

① Early Voltage (V_A) = 80V.

$I_C = 20.60 \text{ mA}$ at $V_{CE} = 2V$.

$$I_C = I_C' \left(1 + \frac{V_{CE}}{V_A}\right)$$

$$\therefore \boxed{I_C'} = 0.60 \times 10^{-3} = I_C' \left(1 + \frac{2}{80}\right)$$

$$\therefore I_C' = 0.585 \text{ mA.}$$

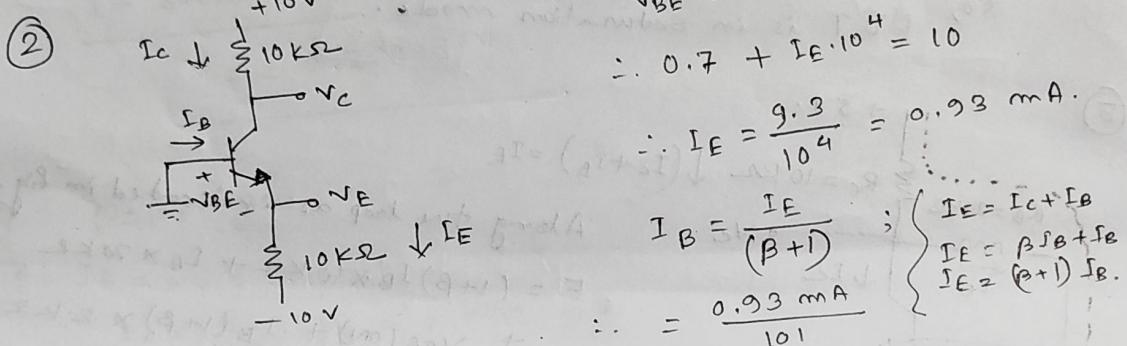
$$\therefore \text{when, } V_{CE} = 5V, I_C = I_C' \left(1 + \frac{5}{80}\right)$$

$$= 0.62 \text{ mA.}$$

$$\text{output resistance } (r_o) = \frac{V_A}{I_C'} = 136.75 \text{ k}\Omega$$

$$V_{BE} + I_E \cdot 10 \times 10^3 - 10 = 0.$$

$$\therefore 0.7 + I_E \cdot 10^4 = 10$$



$$I_E = \frac{9.3}{10^4} = 0.93 \text{ mA.}$$

$$I_B = \frac{I_E}{(\beta+1)} ; \quad \begin{cases} I_E = I_C + I_B \\ I_E = \beta I_B + I_B \\ I_E = (\beta+1) I_B. \end{cases}$$

$$\therefore = \frac{0.93}{101}$$

$$\underline{I_B = 9.2 \text{ mA.}}$$

$$\underline{V_E = 0.7V.}$$

$$\therefore I_C = 0.921 \text{ mA.}$$

$$V_C = 10 - I_C \cdot 10 \text{ k}\Omega$$

$$= 10 - 0.921 \text{ mA} \times 10 \text{ k}\Omega$$

$$\underline{V_C = 0.79 \text{ V.}}$$

③ $V_E = 1V$

$$V_{EB} = 0.7V.$$

$$\therefore V_B = 1 - 0.7 = 0.3V$$

$$\therefore I_B = \frac{0.3}{20 \text{ k}\Omega} = 0.015 \text{ mA.}$$

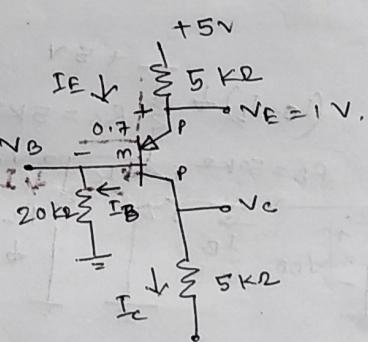
$$\therefore I_E = \frac{5-1}{5 \text{ k}\Omega} = 0.80 \text{ mA.}$$

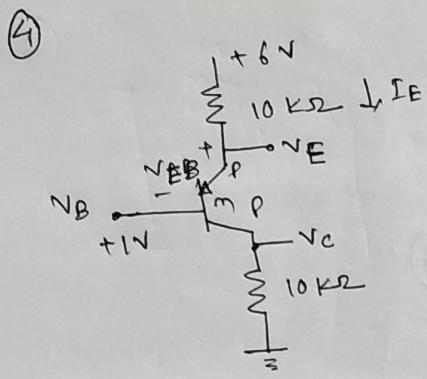
$$\therefore I_C = 0.80 - 0.015 = 0.785 \text{ mA.}$$

$$\therefore V_C = -5 + I_C \cdot 5 \text{ k}\Omega = -1.075 \text{ V.}$$

$$\alpha = \frac{I_C}{I_E} = \underline{0.981}$$

$$\beta = \frac{I_E}{I_B} = \frac{0.785}{0.015} = \underline{52.33}$$





$$V_E = 1V + V_{BE} \quad (\text{assuming } V_{BE} = 0.7V)$$

$$V_E = 1.7V$$

$$\therefore I_E = \frac{6 - 1.7}{10k\Omega}$$

$$= 0.43mA$$

$$I_E = 0.43mA$$

$$\therefore I_C = \frac{\beta}{1+\beta} \cdot 0.43$$

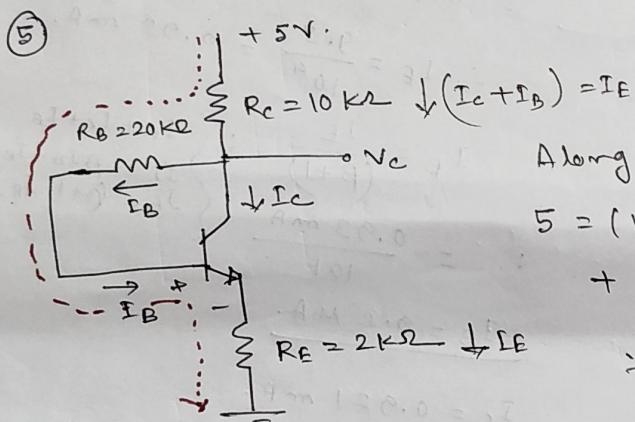
$$= 0.426mA$$

$$\therefore V_C = I_C \times 10k\Omega$$

$$\underline{V_C = 4.26V}$$

So, $V_C > V_B$, Base-collector junction is in ^{forward} ~~reverse~~ bias.

So, BJT is in saturation mode.



Along the loop as indicated in fig.

$$5 = (1+\beta) I_B \times 10k\Omega + I_B \times 2k\Omega$$

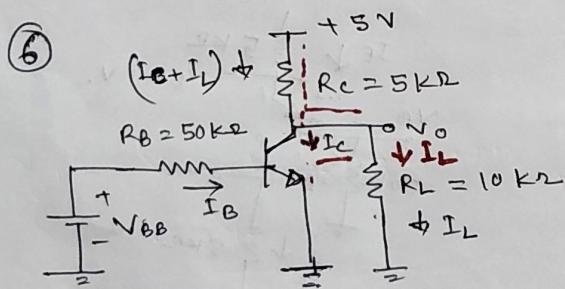
$$+ V_{BE(\text{ON})} + I_B (1+\beta) \times 2k\Omega$$

$$\therefore I_B = 4.61mA$$

$$\therefore I_C = \beta \cdot I_B = 0.346mA$$

$$V_C = 5 - 10k\Omega \times (I_C + I_B)$$

$$\underline{= 1.5V}$$



(a) When, $V_{BB} = 0$.

Cut-off.

$$I_B = 0, I_C = 0$$

$$\therefore V_o = \frac{10}{10+5} \times 5V$$

$$\underline{V_o = 3.33V} \quad (1)$$

(b) When, $V_{BB} = 1V$, $I_B = \frac{V_{BB} - V_{BE(\text{ON})}}{R_B} = \frac{1 - 0.7}{50k\Omega} = 6mA$.

$\therefore I_C = \beta I_B = 75 \times 6mA = 0.45mA$. (assuming BJT in active mode)

KCL in the marked loop,

$$\frac{5 - V_o}{5} = I_C + \frac{V_o}{10}$$

$$\therefore \underline{V_o = 1.83V}$$

, So, BJT is in ~~active~~ active mode as assumed.

(c) When, $V_{BB} = 2V$.

$$I_B = \frac{2 - 0.7}{50} = 26 \text{ mA.}$$

$\therefore I_C = \beta I_B = 1.95 \text{ mA.}$ (if in active mode).

$$\frac{5 - V_o}{5k} = I_C + \frac{V_o}{10k}$$

$$\therefore 1 - 0.2V_o = 1.95 + \frac{V_o}{10} \quad \text{(Amplifier mode)}$$

$$\therefore -0.95 = 0.3V_o.$$

$\therefore V_o < 0.$ (BJT not in active mode, so it is in saturation).

In saturation,

$$V_{CE} = 0.2V = V_o.$$

(7) $\beta = 100$ given.

If $I_B = I_E = 0.1 \text{ mA.}$

$$\therefore I_C = \frac{\beta}{1+\beta} I_E = 0.099 \text{ mA.}$$

$$\therefore V_o = 5 - I_C R_C = 4.50 \text{ V.} \quad (\text{i})$$

(Active mode)

If $I_B = I_E = 0.5 \text{ mA.}$

$$\therefore I_C = \frac{\beta}{1+\beta} I_E = 0.495 \text{ mA.} \quad (\text{Active mode})$$

$$\therefore V_o = 5 - I_C R_C = 2.52 \text{ V.} \quad (\text{ii})$$

If $I_B = I_E = 2 \text{ mA.}$, then BJT is in saturation.

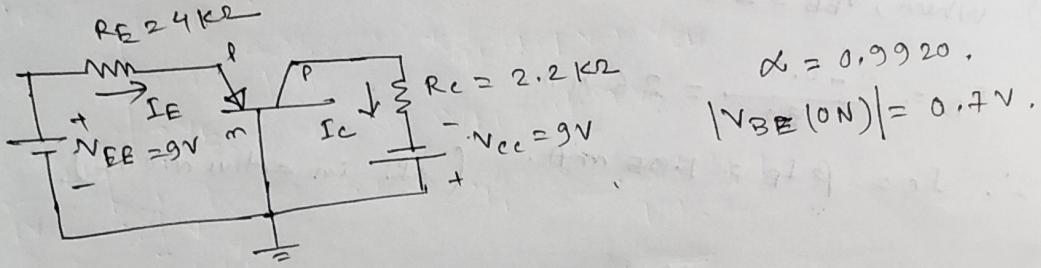
$$\therefore V_{CE} = 0.2V.$$

$$V_{CB} + V_{BE(\text{on})} = 0.2V.$$

$\therefore V_{BE(\text{sat})} = V_{BE(\text{on})}$
(assuming)

$$\therefore V_{CB} = (0.2 - 0.7)V \\ = -0.5V. \quad (\text{iii})$$

(8)



$$\alpha = 0.9920.$$

$$|V_{BE}(ON)| = 0.7V.$$

$$\therefore \text{I/P loop}, \quad 9 = I_E \cdot R_E + V_{EB-ON}$$

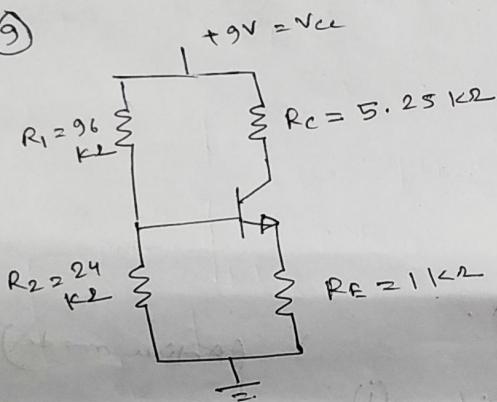
$$\therefore I_E = \frac{9 - 0.7}{4 \text{ k}\Omega} = 2.075 \text{ mA.}$$

$$\therefore I_C = \alpha I_E = 2.06 \text{ mA.}$$

$$\therefore \text{O/P loop}, \quad V_{BC} + I_C R_{CE} = V_{CC}$$

$$\therefore V_{BC} = V_{CC} - I_C R_C = 4.47 \text{ V.}$$

(9)

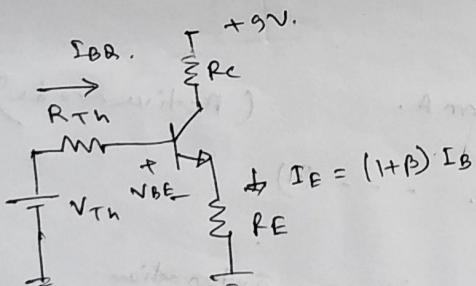


$$\boxed{\beta = 80.}$$

$$R_{TH} = 96 \text{ k}\Omega$$

$$R_{TH} = \frac{96 \times 24}{96+24} = 19.2 \text{ k}\Omega$$

$$\therefore V_{TH} = 9 \times \frac{24}{96+24} = 1.8 \text{ V.}$$



$$V_{TH} = I_{BB.} R_{TH} + V_{BE} + I_E R_E$$

$$\therefore V_{TH} = I_{BB.} R_{TH} + V_{BE} + (1+\beta) I_{BB.} R_E$$

$$\therefore I_{BB.} = \frac{V_{TH} - V_{BE}}{R_{TH} + (1+\beta) R_E}$$

$$= \frac{1.8 - 0.7}{19.2 \text{ k}\Omega + (1+80) 1 \text{ k}\Omega}$$

$$\therefore V_{CEBB} = V_{CC} - I_{CB.} R_C - I_{EB.} R_E$$

$$= 3.50 \text{ V.}$$

$$I_{BB.} = 10.97 \text{ mA.}$$

$$\therefore I_{CB.} = \beta I_{BB.} = 0.878 \text{ mA.}$$

$$\therefore I_{EB.} = (\beta + 1) I_{BB.}$$

$$= 0.888 \text{ mA.}$$

if $\beta = 120$,

$$\therefore I_{BB} = \frac{V_{Th} - V_{BE}}{R_{Th} + (1+\beta) R_E} = \frac{1.8 - 0.7}{19.2 \text{ k} + (1+120) \text{ k}} \\ = 7.84 \text{ mA}$$

$$\therefore I_{CB} = I_{BB} \cdot \beta$$

$$= 120 \times 7.84 \text{ mA} = 0.941 \text{ mA}$$

$$\therefore \text{change in } I_{CB} = \frac{0.941 - 0.878}{0.878} \times 100\% = 7.17\% \quad \text{Ans}$$

$$\therefore \beta = \frac{120 - 80}{80} = 50\%$$

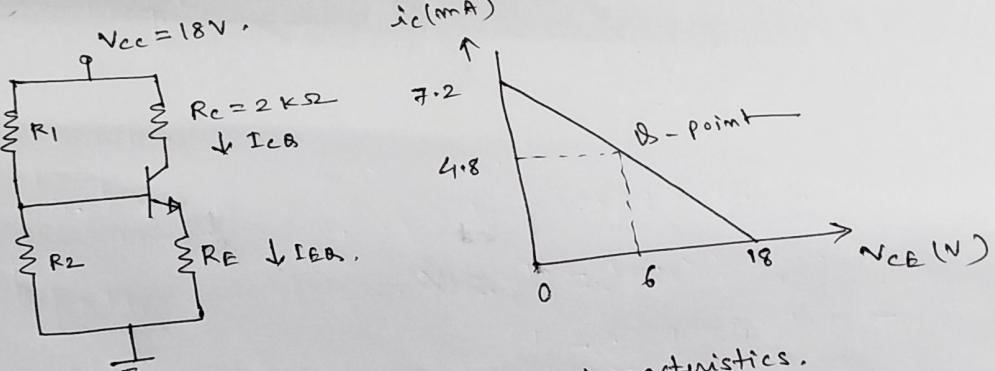
Observation: In voltage divider biasing, though β changes by 50%, the I_{CB} ~~changes~~ only by 7.17%.

So, the circuit is quite stable in terms of change in β .

Now, find V_{CEB} when $\beta = 120$

$$\text{Change in } V_{CEB} = -11\% \quad \text{Ans}$$

(10)



$I_{CB} = 4.8 \text{ mA}$ from DC load line characteristics.

$$I_{EB} = \frac{I_{CB}}{\alpha} = \frac{4.8}{0.9917} = 4.84 \text{ mA}$$

$\beta = 120 \therefore \alpha = 0.9917 \therefore I_{EB} = \frac{I_{CB}}{\alpha} = 4.84 \text{ mA}$

$V_{CEB} = 6 \text{ V}$ from DC load line characteristics.

$$V_{CEB} = 18 - I_{CB} R_c - I_{EB} \cdot R_E$$

$$\therefore 6 = 18 - 4.8 \times 2 - 4.84 \times R_E$$

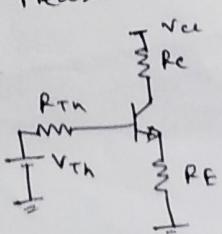
$$\therefore R_E = 0.496 \text{ k}\Omega = 496 \Omega$$

We know that for voltage divider biasing to make the circuit bias stable in terms of change in β ,

$$R_{Th} = 0.1 \times (1+\beta) R_E = 6 \text{ k}\Omega \quad [\text{Please see class note}]$$

where, R_{Th} is the Thévenin's eqn. resistance at base terminal and

$$R_{Th} = (R_1 || R_2) \quad \text{and} \quad V_{Th} = \frac{V_{cc} \cdot R_2}{R_1 + R_2}$$



$$\therefore I_{BB} = \frac{I_{CB}}{\beta} = 0.040 \text{ mA}$$

$$\therefore V_{Th} = I_{BB} \cdot R_{Th} + V_{BE(on)} + RE \cdot IE$$

$$V_{Th} = I_{BB} \cdot R_{Th} + V_{BE(on)} + RE(1+\beta) I_{BB}$$

$$V_{Th} = 0.04 \times R_{Th} + 0.7 + 0.496 \times 0.040 \times 12$$

$$V_{Th} = 3.34 \text{ V.}$$

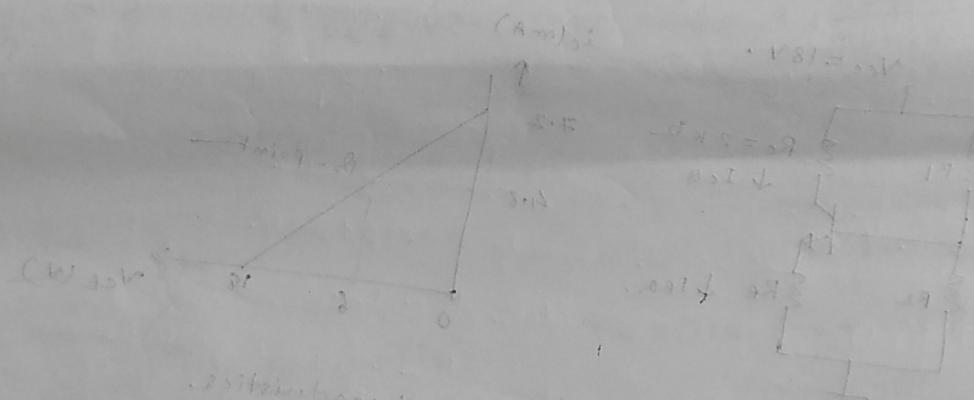
$$V_{Th} = \frac{R_1 \cdot R_2}{R_1 + R_2} \cdot V_{CC} \quad [R_{Th} = \frac{R_1 R_2}{R_1 + R_2}]$$

$$3.34 = 6 \text{ k} \cdot \frac{1}{R_1} \cdot 18 \text{ V.}$$

$$R_1 = 32.3 \text{ k}\Omega$$

$$R_{Th} = \frac{R_1 R_2}{R_1 + R_2} = 6 \text{ k.}$$

$$R_2 = 7.37 \text{ k}\Omega$$



$$A_{voltage} = \frac{V_{out}}{V_{in}} = 63.1 \quad A_{current} = 63.1$$

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Output current is constant in all stages and is 63.1 mA.

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