

EC574 HW8

8.1

$$a) \delta_p = \delta_s = 0.15 \quad \omega_p = 0.5\pi \quad \omega_s = 0.75\pi$$

$$A = -20 \log_{10}(\delta) \quad \delta \text{ means smaller of the passband ripple}$$

$$= -20 \log_{10}(0.15) \approx 16.478 \text{ dB}$$

$$\text{Since } A \leq 21 \quad \text{so } \beta = 0$$

$$N = \frac{A - 8}{2.285(\omega_s - \omega_p)} = \frac{16.478 - 8}{2.285(0.75\pi - 0.5\pi)} = \frac{8.478}{2.285 \times 0.25\pi}$$

$$N \approx 5.39$$

Since filter length must be integer and odd

$$\text{so } N = 7$$

$$b) \delta_p = \delta_s = 0.15 \quad \omega_p = 0.5\pi \quad \omega_s = 0.65\pi$$

$$A = -20 \log_{10}(\delta) = 16.478 \text{ dB}$$

Same as previous, $\beta = 0$.

$$N = \frac{A - 8}{2.285(\omega_s - \omega_p)} = \frac{8.478}{2.285 \times 0.15\pi} \approx 7.87$$

must be integer and odd so $N = 9$

$$c) \delta_p = \delta_s = 0.09 \quad \omega_p = 0.5\pi \quad \omega_s = 0.75\pi$$

$$A = -20 \log_{10}(\delta) = -20 \log_{10}(0.09) \approx 20.916 \text{ dB}$$

$$A < 21, \text{ so } \beta = 0$$

$$N = \frac{A - 8}{2.285(\omega_s - \omega_p)} = \frac{20.916 - 8}{2.285 \times 0.25\pi} \approx 8.22$$

Since N is odd and integer

$$N = 9$$

d) $\delta_p = \delta_s = 0.09$ $\omega_p = 0.5\pi$ $\omega_s = 0.65\pi$

$$A = -20 \log_{10}(8) = -20 \log_{10}(0.09) \approx 20.916 \text{ dB}$$

$$A \leq 21 \text{ so } B = 0$$

$$N = \frac{A - 8}{2.285(\omega_s - \omega_p)} = \frac{20.916 - 8}{2.285 \cdot 0.15\pi} \approx 11.99$$

Since N is integer and odd

$$N = 13$$

8.2

a) We prefer Optimal FIR Filter Design because it yields a more efficient filter with better frequency response control and a shorter filter length. It will be able to minimizing the overall error in passband and stopband.

b) $\delta_p = 0.10$, $\delta_s = 0.05$, $\omega_p = 0.5\pi$, $\omega_s = 0.75\pi$

$2L+1$ = smallest odd integer greater than

$$\frac{\delta_p \delta_s}{\log_{10}(\delta_p \delta_s)} \approx -2.301$$

$$\omega_s - \omega_p = 0.25\pi$$

$$\text{so } \textcircled{1} \approx 5.49 \quad \text{so } 2L+1 = 7$$

$$\frac{-10 \log_{10}(\delta_p \delta_s) - 13}{14.6(\omega_s - \omega_p) / 2\pi}$$

①

c) $N = \frac{A - 8}{2.285(\omega_s - \omega_p)}$ $A = -20 \log_{10}(\delta_s) \approx 26.02 \text{ dB}$

$$N = \frac{A - 8}{2.285 \times (\omega_s - \omega_p)} \approx 11.47$$

N is odd and integer $N = 13$

d) $2L+1=7 \Rightarrow L=3$

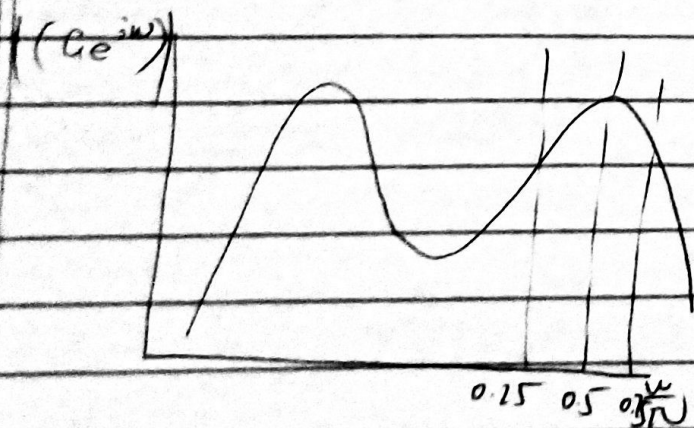
Since $L=3$, the highest power of $\cos(\omega)$ in $R(\omega)$ is $\cos^3(\omega)$

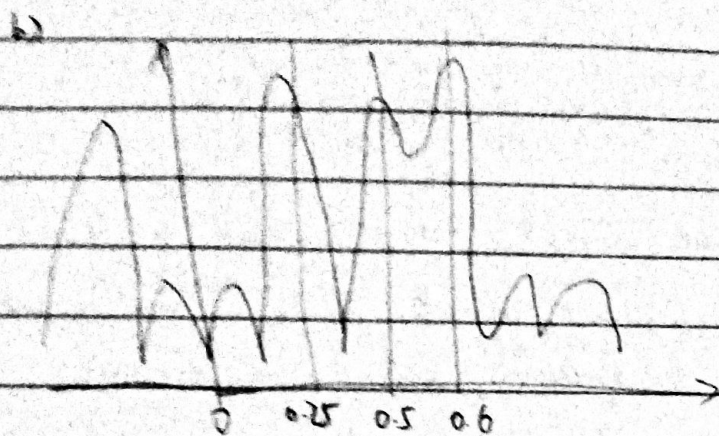
This is because the filter length is $N=7$ and it allows the \cos to have a max power of 3. This is the natural of type I FIR because of the symmetry and length of the impulse response in a Type I FIR filter.

8.3

a) So for $x[n]$ the \cos components will become actually sine in the frequency domain. The max width of the sine is the central lobe and its length depends inversely to N .

So Given $N=4$, the sine will have relatively wide main lobe which will make the three frequency kind of merge into each other.





when $N = 16$
they are separable

c) So when $N \rightarrow \infty$, the rectangular window's frequency response would become ideal needle like aliasing the exact frequency out accurately

8.4.

a) Frequency Resolution

b) Length shape Design