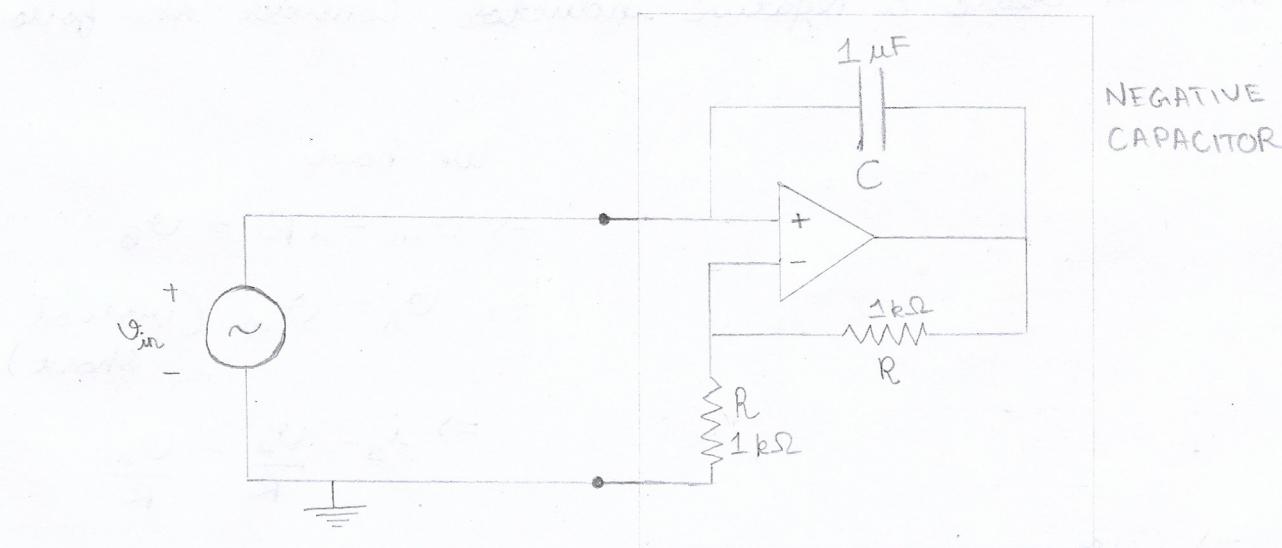


EXPERIMENT 5

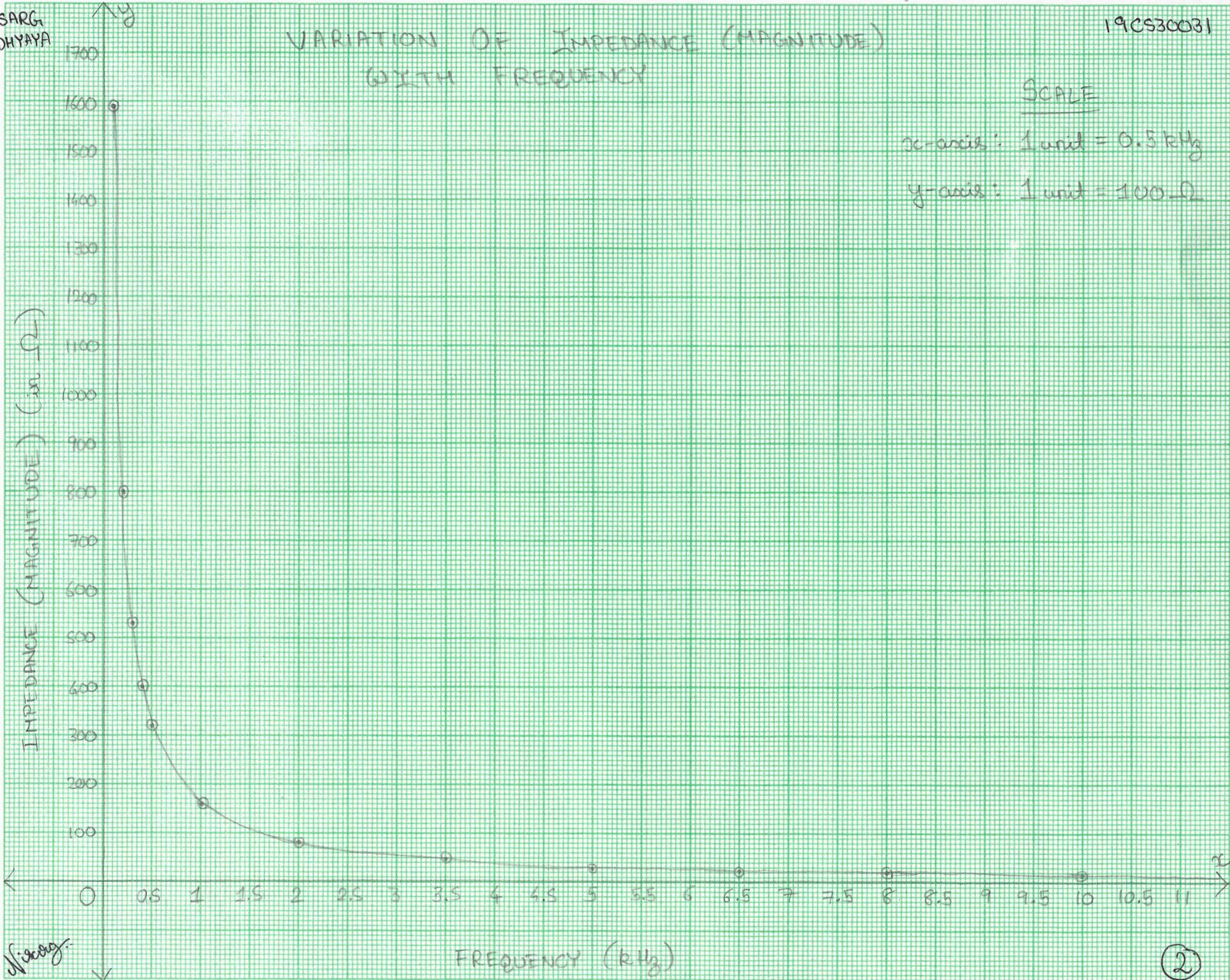
① NEGATIVE IMPEDANCE CONVERTOR

OBSERVATION TABLE ($C = 1 \mu F$)

Frequency (Hz)	Applied voltage magnitude (V)	Current magnitude (A)	Phase of current w.r.t the voltage (degrees)	Calculated complex impedance (Ω)
100	5	3.14×10^3	-90	$1592.36j$
200	5	6.28×10^3	-90	$796.18j$
300	5	9.42×10^3	-90	$530.79j$
400	5	1.26×10^4	-90	$396.83j$
500	5	1.57×10^4	-90	$318.47j$
1000	5	3.14×10^2	-90	$159.24j$
2000	5	6.28×10^2	-90	$79.62j$
3500	5	1.10×10^3	-90	$45.45j$
5000	5	1.57×10^3	-90	$31.85j$
6500	5	2.04×10^3	-90	$24.51j$
8000	5	2.51×10^3	-90	$19.92j$
10000	5	3.14×10^3	-90	$15.92j$

Nisarg:-

(1)

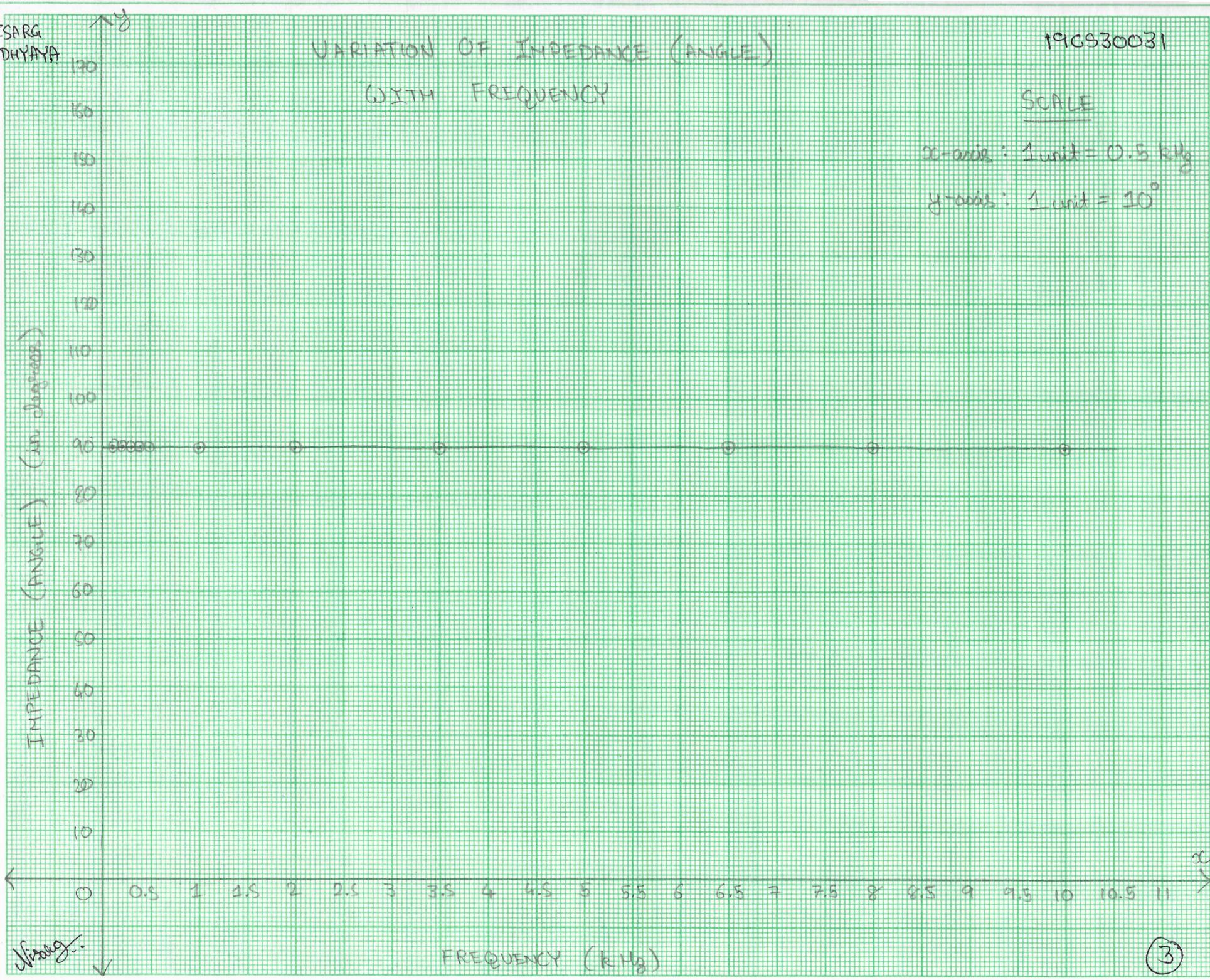
VARIATION OF IMPEDANCE (MAGNITUDE)
WITH FREQUENCY

VARIATION OF IMPEDANCE (ANGLE)

WITH FREQUENCY

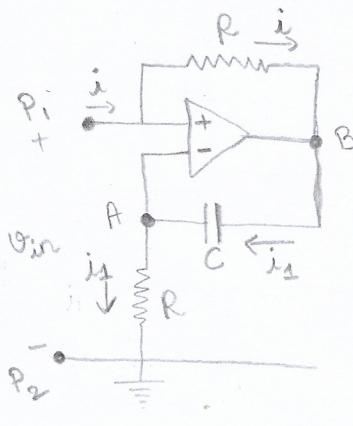
SCALE

X-axis : 1 unit = 0.5 kHz

Y-axis : 1 unit = 10^9 

It is indeed possible to create an inductor using a capacitor using this negative impedance converter with some slight modifications.

First, we will create a negative inductor. Consider the following circuit:



We have

$$\Rightarrow V_{in} - iR = V_B$$

$$\Rightarrow V_A = V_{in} \text{ (virtual short)}$$

$$\Rightarrow i_1 = \frac{V_A}{R} = \frac{V_{in}}{R}$$

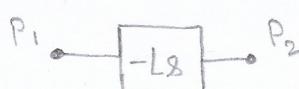
$$\Rightarrow V_A + \frac{i_1}{Cs} = V_B$$

$$\Rightarrow V_{in} + \frac{V_{in}}{Rcs} = V_{in} - iR$$

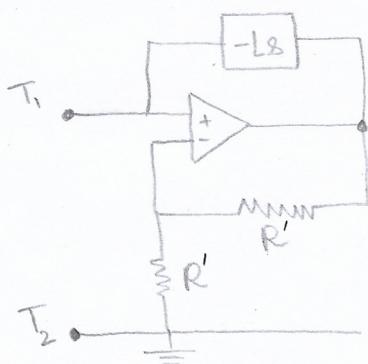
$$\Rightarrow \frac{V_{in}}{i} = Z_{in} = -\underline{\underline{R^2 Cs}}$$

If we take $R^2 C = L$
then $Z_{in} = -L_s$
which is nothing
but a negative
inductor.

We can assume the above to be a two terminal network



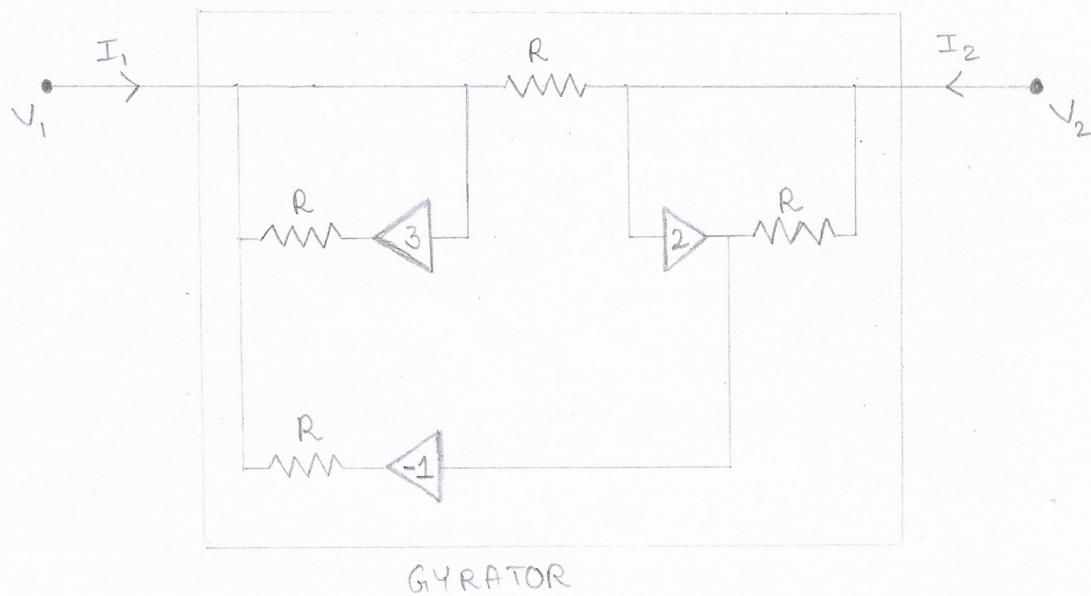
Now using our initial negative impedance converter we can change this to a positive inductor.



Now, the following network between T_1 & T_2 behaves as a positive inductor with inductance

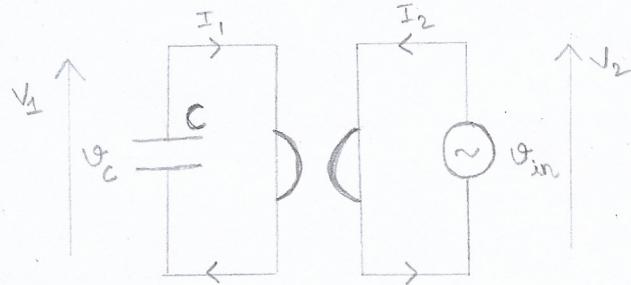
$$\underline{\underline{L = R^2 C}}$$

② GYRATOR



SIMULATING INDUCTOR USING A CAPACITOR

GYRATION RESISTANCE
= R



For a gyrator we can write $V_2 = RI_1$ and $V_1 = -RI_2$

Now $V_2 = V_{in}$ and $V_1 = V_c = \frac{-I_1}{Cs}$

Hence $V_1 = -RI_2 = \frac{-I_1}{Cs} \Rightarrow I_1 = Rcs I_2$

$$\Rightarrow V_2 = V_{in} = RI_1 = R^2 C s I_2$$

$$\Rightarrow Z_{in} = \frac{V_{in}}{I_2} = R^2 C s$$

If we take $R^2 C = L$ then Z_{in} can be treated as an inductor, whose inductance $L = R^2 C$.

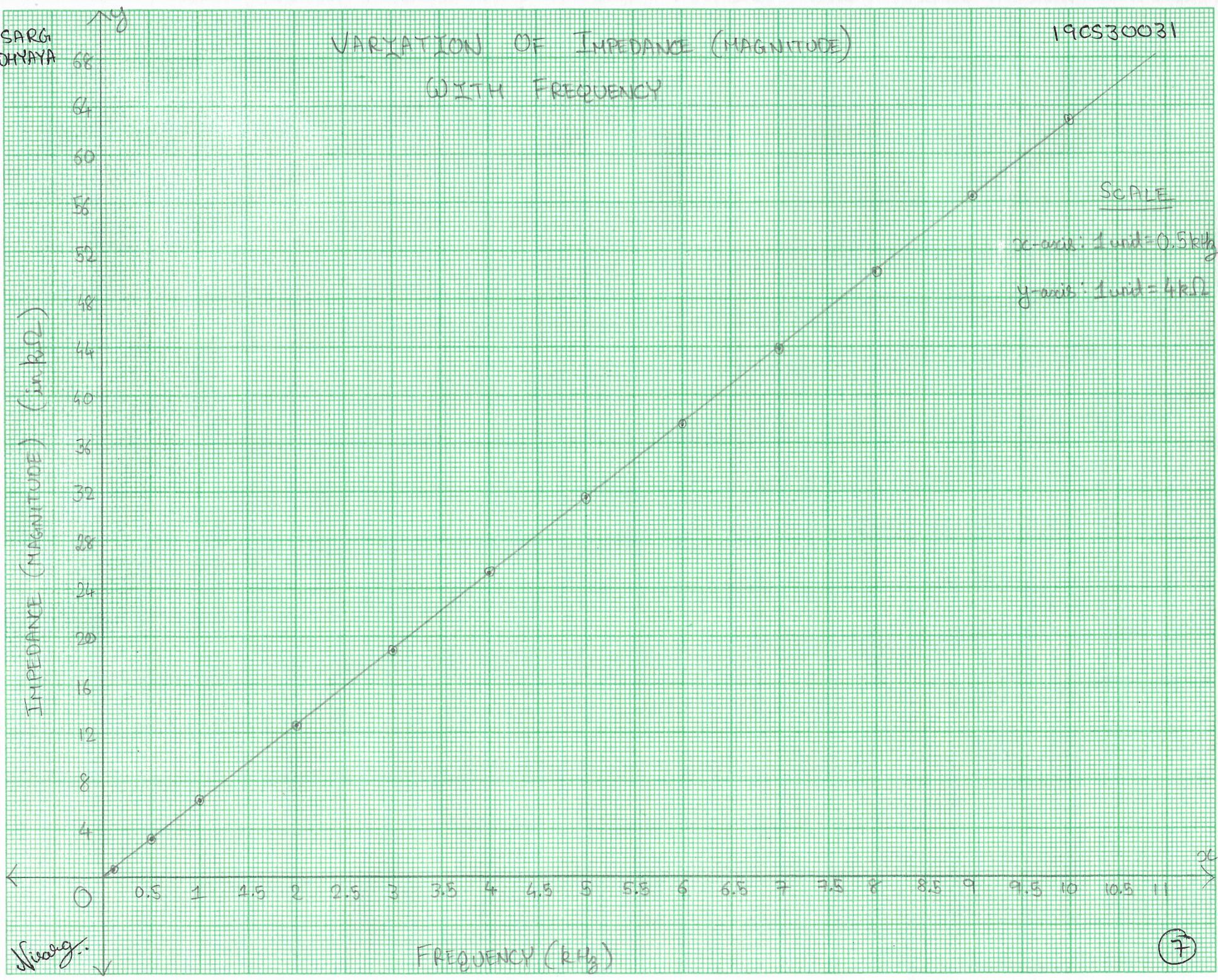
OBSERVATION TABLE ($C = 1 \mu F$, $R = 1 k\Omega$)

Frequency (Hz)	Applied voltage magnitude (V)	Current magnitude (A)	Phase of current w.r.t the voltage (degrees)	Calculated complex impedance (Ω)
100	5	7.96×10^{-3}	-90	$628.14 j$
500	5	1.59×10^{-3}	-90	$3144.65 j$
1000	5	7.96×10^{-4}	-90	$6281.41 j$
2000	5	3.98×10^{-4}	-90	$12562.81 j$
3000	5	2.65×10^{-4}	-90	$18867.92 j$
4000	5	1.99×10^{-4}	-90	$25125.63 j$
5000	5	1.59×10^{-4}	-90	$31446.54 j$
6000	5	1.33×10^{-4}	-90	$37593.98 j$
7000	5	1.14×10^{-4}	-90	$43859.65 j$
8000	5	9.95×10^{-5}	-90	$50251.26 j$
9000	5	8.84×10^{-5}	-90	$56561.09 j$
10000	5	7.96×10^{-5}	-90	$62814.07 j$

(6)

VARIATION OF IMPEDANCE (MAGNITUDE)

(1) WITH FREQUENCY



VARIATION OF IMPEDANCE (ANGLE)

(W)ITH FREQUENCY

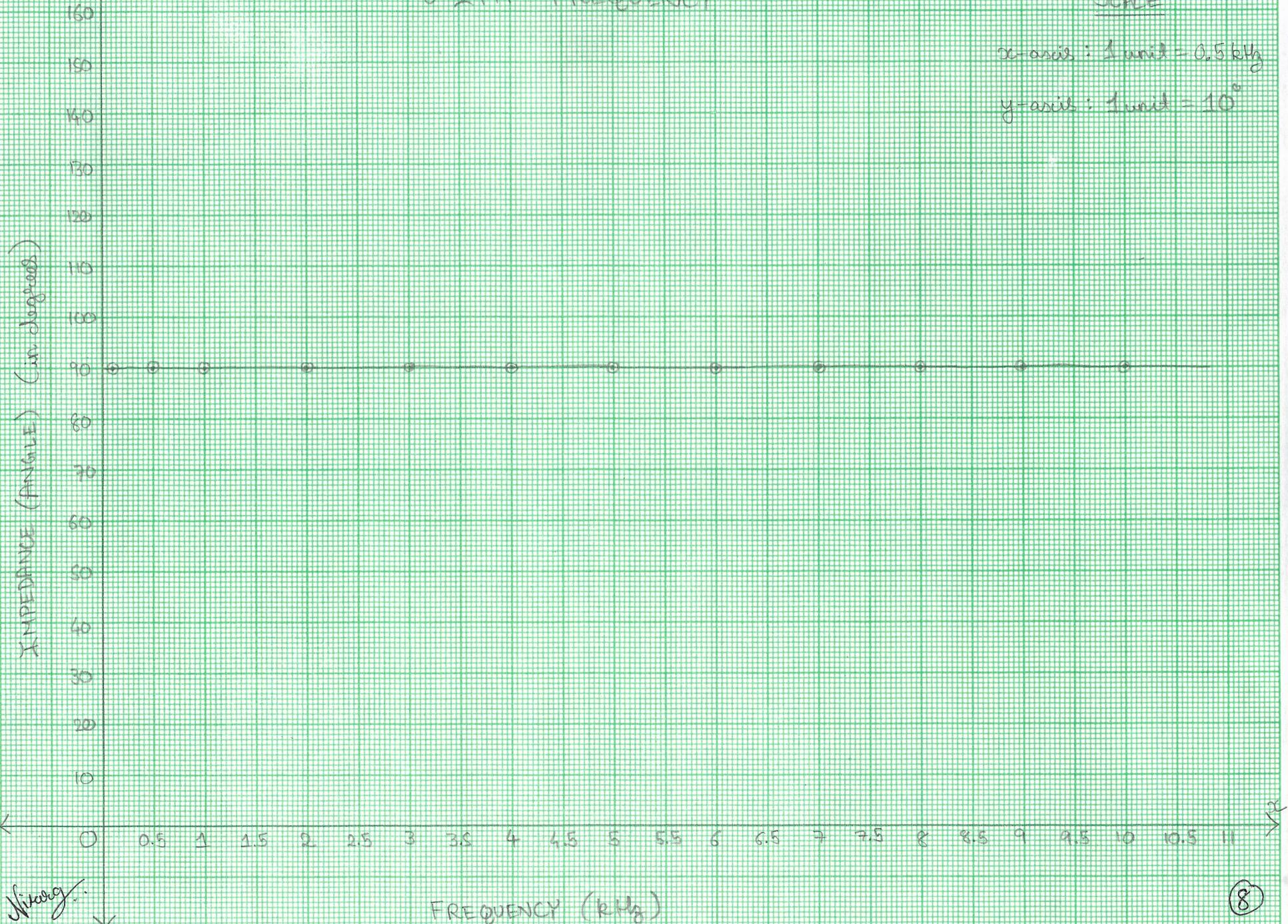
SCALE

x-axis : 1 unit = 0.5 kHz

y-axis : 1 unit = 10°

IMPEDANCE (ANGLE) (in degrees)

Wavelength

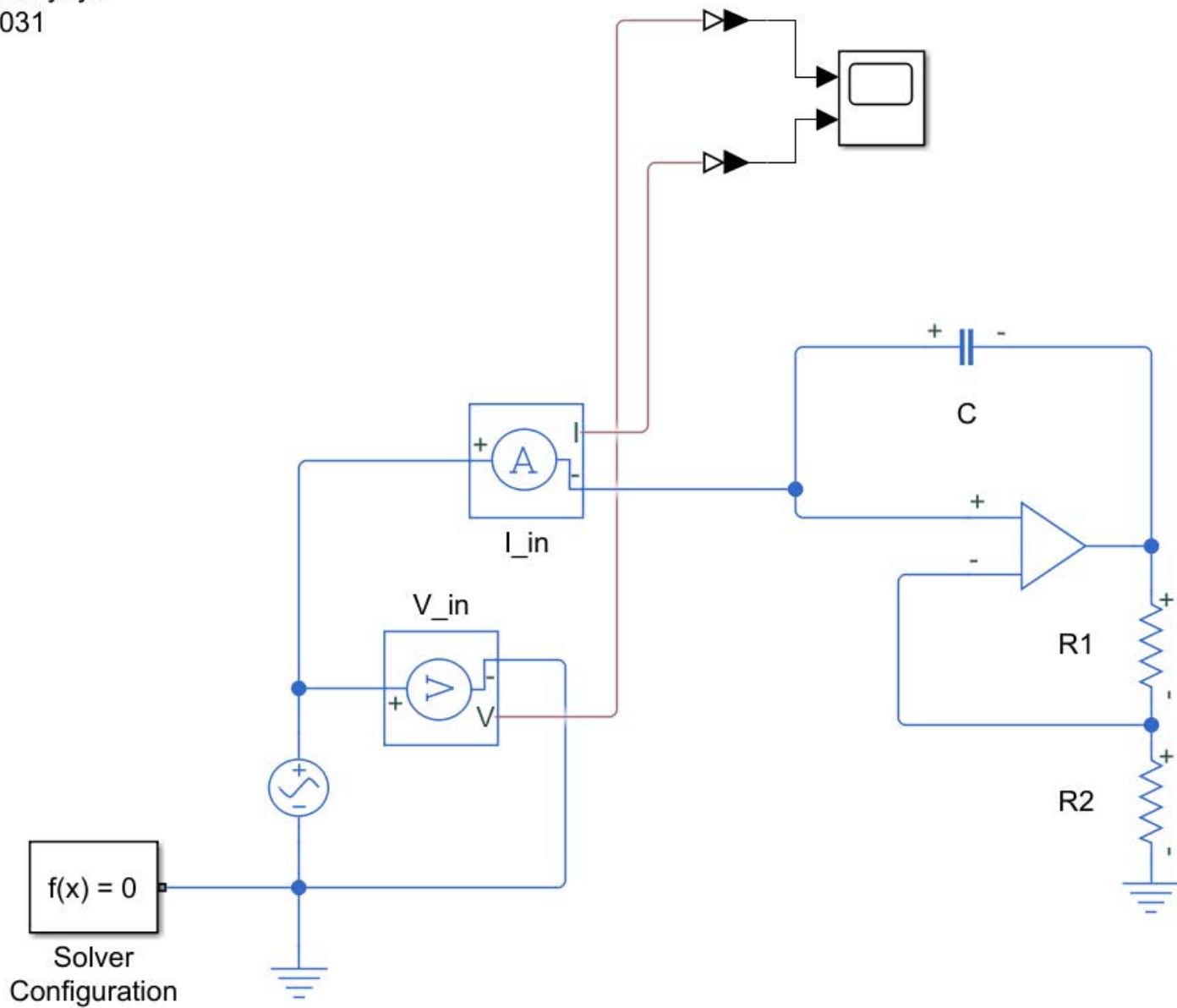


FREQUENCY (kHz)

(8)

Name: Nisarg Upadhyaya

Roll No: 19CS30031



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