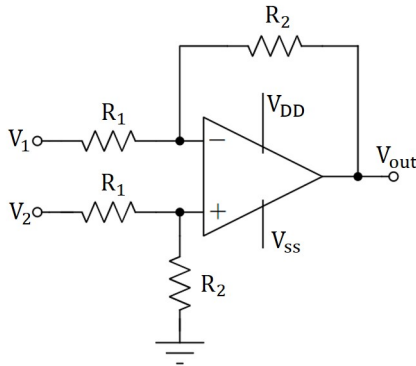


## Quiz 1

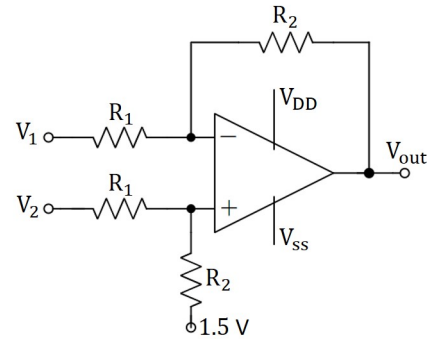
**Total Marks : 10**

**Date: 2<sup>th</sup> September 2023, Time: 9.30 am to 10.30 am.**

**Q1)**



(a) Q1(a) circuit diagram



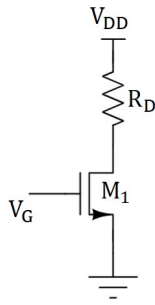
(b) Q1(b) circuit diagram

In the circuit shown in figure 1(a):  $R_1 = 1k\Omega$ ,  $R_2 = 9k\Omega$ . **OpAmp is ideal.**

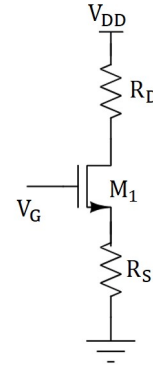
- Derive  $V_{out}$  as a function of  $V_1$  and  $V_2$  if  $V_{DD} = -V_{SS} = 2V$ . [2 marks]
- In the circuit diagram shown in figure 1(b),  $V_{DD} = 3V$  and  $V_{SS} = 0V$ .  $V_1 = 1.5 + V_x$ ,  $V_2 = 1.5 + V_y$ . Derive an expression for  $V_{out}$  as a function of  $V_x$  and  $V_y$ . [2 marks]
- In the question 1(b), if  $V_1 = V_2 = 1.5V$  and resistors  $R_1$  and  $R_2$  are 5% tolerance resistors, i.e.  $R_1 \rightarrow R_1(1 \pm \epsilon)$  and  $R_2 \rightarrow R_2(1 \pm \epsilon)$  where  $\epsilon = 0.05$ , determine the value of  $V_{out}$ . [1 mark]

**Q2)**

- In the circuit diagram shown in figure 2(a),  $R_D = 15k\Omega$ ,  $V_{GST} = 0.3V$ ,  $I_D = 0.1mA$ . Find minimum value of  $V_{DD}$  for which transistor  $M_1$  will be in saturation (pinch-off) region. [1 mark]
- Repeat (a) if resistor as  $R_S = 2k\Omega$  is added to the circuit as shown in circuit shown in figure 2(b). (Rest of the values will be same as the corresponding values in (a).) [1 mark]



(a) Q2(a) circuit diagram



(b) Q2(b) circuit diagram

### Q3)

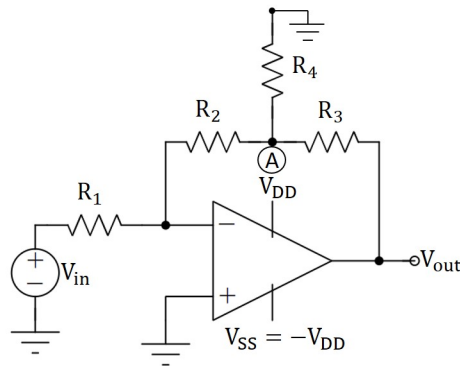


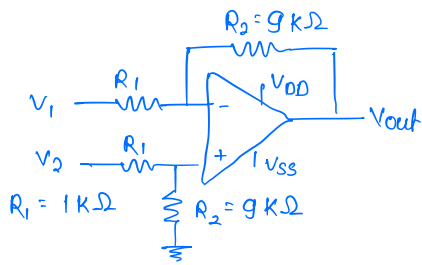
Figure 3: Q3 circuit diagram

In the circuit diagram shown in figure 3, **OpAmp is ideal**.

- (a) Use left side of the circuit, i.e.  $V_{in}$ ,  $R_1$  and  $R_2$  to derive  $V_A$  as a function of  $V_{in}$ . [1 mark]
- (b) Use right side of the circuit, i.e.  $V_{out}$ ,  $R_3$ ,  $R_4$  and  $R_2$  to derive  $V_A$  as a function of  $V_{out}$ . [1 mark]
- (c) Use (a) and (b) to derive  $V_{out}$  as function of  $V_{in}$ . [0.5 mark]
- (d) In your opinion, what is the feature of the circuit? [0.5 mark]

**All the Best!**

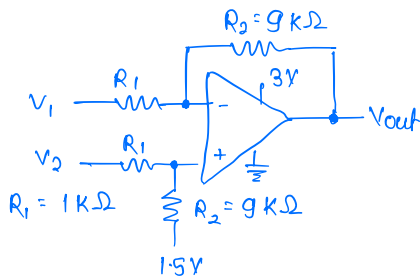
Q1 &gt; (a)



$$V_{out} = \left( \frac{R_2}{R_1 + R_2} \right) \left( 1 + \frac{R_2}{R_1} \right) V_2 + \left( -\frac{R_2}{R_1} \right) \cdot V_1 = \frac{R_2}{R_1} (V_2 - V_1) \quad \text{--- } \langle 1 \text{ mark} \rangle$$

$$V_{out} = 9(V_2 - V_1) \quad \text{--- } \langle 1 \text{ mark} \rangle$$

&lt;b&gt;



$$V_1 = V_x + 1.5$$

$$V_2 = V_y + 1.5$$

$$V_{cm} = \frac{V_{DD} - V_{SS}}{2} = \frac{3 - 0}{2} = 1.5V \quad \text{--- } \langle 0.5 \text{ mark} \rangle$$

$$\begin{aligned} \therefore V_{out} &= V_{cm} + \frac{R_2}{R_1} (V_2 - V_1) \\ &= V_{cm} + \frac{R_2}{R_1} (V_y - V_x) \quad \text{--- } \langle 0.5 \text{ mark} \rangle \end{aligned}$$

$$V_{out} = 1.5 + 9(V_y - V_x) \quad \text{--- } \langle 1 \text{ mark} \rangle$$

$$c) \quad R_1 \rightarrow R_1 \pm 0.05 R_1 \quad \& \quad R_2 \rightarrow R_2 \pm 0.05 R_2$$

We have

$$V_{out} = \left( \frac{R_2'}{R_1' + R_2'} \right) \left( 1 + \frac{R_2''}{R_1'} \right) V_2 + \left( -\frac{R_2''}{R_1'} \right) \cdot V_1 + V_{cm}$$

Lets consider worst case mismatch  $R_2' = (1 + \epsilon) R_2$ ,  $R_1' = (1 - \epsilon) R_1$ ,  $R_2'' = (1 + \epsilon) R_2$  &  $R_1'' = (1 - \epsilon) R_1$

$$\begin{aligned} V_{out} &= \frac{(1 + \epsilon)}{R_1 + R_2} \times \frac{(R_1 + R_2)}{(1 - \epsilon)} V_2 + \frac{(1 + \epsilon)}{(1 - \epsilon)} \frac{R_2}{R_1} \cdot V_1 + V_{cm} \\ &= \left( \frac{1 + \epsilon}{1 - \epsilon} \right) (V_2 - V_1) + V_{cm} \end{aligned}$$

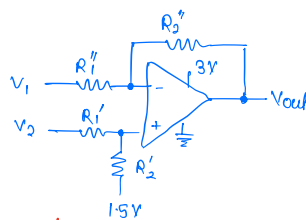
Thus  $V_{out}$  due to mismatch is

$$V_{out} = \left( \frac{1 + \epsilon}{1 - \epsilon} \right) (V_1 - V_2) + V_{cm} \quad \text{--- } \langle 0.5 \text{ mark} \rangle$$

We have  $V_1 = V_2 = 1.5V$  Hence

No matter how mismatched the resistors are, the output voltage remains the same value of  $V_{cm}$

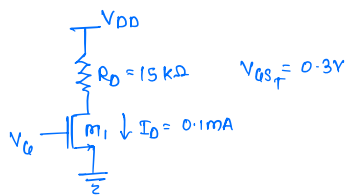
$$V_{out} = V_{cm} = 1.5V \quad \text{--- } \langle 0.5 \text{ mark} \rangle$$



Note\*

The consideration of specific worst case is optional, students can directly write that  $V_{out}$  will remain at  $V_{cm}$  irrespective of mismatch

Q2) <0>



for  $m_1$  to remain in saturation region

$$V_{DS} > V_{DS_T}$$

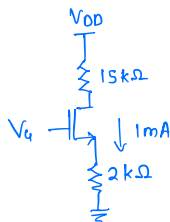
$$\text{i.e. } V_{DD} - I_D \cdot R_D > V_{DS_T} \quad \text{--- } \langle 0.5 \text{ mark} \rangle$$

$$V_{DD} - 0.1 \text{ mA} \times 15 \text{ k}\Omega > 0.3 \text{ V}$$

$$\therefore V_{DD} > 1.8 \text{ V}$$

$$V_{DD_{\min}} = 1.8 \text{ V} \quad \text{--- } \langle 0.5 \text{ mark} \rangle$$

<b>



$$V_{DS} = V_{DD} - I_D \cdot R_D - I_D \cdot R_S \quad \text{--- } \langle 0.5 \text{ mark} \rangle$$

for  $m_1$  to remain in saturation

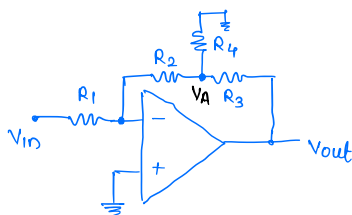
$$V_{DS} > V_{DS_T}$$

$$\therefore V_{DD} - 0.1 \text{ mA} \cdot 15 \text{ k}\Omega - 0.1 \text{ mA} \times 2 \text{ k}\Omega > 0.3 \text{ V}$$

$$V_{DD} > 2 \text{ V}$$

$$V_{DD_{\min}} = 2 \text{ V} \quad \text{--- } \langle 0.5 \text{ mark} \rangle$$

Q3)



$$\langle a \rangle V_A = -\frac{R_2}{R_1} V_{in} \quad \text{--- } \langle 1 \text{ mark} \rangle$$

$$\langle b \rangle V_{out} = \left( 1 + \frac{R_3}{R_4 \parallel R_2} \right) V_A \quad \text{--- } \langle 1 \text{ mark} \rangle$$

$$\langle c \rangle V_{out} = -\left( \frac{R_2}{R_1} \right) \left( 1 + \frac{R_3}{R_4 \parallel R_2} \right) V_{in} \quad \text{--- } \langle 0.5 \text{ mark} \rangle$$

$\langle 0.25 \text{ mark} \rangle$   
 <d> As  $R_4 \parallel R_2 < R_2$ , adding  $R_4$  improves the gain of the inverting amplifier without the need of using wide range of resistance values.  $\leftarrow \langle 0.25 \text{ mark} \rangle$

Alternate Answer

Adding  $R_4$ , amplifies the current flowing into  $R_3$  branch without adding too small or large resistors

$\uparrow$   
 $\langle 0.25 \text{ mark} \rangle$

$\uparrow$   
 $\langle 0.25 \text{ mark} \rangle$