

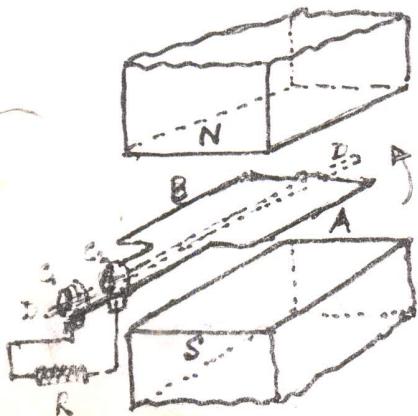
GENERATION OF ALTERNATING EMF

Fig. 1.

Magnetic Field

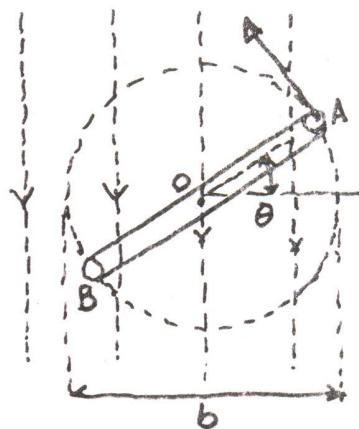


Fig. 2(a)

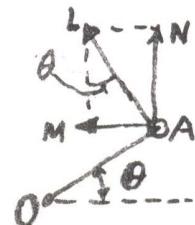


Fig. 2(b)

EMF GENERATED IN ONE SIDE OF LOOP = $B l v \sin \theta$ volts.TOTAL EMF GENERATED IN LOOP = $2 B l v \sin \theta$ volts.where v = tangential velocity of the conductor

$$= \frac{b}{2} \cdot n \cdot 2\pi = \pi b n \text{ metres per second.}$$

 n = speed in rps (revolutions per second) b = Breadth of the loop in metre l = Length of one side of the loop in metre B = Flux density in TeslaTaking $A = lb$ = area of the loop in square metresEMF GENERATED IN THE LOOP = $e = 2\pi B A n \sin \theta$ volts.IF THE LOOP IS REPLACED BY A COIL OF N TURNS IN SERIES,THEN EMF GENERATED IN THE COIL = $2\pi B A N n \sin \theta$ volts.MAXIMUM VALUE OF THE EMF GENERATED = $E_m = 2\pi B A n N$

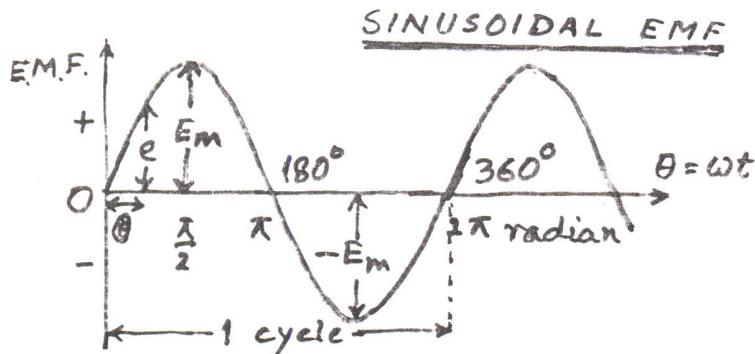
AND INSTANTANEOUS VALUE OF EMF GENERATED IN THE COIL

$$e = E_m \sin \theta = 2\pi B A N n \sin \theta \text{ volts.}$$

Where θ = angular displacement in time t sec.

$$= \omega t = 2\pi n t$$

 ω is angular velocity in rad./sec.

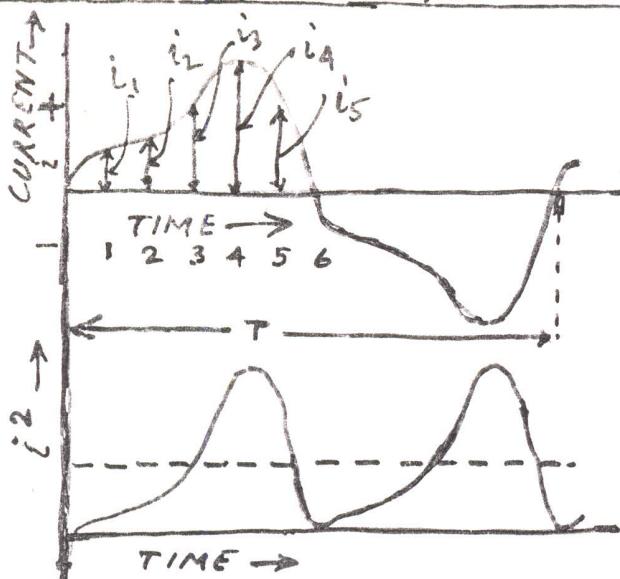


If an a.c. generator has 'p' pairs of poles & its speed is n rps, then
FREQUENCY

$$\begin{aligned} f &= \text{No. of cycles per second} \\ &= \text{No. of cycles per rev.} \\ &\quad \times \text{No. of rev. per second} \\ &= p n \text{ Hz.} \end{aligned}$$

Ex. For a 4-pole Generator to generate a voltage of 50 Hz frequency the speed $n = \frac{f}{p} = \frac{50}{2} = 25$ rps $= 25 \times 60 \text{ rpm} = 1500 \text{ rpm}$ where rpm \Rightarrow revolutions per minute.
For 2-pole machine the speed should be 3000 rpm.

(MEAN, EFFECTIVE) AVERAGE & RMS VALUES OF AN ALTERNATING CURRENT.



* AVERAGE VALUE OVER A HALF CYCLE

$$I_{av} = \frac{i_1 + i_2 + i_3 + \dots + i_n}{n}$$

$$= \frac{\text{AREA OVER HALF CYCLE}}{\text{LENGTH OF BASE OF HALF CYCLE}}$$

$$= \frac{1}{T/2} \int_0^{T/2} i dt$$

* ROOT MEAN SQUARE (R.M.S.) VALUE DEPENDS ON HEATING EFFECT i.e. SQUARE OF THE CURRENT.

$$\text{AVERAGE HEATING EFFECT} \propto \frac{i_1^2 + i_2^2 + \dots + i_n^2}{n}$$

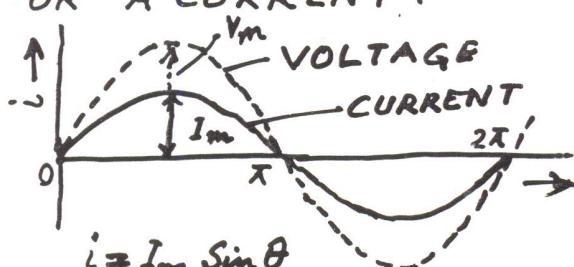
SUPPOSE I BE THE VALUE OF THE DIRECT CURRENT TO PRODUCE SAME HEATING IN THE SAME RESISTOR AS PRODUCED BY ALTERNATING CURRENT; THEN

$$I^2 = \frac{i_1^2 + i_2^2 + \dots + i_n^2}{n}$$

$$\begin{aligned} I &= \sqrt{\frac{i_1^2 + i_2^2 + \dots + i_n^2}{n}} = \sqrt{\frac{\text{AREA OF } i^2 \text{ CURVE}}{\text{LENGTH OF BASE}}} \\ &= \sqrt{\frac{1}{T/2} \int_0^{T/2} i^2 dt} \end{aligned}$$

OVER A HALF CYCLE.

AVERAGE AND RMS VALUES OF A SINUSOIDAL VOLTAGE OR A CURRENT:



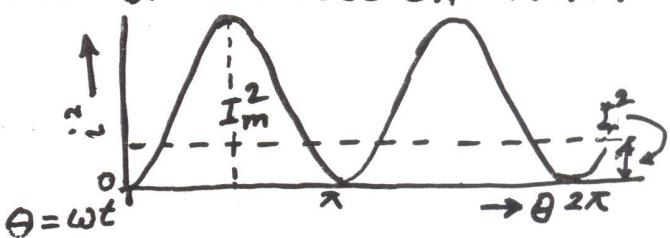
$$i = I_m \sin \theta$$

$$I_{av} = \frac{1}{\pi} \int_0^{\pi} i d\theta$$

$$= \frac{1}{\pi} \int_0^{\pi} I_m \sin \theta d\theta$$

$$= \frac{1}{\pi} \cdot I_m [-\cos \theta]_0^{\pi}$$

$$= \frac{2}{\pi} I_m = 0.637 I_m.$$



$$i^2 = I_m^2 \sin^2 \theta$$

$$I_{rms} = \left[\frac{1}{\pi} \int_0^{\pi} i^2 d\theta \right]^{\frac{1}{2}}$$

$$= \left[\frac{1}{\pi} \int_0^{\pi} I_m^2 \sin^2 \theta d\theta \right]^{\frac{1}{2}}$$

$$= \left[\frac{1}{\pi} \int_0^{\pi} I_m^2 \frac{1}{2}(1 - \cos 2\theta) d\theta \right]^{\frac{1}{2}}$$

$$= \frac{I_m}{\sqrt{2}} = 0.707 I_m.$$

Similarly, $V_{av} = 0.637 V_m$ $V_{rms} = 0.707 V_m$.

Average Value of a sinusoidal current or voltage
= $0.637 \times$ maximum value.

R.M.S. value of a sinusoidal current or voltage
= $0.707 \times$ maximum value.

Form Factor of a sine wave = $\frac{\text{R.M.S. Value}}{\text{Average Value}}$

$$= \frac{0.707 \times \text{maximum value}}{0.637 \times \text{maximum value}} = \underline{\underline{1.11}}$$

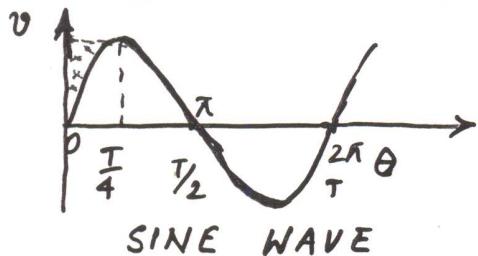
Peak or Crest Factor of a sine wave = $\frac{\text{Maximum Value}}{\text{RMS Value}}$

$$= \frac{\text{maximum value}}{0.707 \times \text{maximum value}} = \underline{\underline{1.414}}$$

NOTE: RMS VALUE IS ALWAYS GREATER THAN AVERAGE EXCEPT FOR A RECTANGULAR WAVE, IN WHICH CASE THE HEATING EFFECT REMAINS CONSTANT SO THAT THE AVERAGE & THE RMS VALUES ARE SAME.

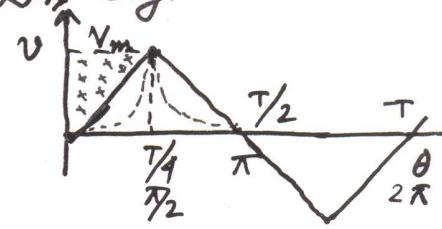
SYMMETRICAL AND UNSYMMETRICAL WAVEFORMS:

THE WAVEFORMS IN WHICH +VE AND -VE HALF CYCLES ARE IDENTICAL ARE REFERRED TO AS SYMMETRICAL WAVE FORM, e.g.



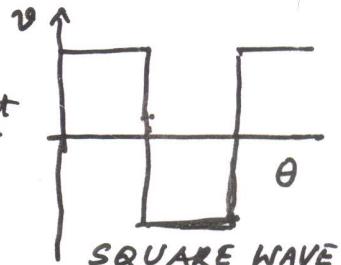
SINE WAVE

(a)



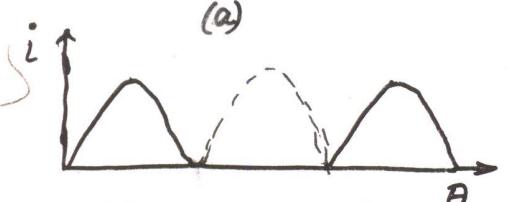
TRIANGULAR WAVE

(b)



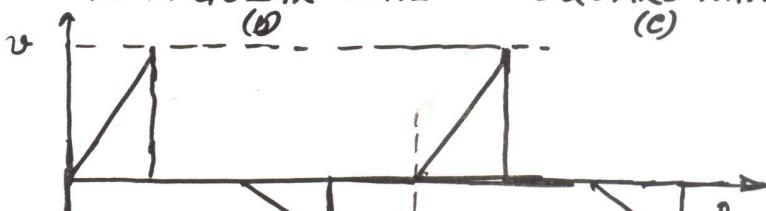
SQUARE WAVE

(c)



HALF WAVE RECTIFIED

(d)



(e)

THE WAVE FORMS (a), (b) & (c) ARE SYMMETRICAL WHEREAS THOSE OF (d) & (e) ARE NON-SYMMETRICAL.

NOTE: i) IN CASE OF UNSYMMETRICAL WAVEFORMS, THE AVERAGE VALUE MUST ALWAYS BE TAKEN OVER THE WHOLE CYCLE.

$$I_{av} = \frac{1}{T} \int_0^T i dt \quad \text{or} \quad V_{av} = \frac{1}{T} \int_0^T v dt.$$

IN CASE OF SYMMETRIC ALTERNATING CURRENT OR VOLTAGE THE AVERAGE VALUE OVER A COMPLETE CYCLE IS ZERO. SO THE AVERAGE VALUE IS TAKEN OVER HALF CYCLE.

ii) TO DETERMINE THE RMS VALUE OF AN ALTERNATING QUANTITY, IT IS IMMATERIAL TO TAKE THE AVERAGE OVER HALF CYCLE OR COMPLETE CYCLE.

&
FOR CASES (a), (b), (c) WE MAY EVEN CONSIDER A SHORTER INTERVAL $T/4$ AS IT IS SYMMETRIC ABOUT THE POINT $T/4$. FOR THE TRIANGULAR WAVE FORM Fig. (b)

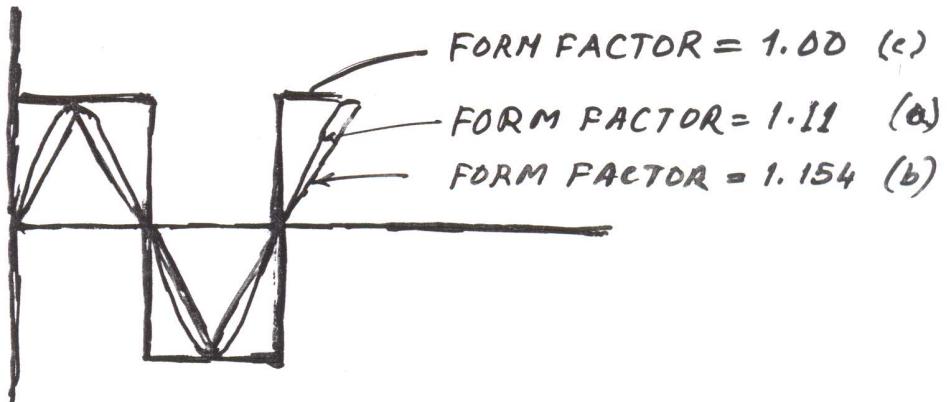
$$V_{av} = \frac{1}{T/4} \int_0^{T/4} \frac{V_m}{T/4} \cdot t dt = \left(\frac{4}{T} \right)^2 V_m \cdot \frac{t^2}{2} \Big|_0^{T/4} = \frac{V_m}{2}.$$

$$V_{rms} = \left[\frac{1}{T/4} \int_0^{T/4} \frac{V_m^2}{(T/4)^2} t^2 dt \right]^{\frac{1}{2}} = \frac{V_m}{\sqrt{3}}.$$

$$\text{FORM FACTOR} = 2/\sqrt{3} = 1.155$$

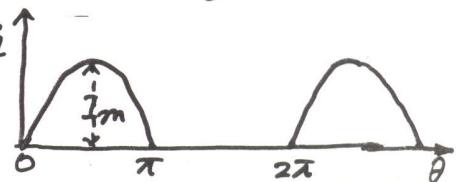
$$\text{PEAK FACTOR} = \sqrt{3} = 1.732.$$

THUS THE FORM FACTORS OF WAVEFORMS SHOWN IN FIGS. (a), (b) & (c) ARE AS FOLLOWS:

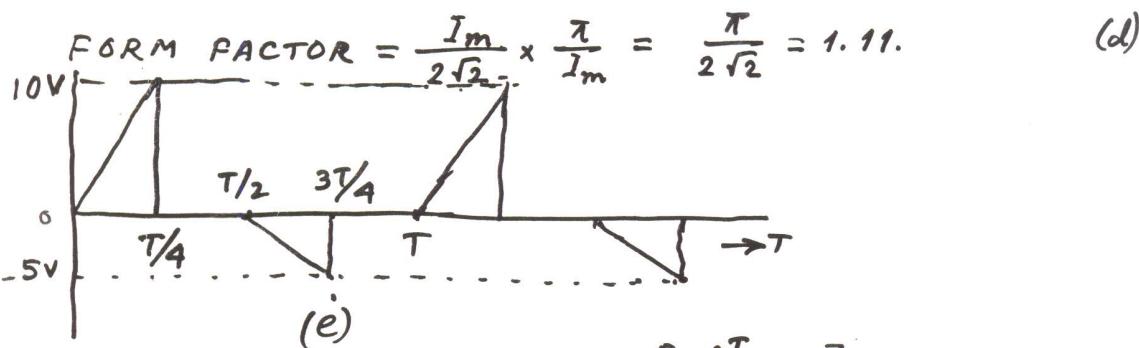


FOR UNSYMMETRICAL WAVES SUCH AS IN FIGS. (d) & (e)

$$\text{AVERAGE VALUE} = \frac{1}{2\pi} \int_0^\pi I_m \sin \theta d\theta = \frac{1}{2\pi} I_m [-\cos \theta]_0^\pi \\ = \frac{1}{2\pi} \cdot 2 I_m = \frac{1}{\pi} \cdot I_m.$$

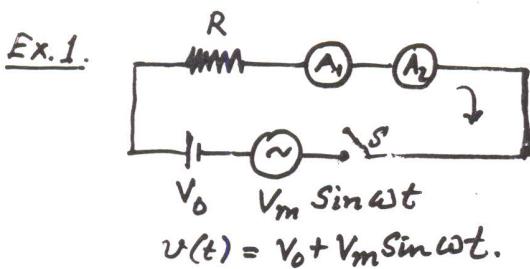


$$\text{RMS VALUE} = \frac{1}{2\pi} \int_0^\pi I_m^2 \sin^2 \theta d\theta = \frac{I_m}{2\sqrt{2}}.$$



$$\text{AVERAGE VALUE} = V_{av} = \frac{1}{T} \left[\int_0^T v dt \right] \\ = \frac{1}{T} \left[\int_0^{T/4} \frac{10}{T/4} \cdot t dt + \int_0^{T/4} \left(\frac{-5}{T/4} \right) t dt \right] \\ = 0.625 \text{ V.}$$

$$\text{RMS VALUE} = V_{RMS} \\ = \sqrt{\frac{1}{T} \int_0^T v^2 dt} = \sqrt{\frac{1}{T} \left[\int_0^{T/4} \left(\frac{10}{T/4} \right)^2 t^2 dt + \int_0^{T/4} \left(\frac{-5}{T/4} \right)^2 t^2 dt \right]} \\ = 3.23 \text{ V.}$$



GIVEN: $V_m = 100 \text{ V}$, $V_0 = 120 \text{ V}$

$$R = 20 \Omega.$$

A_1 = Moving coil ammeter

A_2 = Moving iron ammeter.

WHAT WILL BE READINGS OF A_1 & A_2 ?

THE NET VOLTAGE ACTING IN THE CIRCUIT = $120 + 100 \sin \omega t$ volts.
 SO CURRENT $i(t) = 6 + 5 \sin \omega t$ amps.

$$I_{av} = \frac{1}{T} \int_0^T (6 + 5 \sin \omega t) dt; \quad \omega = \frac{\pi}{2R}$$

$$= 6 \text{ A} = I_{dc}.$$

= READING OF AMMETER A_1

$$I_{rms} = I = \sqrt{\frac{1}{T} \int_0^T (6 + 5 \sin \omega t)^2 dt}$$

$$= \sqrt{\frac{1}{T} \left[\int_0^T (36 + 60 \sin \omega t + 25 \sin^2 \omega t) dt \right]} = \sqrt{36 + \frac{25}{2}}$$

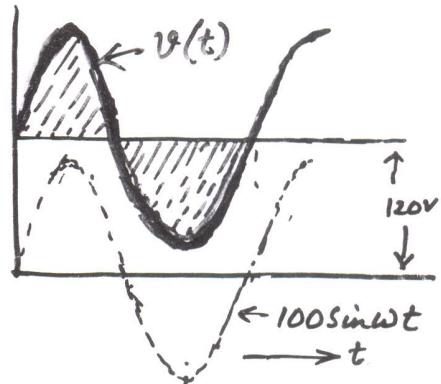
$$\approx 7 \text{ A} = \text{READING OF AMMETER } A_2$$

THUS IF DC IS SUPERIMPOSED ON AC SIGNAL, THE
 RESULTANT RMS VALUE = $\sqrt{I_{dc}^2 + I_{rms(ac)}^2} = \sqrt{6^2 + \left(\frac{5}{\sqrt{2}}\right)^2} \approx 7 \text{ A.}$

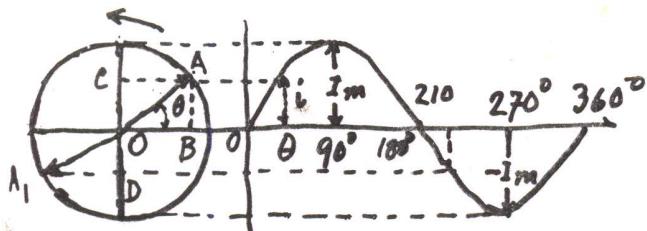
NOTE : i) MOVING COIL INSTRUMENT READS AVERAGE VALUE OVER A PERIOD. THUS FOR AC ALONE IT WILL READ ZERO.
 ii) MOVING IRON, HOT WIRE, ELECTRODYNAMIC & INDUCTION TYPE INSTRUMENTS READ RMS VALUE OF THE SIGNAL.

IMPORTANT TO REMEMBER :

WHEN AN A.C. SUPPLY OF 220 V IS REFERRED, IT IS MEANT ITS RMS VALUE. THE MAXIMUM VALUE IS $220\sqrt{2} \approx 310 \text{ V.}$
 IN CASE OF DC 220V IT IS 220 V ONLY. THUS SHOCK LEVEL OF AC FOR SAME SPECIFIED VOLTAGE IS MORE THAN THAT OF DC VOLTAGE.



REPRESENTATION OF AN SINUSOIDAL SIGNAL BY A PHASOR



LET OA REPRESENT THE MAXIMUM VALUE OF A SINUSOIDAL QUANTITY, SAY, CURRENT, i.e. $OA = I_m$. IT IS BEING ROTATED IN THE ANTI-CLOCKWISE DIRECTION. THE VERTICAL PROJECTION OF OA IS PLOTTED WITH RESPECT TO θ .

IT WILL GENERATE A SINE WAVE. THE VERTICAL PROJECTION OF OA IS $AB = I_m \sin \theta$ (instantaneous value). THE ANGLE θ IS GIVEN BY $\theta = \omega t$ WHERE ω IS THE ANGULAR VELOCITY OF THE PHASOR. AGAIN ω IS GIVEN BY $\omega = 2\pi f$ RADIAN PER SECOND. THUS

$$i = I_m \sin \theta = I_m \sin \omega t = I_m \sin 2\pi ft.$$

LET US NOW CONSIDER TWO QUANTITIES SUCH AS VOLTAGE & CURRENT CAN BE REPRESENTED BY OB & OA.

HERE IT IS SHOWN THAT THE VOLTAGE PHASOR OB IS LEADING THE CURRENT PHASOR OA. BOTH ARE ROTATING AT THE SAME ANGULAR SPEED $\omega = 2\pi f$ RADIANS PER SECOND. THUS WE CAN WRITE TAKING $OB = V_m$.

$$v = V_m \sin(\theta + \phi) = V_m \sin(\omega t + \phi) = V_m \sin(2\pi ft + \phi).$$

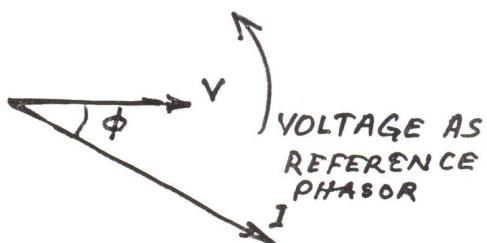
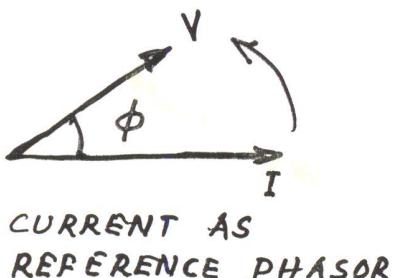
ϕ IS KNOWN AS THE PHASE DIFFERENCE BETWEEN VOLTAGE AND CURRENT. TAKING VOLTAGE AS THE REFERENCE PHASOR ONE CAN WRITE ALSO AS

$$v = V_m \sin \theta = V_m \sin \omega t = V_m \sin 2\pi ft$$

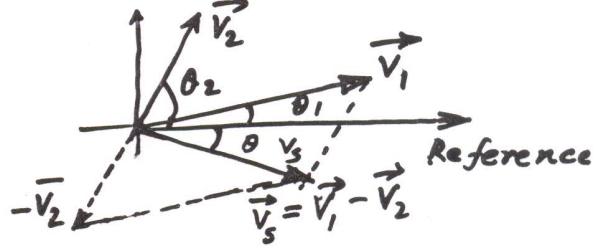
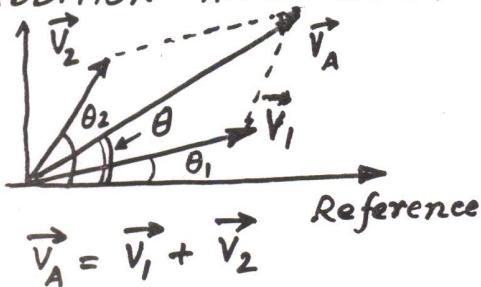
$$i = I_m \sin(\theta - \phi) = I_m \sin(\omega t - \phi) = I_m \sin(2\pi ft - \phi)$$

AND CAN STATE THAT THE CURRENT IS LAGGING BEHIND THE VOLTAGE BY AN ANGLE ϕ .

THE PHASOR DIAGRAMS ARE DRAWN WITH THE RMS VALUES OF SIGNALS. THE PHASOR DIAGRAM CAN BE DRAWN AS EITHER IN(a) OR IN(b).



ADDITION AND SUBTRACTION OF PHASOR QUANTITIES:



PHASORS AS COMPLEX NUMBERS.

THERE ARE THREE EQUIVALENT NOTATIONS OF A PHASOR

POLAR FORM: $\vec{V} = V \angle \theta$

RECTANGULAR FORM: $\vec{V} = V \cos \theta + j V \sin \theta$

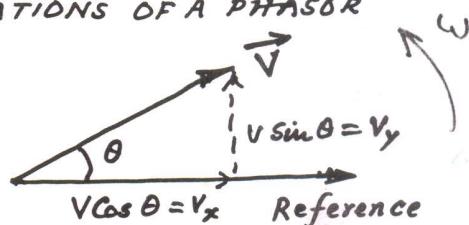
EXPONENTIAL FORM: $\vec{V} = V e^{j\theta}$

(NOTE: $e^{j\theta} = \cos \theta + j \sin \theta = 1 \angle \theta$)

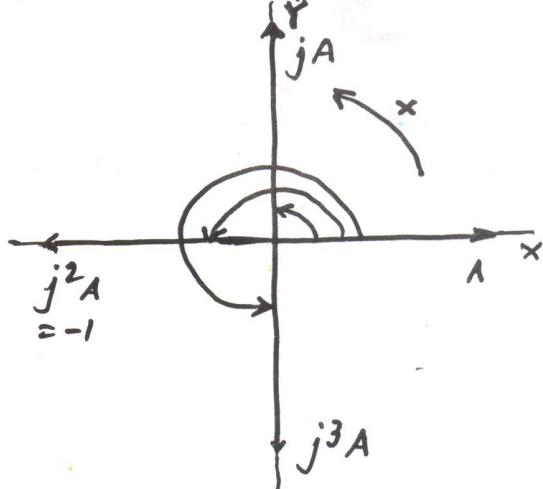
THE OPERATOR j PRODUCES 90° COUNTER CLOCKWISE ROTATION OF ANY PHASOR TO WHICH IT IS APPLIED AS A MULTIPLYING FACTOR.

HENCE $j^2 = -1$ OR $j = \sqrt{-1}$.

SO $\vec{V} = V_x + j V_y$ ALSO.



THE RECTANGULAR FORM OF EXPRESSING A PHASOR IS CONVENIENT FOR ADDITION OR SUBTRACTION OF PHASORS. THE POLAR FORM IS A CONVENIENT WAY TO DISPLAY THE TWO PARTS OF A PHASOR, NAMELY ITS MAGNITUDE AND PHASE ANGLE.



FOR EXPRESSING THE PRODUCT OR QUOTIENT OF TWO PHASORS THE EXPONENTIAL FORM IS CONVENIENT. IF

$$\vec{V}_1 = V_1 e^{j\theta_1} \quad \text{THEN} \quad \vec{V}_1 \vec{V}_2 = V_1 V_2 e^{j(\theta_1 + \theta_2)} = V_1 V_2 \angle \theta_1 + \theta_2.$$

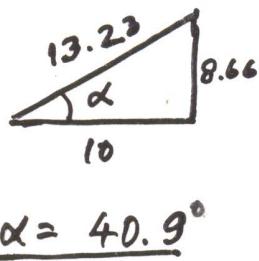
$$\vec{V}_2 = V_2 e^{j\theta_2} \quad \& \quad \frac{\vec{V}_1}{\vec{V}_2} = \frac{V_1}{V_2} e^{j(\theta_1 - \theta_2)} = \frac{V_1}{V_2} \angle \theta_1 - \theta_2.$$

CAUTION: IN AC WORK ALL OPERATIONS OF VOLTAGE AND CURRENT SHOULD BE DONE KEEPING IN MIND THAT THOSE ARE PHASOR QUANTITIES. ALGEBRAIC MANIPULATIONS ARE POSSIBLE ONLY AFTER EXPRESSING THEM AS COMPLEX QUANTITIES.

Ex-2 Add the following currents as waves and as phasors:

$$i_1 = 5 \sin \omega t; \quad i_2 = 10 \sin(\omega t + 60^\circ)$$

As Waves: $i = i_1 + i_2 = 5 \sin \omega t + 10 \sin(\omega t + 60^\circ)$



$$= 5 \sin \omega t + 10 \sin \omega t \cos 60^\circ + 10 \cos \omega t \sin 60^\circ$$

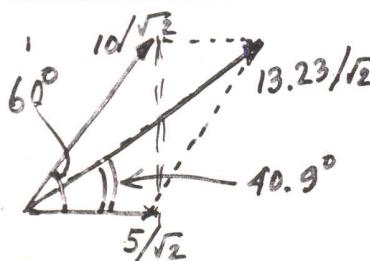
$$= 10 \sin \omega t + 8.66 \cos \omega t.$$

$$= \left[\frac{10}{13.23} \sin \omega t + \frac{8.66}{13.23} \cos \omega t \right] \times 13.23$$

$$= 13.23 [\sin \omega t \cos \alpha + \cos \omega t \sin \alpha]$$

$$= 13.23 \sin(\omega t + \alpha) = 13.23 \sin(\omega t + 40.9^\circ).$$

AS PHASORS:



$$\sum x = \frac{1}{\sqrt{2}}(5 + 10 \cos 60^\circ) = \frac{1}{\sqrt{2}} 10$$

$$\sum y = \frac{1}{\sqrt{2}}(0 + 10 \sin 60^\circ) = \frac{1}{\sqrt{2}} 8.66$$

$$\text{SUM} = \sqrt{(\sum x)^2 + (\sum y)^2} = \sqrt{\frac{1}{2}(10^2 + (8.66)^2)} = 13.23 \frac{1}{\sqrt{2}}$$

$$\alpha = \tan^{-1} \frac{\sum y}{\sum x} = \tan^{-1} \frac{\frac{1}{\sqrt{2}} 8.66}{\frac{1}{\sqrt{2}} 10} = 40.9^\circ.$$

so $i = 13.23 \sin(\omega t + 40.9^\circ)$

WORK OUT THE SUBTRACTION OF $(i_1 - i_2)$ BY BOTH METHODS

Ans: $8.66 \sin(\omega t - 90^\circ)$.

EXPRESSING THE CURRENTS IN RECTANGULAR & POLAR FORMS:

$$\vec{i}_1 = \frac{5}{\sqrt{2}}(1+j0) = \frac{5}{\sqrt{2}} \angle 0^\circ$$

$$\vec{i}_2 = \frac{10}{\sqrt{2}}(\cos 60^\circ + j \sin 60^\circ) = \frac{10}{\sqrt{2}}\left(\frac{1}{2} + j \frac{\sqrt{3}}{2}\right) = \frac{10}{\sqrt{2}} \angle 60^\circ.$$

MULTIPLICATION OF $\vec{i}_1, \vec{i}_2 = \frac{5}{\sqrt{2}} \angle 0^\circ \cdot \frac{10}{\sqrt{2}} \angle 60^\circ = 25 \angle 60^\circ$.

QUOTIENT OF $\vec{i}_1 / \vec{i}_2 = \frac{5}{\sqrt{2}} \angle 0^\circ / \frac{10}{\sqrt{2}} \angle 60^\circ = 0.5 \angle -60^\circ$

Ex. $i_3(t) \rightarrow i_1(t) \rightarrow i_2(t)$ GIVEN: $i_1(t) = 71 \cos \omega t$

$i_2(t) = 100 \sin(\omega t - \frac{\pi}{4})$

FIND $i_3(t)$. SOLVE IT BY PHASOR DIAGRAM. Ans: $50\sqrt{2} \sin \omega t$

STEADY STATE RESPONSE OF R-L-C CIRCUITS TO SINUSOIDAL INPUTS.

ELEMENTARY CIRCUITS:

i) PURELY RESISTIVE CIRCUIT

THE INSTANTANEOUS CURRENT THROUGH THE CIRCUIT

$$i_R = \frac{v}{R} = \frac{V_m}{R} \sin \omega t = I_m \sin \omega t$$

I_m IS THE MAXIMUM VALUE OF THE CURRENT GIVEN BY $I_m = \frac{V_m}{R}$.

RMS VALUE OF THE CURRENT

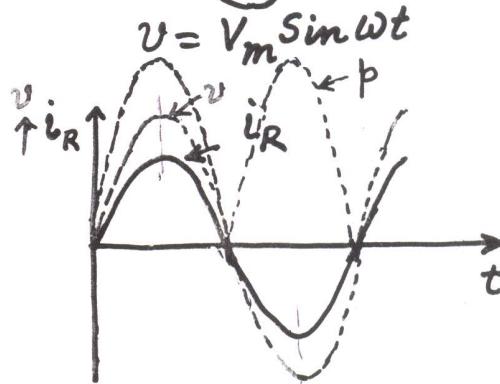
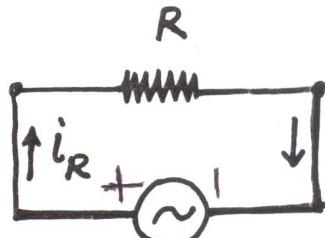
$$\overline{i}_R = \frac{\text{RMS VALUE OF VOLTAGE}}{\text{RESISTANCE}} = \frac{\overline{V}}{R}$$

$$= \frac{V_m / \sqrt{2}}{R} = \frac{I_m}{\sqrt{2}}$$

IN PHASOR NOTATIONS

$$\vec{V} = V \angle 0^\circ = V(1+j0) = V + j0$$

$$\vec{i}_R = I_R \angle 0^\circ = I_R(1+j0) = I_R + j0$$



CURRENT WILL BE IN PHASE WITH VOLTAGE.

THE VOLTAGE AND CURRENT IN AC CIRCUIT IS RELATED BY IMPEDANCE FUNCTION. THE IMPEDANCE FUNCTION MUST TELL TWO IMPORTANT FACTS:

a) THE RATIO OF V_m to I_m (OR \vec{V} to \vec{i}) AND

b) THE PHASE ANGLE BETWEEN THE WAVES OF VOLTAGE AND CURRENT (PHASORS \vec{V} AND \vec{i}).

A SPECIAL TYPE OF NOTATION IS REQUIRED TO SIGNIFY THE TWO PROPERTIES OF IMPEDANCE FUNCTION e.g. Z / ANGLE. Z SIGNIFIES THE MAGNITUDE OF IMPEDANCE AND ANGLE GIVES THE PHASE RELATION OF VOLTAGE & CURRENT.

THE IMPEDANCE FOR A PURE RESISTANCE IS $\vec{Z}_R = R \angle 0^\circ$:

POWER: INSTANTANEOUS POWER $p = v i = V_m I_m \sin^2 \omega t$

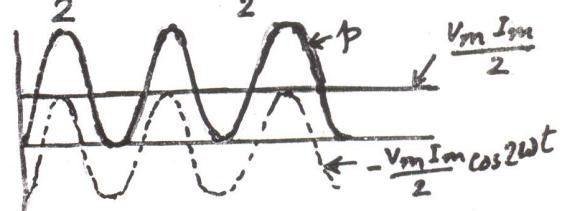
Since $\sin^2 \omega t = \frac{1}{2} - \frac{1}{2} \cos 2\omega t$, $p = \frac{V_m I_m}{2} - \frac{V_m I_m}{2} \cos 2\omega t$

SO AVERAGE POWER $P = \frac{1}{T} \int_0^T p dt$

$$P = \frac{1}{T} \int_0^T \left(\frac{V_m I_m}{2} - \frac{V_m I_m}{2} \cos 2\omega t \right) dt$$

$$= \frac{V_m I_m}{2} = \frac{V_m}{\sqrt{2}} \cdot \frac{I_m}{\sqrt{2}} = V \cdot I$$

ENERGY PRODUCED IS CONVERTED INTO HEAT AND DISSIPATED.



ii) PURELY INDUCTIVE CIRCUIT

$$\text{Since } V = L \frac{di_L}{dt} = V_m \sin \omega t$$

$$di_L = \frac{V_m}{L} \sin \omega t dt$$

$$i_L = -\frac{V_m}{\omega L} \cos \omega t$$

(Constant of integration is zero in the steady state solution as it is symmetrical about x-axis)

$$i_L = \frac{V_m}{\omega L} \sin(\omega t - 90^\circ)$$

$$= I_m \sin(\omega t - 90^\circ).$$

$$I_m = \frac{V_m}{\omega L} \Rightarrow \vec{I}_L = \frac{\vec{V}}{\omega L} = I_L \angle -90^\circ$$

$$\vec{V} = V + j0 \quad \vec{I}_L = 0 - jI_m \quad \vec{Z}_L = \frac{\vec{V}}{\vec{I}_L} = \omega L \angle 90^\circ \quad \vec{I}_L$$

THE MAGNITUDE OF THE ABOVE IMPEDANCE ωL IS CALLED INDUCTIVE REACTANCE AND REPRESENTED BY $X_L = \omega L = 2\pi f L$.

IN A PURELY INDUCTIVE CIRCUIT CURRENT LAGS BEHIND THE VOLTAGE BY 90° .

POWER & ENERGY :

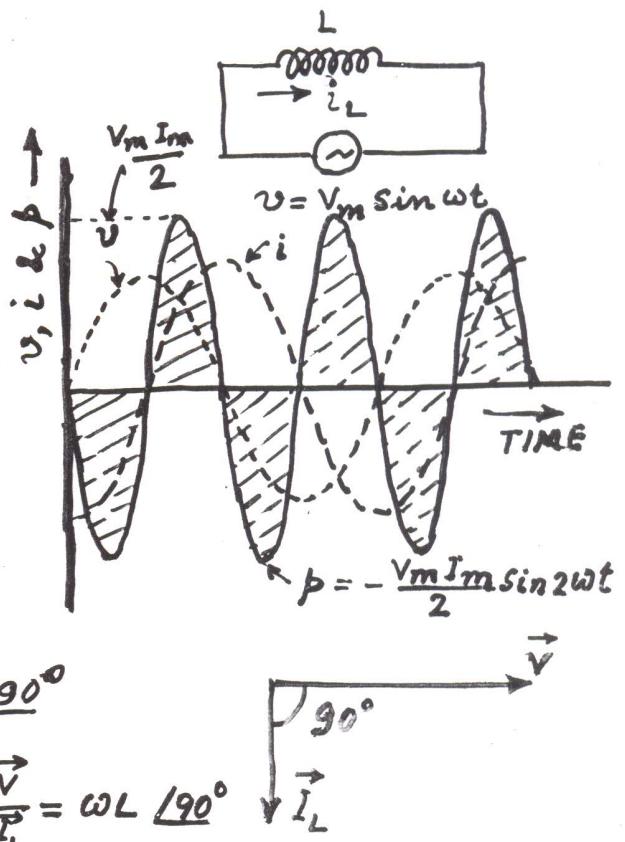
$$\begin{aligned} \text{INSTANTANEOUS POWER} &= p = Vi = [V_m \sin \omega t][I_m \sin(\omega t - 90^\circ)] \\ &= V_m I_m (-\sin \omega t \cos \omega t) \\ &= -\frac{V_m I_m}{2} \sin 2\omega t \end{aligned}$$

AVERAGE POWER

$$P = \frac{1}{T} \int_0^T p dt = 0$$

THE AMOUNT OF ENERGY DELIVERED TO THE CIRCUIT DURING A QUARTER OF A CYCLE $T/2$

$$\begin{aligned} W_L &= \int_{T/4}^{T/2} -\frac{V_m I_m}{2} \sin 2\omega t dt \\ &= \frac{V_m I_m}{2 \left(\frac{4\pi}{T} \right)} \left[\cos \frac{4\pi}{T} t \right]_{T/4}^{T/2} \\ &= \frac{V_m I_m}{2\omega} \quad (\because \omega = \frac{2\pi}{T}) \\ &= \frac{(\omega L I_m) I_m}{2\omega} = \frac{1}{2} L I_m^2 \end{aligned}$$



NOTE:

POWER VARIATION IS A DOUBLE FREQUENCY VARIATION (2ω).

THE AVERAGE POWER ABSORBED IS EQUAL TO ZERO. THE

IMPLICATION IS THAT THE INDUCTIVE ELEMENT RECEIVES ENERGY FROM THE SOURCE DURING ONE QUARTER OF A CYCLE OF THE APPLIED VOLTAGE AND RETURNS EXACTLY THE SAME AMOUNT ENERGY TO THE DRIVING SOURCE DURING THE NEXT ONE-QUARTER OF A CYCLE.

(13)

iii) PURELY CAPACITIVE CIRCUIT.

$$\text{Since } V = \frac{q}{C} = V_m \sin \omega t$$

$$i_c = \frac{dq}{dt} = V_m \omega C \cos \omega t$$

$$= \frac{V_m}{(1/\omega C)} \sin(\omega t + 90^\circ)$$

$$= I_m \sin(\omega t + 90^\circ)$$

$$I_m = \frac{V_m}{(1/\omega C)}$$

$$I_c = \frac{I_m}{\sqrt{2}} = \frac{1}{\sqrt{2}} \frac{V_m}{(1/\omega C)}$$

$$\vec{I}_c = \frac{\vec{V}}{(1/\omega C)} \angle 90^\circ$$

$$\vec{I}_c = I_c \angle 90^\circ$$

IN PHASOR NOTATIONS

$$\vec{V} = V \angle 0^\circ = V + j0$$

$$\vec{I}_c = I_c \angle 90^\circ = 0 + j I_c$$

$$\vec{Z}_c = \frac{\vec{V}}{\vec{I}_c} = \frac{V}{I_c} \angle -90^\circ = Z_c \angle -90^\circ$$

$$= \frac{1}{\omega C} \angle -90^\circ$$

THE MAGNITUDE OF THE ABOVE IMPEDANCE $\frac{1}{\omega C}$ IS CALLED CAPACITIVE REACTANCE AND REPRESENTED BY $X_C = \frac{1}{\omega C} = \frac{1}{2\pi f C}$. IN A PURELY CAPACITIVE CIRCUIT THE CURRENT LEADS THE VOLTAGE BY 90° .

POWER & ENERGY:

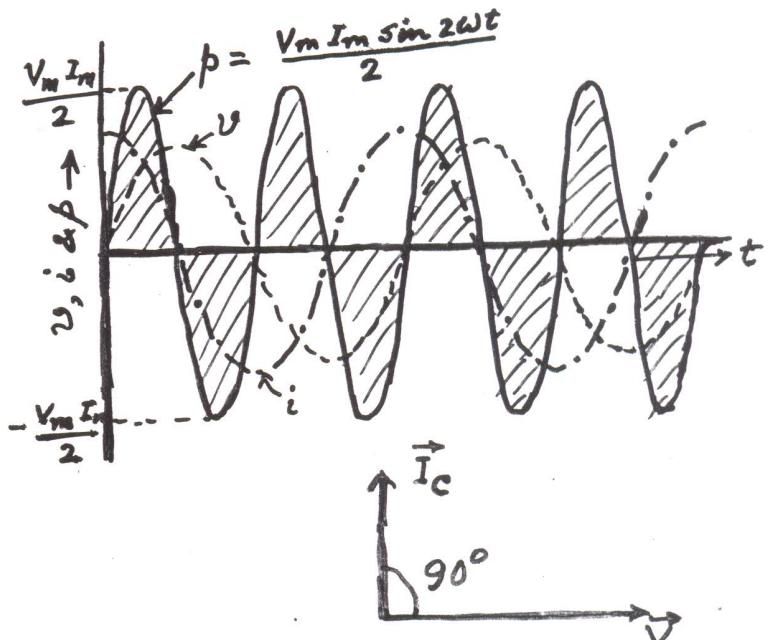
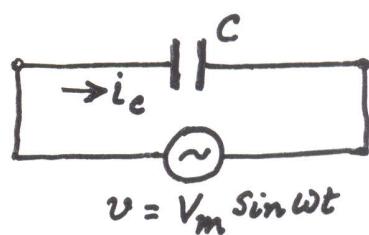
INSTANTANEOUS POWER

$$p = Vi = [V_m \sin \omega t][I_m \sin(\omega t + 90^\circ)]$$

$$= V_m I_m \sin \omega t \cos \omega t = \frac{1}{2} V_m I_m \sin 2\omega t.$$

POWER VARIATION IS AGAIN A DOUBLE FREQUENCY VARIATION (2ω) AND HENCE AVERAGE POWER ABSORBED IS EQUAL TO ZERO.

$$P = \frac{1}{T} \int_0^T p dt = 0$$



HERE, ALSO, THE CAPACITOR RECEIVES ENERGY FROM THE SOURCE DURING THE FIRST QUARTER OF A CYCLE AND RETURNS THE SAME AMOUNT DURING THE SECOND QUARTER CYCLE, etc.

$$W_c = \int_0^{T/4} \frac{V_m I_m}{2} \sin 2\omega t dt$$

$$= \frac{1}{2} \frac{V_m I_m}{\omega} = \frac{1}{2} C V_m^2$$

iv) SERIES R-L CIRCUIT:

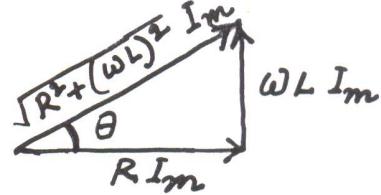
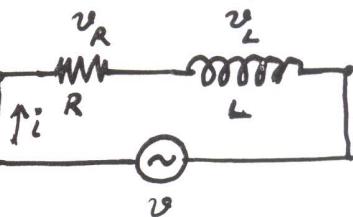
$$\text{let } i = I_m \sin \omega t$$

Instantaneous voltage equation

$$v_R + v_L = v \Rightarrow R i + L \frac{di}{dt} = v$$

$$\text{i.e., } R I_m \sin \omega t + \omega L I_m \cos \omega t = v$$

$$\text{i.e., } I_m \left[\sin \omega t \frac{R}{\sqrt{R^2 + (\omega L)^2}} + \cos \omega t \frac{\omega L}{\sqrt{R^2 + (\omega L)^2}} \right] = \frac{v}{\sqrt{R^2 + (\omega L)^2}}$$



Then

$$I_m \left[\sin \omega t \cos \theta + \cos \omega t \sin \theta \right] = \frac{v}{\sqrt{R^2 + (\omega L)^2}}$$

from which

$$\begin{aligned} v &= I_m \sqrt{R^2 + (\omega L)^2} \sin(\omega t + \theta) \\ &= I_m Z \sin(\omega t + \theta) = V_m \sin(\omega t + \theta) \end{aligned}$$

Thus we get a) $Z = \sqrt{R^2 + (\omega L)^2} = \frac{V_m}{I_m}$, b) $\theta = \tan^{-1} \frac{\omega L}{R}$. and c) v leads i in R-L circuit by θ° , or otherwise i lags behind v by θ°

In Phasor Notations:

$$\begin{aligned} \vec{V} &= \vec{V}_R + \vec{V}_L \\ &= \vec{I} R + \vec{I} j \omega L \\ &= \vec{I} (R + j \omega L) = \vec{I} Z \end{aligned}$$

Impedance:

$$\begin{aligned} \overline{Z}_{R-L \text{ circuit}} &= R + j \omega L = \sqrt{R^2 + (\omega L)^2} \left/ \tan^{-1} \left(\frac{\omega L}{R} \right) \right. \\ &= \sqrt{R^2 + X_L^2} \left/ \tan^{-1} \left(\frac{X_L}{R} \right) \right. \end{aligned}$$

The instantaneous value of the power is given by

$$p = vi = [V_m \sin(\omega t + \theta)] [I_m \sin \omega t]$$

Expanding $\sin(\omega t + \theta)$ and multiplying

$$p = \frac{V_m I_m}{2} \cos \theta - \frac{V_m I_m}{2} [\cos 2\omega t] \cos \theta + \frac{V_m I_m}{2} [\sin 2\omega t] \sin \theta$$

The Average value of the power

$$P_{av} = \frac{1}{T} \int_0^T V_m \sin(\omega t + \theta) I_m \sin \omega t dt = \frac{V_m I_m}{2} \cos \theta.$$

The expression for average power may also be derived as follows:

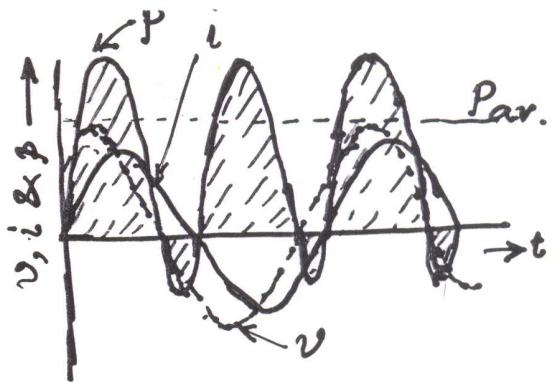
$$P = \frac{1}{T} \int_0^T V_m I_m \sin(\omega t + \theta) \sin \omega t dt$$

$$= \frac{V_m I_m}{T} \int_0^T \frac{1}{2} [\cos \theta - \cos(2\omega t + \theta)] dt$$

The term $\frac{V_m I_m}{2T} \int_0^T \cos \theta dt$

$$= \frac{1}{2} V_m I_m \cos \theta = \frac{V_m}{\sqrt{2}} \cdot \frac{I_m}{\sqrt{2}} \cos \theta = V I \cos \theta.$$

Other term contains $2\omega t$, the average value of which over a complete cycle is zero.



v) SERIES R-C CIRCUIT

Let $i = I_m \sin \omega t$, then

$$v_R + v_C = v$$

$$\text{i.e. } R i + \frac{1}{C} \int i dt = v$$

$$\text{i.e. } R I_m \sin \omega t - \frac{1}{\omega C} I_m \cos \omega t = v$$

$$\text{i.e. } I_m \left[\sin \omega t \frac{R}{\sqrt{R^2 + (\frac{1}{\omega C})^2}} - \cos \omega t \frac{\frac{1}{\omega C}}{\sqrt{R^2 + (\frac{1}{\omega C})^2}} \right] = \frac{v}{\sqrt{R^2 + (\frac{1}{\omega C})^2}}.$$

from which $I_m \sqrt{R^2 + (\frac{1}{\omega C})^2} [\sin \omega t \cos \theta - \cos \omega t \sin \theta] = v$

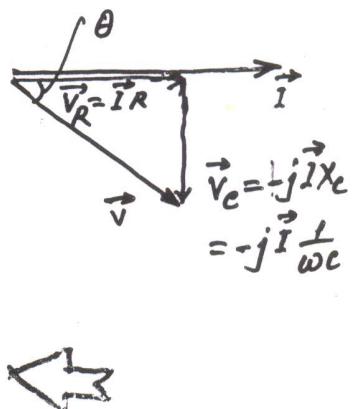
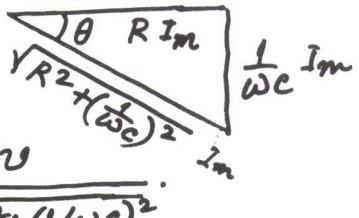
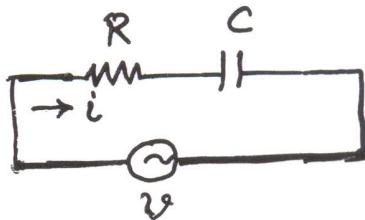
$$\text{i.e. } v = I_m Z \sin(\omega t - \theta) = V_m \sin(\omega t - \theta)$$

Thus we get a) $Z = \sqrt{R^2 + (\frac{1}{\omega C})^2} = \frac{V_m}{I_m}$, b) $\theta = \tan^{-1}\left(\frac{1/\omega C}{R}\right)$
and c) v lags i in R-C circuit by θ° , or
otherwise the current i leads the voltage v by θ° .

In phasor notations.

$$\begin{aligned} \vec{v} &= \vec{v}_R + \vec{v}_C = \vec{I} R - \vec{I} \left(j \frac{1}{\omega C}\right) = \vec{I} \left(R - j \frac{1}{\omega C}\right) \\ &= \vec{I} \vec{Z}. \end{aligned}$$

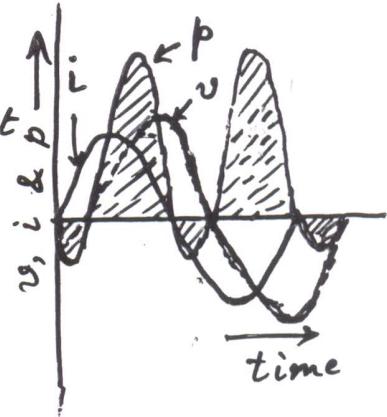
$$\begin{aligned} \text{Impedance: } \vec{Z}_{\text{R-C circuit}} &= R - j \frac{1}{\omega C} \\ &= \sqrt{R^2 + \left(\frac{1}{\omega C}\right)^2} \angle \tan^{-1}\left(\frac{1}{\omega C}\right) \\ &= \sqrt{R^2 + X_C^2} \angle -\tan^{-1}\left(\frac{X_C}{R}\right) \end{aligned}$$



The instantaneous value of the power is given by
 $p = v i = [V_m \sin(\omega t - \theta)][I_m \sin \omega t]$

The average value of the power

$$\begin{aligned} P &= \frac{1}{T} \int_0^T V_m \sin(\omega t - \theta) I_m \sin \omega t dt \\ &= \frac{V_m I_m}{2T} \int_0^T [\cos \theta - \cos(2\omega t - \theta)] dt \\ &= \frac{1}{2} V_m I_m \cos \theta = VI \cos \theta. \end{aligned}$$

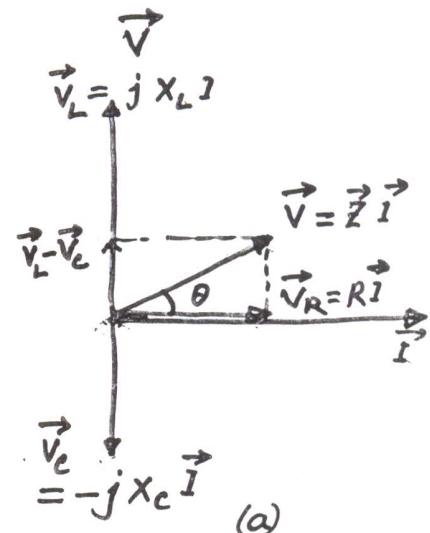
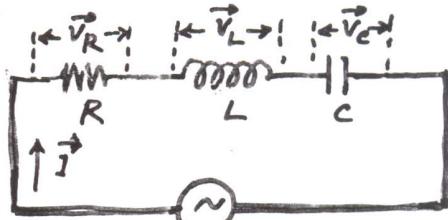


vi) SERIES R-L-C CIRCUIT

IN PHASOR NOTATION

$$\begin{aligned} \vec{V} &= \vec{V}_R + \vec{V}_L + \vec{V}_C \\ &= \vec{I}R + j\vec{I}x_L - j\vec{I}x_C \\ &= \vec{I}[R + j(x_L - x_C)] \end{aligned}$$

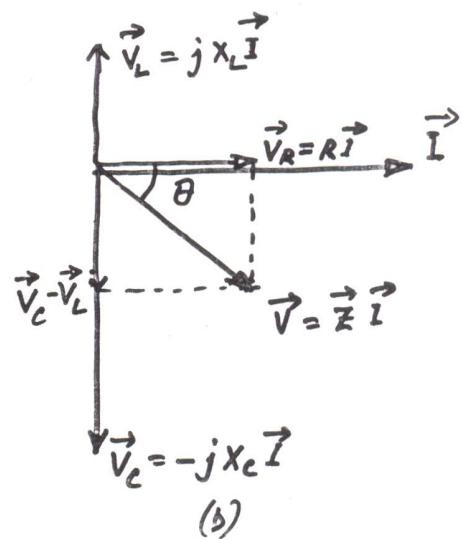
$$\begin{aligned} \vec{Z} &= R + j(x_L - x_C) \\ &= R + j(\omega L - \frac{1}{\omega C}) \\ \vec{Z} &= \sqrt{R^2 + (\omega L - \frac{1}{\omega C})^2} \quad \left/ \tan^{-1} \frac{(\omega L - \frac{1}{\omega C})}{R} \right. \end{aligned}$$



If $V_L > V_C$ the circuit will behave like an R-L circuit and its phasor diagram is shown in Fig (a)

On the other hand if $V_L < V_C$, the circuit will behave like an R-C circuit and its phasor diagram is shown in Fig. (b).

In case (a) the current I drawn will lag the supply voltage V and in case (b) the current I will lead the supply voltage V.



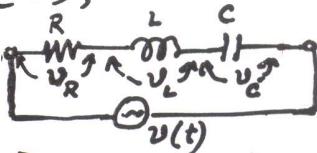
(17)

Ex-2. Given $v(t) = 100 \sin 40t$; $R = 10\Omega$, $L = 0.2H$, $C = 0.0014F$.

i) Determine $i(t)$, $v_R(t)$, $v_L(t)$, $v_C(t)$,

ii) Calculate the power loss,

iii) Show the phasor diagram



Sol. Here $\overline{Z} = R + j(\omega L - \frac{1}{\omega C}) = 10 + j[40 \times 0.2 - \frac{1}{40 \times 0.0014}] \Omega$
 $= (10 - j10) \Omega = 14.14 \angle -45^\circ$

So i) $i(t) = \frac{100}{14.14 \angle -45^\circ} \sin 40t = 7.1 \sin(40t + 45^\circ)$.

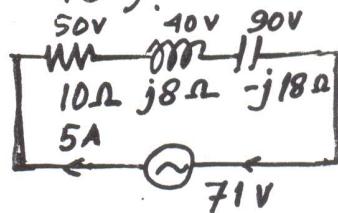
$$v_R(t) = i(t) \cdot R = 71 \sin(40t + 45^\circ)$$

$$v_L(t) = i(t) \cdot (jX_L) = 56.8 \sin(40t + 135^\circ)$$

$$v_C(t) = i(t) (-jX_C) = 127.8 \sin(40t - 45^\circ)$$

In terms of phasors

$$\vec{I} = \frac{\vec{V}}{\overline{Z}} = \frac{(100/\sqrt{2}) \angle 0^\circ}{14.14 \angle -45^\circ} = 5 \angle 45^\circ$$



$$\vec{v}_R = R \vec{I} = 50 \angle 45^\circ, \quad \vec{v}_L = jX_L \vec{I} = 40 \angle 135^\circ$$

$$\vec{v}_C = -jX_C \vec{I} = 90 \angle 45^\circ$$

NOTE: THE VALUES IN TERMS OF PHASOR QUANTITIES ARE ALL RMS VALUES WHEREAS AS TIME FUNCTIONS OR IN INSTANTANEOUS VALUES ARE EXPRESSED WITH MAXIMUM VALUES.

ii) Power loss $= I^2 R = (5)^2 \times 10 W = 250W$. No power loss in inductance or capacitance.

Power factor $\cos \theta = \cos 45^\circ = 0.707$ leading.

iii) Phasor diagram: CIRCUIT IS BEHAVING LIKE A R-C CIRCUIT.

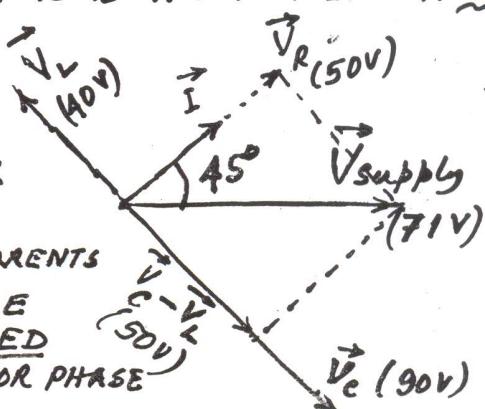
NOTE CAREFULLY

THE ARITHMETIC SUM OF v_R , v_L & v_C IS MUCH GREATER THAN THE APPLIED VOLTAGE.

ALTERNATING VOLTAGES OR CURRENTS OF THE SAME FREQUENCY CAN BE

ADDED BUT THEY MUST BE ADDED

VECTORIZALLY WITH DUE REGARD FOR PHASE RELATION.



VOLT AMPERE , ACTIVE AND REACTIVE POWER

Product of r.m.s. Values of current and applied Voltage is called

APPARENT POWER OR VOLT-AMPERE (VA). A larger unit is KVA or MVA. (Apparent power = VI)

ACTIVE POWER

$$P = VI \cos\theta \text{ (Watt)}$$

$$\therefore \cos\theta = \frac{P}{VI} = \frac{\text{Active Power}}{\text{Volt-Ampere}}$$

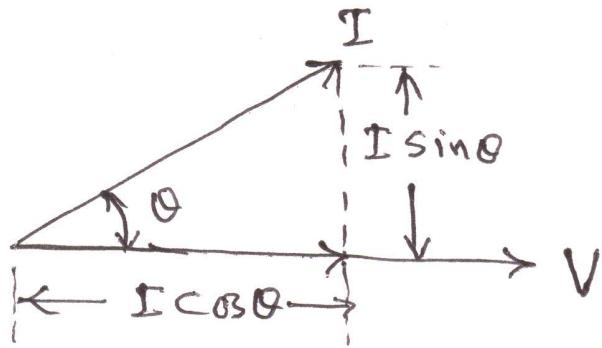
REACTIVE POWER

$$Q = VI \sin\theta \text{ (Var)}$$

Reactive Power Factor = $\sin\theta$

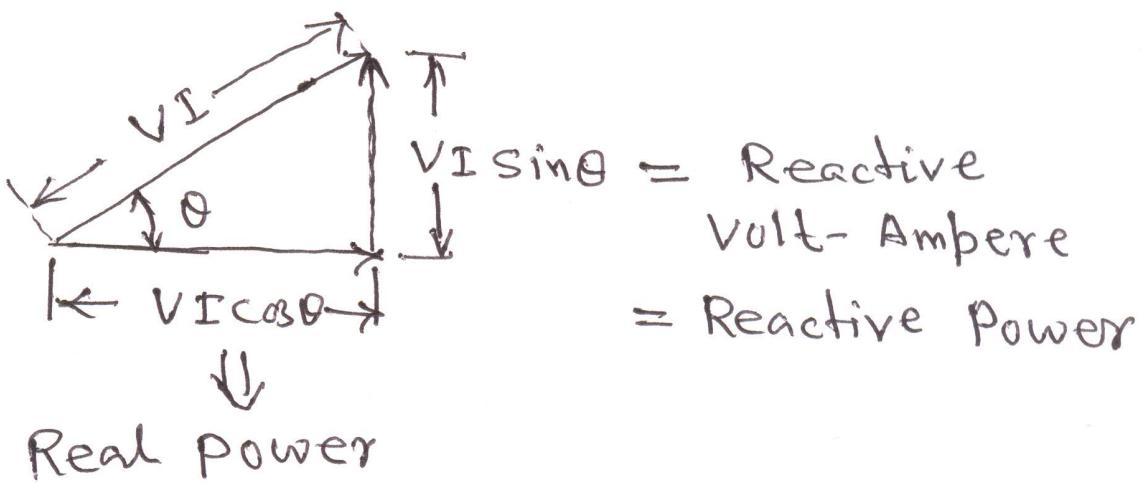
$$= \frac{\text{Reactive Volt-Ampere}}{\text{Volt-Ampere}} = \frac{Q}{VI}$$

(19) (20)



$I \cos \theta \Rightarrow$ active or power component of the current

$I \sin \theta \Rightarrow$ reactive or wattless component of current.

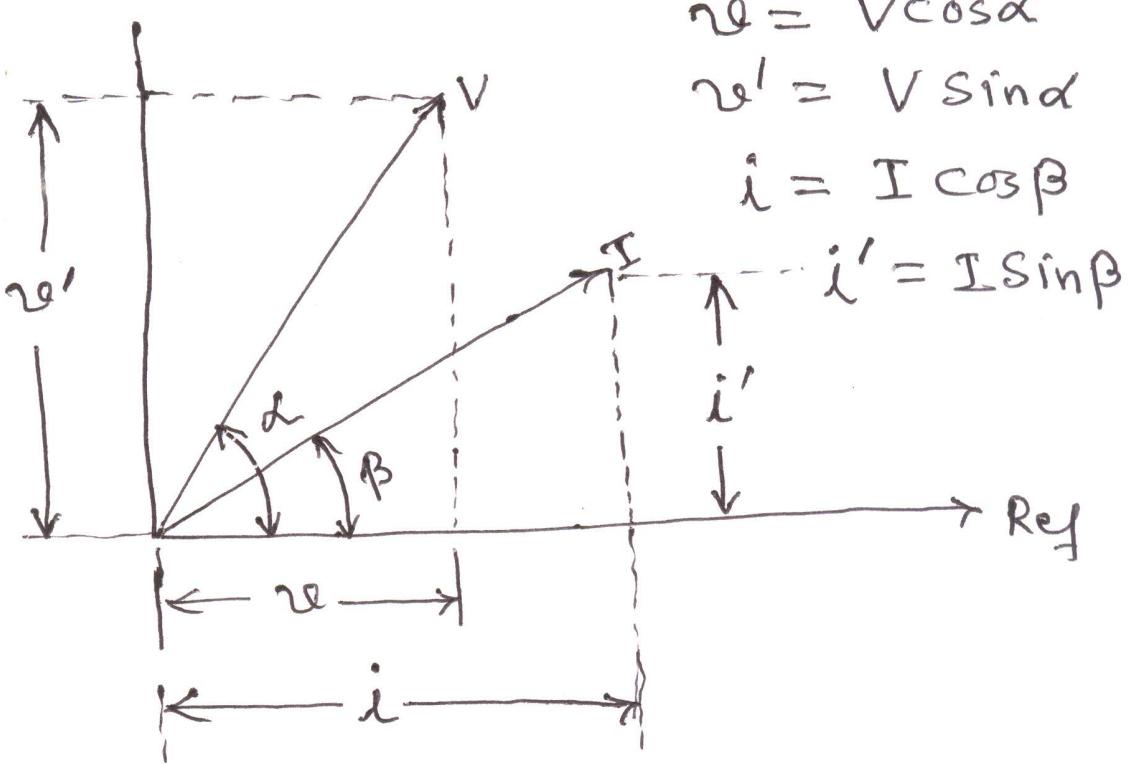


$$\therefore \text{Volt-Ampere} = VI$$

$$= \sqrt{(\text{Real Power})^2 + (\text{Reactive Power})^2}$$

(20) 3

CALCULATION OF POWER USING COMPLEX NOTATION



$$\begin{aligned} \text{Voltage} &= V L \angle \alpha = V (\cos \alpha + j \sin \alpha) \\ &= (v + j v') \end{aligned}$$

$$\begin{aligned} \text{Current} &= I L \angle \beta = I (\cos \beta + j \sin \beta) \\ &= (i + j i') \end{aligned}$$

Phase difference between Voltage and current = $(\alpha - \beta)$

Therefore,

(21) ④

$$\text{Power Factor} = \cos(\alpha - \beta)$$

$$\therefore \text{Active Power} = P = VI \cos(\alpha - \beta).$$

$$\therefore P = VI (\cos \alpha \cos \beta + \sin \alpha \sin \beta)$$

$$\boxed{\therefore P = (re_i + re'i')}$$

$$\text{Reactive Power} = Q = VI \sin(\alpha - \beta)$$

$$\therefore Q = VI (\sin \alpha \cos \beta - \cos \alpha \sin \beta)$$

$$\boxed{\therefore Q = (re'i - rei')}$$

NOW

$$\text{current} = I \angle \beta$$

$$\therefore (\text{current})^* = I \angle -\beta$$

$$\therefore P + jQ = (\text{Voltage}) \times (\text{current})^*$$

(22) (55)

$$\therefore P + jQ = (v + jv') (i + ji')^*$$

$$\therefore P + jQ = (v + jv') (i - ji')$$

$$\therefore P + jQ = (vi + v'i') + j(v'i - vi')$$

$$\therefore P = (vi + v'i')$$

$$Q = (v'i - vi')$$

EXAMPLE - 4.

$$V = (100 + j200) \text{ Volt}$$

$$I = (10 + j5) \text{ Amp}$$

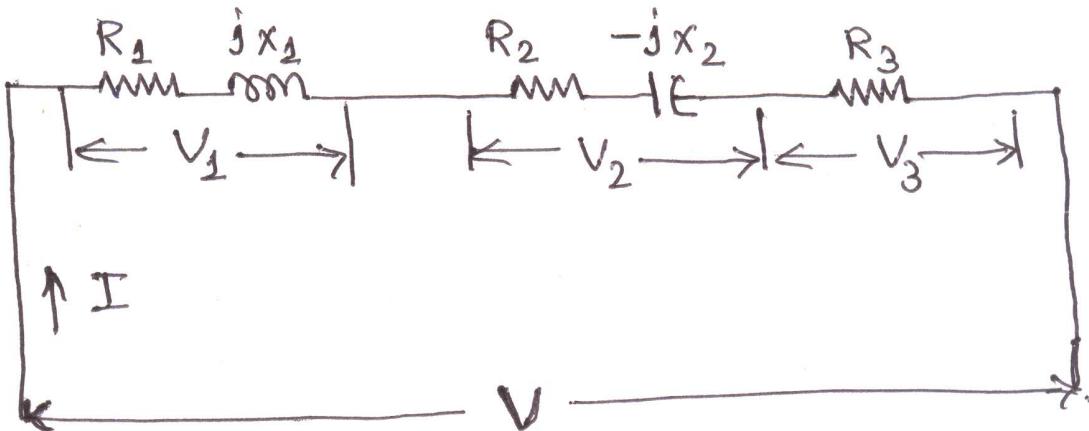
$$\therefore P + jQ = VI^* = (100 + j200) (10 - j5)$$

$$= (2000 + j1500)$$

$$\therefore P = 2000 \text{ Watt} = 2 \text{ kW}$$

$$Q = 1500 \text{ VAR} = 1.5 \text{ kVAR}$$

SINGLE PHASE CIRCUIT ANALYSIS



$$Z_1 = R_1 + jx_1$$

$$Z_2 = R_2 - jx_2$$

$$Z_3 = R_3$$

$$V_1 = IZ_1; \quad V_2 = IZ_2; \quad V_3 = IZ_3$$

$$V = V_1 + V_2 + V_3 = I(Z_1 + Z_2 + Z_3)$$

$$\therefore V = I \left[(R_1 + R_2 + R_3) + j(x_1 - x_2) \right]$$

$$\therefore V = IZ = I(R + jx)$$

Where, $R = (R_1 + R_2 + R_3)$ & $x = (x_1 - x_2)$

(2A) 7

$$Z = \left(\sqrt{R^2 + X^2} \right) \angle \theta$$

$$\theta = \tan^{-1} \left(\frac{X}{R} \right) = \tan^{-1} \left(\frac{X_1 - X_2}{R_1 + R_2 + R_3} \right)$$

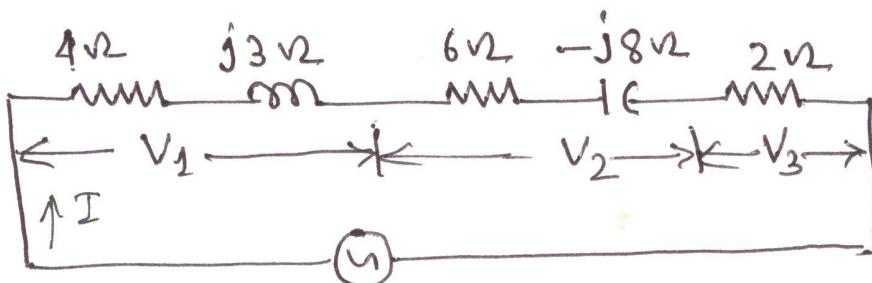
$$I = \frac{V \angle 0^\circ}{\left(\sqrt{R^2 + X^2} \right) \angle \theta} = \frac{V \angle 0^\circ - \theta}{\sqrt{R^2 + X^2}} = \frac{V \angle -\theta}{\sqrt{R^2 + X^2}}$$

\therefore Power factor = $\cos \theta$

$$\rightarrow \tan \theta = \frac{X}{R}; \quad \therefore \frac{\sin \theta}{X} = \frac{\cos \theta}{R} = \frac{1}{\sqrt{R^2 + X^2}}$$

$$\therefore \cos \theta = \frac{R}{\sqrt{R^2 + X^2}}$$

EXAMPLE - 5



$$V = 100\text{ Volt}$$

(25) (Ans)

$$I = \frac{V}{Z} = \frac{100 \angle 0^\circ}{(4+j3) + (6-j8) + 2}$$

$$\therefore I = (7.1 + j2.96) \text{ Amp.} = 7.69 \angle 22.6^\circ \text{ Amp}$$

$$V_1 = IZ_1 = (7.1 + j2.96)(4+j3)$$

$$\therefore V_1 = (19.53 + j33.14) \text{ Volt} = 38.47 \angle 59.5^\circ \text{ Volt}$$

$$V_2 = IZ_2 = (7.1 + j2.96)(6-j8)$$

$$\begin{aligned} \therefore V_2 &= (66.27 - j39.06) \text{ Volt} \\ &= 76.92 \angle -30.5^\circ \end{aligned}$$

$$V_3 = IZ_3 = (7.1 + j2.96) \times 2$$

$$\therefore V_3 = (14.2 + j5.92) \text{ Volt.}$$

$$\therefore V_3 = 15.38 \angle 22.6^\circ \text{ Volt}$$

Check

$$V_1 + V_2 + V_3 = (100 + j0) \text{ Volt.}$$

(26) (4)

$$P_1 = I^2 R_1 = (7.69)^2 \times 4 = 237 \text{ Watt}$$

$$P_2 = I^2 R_2 = (7.69)^2 \times 6 = 355 \text{ Watt}$$

$$P_3 = I^2 R_3 = (7.69)^2 \times 2 = 118 \text{ Watt.}$$

Total Power

$$P = P_1 + P_2 + P_3 = (237 + 355 + 118) \text{ Watt.}$$

$$\therefore P = 710 \text{ Watt.}$$

$$V = 100 \text{ } [100^\circ] \text{ Volt}$$

$$I = 7.69 \text{ } [22.6^\circ] \text{ Amp.}$$

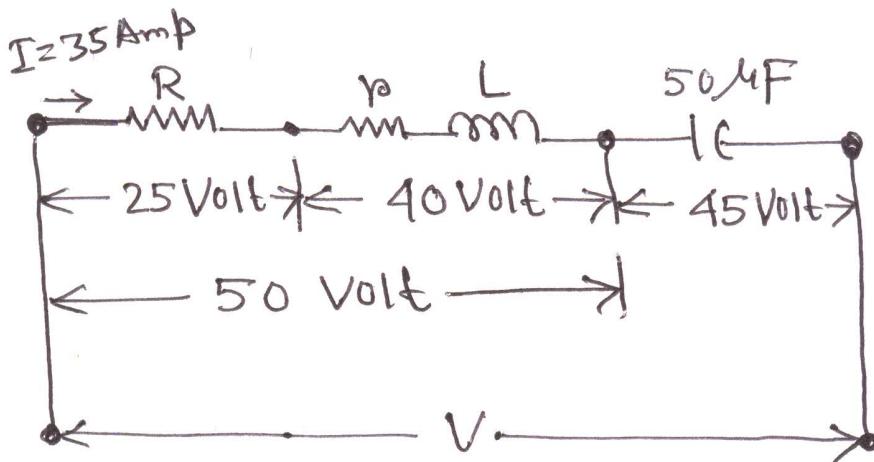
$$\text{Power factor} = \cos(22.6^\circ) = 0.9232$$

(leading)

$$P = V I \cos \theta = 100 \times 7.69 \times 0.9232 \text{ Watt}$$

$$\therefore P = 710 \text{ Watt.}$$

DRAW PHASOR DIAGRAM

EXAMPLE - 6

Determine R , r_o , L , f and V

Soln. $V_1 = 25 \text{ Volt}; V_2 = 40 \text{ Volt};$

$$V_3 = 45 \text{ Volt}.$$

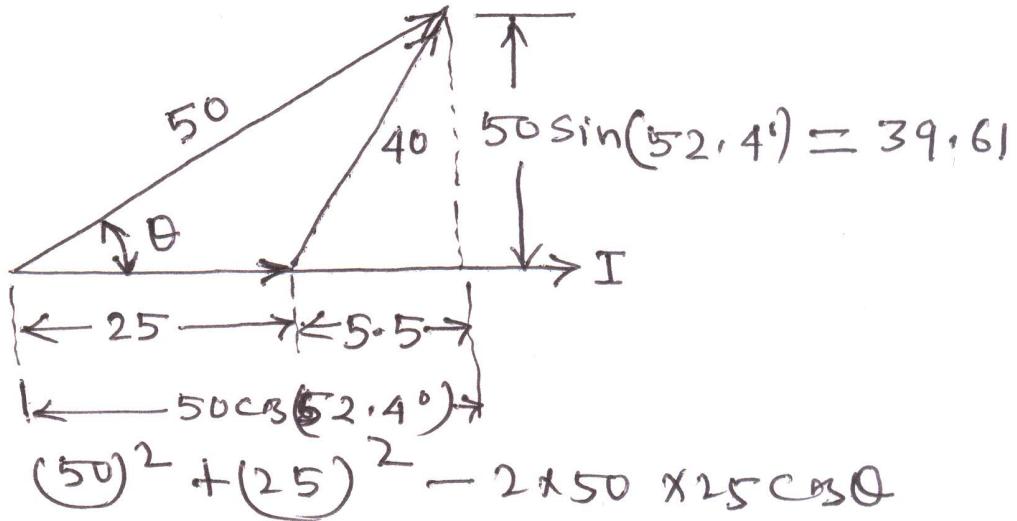
$$\frac{V_3}{X_C} = I = 35$$

$$\therefore \frac{45}{1/\omega C} = 35 \Rightarrow \frac{45}{1/2\pi f C} = 35$$

$$\therefore f = 2.475 \text{ kHz}.$$

$$R = \frac{V_1}{I} = \frac{25}{35} = 0.714 \Omega$$

(28) ~~10~~



$$\therefore \cos\theta = \frac{1525}{2500} = 0.61$$

$$\therefore \theta = 52.4^\circ$$

$$\begin{aligned} 50 \cos(52.4^\circ) \\ = 50 \times 0.61 \\ = 30.5 \end{aligned}$$

Voltage drop across γ
= 5.5 Volt

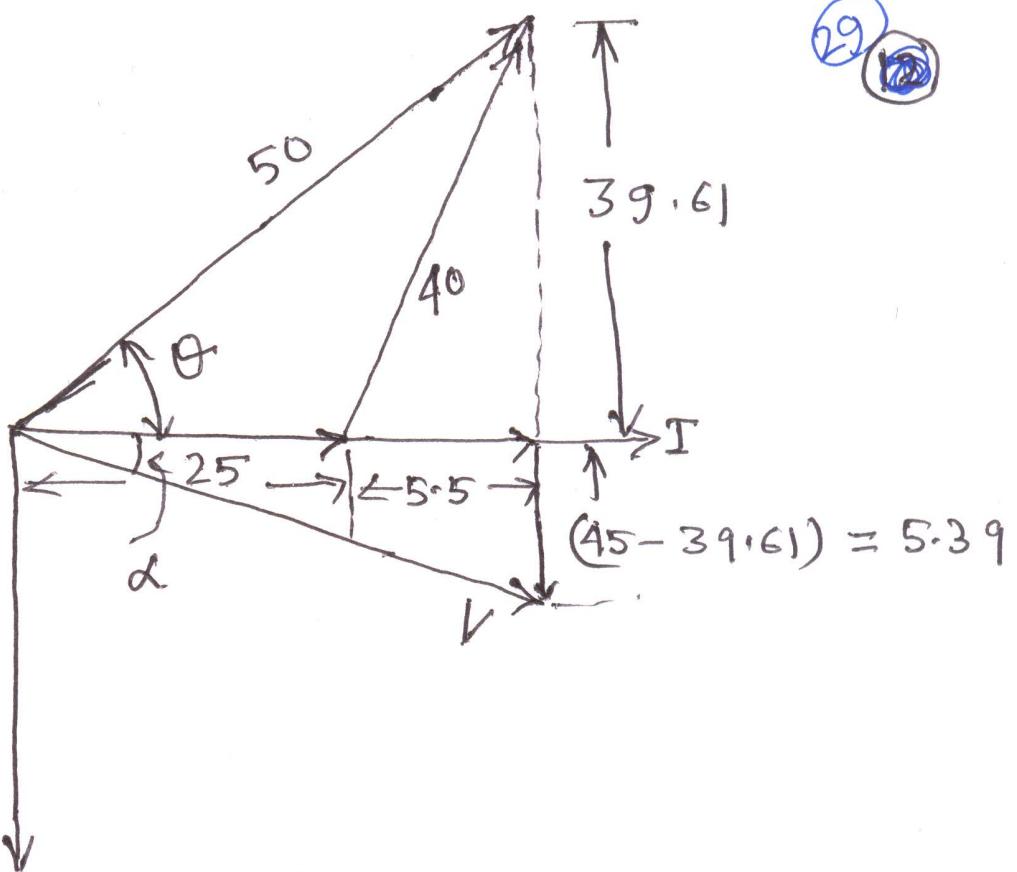
Voltage drop across $L = 39.61$ Volt.

$$I\gamma = 5.5 \quad \therefore 35\gamma = 5.5 \quad \boxed{\therefore \gamma = 0.157 \Omega}$$

$$I \times L = 39.61 \quad \therefore \cancel{I =}$$

$$\therefore 35 \times L \times 2\pi \times \frac{2.475 \times 1000}{1000} = 39.61$$

$$\therefore \boxed{L = 0.073 \text{ mH}}$$



$$V_3 = 45 \text{ Volt}$$

$$\therefore V = \sqrt{(25 + 5.5)^2 + (5.39)^2}$$

$$\therefore V = 31 \text{ Volt. } (\approx 30.972 \text{ Volt})$$

$$\alpha = \tan^{-1} \left(\frac{5.39}{30.5} \right) = 10^\circ$$

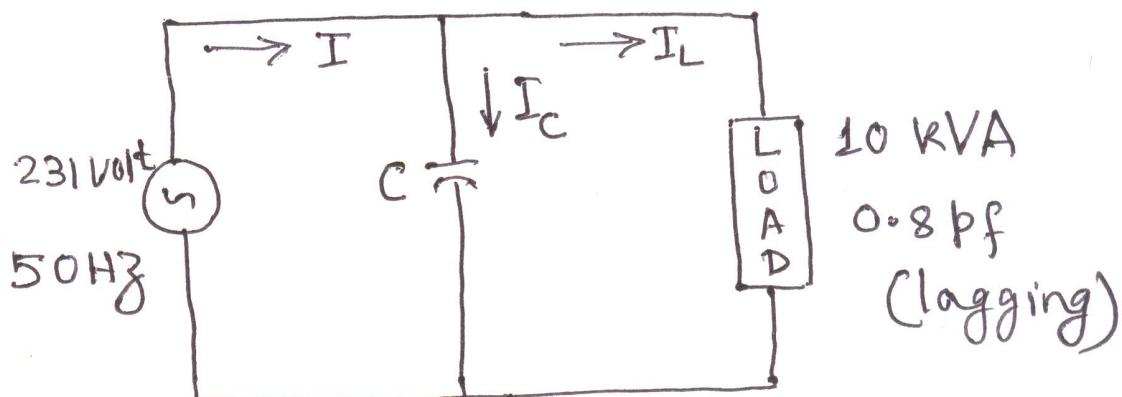
$$\text{Power factor} = \cos \alpha = \cos(10^\circ) = 0.9848$$

$$P = (35)^2 (0.714 + 0.157) = 1.0669 \text{ kW}$$

Also

$$P = VI \cos \alpha = 31 \times 35 \times 0.9848 \approx 1.0669 \text{ kW}$$

$$=$$

EXAMPLE - 7

- i) $C = ?$ to make the overall power factor (i.e., load + shunt capacitor) 0.95 Lagging.
- ii) Current drawn before and after installing the capacitor

Sohm.

Before installing the capacitor, the current is only drawn by the load.

$$\text{i.e., } I_L = \frac{\text{Apparent Power of Load}}{\text{Voltage}}$$

$$\therefore I_L = \frac{10 \times 10^3}{231} = 43.29 \text{ Amp}$$

(31)

(24)

and

$$\cos\theta = 0.8$$

$$\therefore \theta = \cos^{-1}(0.8) = 36.9^\circ \text{ (lagging)}$$

To make overall power factor
0.95 lagging,

$$\cos\delta = 0.95$$

$$\therefore \delta = 18.2^\circ \text{ (lagging)}$$

Since the pure capacitor draws a leading current at 90° with V , I_C is shown \perp to V and leading V by 90°

From phasor diagram

$$X = 43.29 \cos(36.9^\circ)$$

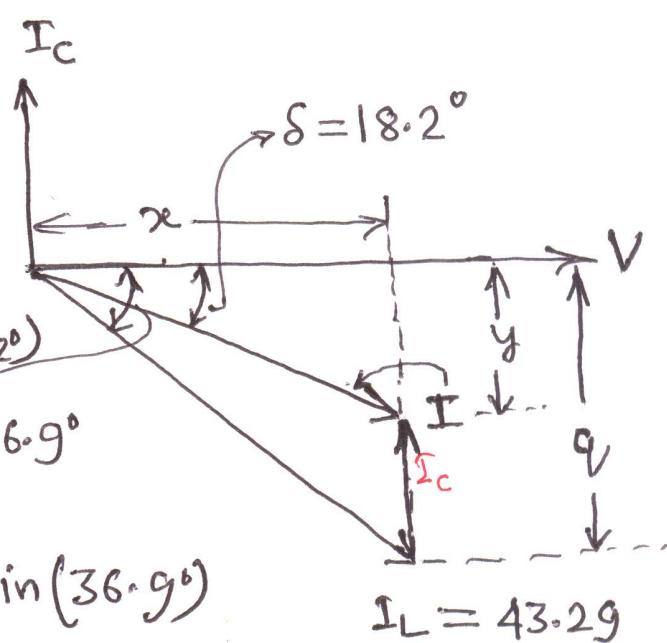
$$= 34.63$$

$$Y = X \tan\delta = 34.63 \tan(18.2^\circ)$$

$$\therefore Y = 11.39$$

$$q_L = I_L \sin\theta = 43.29 \sin(36.9^\circ)$$

$$\therefore q_L = 26.0$$



(32)

Supply Current

$$I = \sqrt{x^2 + y^2} = \sqrt{(34.63)^2 + (11.39)^2}$$

$$\boxed{\therefore I = 36.46 \text{ Amp.}}$$

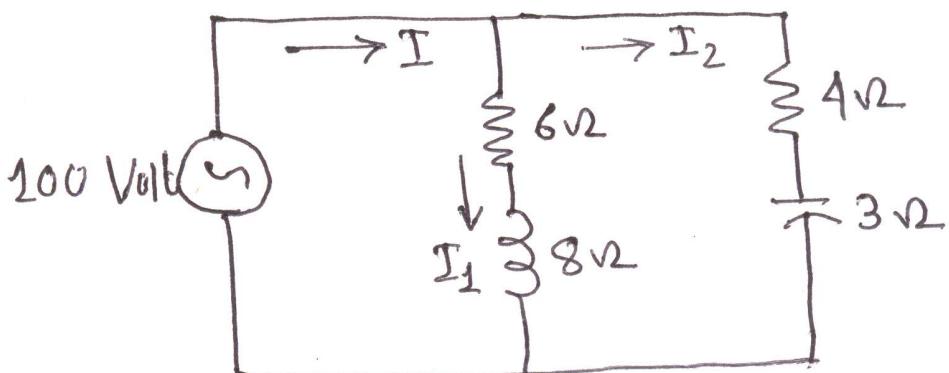
$$I_C = V - y = (26 - 11.39) = 14.61 \text{ Amp}$$

Now $I_C = \frac{V}{(\frac{1}{\omega C})} = \omega CV$

$$\therefore C = \frac{I_C}{\omega V} = \frac{14.61}{2\pi \times 50 \times 231}$$

$$\boxed{\therefore C = 201 \mu F}$$

EXAMPLE - 8



Sohm.

(33)

(10)

$$V = 100 \angle 0^\circ \text{ Volt}$$

$$I_1 = \frac{100 \angle 0^\circ}{6+j8} = (6-j8) = 10 \angle -53.2^\circ \text{ Amp}$$

$$I_2 = \frac{100 \angle 0^\circ}{4-j3} = (16+j12) = 20 \angle 36.8^\circ \text{ Amp}$$

$$I = I_1 + I_2 = (6-j8) + (16+j12) = (22+j4) \text{ Amp}$$

$$\therefore I = 22.35 \angle 10.3^\circ \text{ Amp}$$

$$Y = \frac{1}{Z} = \text{Admittance}$$

$$Y_1 = \frac{1}{Z_1} = \frac{1}{(6+j8)} = (0.06 - j0.08) \text{ mho}$$

$$Y_2 = \frac{1}{Z_2} = \frac{1}{(4-j3)} = (0.16 + j0.12) \text{ mho}$$

$$\rightarrow Y_1 = 0.06 - j0.08 = g_1 - jb_1$$

$$\therefore g_1 = 0.06; b_1 = 0.08$$

$$\rightarrow Y_2 = 0.16 + j0.12 = g_2 - jb_2$$

$$\therefore g_2 = 0.16; b_2 = -0.12$$

$$\frac{1}{Z} = \frac{1}{Z_1} + \frac{1}{Z_2}$$

Q.A

Q.B

$$\therefore Y = Y_1 + Y_2 = (g_1 - jb_1) + (g_2 - jb_2)$$

$$\therefore Y = (g_1 + g_2) - j(b_1 + b_2)$$

$$\therefore Y = g - jb$$

$g \Rightarrow$ Conductance

$b \Rightarrow$ Susceptance

$$g = g_1 + g_2 = (0.06 + 0.16) = 0.22$$

$$b = b_1 + b_2 = 0.08 - 0.12 = -0.04$$

$$\therefore Y = g - jb = (0.22 - j(-0.04))$$

$$\therefore Y = (0.22 + j0.04) \text{ mho}$$

Now

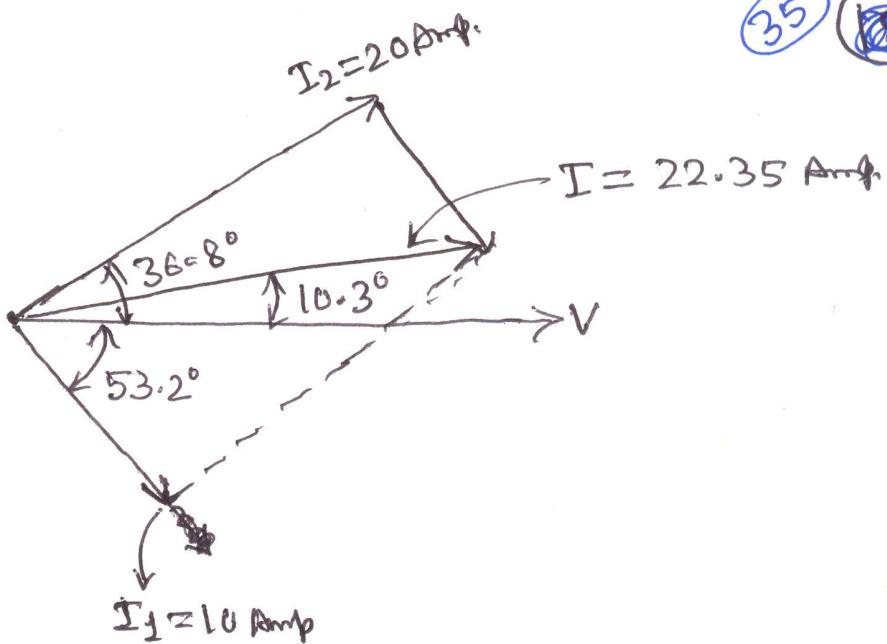
$$I = \frac{V}{Z} = Y V = (100 \angle 0^\circ)(0.22 + j0.04)$$

$$\therefore I = 22.35 \angle 10.3^\circ \text{ Amp}$$

$$I_1 = (100 \angle 0^\circ)(0.06 - j0.08) = (6 - j8) = 10 \angle -53.2^\circ \text{ Amp}$$

$$I_2 = (100 \angle 0^\circ)(0.16 + j0.12) = (16 + j12) = 20 \angle 36.8^\circ \text{ Amp}$$

Phasor diagram



(35) (18)

$$\text{Power factor} = \cos(10.3^\circ) = 0.9838$$

$$\text{Input Power} = VI \cos\theta$$

$$= 100 \times 22.35 \times 0.9838 \\ = 2198.793 \text{ Watt}$$

Also

$$P_1 = I_1^2 R_1 = (10)^2 \times 6 = 600 \text{ Watt}$$

$$P_2 = I_2^2 R_2 = (20)^2 \times 4 = 1600 \text{ Watt.}$$

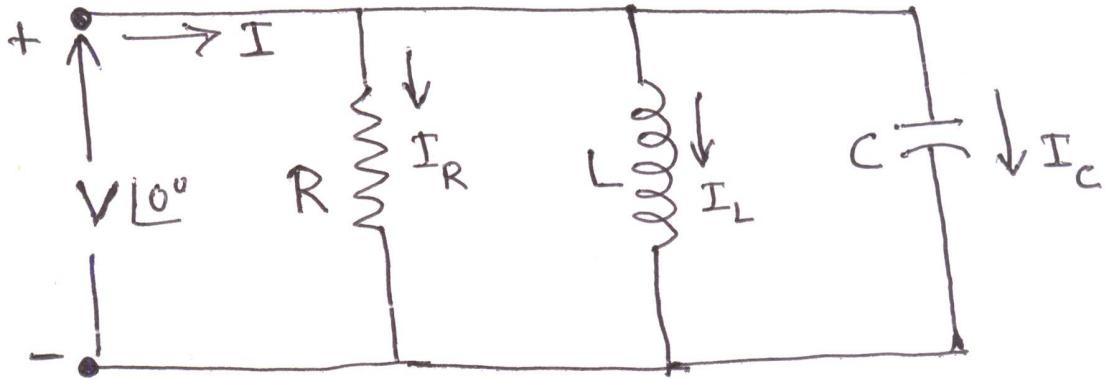
$$\therefore P = P_1 + P_2 = (600 + 1600) = 2200 \text{ Watt}$$

$$VI \cos\theta = 100 \times 22.36 \times 0.98388$$

$$= 2199.96 \approx 2200 \text{ Watt}$$

PARALLEL R-L-C CIRCUIT

(36) (19)



$$I_R = \frac{V [0^\circ]}{R} = \frac{V}{R} [0^\circ]$$

$$I_L = \frac{V [0^\circ]}{jX_L} = \frac{V [0^\circ]}{X_L [-90^\circ]} = \frac{V}{X_L} [-90^\circ]$$

$$I_C = \frac{V [0^\circ]}{-jX_C} = \frac{V [0^\circ]}{X_C [-90^\circ]} = \frac{V}{X_C} [-90^\circ]$$

$$I = I_R + I_L + I_C$$

$$\rightarrow I_R = I_R [0^\circ], \text{ i.e., } I_R = \frac{V}{R}$$

$$\rightarrow I_L = I_L [-90^\circ], \text{ i.e., } I_L = \frac{V}{X_L}$$

$$\rightarrow I_C = I_C [-90^\circ], \text{ i.e., } I_C = \frac{V}{X_C}$$

(20)
37

$$\therefore I = i_R \begin{smallmatrix} 0^\circ \\ \square \end{smallmatrix} + i_L \begin{smallmatrix} -90^\circ \\ \square \end{smallmatrix} + i_C \begin{smallmatrix} 90^\circ \\ \square \end{smallmatrix}$$

$$\therefore I = i_R - j i_L + j i_C = i_R - j(i_L - i_C)$$

$$\therefore I = \left[\sqrt{i_R^2 + (i_L - i_C)^2} \right] \begin{bmatrix} -\tan^{-1} \frac{(i_L - i_C)}{i_R} \\ \square \end{bmatrix}$$

~~$$\therefore I = \left[\sqrt{i_R^2 + (i_L - i_C)^2} \right] \begin{bmatrix} \theta \\ \square \end{bmatrix}$$~~

Where $\theta = -\tan^{-1} \left(\frac{i_L - i_C}{i_R} \right)$

If

$\theta \Rightarrow$ Negative

Resultant current lags the supply voltage

$\theta \Rightarrow$ Positive

Resultant current leads the supply voltage.

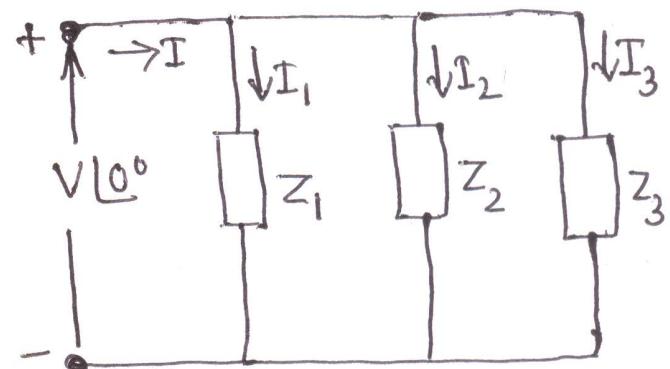
(36) ④

In general

$$I_1 = \frac{V}{Z_1} = Y_1 V$$

$$I_2 = \frac{V}{Z_2} = Y_2 V$$

$$I_3 = \frac{V}{Z_3} = Y_3 V$$



$$I = I_1 + I_2 + I_3 = \frac{V}{Z_1} + \frac{V}{Z_2} + \frac{V}{Z_3}$$

$$\therefore I = (Y_1 + Y_2 + Y_3) V$$

OR

$$I = Y V$$

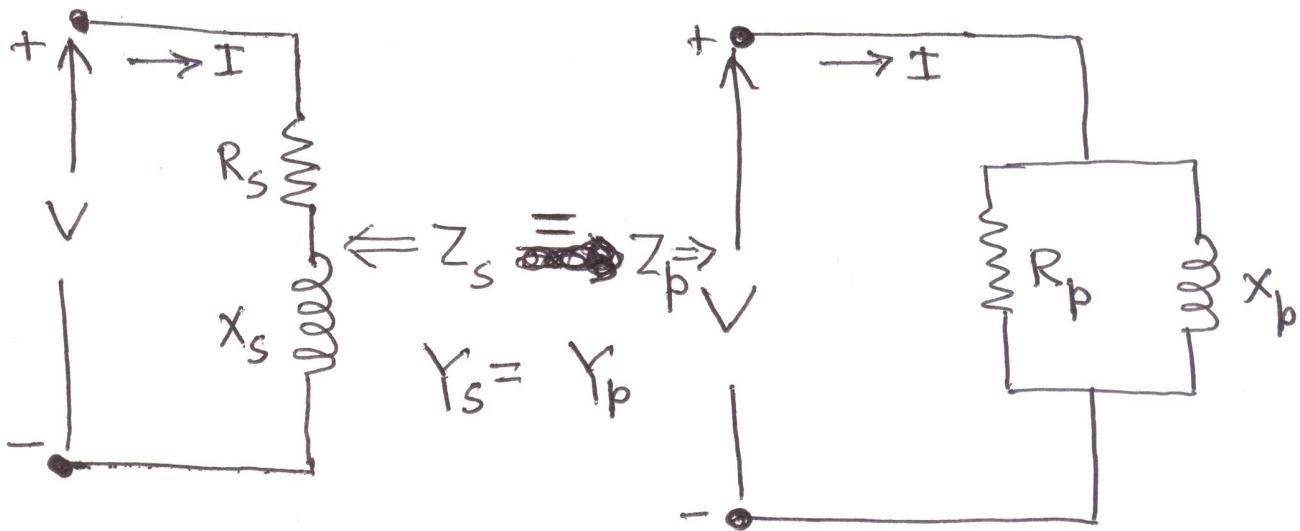
$Y \Rightarrow$ Reciprocal of Impedance
 \Rightarrow Admittance

Admittances are added for parallel branches

For branches in series, Impedances are added

(39) (20)

PARALLEL EQUIVALENT OF A SERIES IMPEDANCE



Equivalent Circuit

$$Y = \frac{1}{Z} = \frac{1}{R_s + jX_s} = \frac{1}{R_p} + \frac{1}{jX_p}$$

↓ ↓

Y_s Y_p

$$\therefore \frac{R_s}{R_s^2 + X_s^2} - j \frac{X_s}{R_s^2 + X_s^2} = \frac{1}{R_p} - j \frac{1}{X_p} = g - jb$$

$$\therefore g = \text{conductance} = \frac{1}{R_p} = \frac{R_s}{(R_s^2 + X_s^2)}$$

$$b = \text{Susceptance} = \frac{1}{X_p} = \frac{X_s}{(R_s^2 + X_s^2)}$$

Also

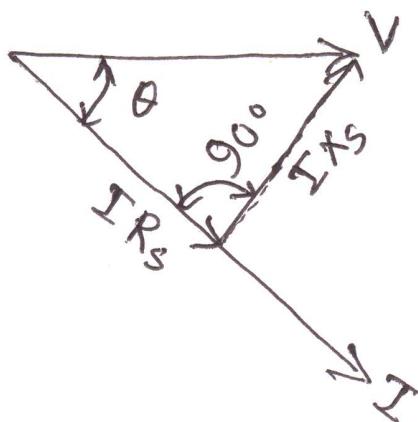
$$R_p = \frac{(R_s^2 + x_s^2)}{R_s}$$

$$x_p = \frac{(R_s^2 + x_s^2)}{x_s}$$

Now

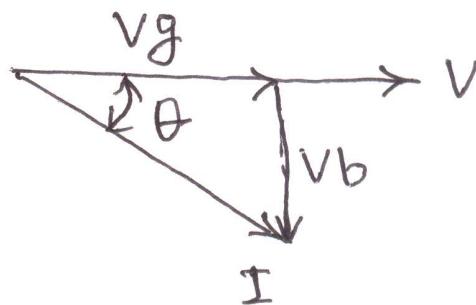
$$\frac{V}{I} = Z_s = R_s + j x_s$$

$$\therefore V = I R_s + j I x_s$$



$$\frac{I}{V} = Y_p = \frac{1}{R_p} - j \frac{1}{x_p} = g - j b$$

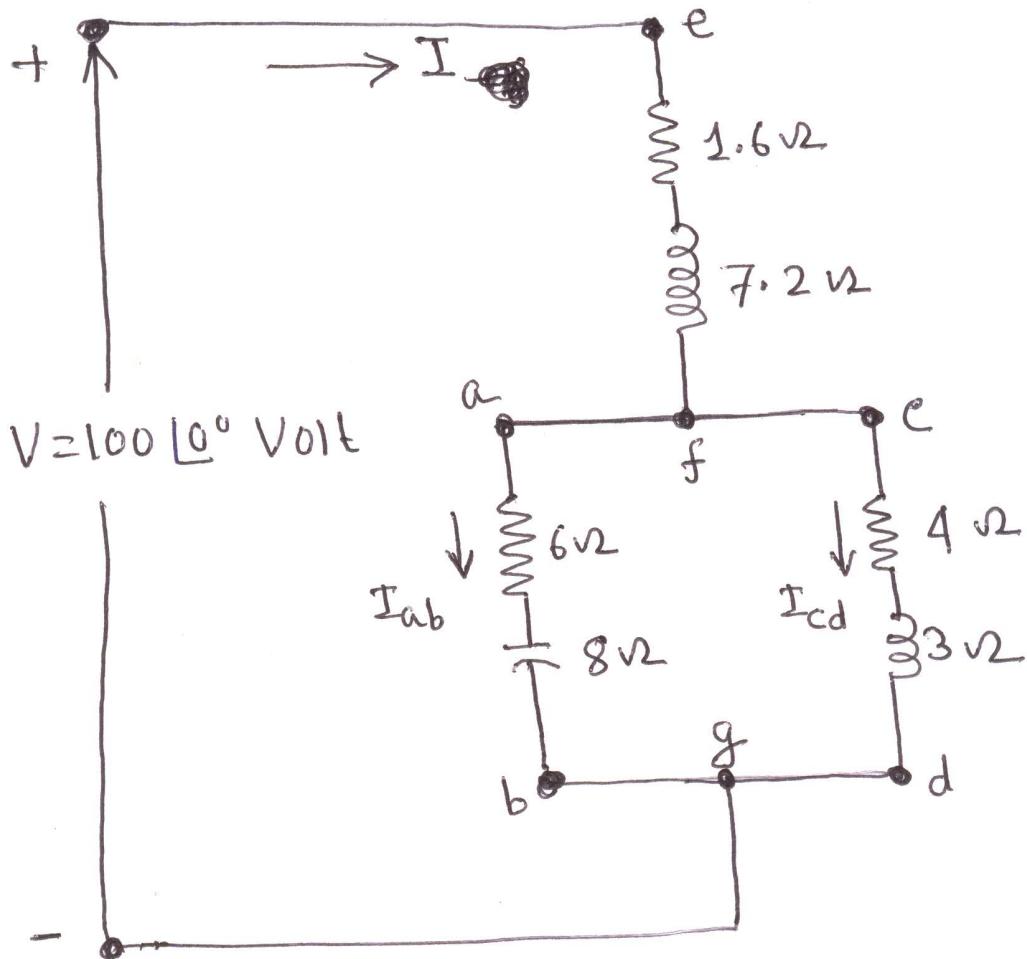
$$\therefore I = Vg - j Vb$$



SERIES-PARALLEL CIRCUIT

(24)
(A)

Ex-9



Determine I_{ef} , I_{ab} , I_{cd} , power and Power Factor.

Soln:

$$Y_{ab} = \frac{1}{Z_{ab}} = \frac{1}{6-j8} = (0.06 + j0.08) \text{ mho}$$

$$Y_{cd} = \frac{1}{4+j3} = (0.16 - j0.12) \text{ mho}$$

$$Y_{fg} = Y_{ab} + Y_{cd} = (0.22 - j0.04) \text{ mho}$$

$$Z_{fg} = \frac{1}{Y_{fg}} = \frac{1}{(0.22 - j0.04)} = (4.4 + j0.8) \Omega$$

(22) (23)

$$Z_{eg} = Z_{ef} + Z_{fg} = (1.6 + j7.2) + (4.4 + j0.8)$$

$$\therefore Z_{eg} = (6 + j8) \Omega$$

$$\therefore I = \frac{V}{Z_{eg}} = \frac{100 [0^\circ]}{6 + j8} = 10 [-53.2^\circ] \text{ Amp}$$

$$\text{Power Factor} = \cos \theta = \cos(53.2^\circ) = 0.60$$

$$P = V I \cos \theta = 100 \times 10 \cos(53.2^\circ)$$

$$\therefore P = 600 \text{ Watt}$$

$$V_{ef} = I \cdot Z_{ef} = I_{ef} Z_{ef} = (6 - j8)(1.6 + j7.2)$$

$$\therefore V_{ef} = 73.8 [24.4^\circ] \text{ Volt}$$

$$\therefore V_{fg} = V - V_{ef} = 100 [0^\circ] - 73.8 [24.4^\circ]$$

$$\therefore V_{fg} = 44.7 [-42.8^\circ] \text{ Volt}$$

$$I_{ab} = V_{ab} Y_{fg} = (0.06 + j0.08) \times 44.7 \angle -42.8^\circ$$

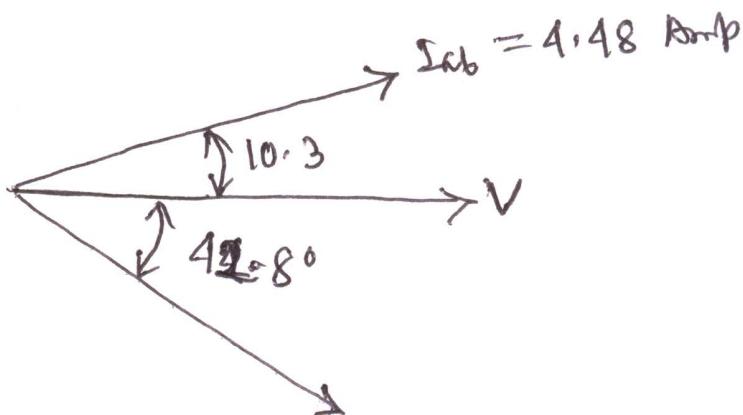
(43) (46)

$$\therefore I_{ab} = 4.48 \angle 10.3^\circ \text{ Amp}$$

$$I_{cd} = V_{fg} \cdot Y_{cd} = 44.7 \angle -42.8^\circ \times (0.16 - j0.12)$$

$$\therefore I_{cd} = 8.95 \angle -79.7^\circ \text{ Amp.}$$

$$P_{ab} = (I_{ab})^2 \cdot R_{ab} = (4.48)^2 \times 6 \approx 120 \text{ Watt.}$$



$$P_{ab} = 120$$

$$V_{fg} = V_{ab} = V_{cd} = 44.7 \text{ volt}$$

$$\begin{aligned}
 P_{ab} &= V_{ab} I_{ab} \cos \theta \\
 &= 44.7 \times 4.48 \cos(42.8^\circ + 10.3^\circ) \\
 &= 120 \text{ Watt.}
 \end{aligned}$$

(AA) (27)

Similarly

$$P_{cd} = (\underline{I}_{cd})^2 \cdot R_{cd} = (8.95)^2 \times 4 = 320 \text{ Watt.}$$

$$P_{ef} = (\underline{I}_{ef})^2 R_{ef}$$

$$= (\underline{I})^2 R_{ef} = (10)^2 \times 1.6 = 160 \text{ Watt.}$$

$$P = P_{ef} + P_{ab} + P_{cd} = (160 + 320 + 120)$$

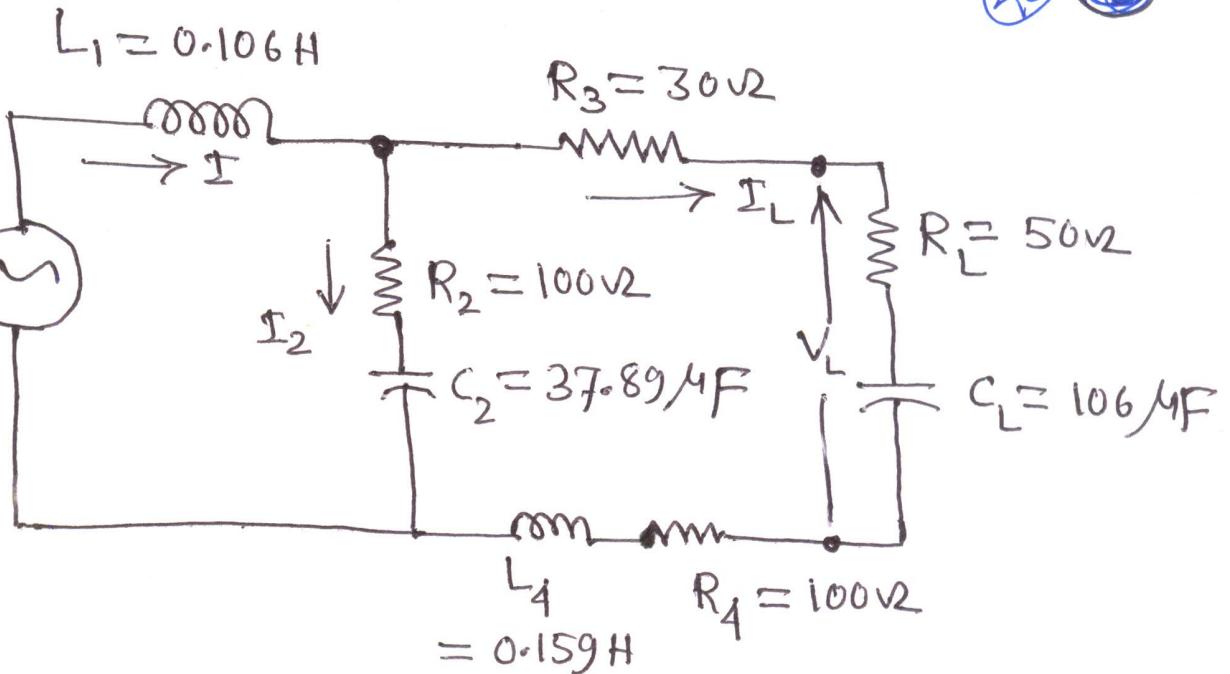
$P = 600 \text{ Watt}$

$$PF_{(ab)} = \frac{6}{\sqrt{6^2+8^2}} = 0.6 \text{ (leading)}$$

$$PF_{(cd)} = \frac{4}{\sqrt{4^2+3^2}} = 0.8 \text{ (lagging)}$$

DRAW PHASOR DIAGRAM.

120 V
60 Hz



Find

(i) I , I_L & V_L

(ii) Find active and reactive power delivered to the load

(iii) Find apparent, active and reactive power of the entire circuit.

Ans:

$$(i) I = 1.5426 \angle -12.614^\circ \text{ Amp}$$

$$I_L = 0.6672 \angle -40.48^\circ \text{ Amp}$$

$$V_L = 37.3 \angle -67.05^\circ \text{ Volt}$$

(46) (20)

(ii) $P_L = 22.25 \text{ Watt}$

$Q_L = -11.126 \text{ VAR} \text{ (capacitive)}$

(iii) Apparent Power = $\frac{185.112}{122.551} \text{ VA}$

$P = \cancel{100.4} \text{ Watt} \quad 180.64 \text{ Watt}$

$Q = \cancel{-40.27} \text{ VAR} \text{ (capacitive)}$

$40.425 \text{ VAR} \text{ (Inductive)}$