

FOURIER TRANSFORMS

s, z are complex no.'s

$e^{st} \rightarrow H(s)e^{st}$ (complex exponentials are eigenfunctions of LTI systems.)
 $z^n \rightarrow H(z)z^n$

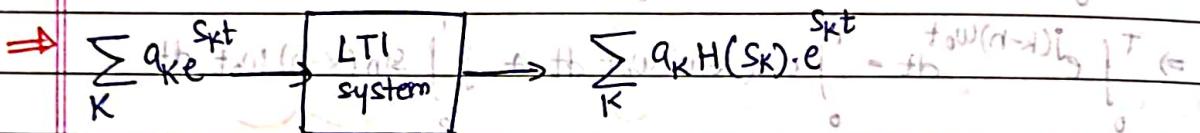
CONT. $\Rightarrow n(t) = e^{st}$ $y(t) = \int h(\tau) e^{s(t-\tau)} d\tau$ converges (assume)

$$(y(t))_{t \geq 0} = \int_{-\infty}^t h(\tau) e^{s(t-\tau)} d\tau = e^{st} \left\{ \int_{-\infty}^0 h(\tau) e^{-s\tau} d\tau \right\} = H(s) e^{st}$$

($y(t)$ is a function of t) $H(s) = \int_{-\infty}^{+\infty} e^{-st} h(t) dt$ (long time behavior at $t \rightarrow \infty$)

DISCRETE $\Rightarrow n[n] = z^n$ $y[n] = \sum_{k=-\infty}^{+\infty} h[k] z^{n-k} = z^n \sum_{k=-\infty}^{+\infty} h[k] z^{-k}$

$$y[n] = \left(\sum_{k=-\infty}^{+\infty} h[k] z^{-k} \right) \cdot z^n \quad \text{ie. } y[n] = H(z) \cdot z^n$$



in continuous time, $s = j\omega$ (Fourier) $\Rightarrow e^{j\omega t}$

in discrete time, $z = e^{j\omega}$ (unit mag.) $\Rightarrow e^{j\omega n}$

e.g. $y(t) = x(t-3) e^{2jt} \rightarrow$ $\boxed{\quad}$ $\rightarrow e^{2j(t-3)} = e^{-6j} \cdot e^{2jt}$

$$H(s) = \int_{-\infty}^{\infty} h(t) e^{-st} dt = \int_{-\infty}^{\infty} \delta(t-3) e^{-st} dt = e^{-s(3)} = e^{-6j}$$

so, $y(t) = H(2j) \cdot e^{2jt}$

PERIODIC SIGNALS

$$n(t) = n(t+T) \quad \text{for all } t$$

T : fundamental period $\omega_0 = 2\pi/T \Rightarrow$ fundamental frequency

e.g. $\cos \omega_0 t, e^{j\omega_0 t}$

$$\Phi_K = e^{jk\omega_0 t} = e^{jK(\frac{2\pi}{T})t}; \quad K=0, \pm 1, \pm 2, \dots$$

\rightarrow each of these $\Phi_K(t)$ has a fundamental freq. as a multiple of ω_0 and each is periodic with period $= T$

$n(t) = \sum_k a_k e^{jk\omega_0 t} \rightarrow$ periodic with period $= T$.

$\int_T n(t) dt = \int_{-T}^T \sum_k a_k e^{jk\omega_0 t} dt = \sum_k a_k \int_{-T}^T e^{jk\omega_0 t} dt = \sum_k a_k \cdot 0 = 0$

\Rightarrow if $n(t)$ is real, $n(t) = n^*(t)$

$$\sum_k a_k e^{jk\omega_0 t} \xrightarrow{\text{conjugate symmetry}} \sum_{k=-\infty}^{\infty} a_k^* e^{-jk\omega_0 t} \xrightarrow{\text{Fourier transform}} \sum_{k=-\infty}^{\infty} a_k^* e^{jk\omega_0 t}$$

$\Rightarrow a_k = a_k^*$ for real $n(t)$

$$n(t) = a_0 + \sum_{k=1}^{\infty} [a_k e^{j\omega_0 t} + a_{-k} e^{-j\omega_0 t}]$$

and as $a_{-k} = a_k^*$

$$n(t) = a_0 + \sum_{k=1}^{\infty} (a_k e^{j\omega_0 t k} + a_k^* e^{-j\omega_0 t k}) = a_0 + \sum_{k=1}^{\infty} 2 \operatorname{Re} \{ a_k e^{j\omega_0 t k} \}$$

Rep. - (2)

$$n(t) = a_0 + 2 \sum_{k=1}^{\infty} a_k \cos(k\omega_0 t + \phi_{0k}) ; a_k = B_k + jC_k$$

Rep. - (3)

$$n(t) = a_0 + 2 \sum_{k=1}^{\infty} B_k \cos(\omega_0 t k) - C_k \sin(\omega_0 t k)$$

Continuous-Time Periodic Signal

$$n(t) e^{-j\omega_0 t} = \sum_{k=1}^{\infty} a_k e^{j(k\omega_0 t)} e^{-j\omega_0 t}$$

Integrating both sides from 0 to T ($= 2\pi/\omega_0$)

$$\int_0^T n(t) e^{-j\omega_0 t} dt = \sum_{k=1}^{\infty} a_k \int_0^T e^{j(k\omega_0 t)} dt$$

$$\Rightarrow \int_0^T e^{j(k-n)\omega_0 t} dt = \int_0^T \cos((k-n)\omega_0 t) dt + j \int_0^T \sin((k-n)\omega_0 t) dt$$

$k \neq n \quad 0 \quad (T = n \times \text{period})$

$$k=n \rightarrow \int_0^T e^{j(k-n)\omega_0 t} dt = \int_0^T e^{j(n-n)\omega_0 t} dt = \int_0^T e^{j(n-n)\omega_0 t} dt = 0 ; k \neq n$$

$$\text{So, } a_k = \frac{1}{T} \int_0^T n(t) e^{-j\omega_0 t} dt \quad (\text{over any length of } T)$$

$$a_0 = \frac{1}{T} \int_0^T n(t) dt$$

$$\text{eg- } n(t) = \sin \omega_0 t$$

$$a_k = \frac{1}{T} \int_0^T \sin \omega_0 t e^{-jk\omega_0 t} dt = \frac{1}{T} \int_0^T \sin \omega_0 t \cos(-k\omega_0 t) dt + j \int_0^T \sin \omega_0 t \sin(-k\omega_0 t) dt$$

$$a_1 = -j/2 \text{ and } a_{-1} = -\frac{1}{2j} \text{ and } a_k = 0, \text{ otherwise}$$

eg(2)

$$n(t) = \begin{cases} 1, & |t| < T_1 \\ 0, & T_1 \leq |t| < T_2 \end{cases}$$

$$a_0 = \frac{1}{T} \int_0^T n(t) dt = \frac{1}{T} \int_{-T_2}^{T_2} n(t) dt = \frac{1}{T} \int_{-T_2}^{T_1} dt = \frac{2T_1}{T}$$

$$a_k = \frac{1}{T} \int_{-T_2}^{T_2} n(t) e^{-jk\omega_0 t} dt = \frac{1}{T} \int_{-T_2}^{T_1} e^{-jk\omega_0 t} dt = -\frac{1}{T} \left[\frac{e^{-jk\omega_0 T_1}}{jk\omega_0} - \frac{e^{jk\omega_0 T_2}}{jk\omega_0} \right] = \frac{2T_1}{T} e^{jk\omega_0 T_2}$$

$$a_k = \frac{2}{jk\omega_0 T} \left\{ \sin(k\omega_0 T_1) \right\} \Rightarrow a_k = \frac{1}{k\pi} \sin(k\omega_0 T_1) ; k \neq 0$$

$$\text{if } T = 4\pi, q_K = \frac{\sin(K \cdot \frac{2\pi}{T} \cdot \frac{T}{4})}{K\pi} = \frac{\sin(K\pi/2)}{K\pi}$$

$$q_1 = q_{-1} = Y\pi, q_3 = q_{-3} = Y_3\pi, q_5 = q_{-5} = Y_5\pi, \dots$$

(1) absolute continuity $\int |x(t)|^2 dt < \infty$
convergence

↳ Dirichlet conditions :- (1) $\int |n(t)| dt < \infty$

$$|q_k| \leq \frac{1}{T} \int |n(t)| e^{-jkw_0 t} dt = \frac{1}{T} \int |n(t)| dt \rightarrow q_k \text{ is finite}$$

e.g. - $n(t) = 1 + 0 < t \leq 1$ with a period = 1 → (not integrable)

- $\cos(\frac{\pi}{T})$ w.p. (1) \leftarrow (2) Finite no. of maxima's & minima's in any single period
(3) Finite no. of discontinuities

PROPERTIES

signal and its Fourier coefficients

1) linearity :- $n(t) \xrightarrow{\text{FS}} q_k$ [same period T]
 $y(t) \xrightarrow{\text{FS}} b_k$ [period T]

any L.C. is also periodic with period T

$$z(t) = A n(t) + B y(t) \xrightarrow{\text{FS}} c_k = A \cdot q_k + B \cdot b_k$$

2) Time shifting :-

Shift a periodic signal → period is retained (CT)

$$(b_k (X(t-t_0))) \xrightarrow{\text{FT}} \frac{1}{T} \int n(t-t_0) e^{-jkw_0 t} dt$$

$$\text{and} = \frac{1}{T} \int n(t) e^{-jkw_0 (t+t_0)} dt = e^{-jkw_0 t_0} \frac{1}{T} \int n(t) e^{-jkw_0 t} dt = e^{-jkw_0 t_0} q_k$$

$$n(t) \xrightarrow{\text{FS}} q_k \quad n(t-t_0) \xrightarrow{\text{FS}} e^{-jkw_0 t_0} q_k$$

magnitude is unaltered, i.e. $|b_k| = |e^{-jkw_0 t_0} \cdot q_k| = |q_k|$

3) Time reversal :- $y(t) = n(-t) + k \text{ f.c.}$

$$n(t) = \sum q_m e^{-jkw_0 t}$$

$$k = -m$$

$$(n(-t)) = \sum q_m e^{jkw_0 t} = a_m \text{ i.e. } a_m = q_m$$

$$n(t) \xrightarrow{\text{FS}} q_k \quad \text{Then } n(-t) \xrightarrow{\text{FS}} q_{-k}$$

(Time) reversal of the corresponding co-efficients

→ if $n(t)$ is even, its Fourier coeff. are also even

→ if $n(t)$ is odd, its F.C. are also odd.

(4)

Time Scaling :- changes the period of the signal.

F-coeff. don't change but the "series" changes

$$n(xt) = \sum a_k e^{j k \omega_0 t}$$

(5)

Multiplication :- $n(t), y(t)$ are periodic with time period T .

$$n(t) y(t) \xleftarrow{\text{P.S.}} h_k = \sum_{l=-\infty}^{+\infty} a_l b_{k-l}$$

convol' of seq sequence
representing $x(t)$ and $y(t)$

multiply Fourier series of $x(t)$ and $y(t)$ \Rightarrow K^{th} harmonic \Rightarrow coeff. in sum of terms as $a_l b_{k-l}$

(6)

Conjugate & Conjugate symmetry :-

$$x(t) \longleftrightarrow a_k \quad x^*(t) \xrightarrow{\text{F.S.}} a_k^*$$

real $n(t) \Rightarrow a_k = a_k^*$ (conjugate symmetry)

if $x(t)$ is real $\Rightarrow a_0$ is real

$$\Rightarrow |a_k| = |a_{-k}|$$

if $n(t)$ is real & even $\Rightarrow a_k = a_{-k} \Rightarrow a_k = a_k^* \Rightarrow a_k$ is real

L F-coeff are real

$n(t)$ is real & odd \Rightarrow F-coeff are purely imaginary & odd.

(7)

Parseval's relation :-

$$\frac{1}{T} \int |x(t)|^2 dt = \sum_{-\infty}^{+\infty} |a_k|^2$$

\Rightarrow Fourier series for sine & cosine

$$a_k = a_{-k}^* \Rightarrow n(t) = a_0 + \sum a_k \cos(k\omega_0 t + \phi_k)$$

$\cos\left(\frac{2\pi}{T} kt\right)$ and $\sin\left(\frac{2\pi}{T} kt\right)$ form a basis

$$\langle x_1, x_2 \rangle = \int x_1 \overline{x_2} dt$$

$$P(t) = \sum A_k \cos(k\omega_0 t + \phi_k) + A_0$$

$$A_k = \frac{1}{T} \int P(t) \cos\left(\frac{2\pi}{T} kt\right) dt$$

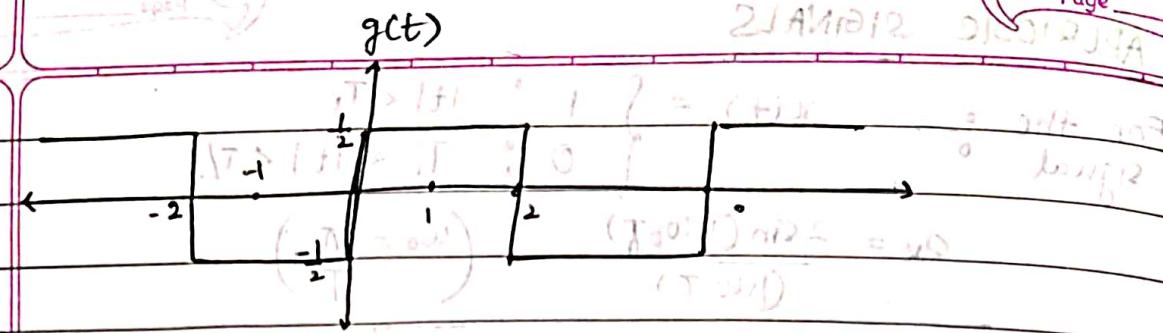
$$A_k \cos\left(\frac{2\pi}{T} kt + \phi_k\right) = \left\{ \frac{2}{T} \int P(u) \cos\left(\frac{2\pi}{T} ku\right) du \right\} \cos\left(\frac{2\pi}{T} kt\right)$$

$$+ \left\{ \frac{2}{T} \int P(u) \sin\left(\frac{2\pi}{T} ku\right) du \right\} \sin\left(\frac{2\pi}{T} kt\right)$$

$$P(t) = A_0 + \frac{2}{T} \sum \left(\int P(u) \cos\left(\frac{2\pi}{T} ku\right) du \right) \cos\left(\frac{2\pi}{T} kt\right) + \frac{2}{T} \sum \left(\int P(u) \sin\left(\frac{2\pi}{T} ku\right) du \right) \sin\left(\frac{2\pi}{T} kt\right)$$

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e.g.



with $T_0 = 4 \quad T_1 = 1 \quad x(t-1) - \frac{1}{2} = g(t)$

$$g(t) = n(t-1) - \frac{1}{2}$$

$$x(t) \longleftrightarrow a_k \quad n(t-1) \longleftrightarrow e^{-j\frac{2\pi k}{T}} \cdot a_k = a_k e^{-jk\pi/2}$$

$$\text{For DC offset, } -y_2 \quad C_K = \begin{cases} 1/2 & K \neq 0 \\ -1/2 & K=0 \end{cases}$$

$$\text{so, } d_K = \begin{cases} a_k e^{-jk\pi/2} & K \neq 0 \\ a_0 - y_2 & K=0 \end{cases} \quad a_k = \frac{\sin(K\pi/2)}{(K\pi/2)}$$

$$d_K = \frac{\sin(K\pi/2)}{K\pi/2} e^{-jk\pi/2}; \quad K \neq 0$$

$$a_0 = \frac{11}{2}$$

⑧ Differentiation :- $x(t) \longleftrightarrow a_k$

$$\frac{dx(t)}{dt} \longleftrightarrow j\omega_0 k a_k = \boxed{j \frac{2\pi}{T} k a_k}$$

$$x(t) = \sum a_k e^{j\omega_0 k t} \quad \frac{dx(t)}{dt} = \sum (j\omega_0 k a_k) e^{j\omega_0 k t}$$

⑨

Integration :- $\int_{-\infty}^t x(t) dt \longleftrightarrow a_k$

$$\int_{-\infty}^t x(t) dt \rightarrow \left(\frac{1}{j\omega_0 k} \right) a_k = \left(\frac{1}{j k (2\pi/T)} \right) a_k$$

eg.- impulse-train \Rightarrow with period T

$$n(t) = \sum_{k=-\infty}^{+\infty} \delta(t - KT)$$

$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} \delta(t) \cdot e^{-j\frac{2\pi k t}{T}} dt = \frac{1}{T} \rightarrow \text{all fourier coeff's are identical.}$$

$$\text{For } s=j\omega, H(j\omega) = \int_{-\infty}^{+\infty} h(t) e^{-j\omega t} dt = e^{j\omega t} \rightarrow H(j\omega) e^{j\omega t}$$

$$\text{For } l \cdot n(t) = \sum_{k=-\infty}^{\infty} a_k e^{j\omega_0 t + k} \rightarrow Y(t) = \sum_{k=-\infty}^{\infty} a_k H(j \cdot k \omega_0) e^{j k \omega_0 t}$$

APERIODIC SIGNALS

Date _____

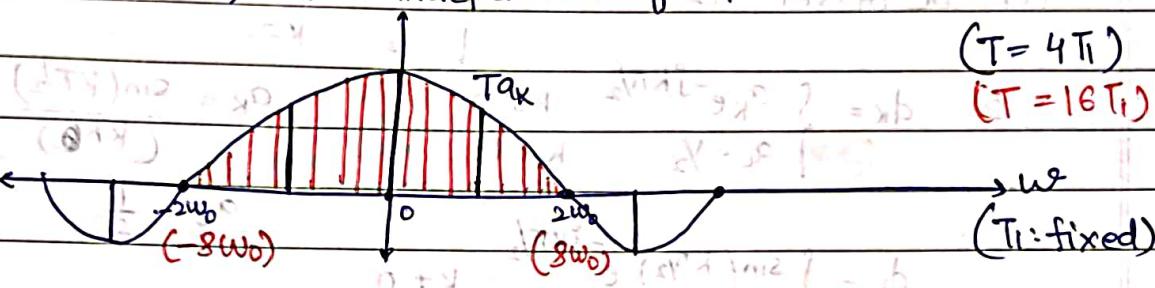
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For the signal $x(t) = \begin{cases} 1 & |t| < T_1 \\ 0 & T_1 < |t| < T/2 \end{cases}$

$$a_K = \frac{2 \sin(K\omega_0 T)}{(K\omega_0 T)} \quad \left(\omega_0 = \frac{2\pi}{T} \right)$$

$$T_{AK} = \frac{2 \sin(\omega T_1)}{\omega} \quad \left(\omega = K\omega_0 \right)$$

- with ω as a cont. variable, they are envelopes of T_{AK} . (coupled)
- for fixed T_1 , it is independent of T .



- as T increases (or $\omega \downarrow$) [$T = 2\pi/\omega_0$] → envelope is sampled closer & closer.
- as $T \rightarrow \infty$, pulse → rectangular pulse : Fourier coeff. approach T_{AK} as $T \rightarrow \infty$

i.e.

Let $n(t)$ be some finite signal ; i.e. $n(t) = 0 \forall |t| > t_1$

↳ we have a periodic signal $\tilde{n}(t)$ for which $n(t)$ is one period and then we choose its period $T \rightarrow \infty$

$n(t)$ $\tilde{n}(t)$

$$\tilde{n}(t) = \sum_{-\infty}^{+\infty} a_k e^{j k \omega_0 t} \quad a_k = \frac{1}{T} \int_{-T/2}^{T/2} \tilde{n}(t) e^{-j k \omega_0 t} dt$$

$n(t) = \tilde{n}(t)$ for $|t| < T/2$

$$T = 0 \quad \omega_0$$

$$a_k = \frac{1}{T} \int_{-\infty}^{\infty} n(t) e^{-j k \omega_0 t} dt = \frac{1}{T} \int_{-T/2}^{T/2} \tilde{n}(t) e^{-j k \omega_0 t} dt$$

defining an envelope $X(j\omega)$ of T_{AK} as

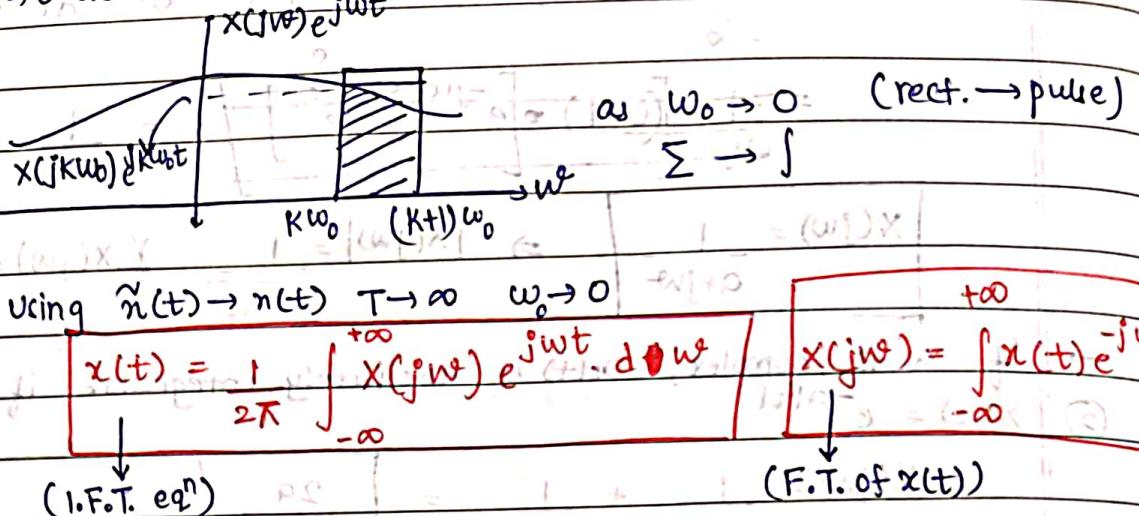
$$X(j\omega) = \int_{-\infty}^{+\infty} x(t) e^{-j \omega t} dt$$

$$so, \quad a_k = \frac{1}{T} X(j k \omega_0)$$

$$so, \quad \tilde{x}(t) = \sum_{k=-\infty}^{+\infty} \frac{1}{T} X(j k \omega_0) e^{j k \omega_0 t}$$

$$\tilde{x}(t) = \frac{1}{2\pi} \sum_{k=-\infty}^{+\infty} X(j k \omega_0) e^{j k \omega_0 t} \quad (\omega_0)$$

as $T \rightarrow \infty$ $\tilde{n}(t) \rightarrow n(t)$ and the carrier eqⁿ becomes a rep' of $n(t)$
 $w_0 \rightarrow 0$ as $T \rightarrow \infty$ and so RHS \rightarrow integral (limit of sum)



so, using $\tilde{n}(t) \rightarrow n(t)$ $T \rightarrow \infty$ $w_0 \rightarrow 0$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(jw) e^{jwt} dw$$

(I.F.T. eqⁿ)

$$X(jw) = \int_{-\infty}^{+\infty} x(t) e^{-jwt} dt$$

(F.T. of x(t))

For aperiodic signals, complex exponentials occur at a continuum of frequencies, and, have an amplitude $(X(jw) \frac{dw}{2\pi})$

F.S. coefficients $\rightarrow X(jw)$ is the spectrum of $n(t)$

a_K (F.coeff.) of some $\tilde{n}(t)$ (periodic, period = T)

$n(t) = \tilde{n}(t)$ for $s \leq t \leq s+T$ (over one period)
 (finite dur') for some s

$$a_K = \frac{1}{T} \int_s^{s+T} \tilde{n}(t) e^{-jwtk} dt = \frac{1}{T} \int_s^{s+T} n(t) e^{-jwtk} dt$$

$$a_K = \frac{1}{T} \int_{-\infty}^{\infty} n(t) e^{-jwtk} dt \Rightarrow a_K = \frac{1}{T} X(jw) |_{w=kw_0}$$

F-coeff. \rightarrow expressed in terms of equally-spaced samples of the Fourier Transform of one period of $\tilde{n}(t)$.

$$\text{if } \tilde{n}(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(jw) e^{jwt} dw$$

$\hat{x}(t) = n(t)$ except for any 't' which is a discontinuity, where \hat{n} is the average on either sides

Dirichlet conditions: if $n(t)$ is absolutely integrable: $\int_{-\infty}^{+\infty} |x(t)| dt < \infty$

2) Finite no. of maxima & minima in any interval

3) Finite no. of discontinuities in any finite interval, each of these must also be finite

eg:-

$$n(t) = e^{-at} \cdot u(t); a > 0$$

$$X(j\omega) = \int_{-\infty}^{\infty} n(t) e^{-j\omega t} dt = \int_0^{\infty} e^{-at} e^{-j\omega t} dt = -\frac{1}{a+j\omega} [e^{-at}]_0^{\infty}$$

$$= -\frac{1}{a+j\omega} [(0-1) - [e^{-j\omega(\infty)} - e^0]] = -\frac{1}{a+j\omega} [0-1]$$

$$X(j\omega) = \frac{1}{a+j\omega} \Rightarrow |X(j\omega)| = \frac{1}{\sqrt{a^2 + \omega^2}} \quad X(j\omega) = -\tan^{-1}\left(\frac{\omega}{a}\right)$$

→ if a is complex, $n(t)$ is absolutely integrable if $\operatorname{Re}(a) > 0$

$$\textcircled{2} \quad x(t) = e^{-at} u(t)$$

$$\textcircled{1.5} \quad X(j\omega) = \frac{1}{a-j\omega} + \frac{1}{a+j\omega} = \frac{2a}{a^2 + \omega^2}$$

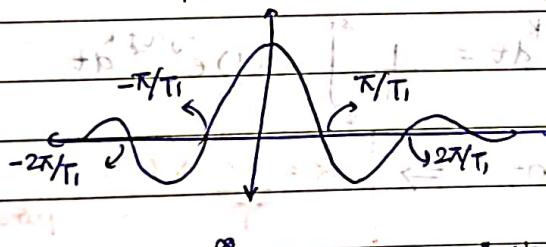
$$\textcircled{3} \quad n(t) = \delta(t) \quad (\text{unit impulse})$$

(unit impulse has a F.T. consisting of equal contrib'n at all frequencies)

$$\textcircled{4} \quad x(t) = \begin{cases} 1 & ; |t| < T_1 \\ 0 & ; |t| > T_1 \end{cases}$$

(rectangular pulse)

$$X(j\omega) = \int_{-T_1}^{T_1} e^{-j\omega t} dt = \frac{1}{j\omega} [e^{j\omega T_1} - e^{-j\omega T_1}] = \frac{2(\sin \omega T_1)}{\omega}$$



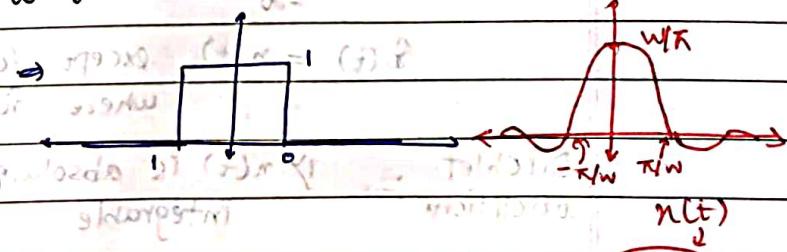
$\hat{n}(t) = n(t)$ ex. at the

points of discontinuity
at $(\pm T_1)$ where

$n(t) \rightarrow \frac{1}{2}$ [avg. on both sides
of discontinuity]

$$\text{eg. } \textcircled{1} \quad X(j\omega) = \begin{cases} 1, & |\omega| < w \\ 0, & |\omega| > w \end{cases}$$

$$n(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} 1 \cdot e^{j\omega t} dw$$



$$\text{symmetric w.r.t. origin} = \frac{1}{2\pi j\omega} \int_{-\infty}^{\infty} [e^{j\omega t} - e^{-j\omega t}] dw = \frac{\sin \omega t}{\pi \omega t}$$

$$= \frac{\sin \omega t}{\pi t}$$

$$\text{sinc}(\theta) = \frac{\sin \pi \theta}{\pi \theta}$$

$$\frac{2 \sin \omega t_1}{\omega} = 2 T_1 \sin c \left(\frac{\omega T_1}{\pi} \right)$$

$$\frac{\sin \omega t}{\pi t} = \frac{\omega}{\pi} \cdot \sin \left(\frac{\omega t}{\pi} \right)$$

as $\omega \uparrow$ in $\sin \omega t$ $\rightarrow x(j\omega)$ becomes broader & peak of $n(t)$ becomes higher and width of 1st peak becomes narrower.

in limit as $\omega \rightarrow \infty$ $x(j\omega) = 1 + \omega$, and $n(t) \rightarrow \text{impulse}(s)$

Inverse reln b/w time and frequency domains.

FOURIER TRANSFORM OF PERIODIC SIGNALS

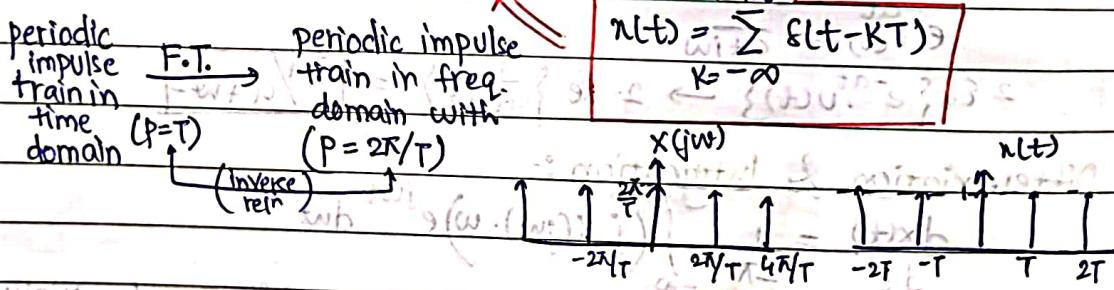
train of impulses in freq. domain with areas \propto F.S. coeff.

$$\text{eg. } X(j\omega) = 2\pi \delta(\omega - \omega_0)$$

$$n(t) = \frac{1}{2\pi} \int 2\pi \delta(\omega - \omega_0) e^{j\omega_0 t} d\omega = e^{j\omega_0 t}$$

$$\text{So, if } X(j\omega) = \sum_{K=-\infty}^{+\infty} 2\pi q_K \delta(\omega - K\omega_0) \Rightarrow n(t) = \sum_{K=-\infty}^{+\infty} q_K e^{j\omega_0 Kt}$$

$$\text{e.g.- } X(j\omega) = \frac{2\pi}{T} \sum_{K=-\infty}^{+\infty} \delta\left(\omega - \frac{2\pi K}{T}\right) \Rightarrow n(t) = \frac{1}{2\pi} \int \frac{2\pi}{T} \delta\left(\omega - \frac{2\pi K}{T}\right) e^{j\omega t} d\omega$$



PROPERTIES

$$X(j\omega) = F\{x(t)\} \text{ and } n(t) = F^{-1}\{X(j\omega)\} \quad n(t) \xleftrightarrow{F} X(j\omega)$$

① Linearity :- $n(t) \xleftrightarrow{F} X(j\omega)$ $y(t) \xleftrightarrow{F} Y(j\omega)$

$$a x(t) + b y(t) \xleftrightarrow{F} a X(j\omega) + b Y(j\omega)$$

② Time Shifting :- $x(t-t_0) \xrightarrow{-\infty} \int x(t-t_0) e^{-j\omega_0 t} dt$

$$x(t-t_0) \xrightarrow{} e^{-j\omega_0 t_0} X(j\omega)$$

$$\text{or } x(t) = \frac{1}{2\pi} \int x(j\omega) e^{j\omega t} d\omega \quad \xrightarrow{\text{Magnitude of F.T.} \rightarrow \text{same phase} \rightarrow \text{Shift by } -\omega t_0}$$

$$n(t-t_0) = \frac{1}{2\pi} \int x(j\omega) e^{j\omega(t-t_0)} dt = e^{-j\omega t_0} F(x(t))$$

③ Conjugation & Conjugate Symmetry :-

$$n(t) \xrightarrow{F} X(jw) \Rightarrow X^*(t) \xrightarrow{F} X^*(-jw)$$

$$X(jw) = \int n(t) e^{-jw t} dt$$

$$X^*(jw) = \int n^*(t) e^{jw t} dt \text{ so, } X^*(-jw) = \int n^*(t) e^{-jw t} dt = F(X^*(t))$$

if $n(t)$ is real $\Rightarrow [X(jw) = X^*(-jw)] \text{ or } [X(-jw) = X^*(jw)]$

so, $\Re\{X(jw)\} = \Re\{X(-jw)\} \rightarrow \text{real part is even f.c. of frequency}$.

and $\Im\{X(jw)\} = -\Im\{X(-jw)\} \rightarrow \text{im. part is an odd f.c. of frequency}$.

$$X(jw) = |X(jw)| e^{j\arg X(jw)}$$

$\xleftarrow{\text{even f.c. of w}}$ $\xrightarrow{\text{odd f.c. of w}}$

$$\Rightarrow n(t) \text{ is real and even, } X(-jw) = \int n(t) e^{-jw t} dt \quad t \rightarrow -t$$

$$\text{if } n(t) = n(-t) = \int_{-\infty}^{\infty} n(-t) e^{-jw t} (-1) dt$$

$$[X(-jw) = X(jw)] \quad [X(-jw) = F(n(t))]$$

$$\text{and } X(jw) = X^*(-jw) = X(-jw)$$

$\Rightarrow n(t)$ is real & even $\Rightarrow X(jw)$ is also real & even.

$n(t)$ is real & odd $\Rightarrow X(jw)$ is purely imaginary

$$\text{so, } n(t) \xrightarrow{F} X(jw)$$

$$(\text{even}) \quad \boxed{\text{Ev}\{n(t)\} \xrightarrow{F} \Re\{X(jw)\}}$$

$$\boxed{\text{Od}\{n(t)\} \xrightarrow{F} j\Im\{X(jw)\}}$$

$$\text{eg. } e^{-at} u(t) \rightarrow e^{-at} u(t) + e^{at} u(-t) = 2 \boxed{\text{Ev}(e^{-at} u(t))}$$

$$e^{-at} u(t) \rightarrow \frac{1}{at+jw}$$

$$2 \text{Ev}\{e^{-at} u(t)\} \rightarrow 2 \cdot \text{Re}\{\frac{1}{at+jw}\} = \frac{2a}{a^2+w^2}$$

④ Differentiation & Integration :-

$$\frac{dx(t)}{dt} = \frac{1}{2\pi} \int_{-\infty}^{\infty} (jw X(jw) \cdot w) e^{jw t} dw$$

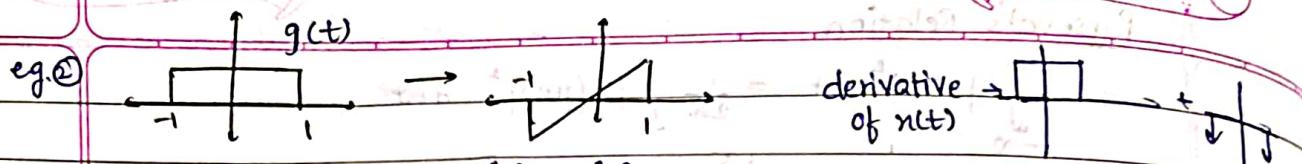
$$\frac{dx(t)}{dt} \xrightarrow{F} (jw) \cdot X(jw) \quad \begin{matrix} \text{diff. in} \\ \text{time} \end{matrix} \rightarrow \begin{matrix} (\text{multiply with } jw) \\ \text{in freq. domain} \end{matrix}$$

$$\frac{1}{jw} \int_{-\infty}^t n(\tau) d\tau \xrightarrow{F} \frac{1}{jw} X(jw) + \pi X(0) s(w)$$

(DC/Average value)

$$\text{e.g. } u(t) \rightarrow G(jw) = 1$$

$$\frac{1}{jw} \int_{-\infty}^t \delta(t) dt \xrightarrow{F} \frac{1}{jw} (1) + \pi (1) \cdot \delta(w) = \frac{1}{jw} + \pi \delta(w)$$



$$G(j\omega) = \frac{2\sin\omega}{\omega} - e^{-j\omega} - e^{j\omega}$$

$$1 \cdot \delta(\omega+1) \rightarrow e^{j\omega} F(\delta) \Rightarrow e^{j\omega} \quad (\text{Integr}^n)$$

$$\text{as } G(0)=0 \rightarrow X(j\omega) = \frac{G(j\omega)}{(j\omega)} \rightarrow \boxed{\frac{2\sin\omega}{j\omega^2} - \frac{2\cos\omega}{j\omega}}$$

F.T.

⑤ Time And Frequency Scaling :-

$$x(t) \xleftrightarrow{F} X(j\omega) \rightarrow \boxed{x(at) \xleftrightarrow{|a|} X(j\omega/a)}$$

$$\text{a. } F\{n(at)\} = \begin{cases} \frac{1}{a} \int_{-\infty}^{\infty} n(\tau) e^{-j(\omega/a)\tau} d\tau & a > 0 \\ -\frac{1}{a} \int_{-\infty}^{\infty} n(\tau) e^{-j(\omega/a)\tau} d\tau & a < 0 \end{cases}$$

$$\text{For } a = -1 \Rightarrow \boxed{x(-t) \xleftrightarrow{F} X(-j\omega)}$$

⑥ DUALITY :-

$$\text{e.g. } g(t) = \frac{2}{1+t^2} \quad n(t) = e^{-|t|} \xleftrightarrow{F} X(j\omega) = \frac{2}{1+\omega^2}$$

$$e^{|t|} = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{+\infty} \left(\frac{2}{1+w^2} \right) e^{j\omega t} dw \Rightarrow 2\pi e^{-|t|} = \int_{-\infty}^{\infty} \left(\frac{2}{1+w^2} \right) e^{-j\omega t} dw$$

$$\text{So, } 2\pi e^{-|t|} = \int_{-\infty}^{\infty} \left(\frac{2}{1+w^2} \right) e^{-j\omega t} dw \xrightarrow{\text{replace } (t \rightarrow -t)} \int_{-\infty}^{\infty} \left(\frac{2}{1+t^2} \right) e^{-j\omega t} dt$$

interchange
names of
t, w

$$\text{So, } \boxed{x(t) \xleftrightarrow{F} X(j\omega)} \\ \boxed{X(jt) \xleftrightarrow{F} 2\pi x(-w)}$$

$$\int \left(\int x(t) e^{-j\omega t} dt \right) e^{j\omega t} dw = F(t)$$

$$2\pi F(-\omega) = \int X(j\omega) e^{-j\omega t} dt$$

$$\xrightarrow{\text{replace names of } t, w} \int x(-jt) e^{-j\omega t} dt = f(t) 2\pi$$

$$\int x(j\omega) e^{-j\omega t} dw \xrightarrow{\text{again replace } \omega} 2\pi x(-w) \rightarrow \int x(-jt) e^{-j\omega t} dt$$

$$\int x(j\omega) e^{-j\omega t} dw \xrightarrow{\text{f(t)}} \int x(jt) e^{-j\omega t} dw \xrightarrow{\text{t} \rightarrow w} \frac{1}{2\pi} \int x(-jt) e^{-j\omega t} dw = f(t) = x(-t)$$

$2\pi x(-\omega) \Rightarrow$ multiply by diff. in
jt in time \Rightarrow freq. domain

$$\boxed{-jt n(t) \xleftrightarrow{F} dX(j\omega)} \xrightarrow{\text{d}\omega} \int -jt n(t) e^{-j\omega t} dt$$

$$\text{e}^{j\omega_0 t} n(t) \xleftrightarrow{F} X(j(\omega - \omega_0))$$

$$\frac{-1}{jt} x(t) + \pi x(0) \delta(t) \xrightarrow{F} \int x(\eta) d\eta$$

Parseval's Relation

$$\int_{-\infty}^{+\infty} |n(t)|^2 \cdot dt = \frac{1}{2\pi} \int |X(jw)|^2 \cdot dw$$

i.e.

$$\int_{-\infty}^{+\infty} n(t) \overline{n(t)} dt = \int_{-\infty}^{+\infty} n(t) \cdot \frac{1}{2\pi} \left[\int_{-\infty}^{+\infty} \overline{x(jw)} e^{-jwt} dw \right] dt$$

$$= \int_{-\infty}^{+\infty} n(t) \cdot \frac{1}{2\pi} \left[\int_{-\infty}^{+\infty} \overline{x(jw)} e^{-jwt} dw \right] \cdot dt$$

reversing the order of integration

$$\begin{aligned} & \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} \frac{1}{2\pi} (n(t)) \overline{x(jw)} e^{-jwt} dw dt \\ &= \int_{-\infty}^{+\infty} \frac{1}{2\pi} \overline{x(jw)} \left[\int_{-\infty}^{+\infty} n(t) e^{-jwt} dt \right] dw \\ &= \frac{1}{2\pi} \int_{-\infty}^{+\infty} \overline{x(jw)} x(jw) \cdot dw = \frac{1}{2\pi} \int_{-\infty}^{+\infty} |x(jw)|^2 \cdot dw \end{aligned}$$

A.45

$\int_{-\infty}^{+\infty} |n(t)|^2 \cdot dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty} |x(jw)|^2 \cdot dw$

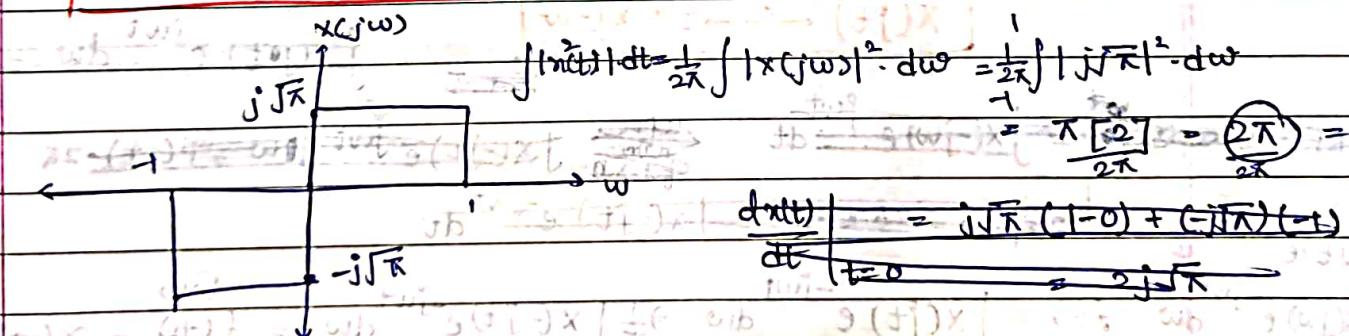
energy per unit time

Total energy \Rightarrow energy - density spectrum

$$\begin{aligned} x(jw) &= \int n(t) e^{-jwt} dt \\ \text{att} &= \frac{1}{2\pi} \int x(jw) e^{jwt} dw \\ \frac{dx(t)}{dt} &= \frac{1}{2\pi} \int x(jw) jw e^{jwt} \cdot dw \end{aligned}$$

$$\left| \frac{dx(t)}{dt} \right|_{t=0} = \frac{1}{2\pi} \int x(jw) \cdot jw \cdot dw$$

e.g.-



$$\left| \frac{dx(t)}{dt} \right|_{t=0} = \frac{1}{2\pi} \int (-j\sqrt{\pi}) jw dw + \frac{1}{2\pi} \int (j\sqrt{\pi})(jw) dw$$

$$= \frac{1}{2\pi} \left[j\sqrt{\pi} \left[\frac{1}{2} (0-1) \right] + (-j\pi) \left[\frac{1}{2} \right] \right] = -\frac{1}{2\sqrt{\pi}}$$

CONVOLUTION PROPERTY

complex exponentials are eigenfunctions of LTI systems.

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega = \lim_{\omega_0 \rightarrow 0} \frac{1}{2\pi} \sum_{k=-\infty}^{+\infty} X(jk\omega_0) e^{jk\omega_0 t}$$

$$e^{jk\omega_0 t} \xrightarrow{\text{LSI system}} H(jk\omega_0) \cdot e^{jk\omega_0 t} \quad H(jk\omega_0) = \int_{-\infty}^{+\infty} h(t) e^{-jkt} dt$$

$$H(j\omega) = F(h(t))$$

↳ frequency response

$$\text{so, } Y(t) = \lim_{\omega_0 \rightarrow 0} \left(\frac{1}{2\pi} \sum_{k=-\infty}^{+\infty} X(jk\omega_0) H(jk\omega_0) e^{jk\omega_0 t} \right) \cdot \omega_0 = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) H(j\omega) e^{j\omega t} d\omega$$

$$\text{so, } Y(j\omega) = X(j\omega) \cdot H(j\omega)$$

$$\Rightarrow y(t) = \int_{-\infty}^{+\infty} x(\tau) h(t-\tau) d\tau$$

$$\begin{aligned} \int_{-\infty}^{+\infty} y(t) e^{-j\omega t} dt &= \int_{-\infty}^{+\infty} \left(\int_{-\infty}^{+\infty} x(\tau) h(t-\tau) d\tau \right) e^{-j\omega t} dt \\ &= \int_{-\infty}^{+\infty} x(\tau) \left(\int_{-\infty}^{+\infty} h(t-\tau) e^{-j\omega(t-\tau)} dt \right) e^{-j\omega\tau} d\tau \\ &\stackrel{\text{Ab. } f((s-a)^+)}{=} \int_{-\infty}^{+\infty} \left(x(\tau) e^{-j\omega\tau} d\tau \right) \left(\int_{-\infty}^{+\infty} h(s) e^{-j\omega s} ds \right) \end{aligned}$$

$$Y(j\omega) = X(j\omega) \cdot H(j\omega)$$

$$h(t) * x(t) \xrightarrow{\text{F}} H(j\omega) \cdot X(j\omega)$$

↳ completely characterizes a LTI system

$$n(t) \xrightarrow{\text{LTI system}} H_1(j\omega) \xrightarrow{\text{LTI system}} H_2(j\omega) \xrightarrow{\text{LTI system}} y(t) \Rightarrow n(t) \xrightarrow{\text{LTI system}} H_1(j\omega) \cdot H_2(j\omega) \xrightarrow{\text{LTI system}} y(t)$$

\Rightarrow LTI system is BIBO stable $\Rightarrow \int_{-\infty}^{+\infty} |h(t)| dt < \infty \rightarrow$ absolutely integrable

$$(1) \quad \mathcal{F}\{x(t)=c\} = C \int_{-\infty}^{+\infty} e^{-j \cdot 2\pi ft} dt \quad \text{ex. when } n=0$$

$$F[A] = \lim_{r \rightarrow \infty} F[A \cdot \text{rect}(t/r)] = A \lim_{r \rightarrow \infty} \gamma \text{sinc}(\omega r/2)$$

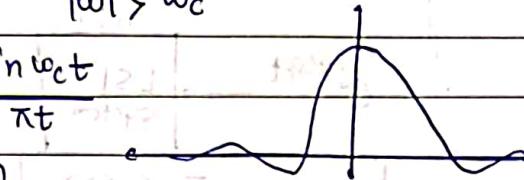
$$\lim_{r \rightarrow \infty} \gamma \text{sinc}(\omega r/2) = \lim_{r \rightarrow \infty} \frac{\sin(\omega r/2)}{(\omega r/2)}$$

$$F[A] \rightarrow 2\pi A \delta(\omega)$$

$$F[e^{j\omega_0 t}] \rightarrow 2\pi \delta(\omega + \omega_0)$$

$$\text{Frequency filter} \Rightarrow H(j\omega) = \begin{cases} 1 & |\omega| \leq \omega_c \\ 0 & |\omega| > \omega_c \end{cases}$$

$$\mathcal{F}^{-1}\{H(j\omega)\} = h(t) = \frac{\sin \omega_c t}{\pi t}$$



i) $h(t)$ is NOT $= 0$ for $t < 0 \rightarrow$ NOT causal

e.g. $h(t) = e^{-at} u(t)$] RC circuit i) causal
 $H(j\omega) = \frac{1}{1+j\omega}$] circuit ii) realizable

e.g. $h(t) = e^{-at} u(t)$ $x(t) = e^{-bt} u(t) \Rightarrow H(j\omega) = \frac{1}{a+j\omega}$ $X(j\omega) = \frac{1}{b+j\omega}$

$$Y(j\omega) = \frac{1}{(a+j\omega)(b+j\omega)} = \frac{A}{a+j\omega} + \frac{B}{b+j\omega}$$

$$Y(j\omega) = \frac{1}{b-a} \left[\frac{1}{a+j\omega} - \frac{1}{b+j\omega} \right] \therefore Y(t) = \frac{1}{b-a} \left[e^{-at} - e^{-bt} \right] u(t)$$

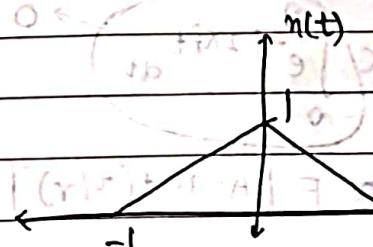
MULTIPLICATION PROPERTY

$$s(t) p(t) \rightarrow R(j\omega) = \frac{1}{2\pi} \int_{-\infty}^{\infty} S(j\theta) P(j(\omega-\theta)) d\theta$$

Some Properties

- (1) $\operatorname{Re}\{X(j\omega)\} = \operatorname{Re}\{X(-j\omega)\} = 0 \Rightarrow n(t) = -n(-t) \Rightarrow$ ODD & real
- (2) $\operatorname{Im}\{X(j\omega)\} = -\operatorname{Im}\{X(-j\omega)\} = 0 \Rightarrow n(t) = n(-t) \Rightarrow$ EVEN & real
- (3) If a real & so T. $e^{j\omega_0 t} X(j\omega)$ is real $\Rightarrow n(t+a)$ is real & even
- (4) $\int_{-\infty}^{\infty} X(j\omega) d\omega = 0 \rightarrow n(0) = 0$
- (5) $\int X(j\omega) \cdot \omega d\omega = 0 \Rightarrow \left. \frac{dX}{dt} \right|_{t=0} = 0$
- (6) $X(j\omega)$ is periodic $\rightarrow n(t)$ is periodic

e.g. Find the fourier transform of



Ans) it is the convolution as $x_1(t) * x_1(t)$ and where $x_1(t) = \begin{cases} 1 & |t| < \frac{1}{2} \\ 0 & \text{else} \end{cases}$

~~24)~~ $x(t) = \begin{cases} 1 & -\frac{1}{2} \leq t \leq \frac{1}{2} \\ 0 & \text{otherwise} \end{cases}$

$\int_{-\infty}^{+\infty} x_1(\tau) x_1(t-\tau) d\tau$

$|t-\tau| \leq \frac{1}{2}$

$t - \frac{1}{2} \leq \tau \leq t + \frac{1}{2}$

$t - \frac{1}{2} \leq t - \tau \leq t + \frac{1}{2}$

$-y_2 \leq \tau \leq y_2$

$-y_2 \leq t - \tau \leq y_2$

$(Tn-1) \leq t \leq Tn$

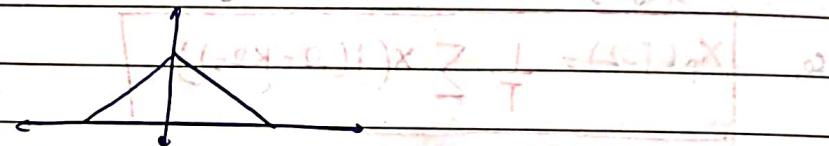
$t - y_2 \leq \tau \leq t + y_2$

$t - \frac{1}{2} \leq \tau \leq t + \frac{1}{2}$

$\int_{t-\frac{1}{2}}^{t+\frac{1}{2}} 1 \cdot dt = \left(t + \frac{1}{2} \right) - \left(t - \frac{1}{2} \right) = 1$

for $t - \frac{1}{2} = -\frac{1}{2} \Rightarrow t = 0$ to $t + \frac{1}{2} = \frac{1}{2} \Rightarrow t = 1$

and $\int_{t-\frac{1}{2}}^{t+\frac{1}{2}} 1 \cdot dt = \left(t + \frac{1}{2} \right) - \left(t - \frac{1}{2} \right) = 1$



$\therefore D = 1/2 \times \text{Area}$ of triangle = $1/2 \times 1 \times 1 = 1/2$

$\therefore F.T. (x(t)) = F.T. (x_1(t) * x_1(t)) = (X_1(jw))^2$

$= \left(\frac{2 \sin(w_0)}{w} \right)^2$

$\# n(t) \xrightarrow{\text{F.S.}} a_k - (2D)jw \xrightarrow{\text{cancelling } (2D)jw}$

$n(t-t_0) \xrightarrow{} a_k e^{-jw_0 t_0}$ $n(t+t_0) \xrightarrow{} a_k e^{jw_0 t_0}$

1) $n(t-t_0) + n(t+t_0) \xrightarrow{} a_k \cdot \{ 2 \cos(w_0 t_0 \cdot k) \}$

2) $E_n(t) = n(t) + \bar{n}(t) \xrightarrow{} (a_k + a_{-k})$

3) $\text{Re}\{n(t)\} = n(t) + \bar{n}(t) \xrightarrow{2} (a_k + a_{-k})$

4) $n(3t-1) \xrightarrow{} \sum a_k e^{jkw_0(3t-1)} = \sum a_k e^{-jkw_0} \cdot e^{jkw_0 t}$
 $(\text{Period} = 3T/3) = |a_k \cdot e^{-jkw_0}|$

$\# n(t) \xrightarrow{} X(jw) \quad n(-t) \xrightarrow{} X(-jw)$

$u(t) = \frac{1}{2} [1 + \text{sgn}(t)]$

$\text{sgn}(t) = \lim_{a \rightarrow 0^+} \left[e^{-at} u(t) - e^{+at} u(-t) \right]$

$F(\text{sgn}(t)) = \frac{1}{jw}$