

Intro to Electronics (Practice Paper-1)

① At thermal equilibrium, conductivity of a semiconductor is given by - $\sigma = n q \mu_m + p q \mu_p$

$$= q \left(\frac{n^2}{p} \mu_m + p \mu_p \right)$$

For maximum resistivity (or min. conductivity).

$$\frac{d\sigma}{dp} = 0.$$

$$\therefore -\frac{n^2}{p^2} \mu_m + \mu_p = 0 \rightarrow p = n i \sqrt{\frac{\mu_m}{\mu_p}}$$

as, $\mu_m > \mu_p$, $p > n i$

i.e. the sample is slightly P-type.

② At, 300K, n_i of Si is $1.5 \times 10^{10} \text{ cm}^{-3}$

Electron mobility (μ_m) = $1350 \text{ cm}^2/\text{V}\cdot\text{s}$.

Hole μ_p = $480 \text{ cm}^2/\text{V}\cdot\text{s}$.

So, resistivity of intrinsic Silicon is, $\rho = \frac{1}{q n_i (\mu_m + \mu_p)}$

$$\rho = \frac{1}{1.6 \times 10^{-19} \times 1.5 \times 10^{10} (1350 + 480)} = [2.3 \times 10^5 \Omega \cdot \text{cm.}]$$

Doping
Assuming complete ionization,
 $n = N_D = 10^{16} \text{ cm}^{-3}$, m-type, $n \gg p$

Resistivity for this case, $\rho = \frac{1}{n q \mu_m} = 0.462 \Omega \cdot \text{cm.}$

So, by adding small amount of impurity (< 1 ppm), resistivity of Si can be changed by many orders.

(3) Resistivity (ρ) = $0.65 \Omega\text{-cm}$
 mobility (H_m) = $1250 \text{ cm}^2/\text{V}\cdot\text{s}$.
 $\rho = \frac{1}{\sigma} = \frac{1}{m q H_m} = 0.65$; $\sigma = \text{conductivity}$
 $m = N_D = \frac{1}{0.65 \times 1.6 \times 10^{-19} \times 1250} \quad ; \quad N_D = \text{dopant concentration.}$

$$N_D = 7.69 \times 10^{15} \text{ cm}^{-3}$$

Drift current density (J) = $m q H_m E$; $E = \text{electric field.}$
 $160 = \frac{1}{e} E$
 $E = 160 \times 0.65 = 104 \text{ V/cm.}$

(4) GaAs P-n junction, $N_A = 5 \times 10^{18} \text{ cm}^{-3}$
 $N_D = 5 \times 10^{16} \text{ cm}^{-3}$ Given, N_A of GaAs
 $= 1.8 \times 10^6 \text{ cm}^{-3}$ at 300K

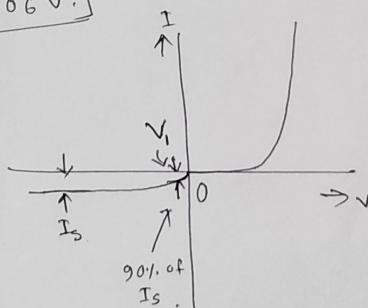
Built-in potential, $V_{bi} = V_T \ln \left(\frac{N_A N_D}{n_i^2} \right) \quad ; \quad V_T = 0.026 \text{ V at } 300\text{K}$
 $V_{bi} = 0.026 \ln \left[\frac{5 \times 10^{18} \times 5 \times 10^{16}}{(1.8 \times 10^6)^2} \right] = 1.37 \text{ V.} = V_{bi}$

(5) For diode, $I = I_s [\exp(\frac{V}{V_T}) - 1]$; assuming $\eta = 1$.
 In reverse bias current is negative and at that particular voltage
 current = $-0.9 I_s$; (90% of reverse saturation current)
 From diode equation, $-0.9 I_s = I_s [\exp(\frac{V_1}{V_T}) - 1]$

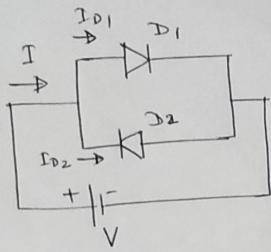
$$\therefore \exp\left(\frac{V_1}{V_T}\right) = 0.1$$

$$V_1 = V_T \ln(0.1)$$

$$V_1 = -0.06 \text{ V.}$$



⑥



Here, D_1 is forward biased
 D_2 is reverse biased.

$$\text{Total current } I = I_1 + I_2$$

$$I = I_s \left[\exp\left(\frac{V}{V_T}\right) - 1 \right] + I_s$$

$$I = I_s \exp\left(\frac{V}{V_T}\right)$$

$$V = V_T \ln\left(\frac{I}{I_s}\right)$$

⑦

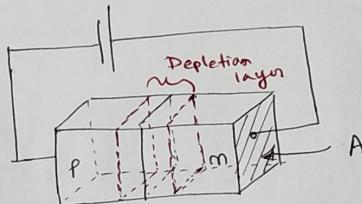
$$C = \frac{\epsilon A}{w}$$

$$C = \frac{\epsilon_0 \epsilon_r}{w} A$$

$$= \frac{11.7 \times 8.854 \times 10^{-12}}{10 \times 10^{-6}}$$

; A = unit area (m^2)

$$= 10.3 \mu\text{F}$$



Introduction to

Electronics (Practice Paper-1)

Given, $V_Z = 6V$.

The Zener diode is in reverse bias.

Voltage across the Zener diode is

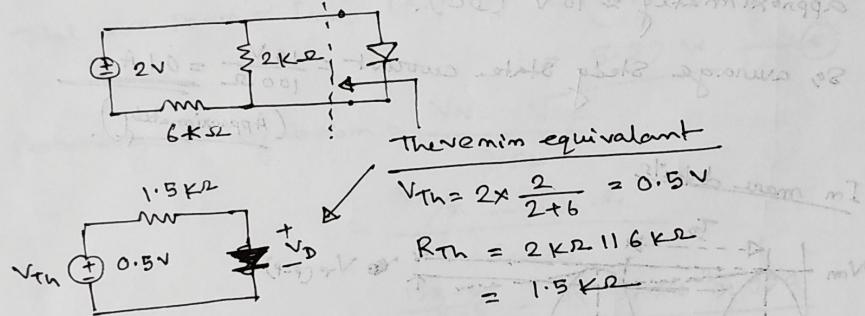
$$= \frac{1k\Omega}{1k\Omega + 1k\Omega} \times 10V = 5V$$

$$= 5V, \text{ but, } V_Z = 6V.$$

So, the Zener diode is in reverse bias, but not in breakdown.

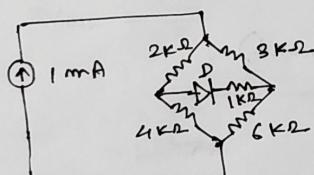
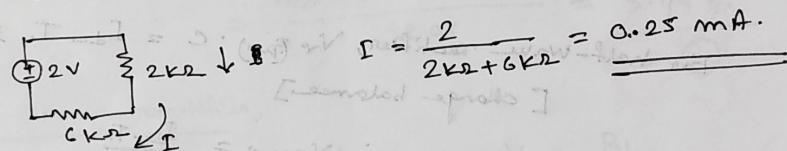
$$\text{So, } V_o = 5V.$$

Let us redraw the circuit as



But, as $V_{Th} = 0.5V$, the diode is OFF.

The circuit reduces to,



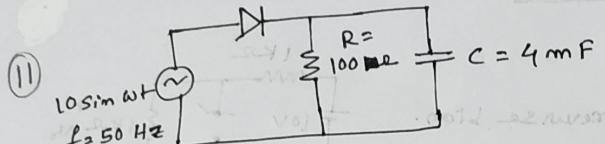
This is a bridge circuit
and the cross arm product is same
i.e. $2 \times 6 = 3 \times 4$

So, the bridge is balanced.

So, no current through the $1k\Omega$ resistor.

$$\text{Current through } 4k\Omega \text{ resistor} = 1 \text{ mA} \times \frac{9}{9+6}$$

$$= 0.6 \text{ mA.}$$



Input signal = $10 \sin \omega t$

Half-wave Rectification.

Frequency = 50 Hz

$$\text{RC time const.} = RC = 100 \times 4 \times 10^{-3} = 400 \text{ mSec.} \gg T_{\text{cycle}} = 20 \text{ mSec}$$

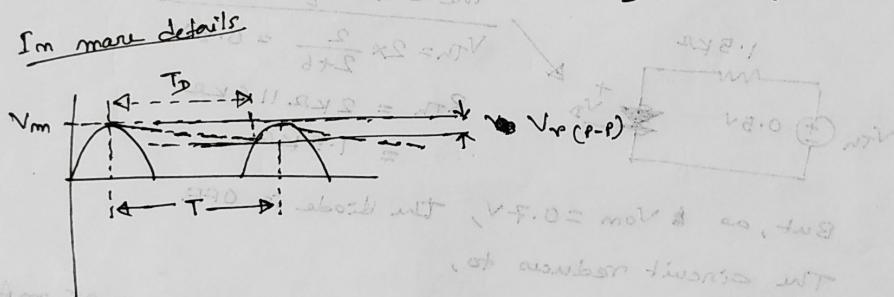
$\therefore V_o = 10 \text{ V}$ (DC)

So, we can say that voltage across the resistor is approximately $\approx 10 \text{ V}$ (DC).

$$\text{So, average steady state current} = \frac{10 \text{ V}}{100 \Omega} = 0.1 \text{ A}$$

(Approximately).

In more details



For half-wave rectifier, $V_r (P-P) \cdot C = I_{dc} \cdot T$

[charge balance]

$$\therefore V_r (P-P) = \frac{I_{dc} \cdot T}{C}$$

$$V_{dc} = V_m - \frac{V_r (P-P)}{2} \quad (\text{approximated})$$

$$\therefore I_{dc} \cdot R = V_m - \frac{I_{dc} \cdot T}{C}$$

$$\therefore I_{dc} \left(R + \frac{1}{2fC} \right) = V_m$$

$$\therefore I_{dc} = \frac{10}{100 + \frac{1}{2 \times 50 \times 4 \times 10^{-3}}} \approx 0.0976 \text{ A}$$

$\therefore I_{dc} = 0.0976 \text{ A}$

Supply, $V_{rms} = 110 \text{ V}$, $V_m = 110 \times \sqrt{2} = 155.56 \text{ V}$

$$(12) \quad \therefore I_{rm} = \frac{V_m}{R_L + R_f}; \quad R_f = 30 \Omega; \quad R_L = 990 \Omega$$

$$= \frac{155.56}{990 + 30} = 0.152 \text{ A} \quad \text{peak load current.} \quad (a)$$

$$I_{dc} = \frac{I_{rm}}{\pi} = \frac{0.152}{\pi} = 48.57 \text{ mA.} \quad (b)$$

$$I_{rms} = \frac{I_{rm}}{2} = \frac{0.152}{2} \text{ A} = 76 \text{ mA.} \quad (c)$$

$$V_{dc} = I_{dc} \cdot R_L = 48.57 \times 990 = 48 \text{ mV} \quad (d)$$

$$\text{Total S.I.P power} = I_{rms}^2 (R_L + R_f) = (76 \times 10^{-3})^2 \times 1020$$

$$\text{Total S.I.P power comes to } = 5.89 \text{ W.} \quad (e)$$

Percentage of regulation = $\frac{V_{NL} - V_{FL}}{V_{FL}} \times 100\%$

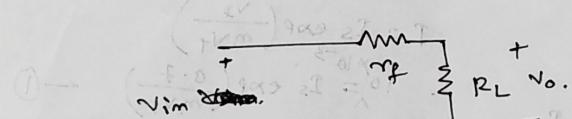
$$= \frac{V_m/\pi - I_{dc} \cdot R_L}{I_{dc} \cdot R_L}$$

$$= \frac{155.56}{3.14} - 48.57 \times 10^{-3} \times 990$$

$$= \frac{49.54 - 48.08}{48.08} \approx 3.04\% \quad (f)$$

For, half-wave rectifier,

$$(13) \quad V_{NL} = \frac{V_m}{\pi} \quad \text{and} \quad V_{FL} = \frac{V_m}{\pi} \times \frac{R_L}{R_L + R_f}$$



$$\therefore \text{percentage of regulation} = \frac{V_{NL} - V_{FL}}{V_{FL}} \times 100\% = \frac{\frac{V_m}{\pi} - \frac{V_m}{\pi} \times \frac{R_L}{R_L + R_f}}{\frac{V_m}{\pi} \times \frac{R_L}{R_L + R_f}} \times 100\%.$$

$$= \frac{V_m}{\pi} \left(\frac{R_f}{R_L + R_f} \right) \times 100\% = \frac{R_f}{R_L} \times 100\%$$

$$= 0.03 \times \left(\frac{30}{990} \right) \times 100\% = 0.009\%$$

For full-wave rectifier, V_{NL} and V_{FL} are double compared to that of half-wave rectifier.

$$\text{Hence, \% of regulation (full-wave)} = \frac{R_f}{R_L} \times 100\%.$$

$$(14) \quad I = I_s \left[\exp\left(\frac{V_D}{mV_T}\right) - 1 \right]$$

$V_{TH} = mV_T$

since, $\exp\left(\frac{V_D}{mV_T}\right) \gg 1$; $\therefore I = I_s \exp\left(\frac{V_D}{mV_T}\right)$

$$\text{So, } \frac{15 \text{ mA}}{0.5 \text{ mA}} = \frac{I_s \exp\left(\frac{0.465}{m \times 0.025}\right)}{I_s \exp\left(\frac{0.340}{m \times 0.025}\right)} = \exp\left(\frac{0.125}{m}\right)$$

$$\therefore 30 = \exp\left(\frac{0.125}{m}\right)$$

$$\therefore \frac{0.125}{m} = 2.303 \log_{10}(30)$$

$$\therefore m = 1.47$$

(15) Three diodes are identical.
so, drop across each diode $= \frac{V_0}{3} = \frac{2}{3} \text{ V}$.



$$I = I_s \exp\left(\frac{V}{V_T}\right) = 10^{-14} \exp\left(\frac{2/3}{0.026}\right)$$

$$\therefore I = 1.366 \text{ mA.} \quad \text{--- (a)}$$

1 mA current is drawn by a load.

$$\therefore \frac{1.366 \text{ mA}}{0.366 \text{ mA}} = \frac{I_s \exp\left(\frac{2/3}{0.026}\right)}{I_s \left[\exp\left(\frac{V'}{0.026}\right) \right]} = \exp\left(\frac{0.66 - V'}{0.026}\right)$$

$$\therefore 1.317 = \frac{0.66 - V'}{0.026}$$

$$\therefore \text{change in o/p vol. } (0.66 - V') = 0.034 \text{ V} = 34 \text{ mV} \quad \text{decrease} \quad (b)$$

$$(16) \quad \begin{array}{l} \text{Circuit diagram: A current source } 10 \text{ mA is connected to a } \\ \text{load resistor } R \text{ through two diodes } D_1 \text{ and } D_2. \text{ Diode } D_1 \text{ is in series with the load, and } \\ \text{diode } D_2 \text{ is in parallel with the load. The output voltage } V_0 \text{ is measured across the load.} \\ \text{Currents } I_1 \text{ and } I_2 \text{ flow through } D_1 \text{ and } D_2 \text{ respectively.} \end{array}$$

$$I = I_s \exp\left(\frac{V_D}{mV_T}\right)$$

$$\therefore 10 \times 10^{-3} = I_s \exp\left(\frac{0.7}{mV_T}\right) \quad \text{--- (1)}$$

$$100 \times 10^{-3} = I_s \exp\left(\frac{0.8}{mV_T}\right) \quad \text{--- (2)}$$

$$\therefore \text{From (1) \& (2),} \\ m = 1.73$$

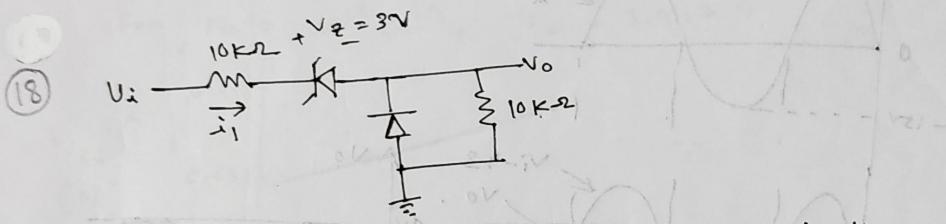
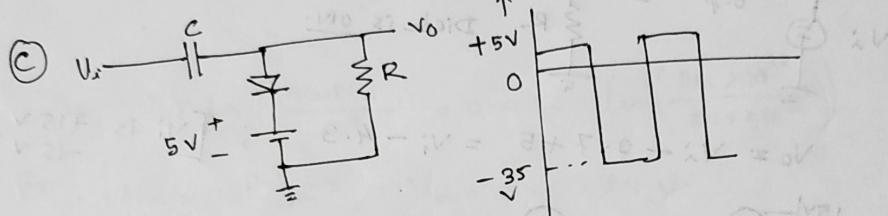
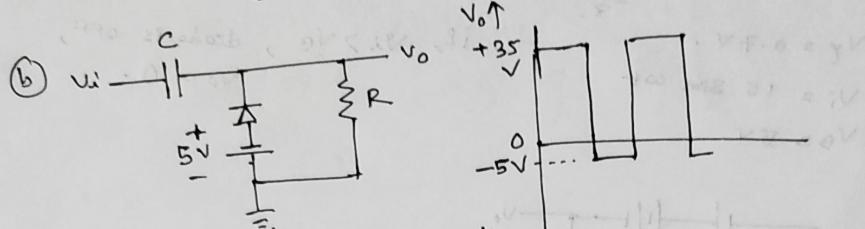
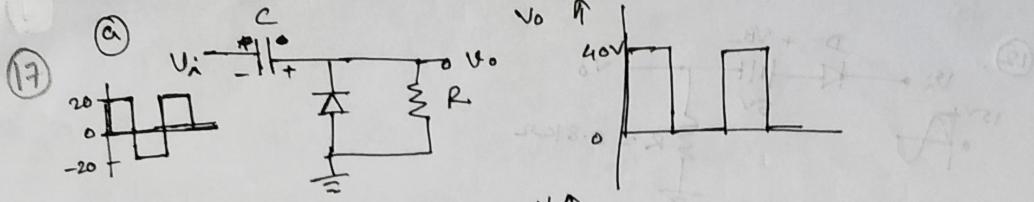
$$V_0 = V_2 - V_1 = mV_T \ln\left(\frac{I_2}{I_1}\right) = 80 \text{ mV.}$$

$$\therefore 1.73 \times 0.025 \times \ln\left(\frac{I_2}{I_1}\right) = 80 \times 10^{-3}$$

$$I_1 + I_2 = 10 \text{ mA.} = 0.01 \text{ A.}$$

$$\therefore I_1 \approx 1.4 \text{ mA}$$

$$\therefore R = \frac{80}{I_1} = 57.1 \Omega$$



For $-10 \leq V_i \leq 0$ V, both diodes are conducting.
 $\therefore V_o = 0$ V; ($\text{as } V_Z = 0$)

For $0 < V_i \leq 3$ V, Zener diode is not in breakdown.
 $\therefore V_o = 0$ V. (all voltage drop occurs across Zener and $i = 0$)

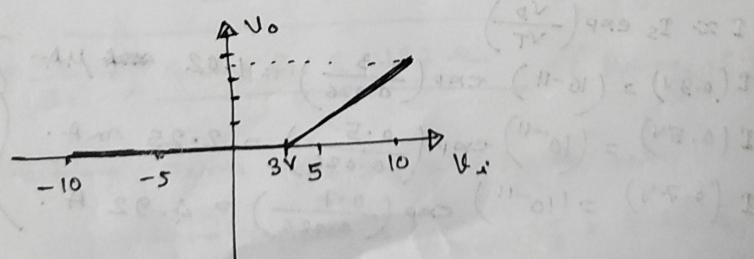
$$\therefore V_o = 0$$
 V.

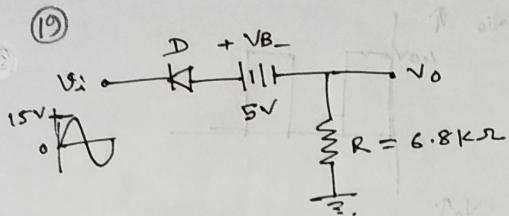
For $V_i > 3$ V, Zener is in breakdown and drop across it is $= V_Z = 3$ V.

$$\therefore i_1 = \frac{V_i - V_Z}{20K\Omega} = \frac{V_i - 3}{20} \text{ mA.}$$

$$\therefore V_o = 10i_1 = \frac{V_i - 3}{2} \text{ V.}$$

$$\text{At, } V_i = 10 \text{ V, } V_o = 3.5 \text{ V.}$$





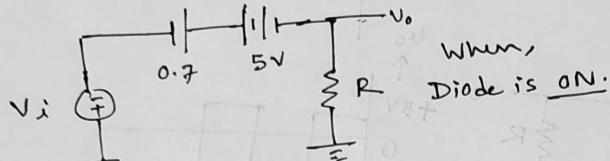
$$V_T = 0.7 \text{ V}.$$

$$V_i = 15 \sin \omega t$$

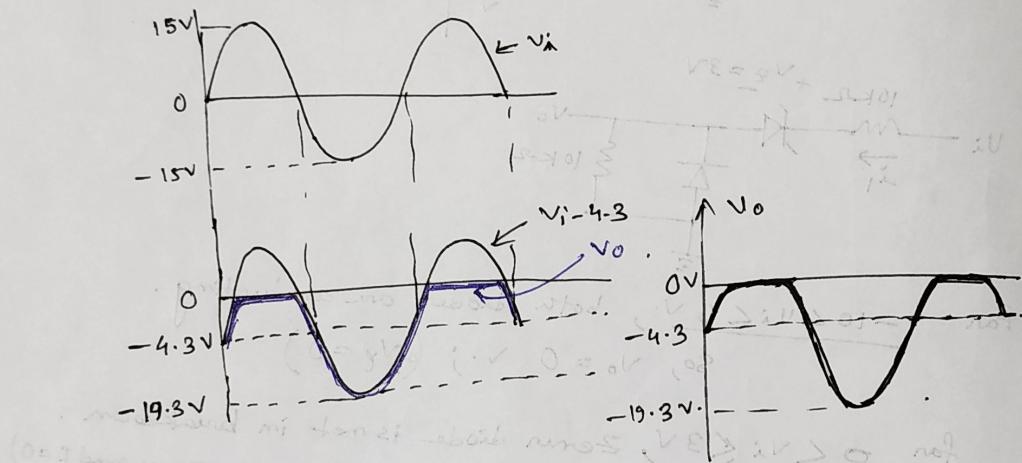
$$V_B = 5 \text{ V}.$$

if, $V_i > V_B$, diode is off,

$$V_o = 0.$$



$$V_o = V_i - 0.7 + 5 = V_i - 4.3 ; [V_i \text{ is } +15 \text{ V} \text{ or } -15 \text{ V}]$$



(Output voltage remains constant when input voltage is zero.)

(20) ~~$I = I_s \exp\left(\frac{V_D}{V_T}\right)$~~ ; $n=1$)

$$\therefore 10I = I_s \exp\left(\frac{V_D + \Delta V}{V_T}\right) ; V_o = 0 \text{ V}$$

$$\therefore \frac{10I}{I} = \frac{\exp\left(\frac{V_D + \Delta V}{V_T}\right)}{\exp\left(\frac{V_D}{V_T}\right)} = \exp\left(\frac{\Delta V}{nV_T}\right) ; V_B = 5 \text{ V}$$

$$\therefore \Delta V = V_T \ln 10 = 59.86 \text{ mV} \approx 60 \text{ mV.} \rightarrow 10 \text{ times}$$

$$\therefore \Delta V = V_T \ln 100 = 119.73 \text{ mV} \approx 120 \text{ mV.} \rightarrow 100 \text{ times.}$$

(21) In forward bias, $V_B = 5 \text{ V}$ and $V_o = 0 \text{ V}$.

$$I \approx I_s \exp\left(\frac{V_D}{V_T}\right)$$

$$I(0.3 \text{ V}) = (10^{-11}) \exp\left(\frac{0.3}{0.026}\right) = 1.02 \text{ mA.}$$

$$I(0.5 \text{ V}) = (10^{-11}) \exp\left(\frac{0.5}{0.026}\right) = 2.25 \text{ mA.}$$

$$I(0.7 \text{ V}) = (10^{-11}) \exp\left(\frac{0.7}{0.026}\right) = 4.92 \text{ A.}$$

See the
change in
current.

In reverse bias..

$$I = I_s \left[\exp\left(\frac{V_D}{V_T}\right) - 1 \right]$$

$$\text{So, } I(-0.2 \text{ V}) = I_s \left[\exp\left(-\frac{0.2}{0.026}\right) - 1 \right] ; I_s = 10^{-11} \text{ A.}$$

$$I(-0.2 \text{ V}) \approx -0.999 \times 10^{-11} \text{ A} \approx -I_s.$$

$$I(-2 \text{ V}) = 10^{-11} \left[\exp\left(-\frac{2}{0.026}\right) - 1 \right]$$

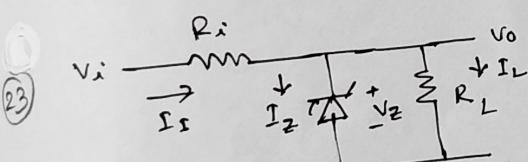
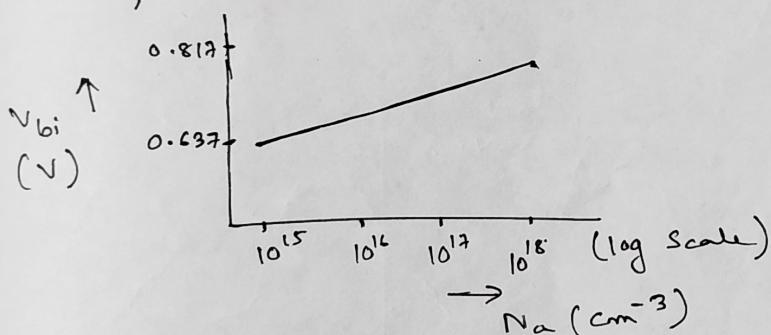
$$I(-2 \text{ V}) \approx -10^{-11} \approx -I_s.$$

[Do for $I_s = 10^{-13} \text{ A.}$]

$$V_{bi} = V_T \ln \left(\frac{N_a N_d}{m^2} \right) = 0.026 \ln \left(\frac{N_a \times 10^{16}}{2.25 \times 10^{20}} \right)$$

(22) For, $N_a = 10^{15} \text{ cm}^{-3}$, $V_{bi} = 0.6374 \text{ V}$.

For, $N_a = 10^{18} \text{ cm}^{-3}$, $V_{bi} = 0.817 \text{ V}$.



$$V_i = 6.3 \text{ V}, V_z = 4.8 \text{ V}, R_z = 12 \Omega$$

$$\therefore I_z = \frac{6.3 - 4.8}{12} = 125 \text{ mA.}$$

$$I_L = I_z - I_z, \quad 5 \text{ mA} \leq I_L \leq 100 \text{ mA.}$$

$$\therefore 25 \text{ mA} \leq I_L \leq 120 \text{ mA} \quad \left. \begin{array}{l} (\text{Range of } I_L) \\ (\text{Range of } R_L) \end{array} \right\} \text{a}$$

$$R_L = \frac{V_z}{I_L} ; 40 \Omega \leq R_L \leq 192 \Omega \quad \left. \begin{array}{l} \\ (\text{Range of } R_L) \end{array} \right\} \text{b}$$

Power rating, $Z_{max} = I_{z, max} \cdot V_z = 480 \text{ mW.}$

Load resistor $= I_{L, max} \cdot V_z = 576 \text{ mW.}$