3.0: INTRODUCTION

So fare, we have analyzed relatively simple circuits by applying Kirchhoff's laws in combination with Ohm's law. This approach can be used for all circuits but as they become more complicated and involve more elements, this direct method becomes very cumbersome. In this chapter, we will apply these laws to develop two powerful techniques that aid in the analysis of complex circuit structures.

- 1. Modal Analysis: Based on a systematic application of Kirchhoff's current law (KCL).
- 2. Mesh Analysis: Based on a systematic application of Kirchhoffls voltage law (KVL).

Using these two techniques, we can analyze any linear circuit by obtaining a set of simultaneous equations that are then solved to obtain the required values of current or voltage. Ex Cramero's rolle is used for solving simultaneous equations which allows to calculate circuit variables as a quotient of determinants.

3.1: NODAL ANALYSIS

Nodal analysis gives a general technique for analyzing circuits using node voltages as the circuit learnables. In this section, we shall assume that circuits do not

- 1. Select one node as the reference node.

 Assign voltages to the remaining n-1 nodes,
 i.e., 21, 22, ----, 21. The voltages are
 referenced with respect to the reference node.
 - 2. Apply KCL to each of the n-1 nonreference node. Express the branch currents in terms of node voltages by using Ohm's law.
 - 3. Solve the simultaneous equations to obtain the unknown node voltages.

We shall now explain these three steps systematically.

The first step in the nodal analysis is to select a node as the reference or datum node. The reference node is assumed to have zero potential and is commonly known as ground. A reference node is indicated by any of the three symbols as shown in Fig. 3.1. The type of ground shown in Fig. 3.1(a) is a known as chassis ground—and is used in devices where the chassis, case or enclosure acts and as a reference point for

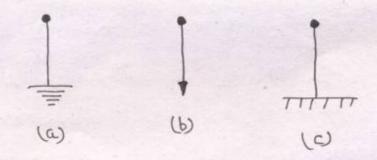


Fig. 3.1: Different symbols for indicating a reference node

(a) chassis ground (b) common ground
(c) ground

all circuits. When the potential of the earth is used as reference, we use the earth ground as shown in Fig. 3.1(b) or 3.1(c). In this book, we shall always use the symbol of Fig. 3.1(a).

After selecting a reference node, assign voltage designations to nonreference nodes. For example, consider the circuit of Fig. 3.2(a). Node o is the

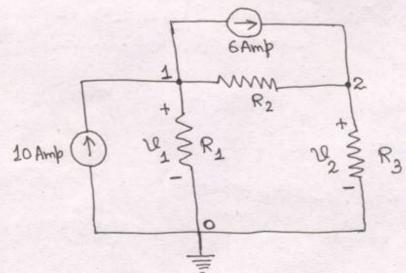


Fig. 3.2(a): Circuit with two independent current sources

reference node (2 = 0). Nodes 1 and 2 are a assigned voltages 21 and 22, respectively. A node voltage is defined as the voltage rise from the reference node to a nonreference node.

The Second step in the nodal analysis is to apply KCL to each nonreference node in the circuit. For further explanation, circuit of Fig. 3.2(a) is redrawn in Fig. 3.2(b) to avoid butting too much information on the same circuit.

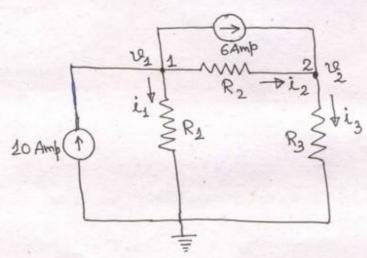


Fig. 3.2(b): Circuit of Fig. 3.2(a) is redrawn with useful information.

Applying KCL at node 1, we have,

$$10 = 6 + i_1 + i_2 - - - (3.1)$$

At node 2

$$6 + i_2 = i_3 - \dots (3.2)$$

To obtain the unknown currents i, iz and is in terms of mode voltages, we apply ohm's law.

From Fig. 3.26), we have,

$$i_1 = \frac{v_1 - o}{R_1} = \frac{v_1}{R_1} - \cdots (3.3)$$

$$i_2 = \frac{v_1 - v_2}{R_2}$$
 ---- (3.4)

$$i_3 = \frac{v_2 - o}{R_3} = \frac{v_2}{R_3}$$
 - - (3.5)

substituting eqns. (3.3) and (3.4) in eqns. (3.1) results, in

$$10 = 6 + \frac{u_1}{R_1} + \frac{u_1 - u_2}{R_2}$$

$$= \left(\frac{1}{R_1} + \frac{1}{R_2}\right) v_1 - \frac{1}{R_2} v_2 = 4 - \dots (3.6)$$

Substituting egms. (3.4) and (3.5) in egms. (3.2), we get,

$$6 + \frac{11-712}{R_2} = \frac{12}{R_3}$$

$$\frac{1}{R_2} u_1 + \left(\frac{1}{R_2} + \frac{1}{R_3}\right) u_2 = 6 - (3.7)$$

The third step in the nodal analysis is to solve for the node voltages. If we apply KCL to n-1 nonreference node, we will obtain n-1 simultaneous equations such as equal (3.6) and (3.7). For the circuit of Fig. 3.2, we can easily obtain the apply of and 12 by solving equal (3.6) and (3.7) using any standard method, such as the elimination method, substitution method, matrix inversion on cramer's rule.

To use the matrix inversion or cramer's rule, 6 we must be put 1 equations in matrix form. For example, equal (3.7) can be written in matrix form. as

$$\begin{bmatrix} \frac{1}{R_{1}} + \frac{1}{R_{2}} & -\frac{1}{R_{2}} \\ -\frac{1}{R_{2}} & \frac{1}{R_{2}} + \frac{1}{R_{3}} \end{bmatrix} \mathcal{N}_{2} = \begin{bmatrix} 4 \\ 6 \end{bmatrix}$$

$$(3.8)$$

Eqm. (3.8) can be solved to get 21, and 212,

3.2: SIMULTANEOUS EQUATIONS AND CRAMER'S RULE

Consider a set of simultaneous equations having the form

$$a_{11}x_{1} + a_{12}x_{2} + \cdots + a_{1n}x_{n} = b_{1}$$

$$a_{21}x_{1} + a_{22}x_{2} + \cdots + a_{2n}x_{n} = b_{2}$$

$$\vdots$$

$$\vdots$$

$$a_{n1}x_{1} + a_{n2}x_{2} + \cdots + a_{nn}x_{n} = b_{n}$$

$$a_{n1}x_{1} + a_{n2}x_{2} + \cdots + a_{nn}x_{n} = b_{n}$$

Eqn. (3.9) can be written in matrix form as

0

$$\begin{bmatrix} a_{11} & a_{12} & \cdots & a_{1n} \\ a_{21} & a_{22} & \cdots & a_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ a_{n1} & a_{n2} & \cdots & a_{nn} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} = \begin{bmatrix} b_1 \\ b_2 \\ \vdots \\ b_n \end{bmatrix}$$

$$(3.10)$$

Eqn. (3.10) can be but in a compact form as

Where

$$A = \begin{bmatrix} a_{11} & a_{12} & \cdots & a_{1n} \\ a_{21} & a_{22} & \cdots & a_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ a_{n1} & a_{n2} & \cdots & a_{nn} \end{bmatrix}$$

$$X = \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix}, \quad B = \begin{bmatrix} b_1 \\ b_2 \\ \vdots \\ b_n \end{bmatrix}$$

A is square (n xn) matrix while X and B B are column (n x1) modrices.

There are several methods for solving eqn. (3.10). These include back substitution, Gaussian elimination, matrix inversion, Cramer's rule and numerical analysis.

Cramer's Rule

Cramer's rule can be used to solve the simultaneous equations. According to Cramer's rule, the solution of eqn. (3.10) is:

$$\chi_{1} = \frac{\Delta_{1}}{\Delta}$$

$$\chi_{2} = \frac{\Delta_{2}}{\Delta}$$

$$\chi_{n} = \frac{\Delta_{n}}{\Delta}$$

$$(3.12)$$

Where the D's are the determinants given by

$$\Delta = \begin{vmatrix} a_{11} & a_{12} & \cdots & a_{1n} \\ a_{21} & a_{22} & \cdots & a_{2n} \\ \vdots & \vdots & \vdots & \vdots \\ a_{n1} & a_{n2} & \cdots & a_{nn} \end{vmatrix}, \quad \Delta_1 = \begin{vmatrix} b_1 & a_{12} & \cdots & a_{1n} \\ b_2 & a_{22} & \cdots & a_{2n} \\ \vdots & \vdots & \vdots & \vdots \\ b_n & a_{n2} & \cdots & a_{nn} \end{vmatrix}$$

$$\Delta_{2} = \begin{vmatrix} a_{11} & b_{1} & \cdots & a_{1n} \\ a_{21} & b_{2} & \cdots & a_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ a_{n1} & b_{n} & \cdots & a_{nn} \end{vmatrix}, \dots \Delta_{n} = \begin{vmatrix} a_{11} & a_{12} & \cdots & b_{1} \\ a_{21} & a_{22} & \cdots & b_{2} \\ \vdots & \vdots & \ddots & \vdots \\ a_{n1} & a_{n2} & \cdots & b_{n} \end{vmatrix}$$

Note that Δ is the determinant of matrix A and Δ_{k} is the determinant of the matrix formed by replacing the k-th column of matrix A by B.

From eqn.(3.12), it is evident that Cramer's rule applies only when $\Delta \neq 0$. When $\Delta = 0$, the set of equations has no unique solution, because the equations are linearly dependent.

The value of the determinant Δ can be obtained by expanding along the first row:

$$\Delta = \begin{bmatrix} a_{11} & a_{12} & a_{13} & - - - a_{1n} \\ a_{21} & a_{22} & a_{23} & - - - a_{2n} \\ a_{31} & a_{32} & a_{33} & - - - a_{3n} \\ \vdots & \vdots & \vdots & \vdots \\ a_{n1} & a_{n2} & a_{n3} & - - - a_{nn} \end{bmatrix}$$

$$= \alpha_{11} M_{11} - \alpha_{12} M_{12} + \alpha_{13} M_{13} + \dots + (-1)^{1+n} \alpha_{1n} M_{1n}$$

$$- \cdot \cdot (3.14)$$

where the minor Mij is an $(n-1) \times (n-1)$ determinant of the mostrix formed by striking out the i-th row and j-th column.

The value of Δ may also be obdained by expanding along the first column:

$$\Delta = \alpha_{11} M_{11} - \alpha_{21} M_{21} + \alpha_{31} M_{31} + \cdots + (-1)^{n+1} \alpha_{n1} M_{n1}$$

$$\Delta = \begin{vmatrix} a_{11} & a_{12} \\ a_{21} & a_{22} \end{vmatrix} = a_{11} a_{22} - a_{12} a_{21} - a_{12} a_{21} - a_{12} a_{21}$$

For a 3 x 3 modrix

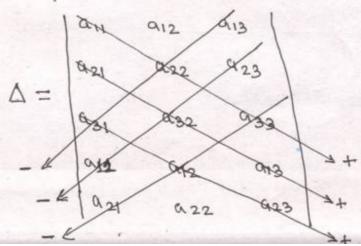
$$\Delta = \begin{vmatrix} a_{11} & a_{12} & a_{13} \\ a_{21} & a_{22} & a_{23} \\ a_{31} & a_{32} & a_{33} \end{vmatrix}$$

$$\triangle = \left[\frac{\alpha_{11} (-1)^{2}}{\alpha_{32}} \right]^{\frac{\alpha_{22}}{\alpha_{33}}} + \left[\frac{\alpha_{21} (-1)^{3}}{\alpha_{32}} \right]^{\frac{\alpha_{12}}{\alpha_{33}}} + \left[\frac{\alpha_{13}}{\alpha_{32}} \right]^{\frac{\alpha_{13}}{\alpha_{33}}} + \left[\frac{\alpha_{13}}{\alpha_{22}} \right]^{\frac{\alpha_{13}}{\alpha_{23}}}$$

$$\triangle = \alpha_{11} q_{22} \alpha_{33} + \alpha_{21} \alpha_{32} \alpha_{13} + \alpha_{31} \alpha_{12} \alpha_{23}$$

$$- \alpha_{13} q_{22} \alpha_{31} - q_{23} \alpha_{32} \alpha_{11} - \alpha_{93} \alpha_{12} \alpha_{21} - \dots (3.17)$$

A simple method of obtaining the determinant of a 3x3 modrix is by repeating the first two rows and multiplying the terms diagonally as follows:



$$\Delta = a_{11} a_{22} a_{33} + a_{21} a_{32} a_{13} + a_{31} a_{12} a_{23}$$

$$- a_{13} a_{22} a_{31} - a_{23} a_{32} a_{11} - a_{33} a_{12} a_{21} - (3.18)$$

Determine the node voltages in the circuit Ex-3.1: shown in Fig. 3.3(a).

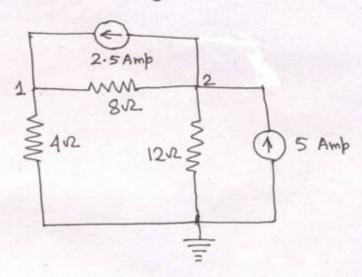


Fig. 3.3(a): Circuit for Ex-3.1

Som.

Fig. 3.3(b) whows the circuit for analysis of Fig. 3.3(a).

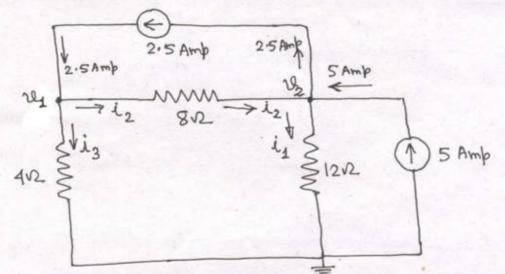


Fig. 3.36): Circuit for analysis of original circuit shown in Fig. 3.3(a).

At node 1, applying KCL and Ohm's law gives.

$$2.5 = i_2 + i_3 - \frac{2i_1 - 2i_2}{8} + \frac{2i_1 - 0}{4}$$

At mode 2,

$$2.5 + \frac{12_1 - 12_2}{8} = \frac{12_2 - 0}{12}$$

By solving eqns.(i) and (ii), we get $V_1 = 13.33$ Volt, $V_2 = 20$ Volt.

EX-3.2: Calculate the node Voltages in the circuit shown in Fig. 3.419).

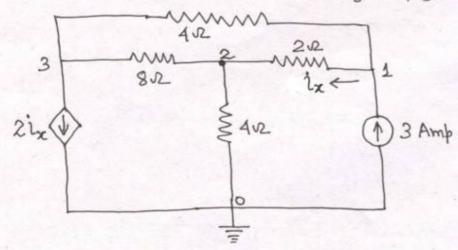


Fig. 3.4(a): Circuit for example - 3.2

Solm.

Fig. 3.4(b) Shows the circuit for analysis of Fig. 3.4(a).

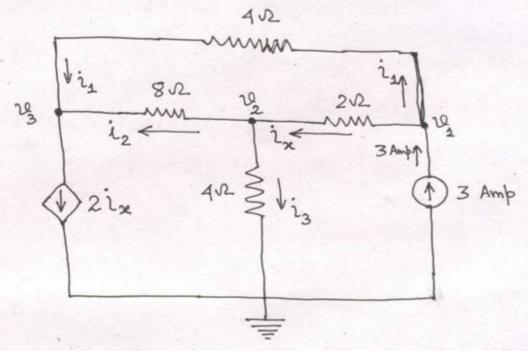


Fig. 3.4(b): Circuit for analysis of Fig. 3.4(a).

At node 1,

$$\frac{2 \cdot 2 \cdot - 2 \cdot 3}{4} + \frac{2 \cdot 1 - 2 \cdot 2}{2}$$

At node 2,

$$\frac{1. \quad u_{1} - u_{2}}{2} = \frac{u_{2} - u_{3}}{8} + \frac{u_{2} - 0}{4}$$

At node 3,

$$l_1 + l_2 = 2l_x$$

$$\frac{u_1-u_3}{4} + \frac{u_2-u_3}{8} = \frac{2(u_1-u_2)}{2}$$

Use Cramer's rule, we put egns. i), (ii) and (iii)

$$\begin{bmatrix} 3 & -2 & -1 \end{bmatrix} \begin{bmatrix} u_4 \\ -4 & 7 & -4 \end{bmatrix} \begin{bmatrix} u_2 \\ 2 & -3 \end{bmatrix} = \begin{bmatrix} 12 \\ 0 \end{bmatrix}$$

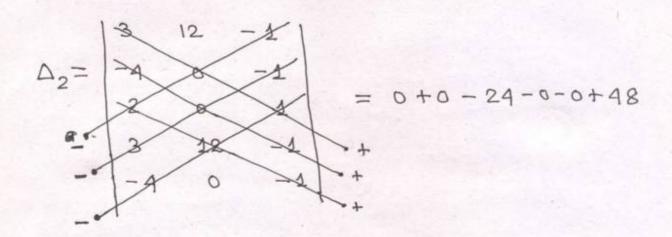
From com. (W), we obsonin

$$u_1 = \frac{\Delta_1}{\Delta}$$
, $u_2 = \frac{\Delta_2}{\Delta}$, $u_3 = \frac{\Delta_3}{\Delta}$

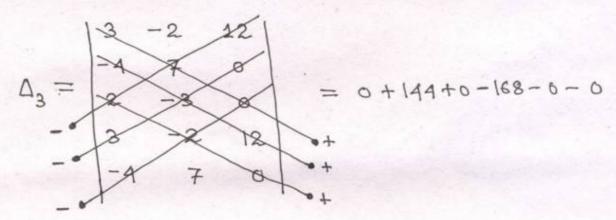
$$\Delta = \begin{vmatrix} 3 & -2 & -1 \\ -4 & 7 & -1 \end{vmatrix} = \begin{vmatrix} 9 & -2 & 7 \\ -4 & 7 & -1 \end{vmatrix} = \begin{vmatrix} -4 & 7 & -1 \\ 2 & -3 & 1 \end{vmatrix} = \begin{vmatrix} -4 & 7 & -1 \\ -4 & 7 & -1 \end{vmatrix}$$

. D= 21-12+4+14-9-8=10

$$\Delta_{1} = \begin{vmatrix} 12 & -2 & -4 \\ 0 & 7 & -1 \\ 0 & -3 & 2 \end{vmatrix}$$



$$\Delta_2 = 24$$



$$\Delta_3 = -24$$

Mus, we obtain

$$\mathcal{Q}_{1} = \frac{\Delta_{1}}{\Delta} = \frac{48}{10} = 4.8 \text{ Volt}$$

$$\mathcal{Q}_{2} = \frac{\Delta_{2}}{\Delta} = \frac{24}{10} = 2.4 \text{ Volt}$$

$$\mathcal{Q}_{3} = \frac{\Delta_{3}}{\Delta} = \frac{-24}{10} = -2.4 \text{ Volt}.$$

3.3: NODAL ANALYSIS WITH VOLTAGE

(16)

SOURCES

We will now consider how voltage sources affect modal analysis. For the purpose of explanation consider Fig. 3.5.

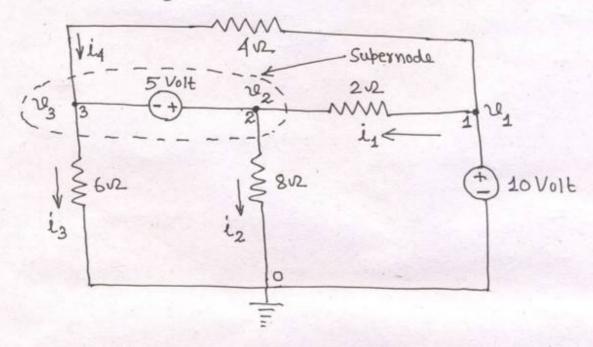


Fig. 3.5: A simple circuit with a supernode

In Fig.3.5, a voltage source is connected between the reference node (node o) and a nonreference mode (node 1). Therefore, voltage at the nonreference node equal to the voltage of the voltage source. Hence,

If the voltage source (dependent or independent) is connected between two nonreference nodes, the two nonreference nodes form a supernode or generalized node. A supernode requires the

application of both KCL and KVL to determine (17) the node voltages.

In Fig. 3.5 nodes 2 and 3 form a supernode. An independent voltage source is connected between nonreference nodes 2 and 3.

now we apply KCL at supernode, $i_1 + i_4 = i_2 + i_3 - \cdots (3.20)$

$$\frac{\mathcal{U}_1 - \mathcal{U}_2}{2} + \frac{\mathcal{U}_1 - \mathcal{U}_3}{4} = \frac{\mathcal{U}_2 - 0}{8} + \frac{\mathcal{U}_3 - 0}{6} - \cdots (3.21)$$

To apply Kirchhoff's voltage law to the supermode in Fig. 3.5, the circuit is redrawn in Fig. 3.6.

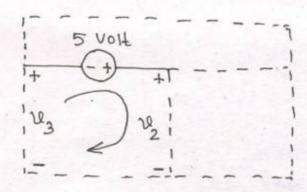


Fig. 3.6: Applying KVL to a supernode in the clockwise direction.

Going around the loops in clockwise direction, we have,

$$-5 + u_2 - u_3 = 0$$

$$\therefore u_2 - u_3 = 5 - \dots (3.22)$$

From egms. (3.19), (3.21) and (3.22), we obtain the mode voltages. Therefore, a supernode has the following properties;

- 1. A supernode requires the application of both KCL and KVL.
 - 2. A supermode has no voltage of its own.
 - 3. The voltage source inside the supernode provides a constraint equation needed to solve for the node voltages.

EX-3.3: Determine the node voltages of the circuit whown in Fig. 3.6.

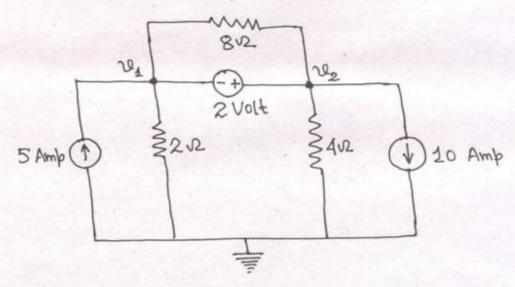


Fig. 3.6: Circuit for Ex-3.3

Soln.

For the purpose of understanding the problem, two circuits are whom in Figs 3.7(a) and (b) for applying KCL to the supernode and KVL to the loop.

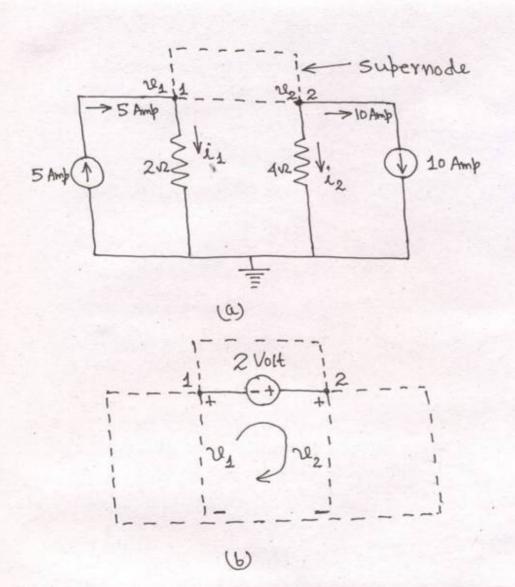


Fig. 3.7: Applying (a) KCL to the supernode

(b) KVL to the loop in the clockwise direction

Applying KCL to the supernode supernode as shown in Fig. 3.7(a),

$$5 = \frac{1}{1} + \frac{1}{2} + 10$$

$$5 = \frac{1}{10} + \frac{1}{2} + \frac{1}{2} + 10$$

$$2 \cdot 1 + 1 \cdot 2 = -20 - \dots$$

$$2 \cdot 1 + 1 \cdot 2 = -20 - \dots$$

Applying KVL to the loop in the clockwise direction on whown in Fig. 3.76.

20

Solving egms. (i) and (ii), we obtain,

Note that 8 12 resistor does not make any difference because it is connected across the supernode.

EX-3.4: Determine is, is and is of the circuit on whown in Fig. 3.8 using node-voltage method.

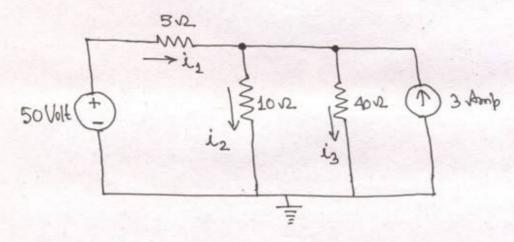


Fig. 3.8; Circuit of Ex-3.4

Som.

Circuid of Fig. 3.8 has two essential nodes: one nonreference node and one reference node. Fig. 3.9 shows these decisions.

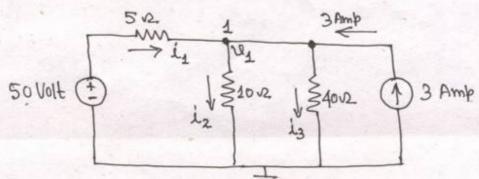


Fig. 3.9: Circuit of Fig. 3.8 is redrawn for analysis

At mode 1,

$$\frac{\mathcal{U}_{1}-50}{5}+\frac{\mathcal{U}_{1}}{10}+\frac{\mathcal{U}_{1}}{40}-3=0$$

Hence,

$$i_1 = \frac{50 - 40}{5} = 2 \text{ Amp}.$$

$$i_2 = \frac{40}{10} = 4 \text{ Amp}.$$

$$i_3 = \frac{40}{40} = 1 \text{ Amp}.$$

EX-3.5: Determine the node voltages and bowers dissipoded in 5 12 resistor in the circuit whown in Fig. 3.10.

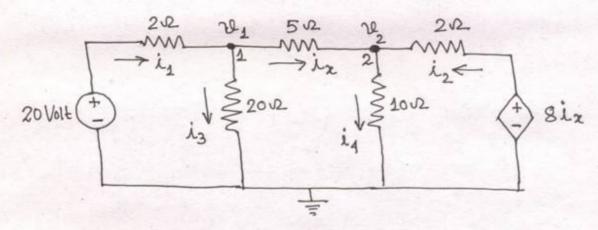


Fig. 3.10: Circuit for EX-3.5

$$i_1 = i_3 + i_2$$

$$\frac{20-v_1}{2} = \frac{v_1-0}{20} + \frac{v_1-v_2}{5} - \dots (i)$$

At node 2,

$$\frac{v_2 - 0}{10} = \frac{v_1 - v_2}{5} + \frac{8i_2 - v_2}{2} - \cdots (ii)$$

Also

$$i_x = \frac{2l_1 - 2l_2}{5}$$
 --- (iii)

From egns. (ii) and (iii), we get

$$\frac{v_2}{10} = \frac{v_1 - v_2}{5} + 4\left(\frac{v_1 - v_2}{5}\right) - \frac{v_2}{2}$$

Solving egns. (i) and (iv), we obtain, $2l_1 = 16$ Volt, and $2l_2 = 10$ Volt.

$$i_{x} = \frac{v_{1} - v_{2}}{5} = \frac{16 - 10}{5} = 1.2 \text{ Amp}$$

Ex-3.6: Determine the node voltages of the circuit as whomin in Fig. 3.11.

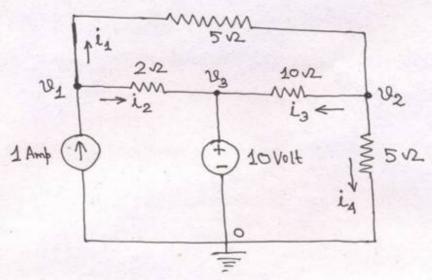


Fig. 3.11: Circuit for Ex-3.6

At node 1,

$$i_1 + i_2 = 1$$

$$\frac{1}{5} + \frac{2}{2} = 1$$

At node 2,

$$\frac{2}{5} = \frac{2}{10} + \frac{2}{5} = \frac{2}{5}$$

$$= -0.2 \cdot 2 \cdot 1 + 0.5 \cdot 2 = 1 - -(ii)$$

Solving egms. (i) and (ii), we obtain 2 = 10.32 Volt; 2 = 6.13 Volt. Ex-3.7: Determine the node voltages of the 24 circuit as whown in Fig. 3.12.



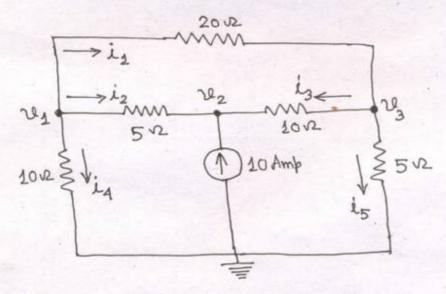


Fig.3.12: Circuit for Ex-3.7

Salm.
At mode 1,

$$\frac{2}{20} + \frac{2}{5} + \frac{2}{10} = 0$$

At mode 2,

$$\frac{1}{5} + \frac{v_3 - v_2}{10} + 10 = 0$$

At node 3,

$$i_1 = i_3 + i_5$$

$$\frac{v_1 - v_3}{20} = \frac{v_3 - v_2}{10} + \frac{v_3}{5}$$

21 = 45. 45 Volt; 2=72.73 Volt; 2=27.27 Volt.

Ex-3.8: Using node voltage method, determine the currents as of the circuit or whown in Fig. 3.13.

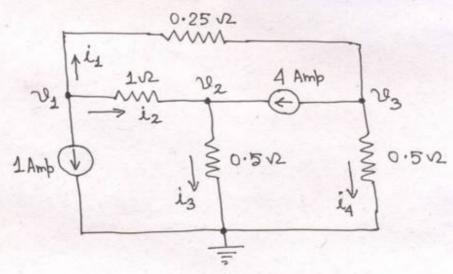


Fig. 3.13: Circuit for EX-3.8

Solm.

At node 1,

$$i_1 + i_2 + 1 = 0$$
 $\frac{v_1 - v_3}{0.25} + \frac{v_1 - v_2}{1} + 1 = 0$

At node 2,

$$i_2 + 4 = i_3$$

$$= \frac{10_1 - 10_2}{1} + 4 = \frac{10_2}{0.5}$$

At mode 3,

$$\frac{1}{0.25} = 4 + \frac{1}{0.5}$$

Solving eqns.(i), (ii) and (iii), we get $\mathcal{V}_1 = -\frac{7}{6} \text{ Voit}; \ \mathcal{V}_2 = \frac{17}{18} \text{ Voit}; \ \mathcal{V}_3 = -\frac{13}{9} \text{ Voit}$

$$i_{1} = \frac{2l_{1} - 2l_{3}}{0.25} = 4\left(-\frac{7}{6} + \frac{13}{9}\right) = \frac{10}{9} \text{ Amp}$$

$$i_{2} = \frac{2l_{1} - 2l_{2}}{1} = \left(-\frac{7}{6} - \frac{17}{18}\right) = -\frac{19}{9} \text{ Amp}$$

$$i_{3} = 22l_{2} = 2 \times \frac{17}{18} = \frac{17}{9} \text{ Amp}$$

$$i_{4} = 22l_{3} = 2 \times \left(-\frac{13}{9}\right) = -\frac{26}{9} \text{ Amp}.$$

of the circuit as whom in Fig. 3.14.

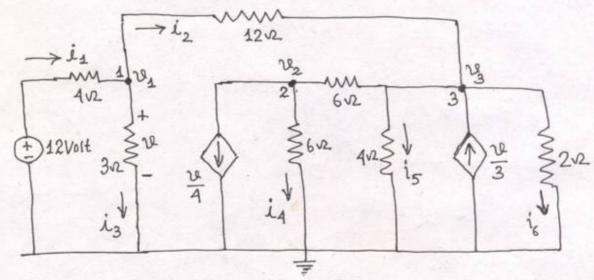


Fig. 3.14: Circuit for EX-3.9

At mode 1,

$$i_1 = i_2 + i_3$$

$$\frac{12-v_1}{4} = \frac{v_1-v_3}{12} + \frac{v_1-v_3}{3}$$

Note that 21 = 2

$$\frac{12-12}{4} = \frac{12-12}{12} + \frac{12}{3}$$

At node 2,

22, - 22 20 22-

$$\frac{2^{2} - 2^{2}}{6} = \frac{12}{4} + \frac{2^{2}}{6}$$

At node 3,

$$= \frac{12_3 - 12_2}{6} + \frac{12_3}{4} + \frac{12_3}{2} = \frac{12}{3} + \frac{12 - 12_3}{12}$$

Solving equs. (i), (ii) and (iii), we obtain $2 = 2 \cdot 4 \cdot 686$ Volt; $2 = -2 \cdot 769$ Volt; $3 = 1 \cdot 491$ Volt.

Ex-3.10: Determine U1, 12 and 29 of the circuit (28) shown in Fig. 3.15 Using nodal analysis.

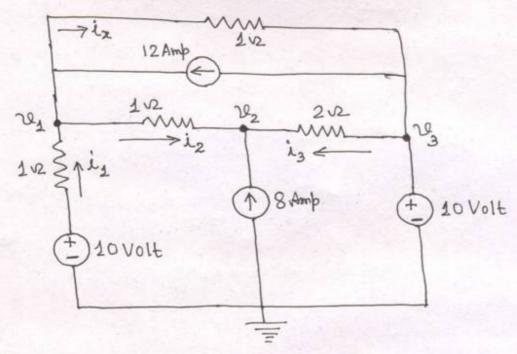


Fig. 3.15: Circuit for Ex-3.10

At node 1,

$$\frac{20-2l_1}{1}+12=\frac{2l_1-2l_2}{1}+\frac{2l_1-2l_3}{1}.$$

$$22-29_1=229_1-19_2-10$$

At node 2,

$$\frac{2l_1 - 2l_2}{1} + 8 + \frac{2l_3 - 2l_2}{2} = 0$$

$$2 \cdot 2 \cdot 1 - 2 \cdot 2 + 8 + \frac{10 - 10^2}{2} = 0$$

Solving eqn. (i) and (ii), we obtain, $v_1 = 17.428 \text{ Volt}$; $v_2 = 20.285 \text{ Volt}$.

current in
$$=\frac{2l_1-2l_3}{1}=(17.428-10)=7.428 \text{ Amp.}$$

$$i_2=\frac{2l_1-2l_2}{1}=(17.428-20.285)=-2.857 \text{ Amp.}$$

$$i_3=\frac{2l_3-2l_2}{2}=\frac{10-20.285}{2}=-5.142 \text{ Amp.}$$

EX-3.11: Determine 202 using nodal analysis of the circuit whown in Fig. 3.16.

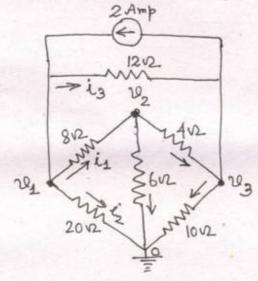


Fig. 3.16: Circuit for Ex-3.11

Solm. At node 1,

$$\frac{2i_1-2i_2}{8} + \frac{2i_1-0}{20} + \frac{2i_1-2i_3}{12} = 2$$
 $312i_1 - 152i_2 - 102i_3 = 240 - - \cdot (i)$

Similarly of node 2,

 $-32i_1 + 132i_2 - 62i_3 = 0 - - (ii)$

Similarly of node 3,

 $52i_1 + 152i_2 - 262i_3 = 120 - (iii)$

Solving Eqns.(i) (ii) and (iii), we get,

 $2 = 0.0$; This means bridge circuit is balanced.