

FALL 2021 EC516 PRACTICE TEST 1

Closed Book; Total Time=60 minutes; No collaboration with Anyone; No Electronics (such as cell phones, calculators etc.

Problem 1 (20 Points)

Consider a digital filter **F** with impulse response $h[n] = \sum_{k=0}^9 2\delta[n - k]$

- (a) What are the locations of the poles of the system function, $H(z)$, of the filter **F**? Justify your answer.
- (b) For what frequencies in the range $0 \leq \omega \leq \pi$, is it guaranteed that the frequency response of filter **F** is zero? Justify your answer,

Problem 2 (20 Points)

Consider a filter **G** with corresponding frequency response $H(e^{j\omega}) = e^{-j4\omega} \left(\frac{\sin(\frac{5\omega}{2})}{\sin(\frac{\omega}{2})} \right)^2$.

- (a) If the filter **G** is provided the input $x[n] = u[n + 3]$, for what values of n is it *guaranteed* that the output signal $y[n]$ is equal to zero? Justify your answer.
- (b) Draw a flowgraph for filter **G** as a cascade of second order sections. Show your work.

Problem 3 (20 Points)

- (a) Determine a non-recursive difference equation for a digital FIR filter **H** whose magnitude of the frequency response is zero at frequencies $\omega = 0$ and $\omega = \pi$ and is non-zero in the frequency range $0 < \omega < \pi$. Justify your answer.
- (b) Determine a recursive difference equation for the digital FIR filter **H** in the previous part of this problem. Justify your answer.

Problem 4 (20 Points)

- (a) Draw the Pole-Zero plot for an IIR filter **S** whose input and output are related by the following difference equation:

$$y[n] = 0.25y[n - 4] + x[n] + x[n - 2]$$

- (b) Would you describe a digital filter **T** obtained by applying Bilinear transformation to an analog filter with frequency response $H(j\omega) = 1/(j\omega + 5000)$ as a lowpass filter? Justify your answer.

Problem 5 (10 Points)

Sketch the phase of the DTFT of the signal specified as $g[n] = \delta[n] + \delta[n - 4]$ where $\delta[n]$ is the unit impulse. *Justify your answer*

Problem 6 (10 points)

During the discovery process for a digital filtering application, it is reported that the filter is required to have linear phase. What does that imply for the filter design process? Explain your answer.

EC516 FORMULA SHEET MIDTERM EC516 FALL 2024

Unit Step

$$u[n] = \begin{cases} 1 & n \geq 0 \\ 0 & \text{otherwise} \end{cases}$$

Unit Impulse

$$\delta[n] = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases}$$

Complex Exponentials and Sinusoids

$$e^{j\omega n} = \cos(\omega n) + j \sin(\omega n) \quad \cos(\omega n) = (1/2)(e^{j\omega n} + e^{-j\omega n}) \quad \sin(\omega n) = (1/2j)(e^{j\omega n} - e^{-j\omega n})$$

$$\text{DT Convolution : } y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k] \quad \text{CT Convolution : } y(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau)d\tau$$

$$\text{FSF : } \sum_{k=0}^{N-1} \alpha^k = \begin{cases} \frac{1-\alpha^N}{1-\alpha} & ; \alpha \neq 1 \\ N & ; \alpha = 1 \end{cases}$$

$$\text{ISF : } \sum_{k=0}^{\infty} \alpha^k = \frac{1}{1-\alpha} \quad ; |\alpha| < 1$$

$$\text{DTFT : } X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}; \quad \text{Inverse DTFT : } x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega})e^{j\omega n}d\omega$$

Basic DTFT Properties:

$$\begin{aligned} x[n-n_0] &\Leftrightarrow e^{-j\omega n_0} X(e^{j\omega}) & e^{j\omega_0 n} x[n] &\Leftrightarrow X(e^{j(\omega-\omega_0)}) & x^*[n] &\Leftrightarrow X^*(e^{-j\omega}) \\ x[-n] &\Leftrightarrow X(e^{-j\omega}) & x[n] * h[n] &\Leftrightarrow X(e^{j\omega})H(e^{j\omega}) \end{aligned}$$

Common DTFT Pairs

$$e^{j\omega_0 n} \Leftrightarrow 2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - k\omega_0)$$

$$\delta[n-n_0] \Leftrightarrow e^{-j\omega n_0} \quad u[n] - u[n-N] \Leftrightarrow \frac{\sin(\omega N/2)}{\sin(\omega/2)} e^{-j\omega(N-1)/2}$$

$$\frac{\sin \omega_0 n}{\pi n} \Leftrightarrow \begin{cases} 1 & 0 \leq |\omega| \leq \omega_0 \\ 0 & \omega_0 < |\omega| \leq \pi \end{cases}$$

z-transform:

$$X(z) = \sum_{n=-\infty}^{\infty} x[n]z^{-n}$$

Properties of z-transform:

$$x[n - n_0] \Leftrightarrow z^{-n_0} X(z) \quad x[-n] \Leftrightarrow X(z^{-1}) \quad x^*[n] \Leftrightarrow X^*(z^*)$$

$$x[n] * h[n] \Leftrightarrow X(z) \times H(z)$$

Basic z-transform pair: Decaying Exponential $a^n u[n] \Leftrightarrow \frac{1}{1 - az^{-1}}$ ROC: $|z| > |a|$

Recursive Difference Equation

$$y[n] = \sum_{k=1}^N a_k y[n-k] + \sum_{m=0}^M b_m x[n-m]$$

Non-Recursive Difference Equation

$$y[n] = \sum_{k=0}^M b_k x[n-k]$$

FIR filters

A FIR filter has an impulse response of length N such that $h[n] = 0$ for $n < 0$ and for $n > N-1$

IIR filters

Impulse response $h[n] = 0$ for $n < 0$ and system function $H(z)$ is rational and has at least one pole not at the origin.

Linear Phase FIR Filters:

Type I: Odd Length, Symmetric
 Type II: Even Length, Symmetric
 Type III: Odd Length, Anti-symmetric
 Type IV: Even Length, Anti-symmetric

N-Point Signal

$$h[n] = 0 \text{ for } n < 0 \text{ and for } n > N-1$$

Bilinear Transformation

$$H_d(z) = H_a\left(\frac{1 - z^{-1}}{1 + z^{-1}}\right)$$