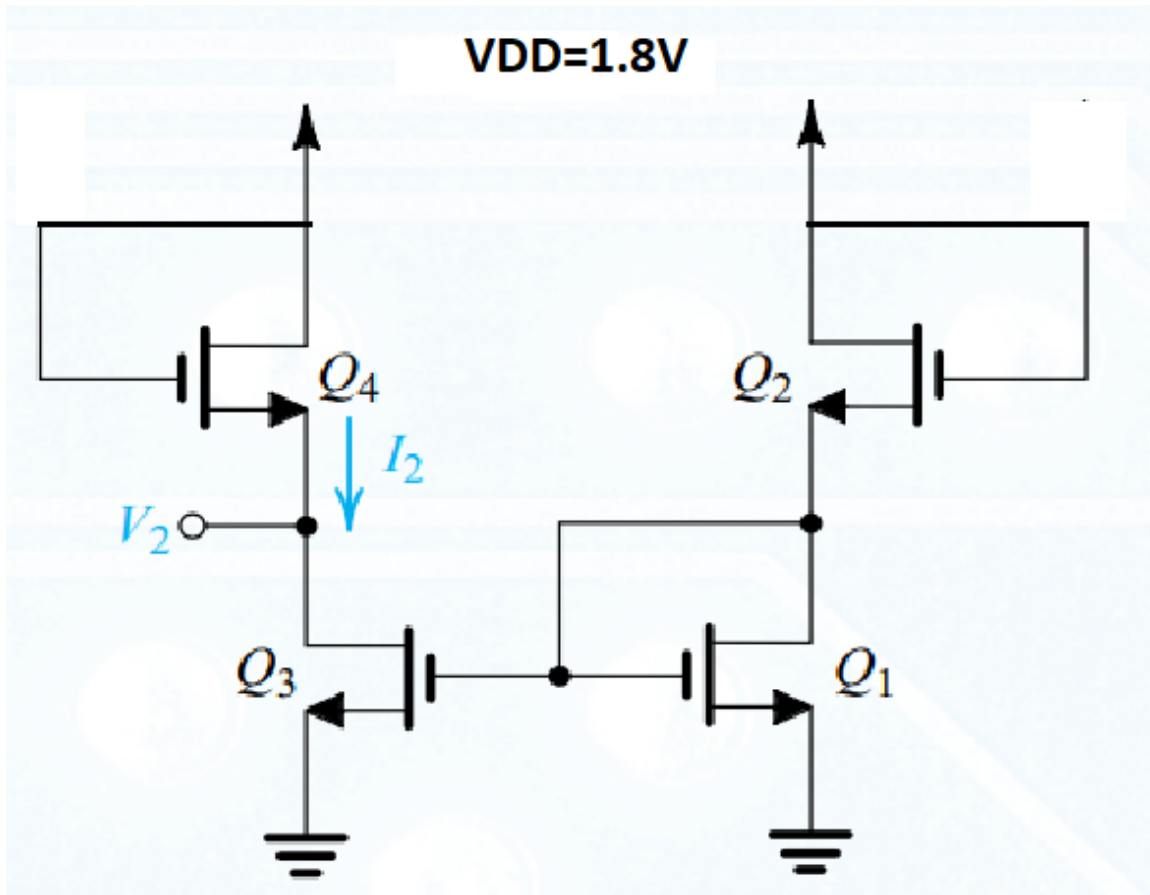


Figure 3: Circuit for Q1

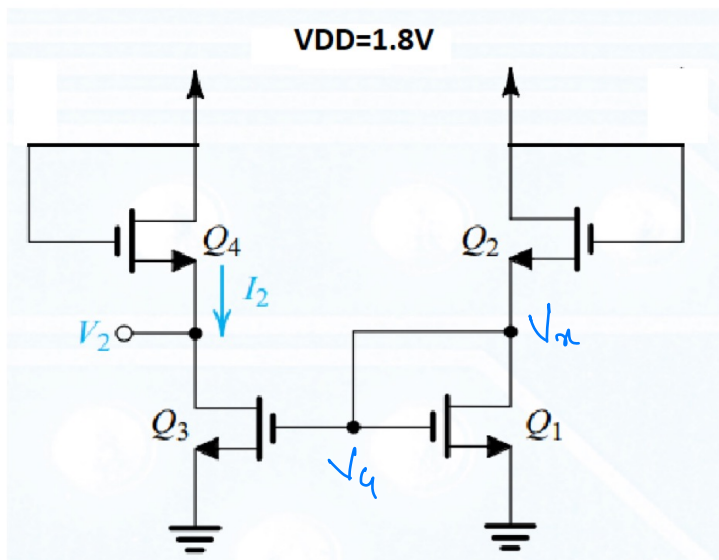


Figure 3: Circuit for Q1

Equating currents in Q_2 and Q_1

$$\frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{DD} - V_x - V_t)^2 = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_x - V_t)^2$$

$$V_{DD} - V_x - V_t = V_x - V_t$$

$$V_x = \frac{V_{DD}}{2} = 0.9$$

$$\text{So } V_g = 0.9$$

As Q_4 has to pass same current as Q_3

$$(V_{GS})_{Q_4} = (V_{GS})_{Q_3}$$

$$V_{DD} - V_2 = V_{GS}$$

$$V_2 = V_{DD} - V_{GS} = 0.9$$

$$\begin{aligned} I_2 &= \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{th})^2 \\ &= \frac{1}{2} \frac{40 \mu A}{V^2} 10 (0.9 - 0.5)^2 \end{aligned}$$

$$I_2 = 32 \mu A$$

If Q_3 and Q_4 W changes to $100 \mu m$
then I_2 increases by 10 times

$$I_2 = 320 \mu A$$