Sparse Signal Processing Lecture 4 — Fourier interpolation and the linear model

Nicolas Keriven (slides by Clément Elvira)

Outline

N. Keriven: CNRS researcher at IRISA, SIROCCO team.

https://nkeriven.github.io

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- $\,\blacktriangleright\,$ particularize the linear model to an infinite-dimensional setting ;
- introduce a linear target set;
- derive a decoder achieving exact recovery;

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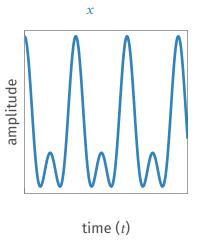
Material at https://github.com/nkeriven/ENS-signal (subject to changes, check every now and then)

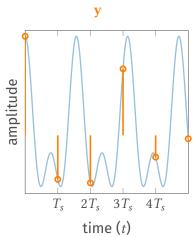
References

- ► Mallat, 2012. A Wavelet Tour of Signal Processing: the Sparse Way
- ► F Golse polycopié: http://www.math.polytechnique.fr/ ~golse/MAT431-10/POLY431.pdf
- ► Wikipédia!

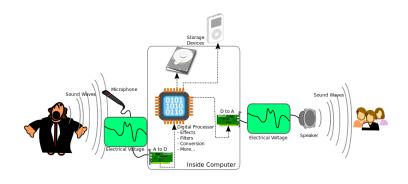
Considered Task

Recovering a signal x from **infinitely** many function evaluations



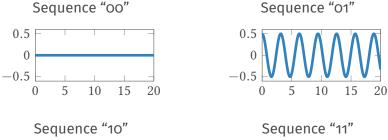


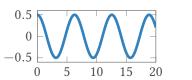
Application 1: sampling and reconstructing audio recording



Important component of sound perception: the frequency

Application 2: (simplified) transmitters





Sequence "11"

0.5
0
-0.5
0
5
10
15
20

Support of information: electromagnetic (sinusoidal at a given frequency) waves

Wrapping up: main ingredients

1. The sensing operator: The (linear) operator defined by

$$M \colon \mathcal{L}^2(\mathbf{R}) \longrightarrow \mathbf{R}^{\mathbf{Z}}$$
$$x \longmapsto \{x(jT_s)\}_{i \in \mathbf{Z}}$$

2. The target set: To be defined

Related to some notion of frequency

- 3. The decoder: To be defined
- 4. The accuracy criterion: "noiseless exact recovery", i.e.

$$\forall x \in \mathcal{X}_{\text{target}}, \quad D(M(x)) = x$$



Which mathematical tool can be used to emphasize the frequencies characterizing a signal?



Which mathematical tool can be used to emphasize the frequencies characterizing a signal?



The Fourier transform

Rappels: The Fourier transform

Definition

The Fourier transform is a linear operator defined by

$$\mathcal{F} \colon \mathcal{L}^{1}(\mathbf{R}) \longrightarrow \mathcal{L}^{1}(\mathbf{R})$$
$$x \longmapsto \left(f \mapsto \int_{-\infty}^{+\infty} x(t) e^{-i2\pi f t} dt \right)$$

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- ightharpoonup Sometimes defined without the 2π (but less clean, multiplicative constants everywhere)
- $ightharpoonup \mathcal{F}x$ is often written \hat{x}

An example

Exercice

Evaluate the Fourier transform of

$$\Pi_{[-a/2,a/2]}:t\longmapsto egin{cases} 1 & ext{if } t\in[-rac{a}{2},+rac{a}{2}] \ 0 & ext{otherwise} \end{cases}$$

An example

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Answer

$$\mathcal{F}\Pi(f) = a \frac{\sin(\pi a f)}{\pi a f} \stackrel{\text{def.}}{=} a \operatorname{sinc}(a f)$$

Rappels: Inverse Fourier transform

Definition

The conjugate Fourier transform is the linear operator defined by

$$\overline{\mathcal{F}} \colon \mathcal{L}^{1}(\mathbf{R}) \longrightarrow \mathcal{L}^{1}(\mathbf{R})$$
$$x \longmapsto \left(f \mapsto \int_{-\infty}^{+\infty} x(t) \mathrm{e}^{\mathrm{i}2\pi f t} \mathrm{d}t \right)$$

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Theorem: Inversion

If $x \in \mathcal{L}^1(\mathbf{R})$ is such that $\mathcal{F}x \in \mathcal{L}^1(\mathbf{R})$ then

$$\overline{\mathcal{F}}[\mathcal{F}x] \stackrel{\text{a.e.}}{=} x$$

with equality at every point of continuity.



Pbm: $x \in \mathcal{L}^1$ does **not** imply $\mathcal{F}x \in \mathcal{L}^1$! (ex: door function and cardinal sinus)



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 \rightarrow Solution: extension to \mathcal{L}^2 and Parseval identity.



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Pbm: the classical Fourier integral is not defined for $\mathcal{L}^2...$

→ Solution: extension by *density*.

Parseval's identity

Parseval's identity

If $f, g \in \mathcal{L}^1 \cap \mathcal{L}^2$, $\langle f, g \rangle_{L^2} = \langle \mathcal{F}f, \mathcal{F}g \rangle_{L^2}$ (recall that $\langle f, g \rangle_{L^2} = \int f(t) \overline{g(t)} dt$)

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Consequence: $\|f\|_{L^2} = \|\mathcal{F}f\|_{L^2}$ and $f \in \mathcal{L}^1 \cap \mathcal{L}^2 \Rightarrow \mathcal{F}f \in \mathcal{L}^2$. Now we just need to extend the domain of definition of \mathcal{F}

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Consequence: $||f||_{L^2} = ||\mathcal{F}f||_{L^2}$ and $f \in \mathcal{L}^1 \cap \mathcal{L}^2 \Rightarrow \mathcal{F}f \in \mathcal{L}^2$. Now we just need to extend the domain of definition of \mathcal{F}

We are going to extend Fourier to \mathcal{L}^2 by *density* of $\mathcal{L}^1 \cap \mathcal{L}^2$ within \mathcal{L}^2

$$o$$
 i.e., $\forall \epsilon > 0, f \in \mathcal{L}^2, \exists g \in \mathcal{L}^2 \cap \mathcal{L}^1$ s.t. $\|f - g\|_{\mathcal{L}^2} \leq \epsilon$

- ▶ Let $f \in \mathcal{L}^2$.
- ▶ By density of $\mathcal{L}^2 \cap \mathcal{L}^1$ in \mathcal{L}^2 , there is a sequence $f_n \in \mathcal{L}^2 \cap \mathcal{L}^1$ such that $\|f_n f\|_{\mathcal{L}^2} \to 0$

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- ▶ f_n is a Cauchy sequence, and by Parseval's identity $||f_n f_p||_{\mathcal{L}^2} = ||\mathcal{F}(f_n f_p)||_{\mathcal{L}^2}$, hence $\mathcal{F}f_n \in \mathcal{L}^2$ is also a Cauchy sequence

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- ▶ Since \mathcal{L}^2 is a Hilbert space and **is therefore complete**, $\mathcal{F}f_n$ is a convergent sequence.
- ▶ We call its limit the "Fourier transform of f" denoted by $\mathcal{F}f$. It satisfies all the usual properties of the Fourier transform (see next slide)

Properties of Fourier Transform (...easy, or admitted)

- ► f symmetric $\Rightarrow \mathcal{F}f$ real
- ightharpoonup f anti-symmetric $\Rightarrow \mathcal{F}f$ imaginary
- ► $f \text{ real} \Rightarrow \mathcal{F}f \text{ Hermitian}$
- $lackbox{}{} f$ imaginary $\Rightarrow \mathcal{F} f$ (conjugate) anti-symmetric

Properties of Fourier Transform (...easy, or admitted)

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- ▶ $f \text{ real} \Rightarrow \mathcal{F}f \text{ Hermitian}$
- f imaginary $\Rightarrow \mathcal{F}f$ (conjugate) anti-symmetric
- $ightharpoonup \mathcal{F}(f\star g)(t) = \mathcal{F}f(t)\mathcal{F}g(t)$ (convolution thm)
- $\blacktriangleright \langle f, g \rangle_{L^2} = \langle \mathcal{F}f, \mathcal{F}g \rangle_{L^2}$ (Parseval identity)
- ► $\mathcal{F} \frac{df}{dt}(t) = 2i\pi t \mathcal{F} f(t) \Rightarrow \text{if } f \text{ is } k \text{ times differentiable, } \mathcal{F} f \text{ decrease faster than } 1/t^k$

Properties (summary)

Property	Function	Fourier transform
	x(t)	$\hat{x}(f)$
Inverse	$\hat{x}(t)$	x(-f)
Convolution	$x_1 \star x_2(t)$	$\hat{x}_1\hat{x}_2(f)$
Multiplication	$x_1(t)x_2(t)$	$\hat{x}_1 \star \hat{x}_2(f)$
Translation	x(t-u)	$e^{-2i\pi uf}\hat{x}(f)$
Modulation	$e^{2i\pi\xi t}x(t)$	$\hat{x}(f-\xi)$
Scaling	x(t/s)	$ s \hat{x}(sf)$
Time derivatives	$x^{(p)}(t)$	$(2i\pi f)^p \hat{x}(f)$
Freq. derivatives	$(-2i\pi t)^p x(t)$	$\hat{x}^{(p)}(f)$
Complex conjugate	$\overline{x}(t)$	$\overline{\hat{x}}(-f)$

Uncertainty principle



Can we construct a function that is **highly localised in space**, whose Fourier transform is **highly localised in frequency**?

Uncertainty principle



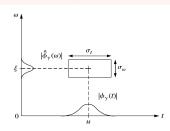
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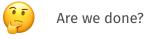
Heisenberg's uncertainty principle

For $x \in \mathcal{L}^2(\mathbf{R})$, define the two density functions

$$p_x=rac{|x(t)|^2}{\|x\|_{L^2}},\quad p_{\hat{x}}=rac{|\hat{x}(f)|^2}{\|x\|_{L^2}}$$
 and their respective variance σ_x^2 , $\sigma_{\hat{x}}^2$. Then

$$\sigma_{\mathbf{r}}^2 \sigma_{\hat{\mathbf{r}}}^2 \geq 1/4$$







Are we done?



The function of interest, namely $t \longmapsto \cos(2\pi f t)$ does not belong to $\mathcal{L}^1(\mathbf{R})!$

nor $\mathcal{L}^2(\mathbf{R})$



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nor $\mathcal{L}^2(I\!\!R)$

We need a "bigger" space to work with!





Are we done?



The function of interest, namely $t \longmapsto \cos(2\pi f t)$ does not belong to $\mathcal{L}^1(\mathbf{R})!$

nor $\mathcal{L}^2(\mathbf{R})$

We need a "bigger" space to work with!



 \rightarrow Space of "distributions", defined by *duality*.

We need the smallest dual space as possible, to make the biggest distribution space as possible!

Reminder (?) on distributions



Tests functions

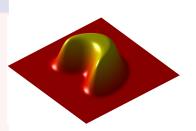
Definition

A test function refers to any infinitely differentiable function

$$\varphi \colon \mathbf{R} \longrightarrow \mathbf{C}$$

witch is compactly supported.

The set of test functions is denoted $\mathcal{D}(\mathbf{R})$.



Tests functions

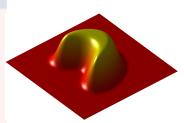
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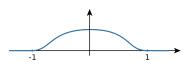
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The set of test functions is denoted $\mathcal{D}(\mathbf{R})$.



Example: (standard one) the function

$$t \longmapsto \begin{cases} \exp\left(-\frac{1}{1-t^2}\right) & \text{ if } t \in [-1,1] \\ 0 & \text{ otherwise} \end{cases}$$



can be proved to be a test function.

Topology on $\mathcal{D}(\mathbf{R})$

Definition: convergence in $\mathcal{D}(\mathbf{R})$

Let $\{\varphi_n\}_{n\in\mathbb{N}}$ be a family of test function.

We say that $\varphi_n \to \varphi \in \mathcal{D}(\mathbf{R})$ if and only if

- ▶ there exists an integer N and a compact set K such $n \ge N \Rightarrow$ support $(\varphi_n \varphi) \subset K$
- ▶ for all $r \in \mathbb{N}$.

$$\left\| \varphi_n^{(r)} - \varphi^{(r)} \right\|_{\infty} \to 0$$

 \blacktriangleright i.e., supremum norm for φ and all its derivatives.

Distributions

Definition

We refer to "distribution" as any continuous linear functional on $\mathcal{D}(\mathbf{R})$.

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Remarks

► Said differently, the space of distributions is the (topological) dual space of $\mathcal{D}(\mathbf{R})$, denoted by $\mathcal{D}'(\mathbf{R})$;

Distributions

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We refer to "distribution" as any continuous linear functional on $\mathcal{D}(\mathbf{R})$.

Remarks

- ► Said differently, the space of distributions is the (topological) dual space of $\mathcal{D}(\mathbf{R})$, denoted by $\mathcal{D}'(\mathbf{R})$;
- \blacktriangleright When evaluating a distribution T at φ , we write

$$\langle T, \varphi \rangle$$

 \blacktriangleright By extension, we say that two distributions T, T' are equal iff

$$\langle T, \varphi \rangle = \langle T', \varphi \rangle \qquad \forall \varphi \in \mathcal{D}(\mathbf{R})$$



Distribution are often referred to as "generalized functions".

This can nevertheless be a misleading terminology, as distributions do not take scalar as inputs. (...but test functions)



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This can nevertheless be a misleading terminology, as distributions do not take scalar as inputs. (...but test functions)



Besides mathematical benefits, there is a physical intuition behind this reference

A compactly supported test functions can be seen as a model of a physical system measuring some quantity

Regular distributions

Definition/Proposition

If g is any locally integrable function, then the linear form T_g defined as

$$\forall \varphi \in \mathcal{D}(\mathbf{R}), \qquad \langle T_g, \varphi \rangle \stackrel{\mathsf{def.}}{=} \int g(t) \varphi(t) \mathrm{d}t$$

is a distribution.

Such a distribution is referred to as "regular distribution".

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is a distribution.

Such a distribution is referred to as "regular distribution".

In short: Linearity is fine Continuity: letting $\varphi_n \to \varphi$, there exists a compact K that contains all supports Then $|\langle T_g, \varphi_n - \varphi \rangle| \leq \int g(t) \mathrm{d}t \|\varphi_n - \varphi\|_{\infty}$

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Remark

- ► Hence distributions is a "bigger" space...
- ▶ If $T_{g_1} = T_{g_2}$ then $g_1 \stackrel{\text{a.e.}}{=} g_2$

Singular distributions



The terminology "singular distributions" refers to any distribution that is not regular

Singular distributions



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Example 1: The dirac function δ_t defined $\forall t \in \mathbf{R}$ as

$$\forall \varphi \in \mathcal{D}(\mathbf{R}), \qquad \langle \delta_t, \varphi \rangle = \varphi(t)$$

since
$$|\langle \delta_t, \varphi_n - \varphi \rangle| \leq ||\varphi_n - \varphi||_{\infty} \to 0$$

Singular distributions



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since $|\langle \delta_t, \varphi_n - \varphi \rangle| \leq ||\varphi_n - \varphi||_{\infty} \to 0$

Example 2: The dirac comb \coprod_{T_s} defined as

$$\forall \varphi \in \mathcal{D}(\mathbf{R}), \qquad \langle \mathrm{III}_{T_{\mathtt{s}}}, \varphi \rangle = \sum_{j \in \mathbf{Z}} \langle \delta_{jT_{\mathtt{s}}}, \varphi \rangle = \sum_{j \in \mathbf{Z}} \varphi(jT_{\mathtt{s}})$$

Letting $\varphi_n \to \varphi$, there exists a compact K that contains all supports, and $|\langle \delta_t, \varphi_n - \varphi \rangle| \leq K_{T_s} \|\varphi_n - \varphi\|_{\infty} \to 0$, where K_{T_s} is the number of jT_s contained in K



Regular distributions provide intuition on how to define many concepts related to distributions

derivation, convolution, Fourier transform to name but a few

Derivation of a distribution

For regular distributions: let g be a differentiable locally integrable function. Then g' is also locally integrable and

$$\langle T_{g'}, \varphi \rangle = \int g'(t)\varphi(t)dt$$

 $= -\int g(t)\varphi'(t)dt$ (IBP)
 $= -\langle T_{\sigma}, \varphi' \rangle$

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 $= -\int g(t)\varphi'(t)dt$ (IBP)
 $= -\langle T_g, \varphi' \rangle$

Definition

Let T be a distribution. The derivative of T is the distribution $T^{(1)}$ defined by

$$\forall \varphi \in \mathcal{D}(\mathbf{R}), \qquad \langle T^{(1)}, \varphi \rangle \stackrel{\text{def.}}{=} -\langle T, \varphi' \rangle$$

A second example: Fourier transform

Let g be an integrable function (so that T_g is a regular distribution). Then \mathcal{F}_g is also integrable and

$$\langle T_{\mathcal{F}g}, \varphi \rangle = \int \mathcal{F}g(f)\varphi(f)df$$

$$= \int \left(\int g(t)e^{-i2\pi ft}dt \right)\varphi(f)df$$

$$= \int g(t)\left(\int \varphi(f)e^{-i2\pi ft}df \right)dt$$

$$\stackrel{(?)}{=} \langle T_g, \mathcal{F}\varphi \rangle$$

Definition

Let T be a distribution. The Fourier of T is the distribution $\mathcal{F}\mathit{T}$ defined by

$$\forall \varphi \in \mathcal{D}(\mathbf{R}), \qquad \langle \mathcal{F}T, \varphi \rangle \stackrel{\text{def.}}{=} \langle T, \mathcal{F}\varphi \rangle$$

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► (need complex-valued test functions to technically make sense)

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► (need complex-valued test functions to technically make sense)



Unfortunately, in general $\mathcal{F}\varphi$ is not compactly supported, it may not be properly defined

Definition

Let ${\it T}$ be a distribution. The Fourier of ${\it T}$ is the distribution ${\it FT}$ defined by

$$\forall \varphi \in \mathcal{D}(\mathbf{R}), \qquad \langle \mathcal{F}T, \varphi \rangle \stackrel{\mathsf{def.}}{=} \langle T, \mathcal{F}\varphi \rangle$$

► (need complex-valued test functions to technically make sense)



Unfortunately, in general $\mathcal{F}\varphi$ is not compactly supported, it may not be properly defined

The set of distributions is too big a space!!

Tempered distribution

We will **extend** the set of test functions...



Tempered distribution



We will **extend** the set of test functions...

... which will **reduce** the set of distributions (by duality).

Tempered distributions

Definition: Schwartz space

We say that a function g is a Schwartz function if it is infinitely differentiable and for all $k, q \in \mathbb{N}$

$$\lim_{t \longmapsto \pm \infty} |t^k \varphi^{(q)}(t)| = 0$$

The set of Schwartz functions is denoted $S(\mathbf{R})$.

► Instead of compact support, we take **fast decrease** (faster than any polynomial).

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Definition: Tempered distribution

We call "tempered distribution" any element of the (topological) dual space of the Schwartz space $S'(\mathbf{R})$.

"Topological": **continuous** linear functionals

► Convergence in $\mathcal{S}(\mathbf{R})$ is defined through the family of semi-norms $\mathcal{N}_p(\varphi) = \sum_{k,q \leq p} \sup_t |t^q \varphi^{(q)}(t)|$, i.e. φ_n converge to φ if $\forall p$, $\mathcal{N}_p(\varphi_n - \varphi) \to 0$

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- ► For a tempered distribution T, there is p, C such that $\forall \varphi \in \mathcal{S}(\mathbf{R})$, $\langle T, \varphi \rangle \leq C \mathcal{N}_p(\varphi)$ (continuity)

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- ► For a tempered distribution T, there is p, C such that $\forall \varphi \in \mathcal{S}(\mathbf{R})$, $\langle T, \varphi \rangle \leq C \mathcal{N}_p(\varphi)$ (continuity)
- ▶ Any test function is a Schwartz function: $\mathcal{D}(\mathbf{R}) \subset \mathcal{S}(\mathbf{R})$

- ► Convergence in $\mathcal{S}(\mathbf{R})$ is defined through the family of semi-norms $\mathcal{N}_p(\varphi) = \sum_{k,q \leq p} \sup_t |t^q \varphi^{(q)}(t)|$, i.e. φ_n converge to φ if $\forall p, \ \mathcal{N}_p(\varphi_n \varphi) \to \mathbf{0}$
- ► For a tempered distribution T, there is p, C such that $\forall \varphi \in \mathcal{S}(\mathbf{R})$, $\langle T, \varphi \rangle \leq C \mathcal{N}_p(\varphi)$ (continuity)
- lacktriangle Any test function is a Schwartz function: $\mathcal{D}(R)\subset\mathcal{S}(R)$
- ▶ (hence) any tempered distribution is a distribution: $S'(\mathbf{R}) \subset \mathcal{D}'(\mathbf{R})$

Fourier and tempered distributions

Thm

► Thm: All tempered distribution **admit a Fourier transform** defined by

$$\forall \varphi \in \mathcal{S}(\mathbf{R}), \qquad \langle \mathcal{F}T, \varphi \rangle \stackrel{\text{def.}}{=} \langle T, \mathcal{F}\varphi \rangle$$

► The Fourier transform of a Schwartz function is **also** a Schwartz function

Hint: link between differentiability and polynomial decrease of the Fourier transform...

Classical tempered distribution



Unlike regular distributions, all functions do not correspond to tempered distributions! Need all derivative to grows "as fast as" polynomials...

Classical tempered distribution



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Still:

- ▶ The Dirac function δ_t is a tempered distribution
- ► The Dirac comb is a tempered distribution
- ▶ the function $t \mapsto \cos(2\pi f t)$ is a tempered distribution (as all bounded functions)

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- ▶ The Dirac function δ_t is a tempered distribution
- ► The Dirac comb is a tempered distribution
- ▶ the function $t \mapsto \cos(2\pi f t)$ is a tempered distribution (as all bounded functions)
- \rightarrow that's all we need in this course...?



In signal processing, it is folklore to believe that distributions are always well-behaved

That it all operations are legitimate

It it nevertheless not always true



In this lecture (and companion lab) it will nevertheless be true

Some justifications will be out of scope and left in exercise

Some properties of (tempered) distributions

Some Fourier transform (1)

$$\langle \mathcal{F}\boldsymbol{\delta_a}, \varphi \rangle = \langle \delta_a, \mathcal{F}\varphi \rangle$$

$$= (\mathcal{F}\varphi)(a)$$

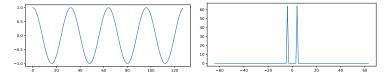
$$= \int \varphi(t) e^{-2i\pi at} dt$$

$$= \langle T_{t \mapsto e^{-2i\pi at}}, \varphi \rangle$$

Some Fourier transform (2)

Some Fourier transform (3)

$$\begin{split} \langle \mathcal{F} T_{t \mapsto \cos(2\pi a t)}, \varphi \rangle &= \langle \mathcal{F} T_{t \mapsto 0.5(\mathrm{e}^{2\pi a t} + \mathrm{e}^{-2\pi a t}, \varphi)} \\ &= \frac{1}{2} \langle \mathcal{F} T_{t \mapsto \mathrm{e}^{2\pi a t}}, \varphi \rangle + \frac{1}{2} \langle \mathcal{F} T_{t \mapsto \mathrm{e}^{-2\pi a t}}, \varphi \rangle \\ &= \frac{1}{2} (\langle \delta_a, \varphi \rangle + \langle \delta_{-a}, \varphi \rangle) \end{split}$$



Some Fourier transform (3)

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Recall that
$$\mathcal{F}[Re(f)](\xi) = \frac{1}{2}(\mathcal{F}f(\xi) + \mathcal{F}f(-\xi))$$



Have we achieved our initial objective?



Have we achieved our initial objective?



It is folklore to provide a graphical representation of a Fourier transform

Another example: the "door" distribution

We define $\Pi_{[-a/2,a/2]}$ such that

$$orall arphi \in \mathcal{S}(\mathbf{R}), \qquad \langle \Pi_{[-a/2,a/2]}, arphi
angle = \int_{-a/2}^{a/2} arphi(t) \mathrm{d}t$$

Question: Evaluate the Fourier transform of $\Pi_{[-a/2,a/2]}$.

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Question: Evaluate the Fourier transform of $\Pi_{[-a/2,a/2]}$.

$$\begin{split} \langle \mathcal{F}\Pi_{[-a/2,a/2]},\varphi\rangle &= \langle \Pi_{[-a/2,a/2]},\mathcal{F}\varphi\rangle \\ &\stackrel{\mathsf{Fub.}}{=} \int \varphi(t) \int_{-a/2}^{a//2} \mathrm{e}^{-2\mathrm{i} f t} \mathrm{d} f \mathrm{d} t \\ &= \langle T_{a \operatorname{sinc}(a \cdot)},\varphi\rangle \end{split}$$



Does it ring a bell?

Compatibility with the "standard" Fourier transform

Theorem (admitted)

If $x \in \mathcal{L}^1(\mathbf{R})$ (or $\mathcal{L}^2(\mathbf{R})$) then

$$\mathcal{F}T_{r}=T_{\mathcal{F}r}$$

Towards an inverse Fourier transform

Definition: Conjugate Fourier transform

Let T be a tempered distribution. The conjugate Fourier of T is the distribution $\overline{\mathcal{F}}T$ defined by

$$\forall \varphi \in \mathcal{S}(\mathbf{R}), \qquad \langle \overline{\mathcal{F}}T, \varphi \rangle \stackrel{\text{def.}}{=} \langle T, \overline{\mathcal{F}}\varphi \rangle$$

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Theorem: Inversion formula

Let T be a tempered distribution.

Then, its Fourier transform satisfies

$$T = \overline{\mathcal{F}}[\mathcal{F}T]$$

In other words, the standard inversion formula holds in the sense of tempered distributions

Convolution

Definition

Let T_1, T_2 .

Their convolution (if it exists) is the distribution $T_1 * T_2$ defined as

$$\forall \varphi \in \mathcal{S}(\mathbf{R}), \qquad \langle T_1 * T_2, \varphi \rangle \stackrel{\mathsf{def.}}{=} \langle T_1, \varphi_{T_2} \rangle$$

where

$$\forall t \in \mathbf{R}, \qquad \varphi_{T_2}(t) = \langle T_2, \varphi(\cdot + t) \rangle$$

 $(T_1, T_2 \text{ can be exchanged in the definition})$

Example: of $T_1 = \delta_a$

$$\forall \varphi \in \mathcal{S}(\mathbf{R}), \qquad \langle \delta_a * T_2, \varphi \rangle = \langle T_2, \varphi(\cdot - a) \rangle$$

ightarrow Translation of the distribution

Support of a distribution

Definition

Let $T \in \mathcal{S}(\mathbf{R})'$.

Define Ω as the largest open set of **R** such that

$$support(\varphi) \subseteq \Omega \implies \langle T, \varphi \rangle = 0$$

By extension, the (closed) set $\operatorname{support}(T) \stackrel{\text{def.}}{=} \Omega^c$ is called the **support** of the distribution

Support of a distribution

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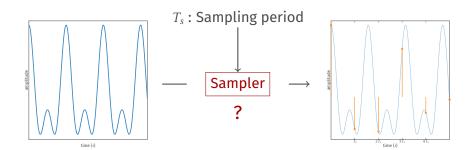
Examples:

- ▶ support(δ_a) = {a}
- ightharpoonup support $(\delta_a + \delta_{-a}) = \{-a, a\}$

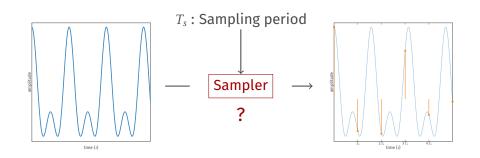
(with new ingredients)

Back to our sampling problem

Ideal Sampling



Ideal Sampling



Observation model: $\equiv \mathbf{x} \times \mathsf{Dirac}$ comb with period T_s :

$$x_e \stackrel{\text{def.}}{=} \sum_s x(jT_s) \delta_{jT_s} \stackrel{?}{=} \coprod_{T_s} x$$

Such a definition is sounded

Is the distribution $\sum_{j\in\mathbb{Z}} x(jT_s)\delta_{jT_s}$ sounded?



Said differently, does the "distribution-function product" $\coprod_{T_n} x$ mean something?

Such a definition is sounded

Is the distribution $\sum_{j \in \mathbb{Z}} x(jT_s) \delta_{jT_s}$ sounded?

Said differently, does the "distribution-function product" $\coprod_{T_s} x$ mean something?

Theorem

Let $T \in \mathcal{S}(\mathbf{R})'$ and $x \in \mathcal{C}^{\infty}(\mathbf{R})$ be such that

$$\forall r \geq 0, \exists C_r > 0 \text{ and } n_r \in \mathbb{N}: \quad |x^{(r)}(t)| \leq C_r (1+|t|)^{n_r}.$$

Then the product Tx defined as

$$\forall \varphi \in \mathcal{S}(\mathbf{R}), \qquad \langle \mathit{Tx}, \varphi \rangle \stackrel{\mathsf{def.}}{=} \langle \mathit{T}, \mathit{x} \varphi \rangle$$

is a tempered distribution

Since $x\varphi \in \mathcal{S}(\mathbf{R})$.

The set of such functions is denoted $\mathcal{O}(R)$

Wrapping up: main ingredients

1. The sensing operator: The (linear) operator defined by

$$M \colon \mathcal{O}(\mathbf{R}) \longrightarrow \mathbf{R}^{\mathbf{Z}}$$
 $M \colon \mathcal{O}(\mathbf{R}) \longrightarrow \mathcal{S}(\mathbf{R})'$
 $x \longmapsto \{x(jT_s)\}_{j \in \mathbf{Z}}$ or $x \longmapsto \coprod_{T_s} x$

- 2. The target set:
- 3. The decoder:

$$\forall x \in \mathcal{X}_{\text{target}}, \quad D(M(x)) = x$$

Question: Does the sampling process affect the spectrum?

$$\mathcal{F}[\coprod_{T_s} x] =$$

Two (rather technical) admitted results

Fourier transform of a product (admitted)

Let x be such that T_x has a compactly supported distribution.

Then
$$\mathcal{F}[\coprod_{T_{i}}x]=\mathcal{F}[\coprod_{T_{i}}]*\mathcal{F}[T_{x}]$$

(The result holds under broader assumptions)

Two (rather technical) admitted results

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$$\mathcal{F}[\coprod_{T_s} x] = \mathcal{F}[\coprod_{T_s}] * \mathcal{F}[T_x]$$

(The result holds under broader assumptions)

Fourier transform of a dirac comb (admitted)

$$\mathcal{F}[\mathrm{III}_{T_s}] = rac{1}{T_c} \mathrm{III}_{^1/T_s}$$

Question: Does the sampling process affect the spectrum?

$$\mathcal{F}[\coprod_{T_s} x] \stackrel{\mathsf{Result 1.}}{=} \mathcal{F}[\coprod_{T_s}] * \mathcal{F}[T_x]$$

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$$\mathcal{F}[\coprod_{T_s} x] \stackrel{\mathsf{Result 1.}}{=} \mathcal{F}[\coprod_{T_s}] * \mathcal{F}[T_x]$$

$$\stackrel{\mathsf{Result 2.}}{=} F_s \coprod_{F_s} * \mathcal{F}[T_x]$$

where $F_s = \frac{1}{T_s}$ is the sampling frequency

Question: Does the sampling process affect the spectrum?

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$$\stackrel{\text{Distribu.}}{=} F_s \sum_{i \in \mathbf{Z}} (\delta_{jF_s} * \mathcal{F}[T_x])$$

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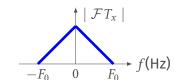
Sampling in the time domain ⇒ Periodic spectrum

Spectrum of band-limited signals

Band-limited signals

A band-limited signal x is a signal whose Fourier transform density has **bounded support**

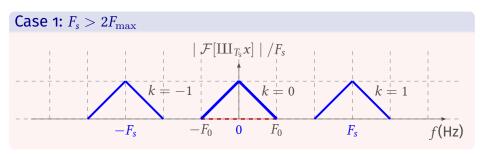
Let \mathbf{x} be a band-limited signal whose TF has support $[-F_0, F_0]$

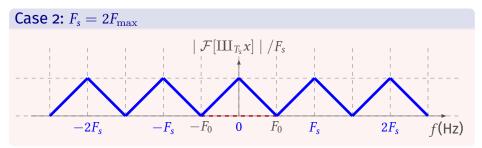


Question: Spectrum of $\coprod_{T_n} \mathbf{x}$?

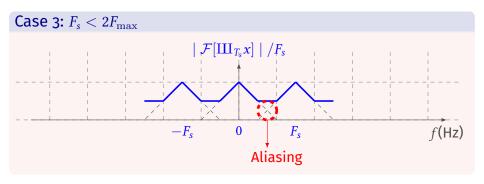
$$\mathcal{F}[\coprod_{T_s} \mathbf{x}] = F_s \sum_{j=1}^{+\infty} \delta_{jF_s} * \mathcal{F} T_x$$

Spectrum of a band-limited signal





Spectrum of a band-limited signal



The signal is **deteriorated** \implies **Loss** of information

Shannon-Nyquist sampling theorem

Question: Necessary condition on F_s such that no information is lost?

Nyquist-Shannon sampling theorem

If a function x(t) contains no frequencies higher than F_{max} hertz, it is **completely determined** by giving its ordinates at a series of points spaced $T_s \geq \frac{1}{2E_{\text{max}}}$ seconds apart.

→ Sufficient condition

 $F_{\rm s}=2F_{\rm max}$ is dubbed "Nyquist frequency"

Interpolation formula

Interpolation formula

If $F_s \geq 2F_{
m max}$, then T_x is equal to the regular distribution spanned by

$$t \longmapsto \sum_{j \in \mathbb{Z}} x(jT_s) \operatorname{sinc}(F_s(t-jT_s))$$

(Informal) proof: if $F_s > 2F_{\rm max}$ the spectrum of x can be recovered with an ideal low pass filter



This operation (operating on the spectrum) is common in signal processing and often referred to as "filtering"

Interpolation formula

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If $F_s \geq 2F_{\max}$, then T_x is equal to the regular distribution spanned by

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(Informal) proof: if $F_s > 2F_{\text{max}}$ the spectrum of x can be recovered with an ideal low pass filter

$$\mathcal{F}T_{x} \stackrel{\mathcal{S}(\mathbf{R})'}{=} \mathcal{F}[\coprod_{T_{s}} x] \cdot \frac{1}{F_{s}} \prod_{F_{s}/2} (f)$$

$$\Rightarrow$$

$$T_{x} \stackrel{\mathcal{S}(\mathbf{R})'}{=} \coprod_{T_{s}} x * T_{\operatorname{sinc}(2\pi F_{s} \cdot)}$$

$$\stackrel{\mathcal{S}(\mathbf{R})'}{=} \sum_{j \in \mathbf{Z}} x[jT_{s}] \delta_{jT_{s}} * T_{\operatorname{sinc}(2\pi F_{s} \cdot)}$$

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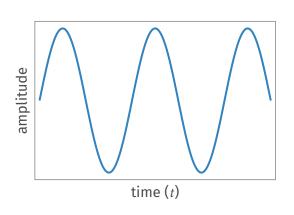
- 2. **The target set**: The set of functions which admit a Fourier transform which is compactly supported in $[-F_s, F_s]$
- 3. The decoder:

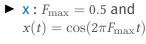
$$M: \quad \mathbf{R}^{\mathbf{Z}} \longrightarrow \mathcal{S}' \\ \{y_j\}_{j \in \mathbf{Z}} \longmapsto \sum_{i \in \mathbf{Z}} y_i \operatorname{sinc}(F_s(\cdot - jT_s)) \quad \text{if} \quad F_s \ge 2F_{\max}$$

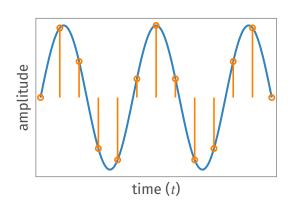
4. The accuracy criterion: "noiseless exact recovery", i.e.

$$\forall x \in \mathcal{X}_{\text{target}}, \quad D(M(x)) \stackrel{\mathcal{D}(\mathbf{R})'}{=} T_x$$

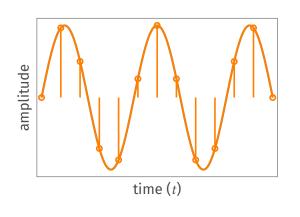
Conclusion



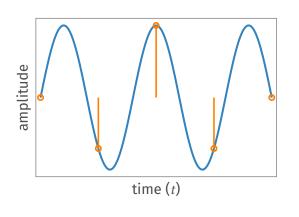




- $ightharpoonup x: F_{\max} = 0.5 \text{ and}$ $x(t) = \cos(2\pi F_{\max} t)$
- $T_s = 0.4 \Longrightarrow$ $F_s = 2.5 > 2F_{\text{max}}$

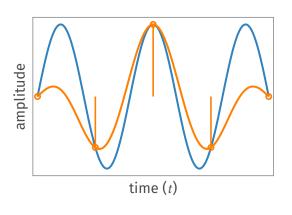


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- ► **x** : $F_{\text{max}} = 0.5$ and $x(t) = \cos(2\pi F_{\max} t)$
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This is a legitimate question



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Rappels: a linear model is characterized by

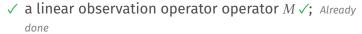
 \checkmark a linear observation operator operator $M \checkmark$; Already done





This is a legitimate question

Rappels: a linear model is characterized by





 \checkmark a vector space as choices for $\mathcal{X}_{\text{target}}$; By linearity of the Fourier transform



This is a legitimate question

Rappels: a linear model is characterized by



- \checkmark a linear observation operator operator $M \checkmark$; Already done
- \checkmark a vector space as choices for $\mathcal{X}_{\text{target}}$; By linearity of the Fourier transform
- ► (an orthogonal projection as "best" decoder)

 In the finite-dimensional case

A concluding remark

If we restrict ourself to signals $x \in \mathcal{L}^2(\mathbf{R})$

$$ightharpoonup \mathcal{L}^2(\mathbf{R})$$
 is a Hilbert space equipped with the inner product

$$\langle x_1, x_2 \rangle = \int x_1(t) \overline{x_2}(t) dt$$

► The family of function

The family of function
$$\forall j, \quad m_j: t \mapsto \frac{1}{\sqrt{T}} \operatorname{sinc}\left(\frac{t-jT_s}{T_s}\right)$$

is an orthonormal basis of the set of functions with Fourier

transform supported in $[-F_s, F_s]$

 \blacktriangleright For all signals x with Fourier transform supported in $[-F_s, F_s]$,

 $\langle m_i, x \rangle = x(iT_s)$

▶ The proposed decoder thus rewrites
$$t \mapsto \sum \langle m_j, x \rangle m_j(t)$$



It is an orthogonal projection!

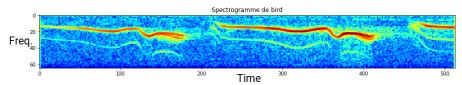
Bonus: Time-Frequency

analysis, wavelets

Time-frequency analysis

► In most domains (audio processing...), the frequency spectrum *itself* evolves in time.

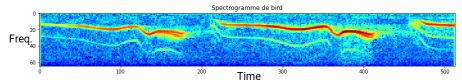




Time-frequency analysis

- ► In most domains (audio processing...), the frequency spectrum *itself* evolves in time.
- ► We want to analyze the frequency component *localized in time*, hence "time-frequency analysis"

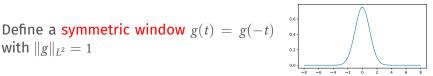




► Popular tool in audio: Windowed Fourier Transform, also called Short Time Fourier Transform (STFT) in its discrete version.

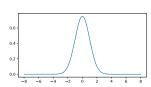
with $||g||_{L^2} = 1$

▶ Popular tool in audio: Windowed Fourier Transform, also called **Short Time Fourier Transform** (STFT) in its discrete version.



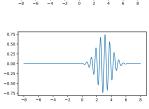
► Popular tool in audio: Windowed Fourier Transform, also called Short Time Fourier Transform (STFT) in its discrete version.

Define a symmetric window g(t) = g(-t) with $\|g\|_{L^2} = 1$



Time-frequency atoms are localized in time and frequency:

$$g_{u,\xi}(t) = \underbrace{e^{2i\pi\xi t}}_{\text{Freq. } \mathcal{E}} \underbrace{g(t-u)}_{\text{Times}}$$



The Windowed Fourier Transform is defined as

$$Sx(u,\xi) = \langle x, g_{u,\xi} \rangle_{L^2} = \int x(t)g(t-u)e^{-2i\pi\xi t}dt$$

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 \blacktriangleright Analyze the signal x around time u and frequency f

The **Windowed Fourier Transform** is defined as

$$Sx(u,\xi) = \langle x, g_{u,\xi} \rangle_{L^2} = \int x(t)g(t-u)e^{-2i\pi\xi t}dt$$

► Analyze the signal x around time u and frequency f

By Parseval's identity,

$$Sx(u,\xi) = \langle \hat{x}, \hat{g}_{u,\xi} \rangle_{L^2}$$

and by the shift formula

$$\hat{g}_{u,\xi}(f) = \hat{g}(f-\xi)e^{-2i\pi u(f-\xi)}$$

- ▶ since g is real symmetric, \hat{g} is real symmetric (ex: for g Gaussian, \hat{g} is Gaussian!)
- ► Also performs a windowed transform of \hat{x} around ξ (and "inverse frequency" u)

$$Sx(u,\xi) = \langle x, g_{u,\xi} \rangle_{L^2} = \langle \hat{x}, \hat{g}_{u,\xi} \rangle_{L^2}$$

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Considering $\mathbb{P}=|g_{u,\xi}|^2$ and $\hat{\mathbb{P}}=|\hat{g}_{u,\xi}|^2$ as **probabilities densities** (since $\|g_{u,\xi}\|_{L^2}=\|g_{u,\xi}\|_{L^2}=1$):

$$\mathbb{E}(\mathbb{P}) = u$$
 $\operatorname{Var}(\mathbb{P}) = \sigma_{\mathbb{P}}^2$ $\operatorname{Var}(\mathbb{P}) = \sigma_{\hat{\mathbb{D}}}^2$

$$Sx(u,\xi) = \langle x, g_{u,\xi} \rangle_{L^2} = \langle \hat{x}, \hat{g}_{u,\xi} \rangle_{L^2}$$

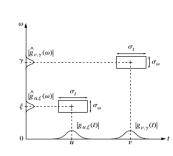
Considering $\mathbb{P}=|g_{u,\xi}|^2$ and $\hat{\mathbb{P}}=|\hat{g}_{u,\xi}|^2$ as **probabilities densities** (since $\|g_{u,\xi}\|_{L^2}=\|g_{u,\xi}\|_{L^2}=1$):

$$\mathbb{E}(\mathbb{P}) = u$$
 $\operatorname{Var}(\mathbb{P}) = \sigma_{\mathbb{P}}^2$ $\operatorname{Var}(\mathbb{P}) = \sigma_{\hat{\mathbb{P}}}^2$

and by Heisenberg's thm:

$$\sigma_{\mathbb{P}}^2 \sigma_{\hat{\mathbb{P}}}^2 \geq 1/4$$

with equality if and only if g is (any) **Gaussian**.



Reconstruction

Thm: inversion formula

If $x \in \mathcal{L}^2$,

$$x(t) = \iint Sx(u,\xi)g(t-u)e^{2i\pi\xi t}d\xi du$$

and

and
$$\|x\|_{L^2}^2 = \int\!\!\int |Sx(u,\xi)|^2 \mathrm{d}\xi \mathrm{d}u$$

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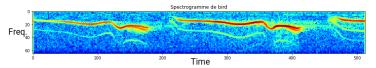
and

$$||x||_{L^2}^2 = \iint |Sx(u,\xi)|^2 \mathrm{d}\xi \mathrm{d}u$$

- ► no information is lost (true "transform")
- $ightharpoonup Sx \in \mathcal{L}^2(\mathbf{R}^2)$

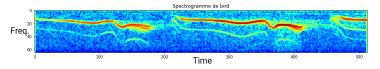
Example of applications

Spectrogram $|Sx(u,\xi)|^2$ used a lot in audio.

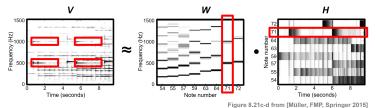


Example of applications

Spectrogram $|Sx(u,\xi)|^2$ used a lot in audio.



Ex: wav2midi with sparse matrix factorization.

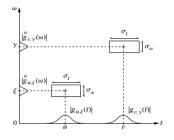


More precisely, "Non-negative matrix factorization" (NMF), a low-rank factorization of a non-neg. matrix (the spectrogram) into non-neg. components, frequency/activation. See https://www.audiolabs-erlangen.de/resources/MIR/FMP/C8/C8S3_NMFbasic.html

Problem with WFT



- ► The WFT analyzes content with the same window length
- ► Element can have different durations... (especially low-frequency)
- ► Human hearing is spread "logarithmically" in frequency



Wavelet transform

A wavelet is a normalized function $\psi \in \mathcal{L}^2(\mathbf{R})$ with a zero average:

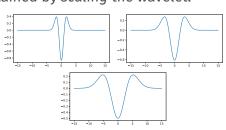
$$\int \psi(t)dt = 0, \quad \|\psi\| = 1$$

Time-frequency atoms are obtained by scaling the wavelet:

$$\psi_{u,s}(t) = \frac{1}{\sqrt{s}}\psi\left(\frac{t-u}{s}\right)$$

Wavelet transform:

$$Wx(u,s) = \langle x, \psi_{u,s} \rangle$$



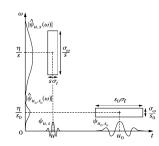
Assume ψ is analytic for simplicity (i.e. zero negative frequency, see Mallat book).

Assume ψ is analytic for simplicity (i.e. zero negative frequency, see Mallat book).

Considering $\mathbb{P}=|\psi_{u,s}|^2$ and $\hat{\mathbb{P}}=|\hat{\psi}_{u,s}|^2$ as **probabilities densities,** and $\mathbb{P}_1,\hat{\mathbb{P}}_1$ the same for reference s=1 (u does not matter). Then:

$$\begin{split} \mathbb{E}(\mathbb{P}) &= u & \mathsf{Var}(\mathbb{P}) = s^2 \mathsf{Var}(\mathbb{P}_1) \\ \mathbb{E}(\hat{\mathbb{P}}) &= \mathbb{E}(\hat{\mathbb{P}}_1)/s & \mathsf{Var}(\mathbb{P}) = \mathsf{Var}(\mathbb{P}_1)/s^2 \end{split}$$

The "right" multiplicative scale! (for human ears)



Scalogram



