

Experiment 1-B

Transient Response of RLC Circuit
with Initial Conditions.

Aim: To determine the transient response of an RLC network in terms of the parameters σ , ξ , ω , ω_0 and the initial conditions $i_L(0^-)$, $v_C(0^-)$.

Theory: ξ is the normalised damping constant defined as -

$$\xi = \frac{\alpha}{\omega_0}$$

where α is $R/2L$ and ω_0 is the natural frequency $1/\sqrt{LC}$.

Under damping conditions, the frequency of oscillation drops down to $\omega_0 \sqrt{1-\xi^2}$ where ξ is damping constant of that network.

Theoretically, RLC networks with non zero I₀, initial conditions can be solved by formulating a differential eqn based on the I-V characteristics of the components R, L, C.

$$V_R = I_R R \quad (\text{Ohm's law})$$

$$i_C(t) = C \frac{dV_C}{dt}$$

$$V_L(t) = L \frac{di_L}{dt}$$

Besides, the RLC network can be transformed to an s -domain network and all the components R, L, C can be treated as nothing but resistances which follow Ohm's law, but have their values in s -domain.

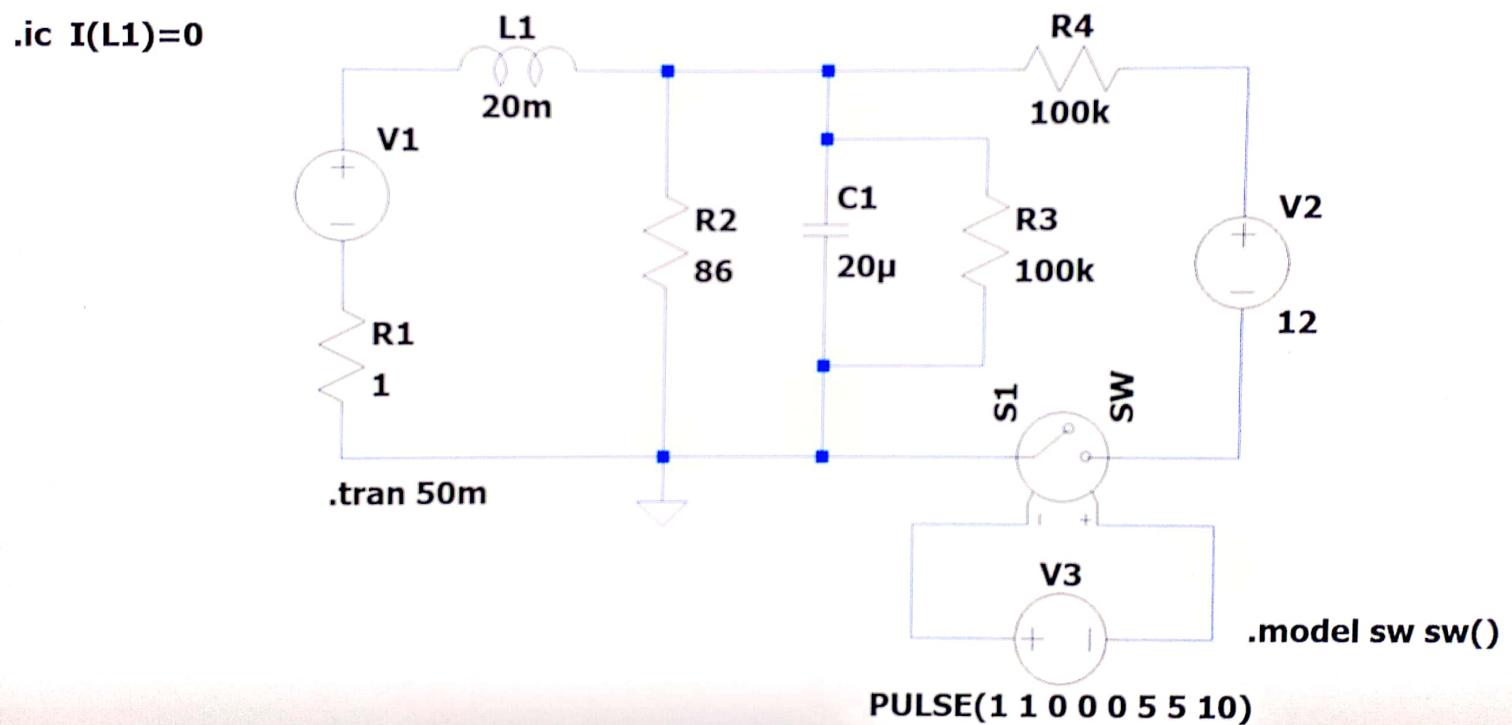
For a circuit with non-zero initial conditions however a constant voltage or current source might have to be imagined to write the resistances/impedances of L and C in s -domain as Ls and $\frac{1}{Cs}$ respectively.

These techniques will be studied more deeply in the calculations part.

LTSpice Circuit Diagrams:

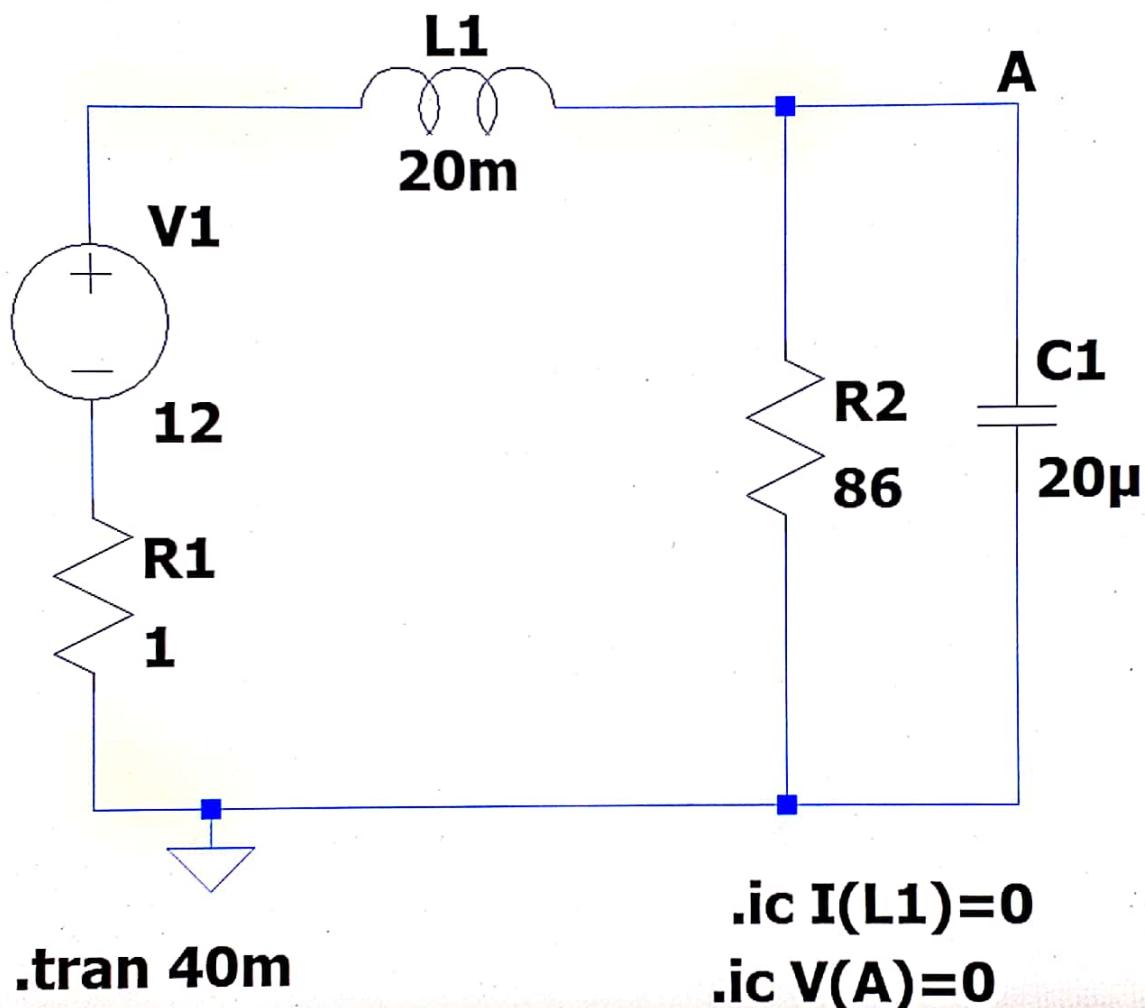
The circuit diagrams for this experiment follow from the next page.

EXPERIMENT O2 - PART B
CIRCUIT O1



EXPERIMENT O2 - PART B

CIRCUIT O2



Observations:

Part 01 - The voltage source V_1 was switched on and the transient was captured across the $20\mu F$ capacitor. The waveform is shown on the next page.

$$T \text{ (time constant)} = 6.12 \text{ ms.}$$

Part 02 - The value of R_2 was changed to 75Ω to adjust the current through 20mH inductor to less than 0.75A .

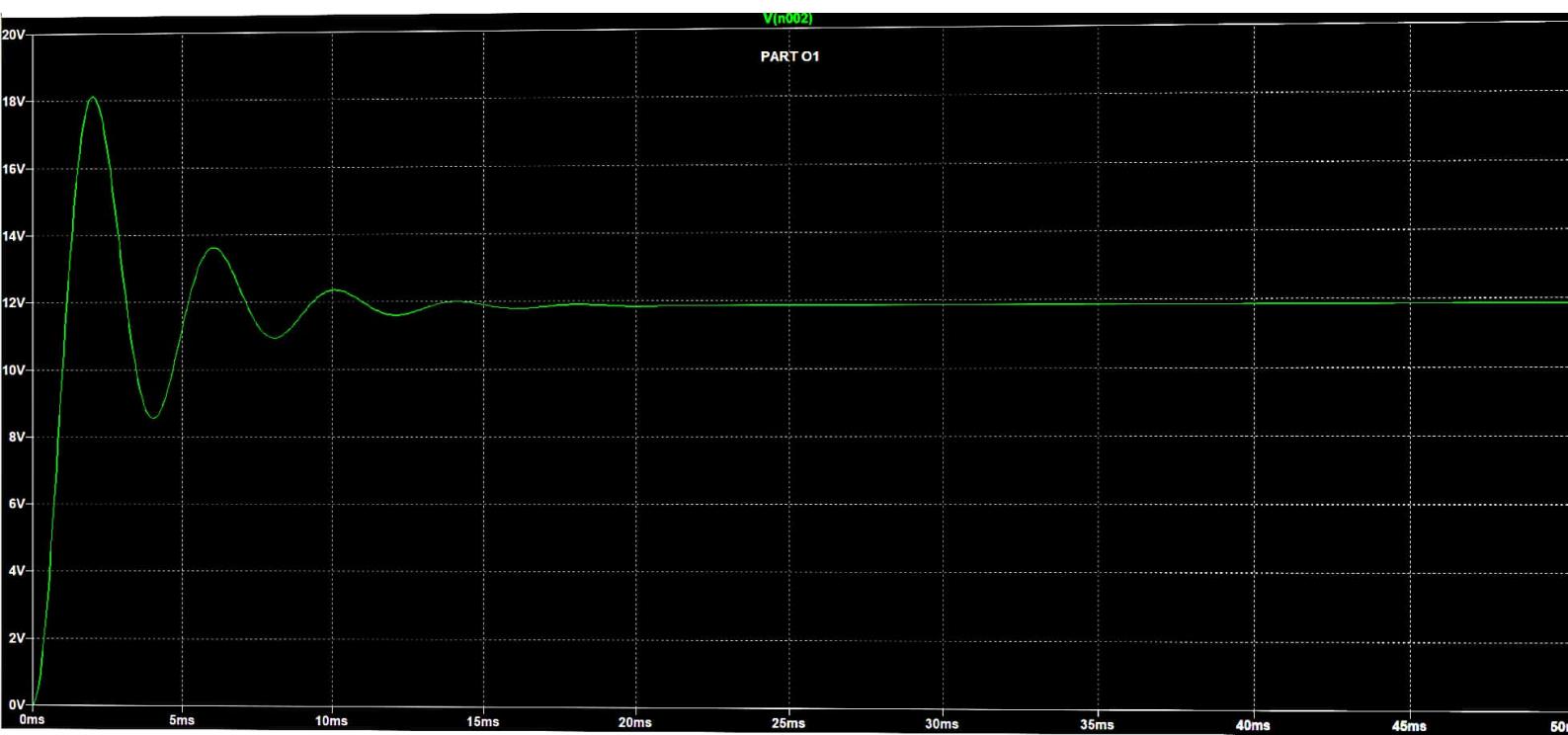
The transient of the inductor current was recorded via capturing the output across 1Ω shunt resistance.

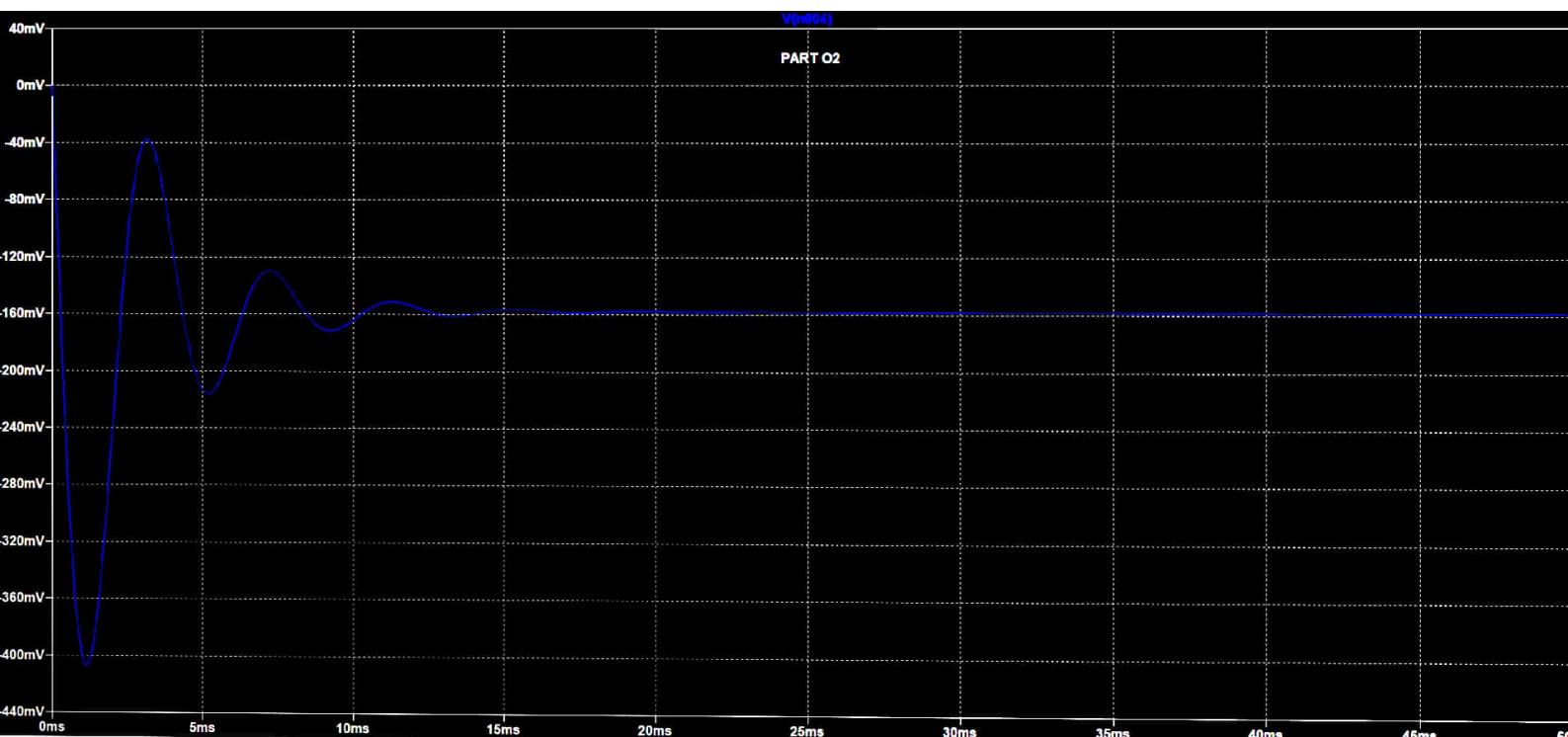
The waveform is shown on the next page.

$$T \text{ (time constant)} = \frac{3.31 \text{ ms}}{6.25 \text{ steps}}$$

Part 03 - The voltage across the capacitor was captured for different values of $V_c(0^-)$. ($C = 20\mu F$).

$V_c(0^-)$	Max overshoot	f(Hz)	Time Constant	$V_c(\infty)$
12V	+3.28V	251.26	2.59ms	11.86V
5V	+163.06V	248.14	2.91ms	11.86V
2.5V	+222.45V	245.70	3.94ms	11.86V





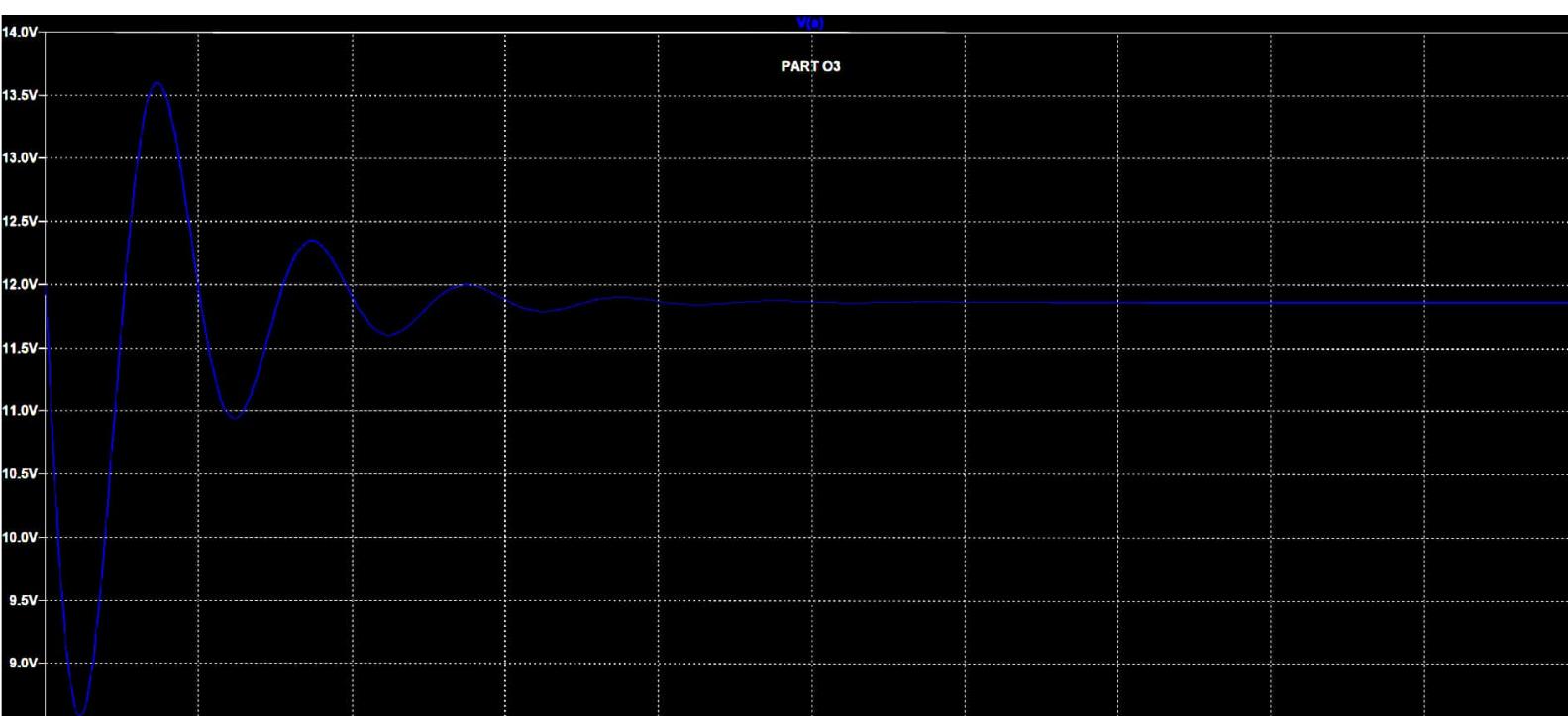
Part 04 The voltage across the capacitor was captured for different capacitances ($v_c(0^-) = 5V$, $R_2 = 86\Omega$).

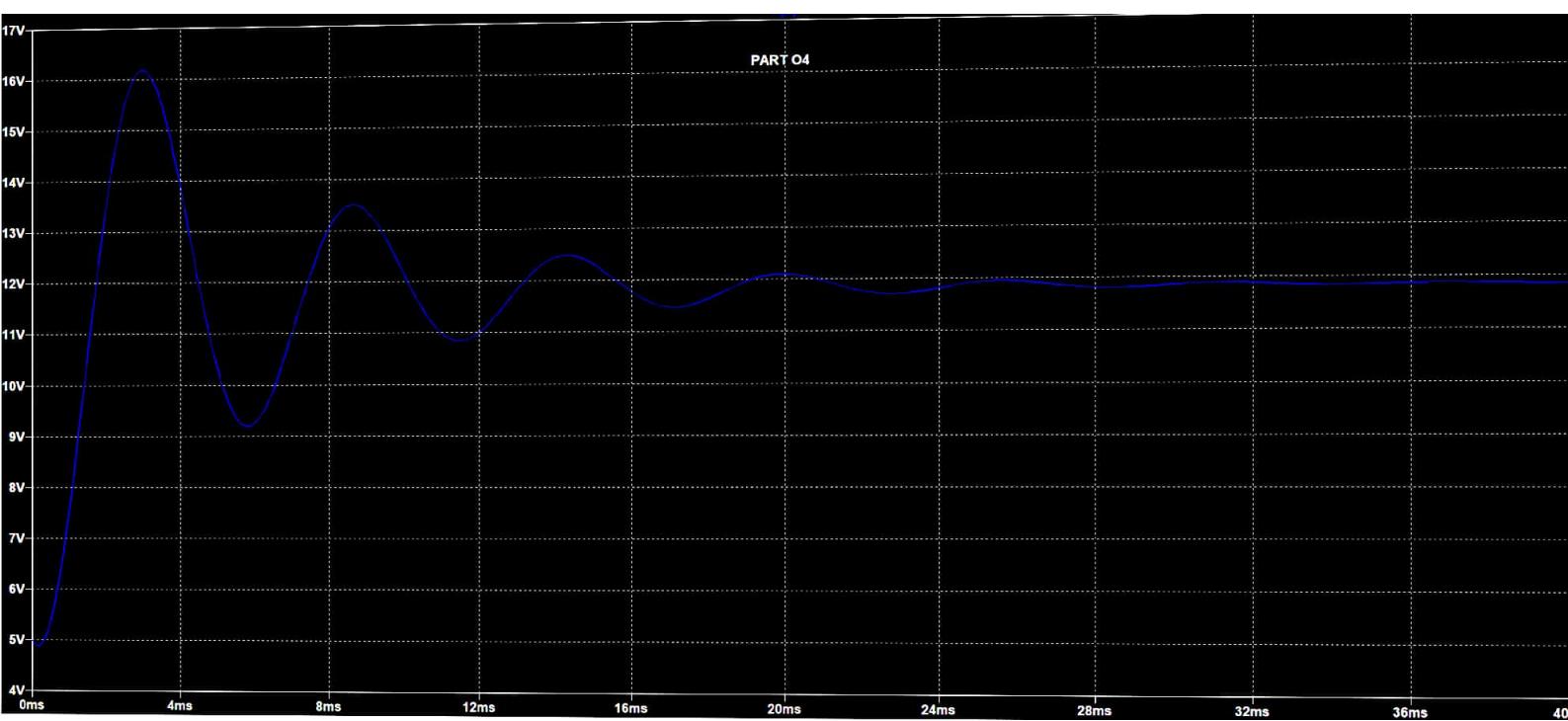
C	Max Overshoot	f(Hz)	Time Constant	$v_c(\infty)$
40μF	+4.31V	173.61	8.62 ms	11.86V
60μF	+4.56V	142.25	10.47 ms	11.86V
100μF	+4.78V	111.23	17.91 ms	11.86V

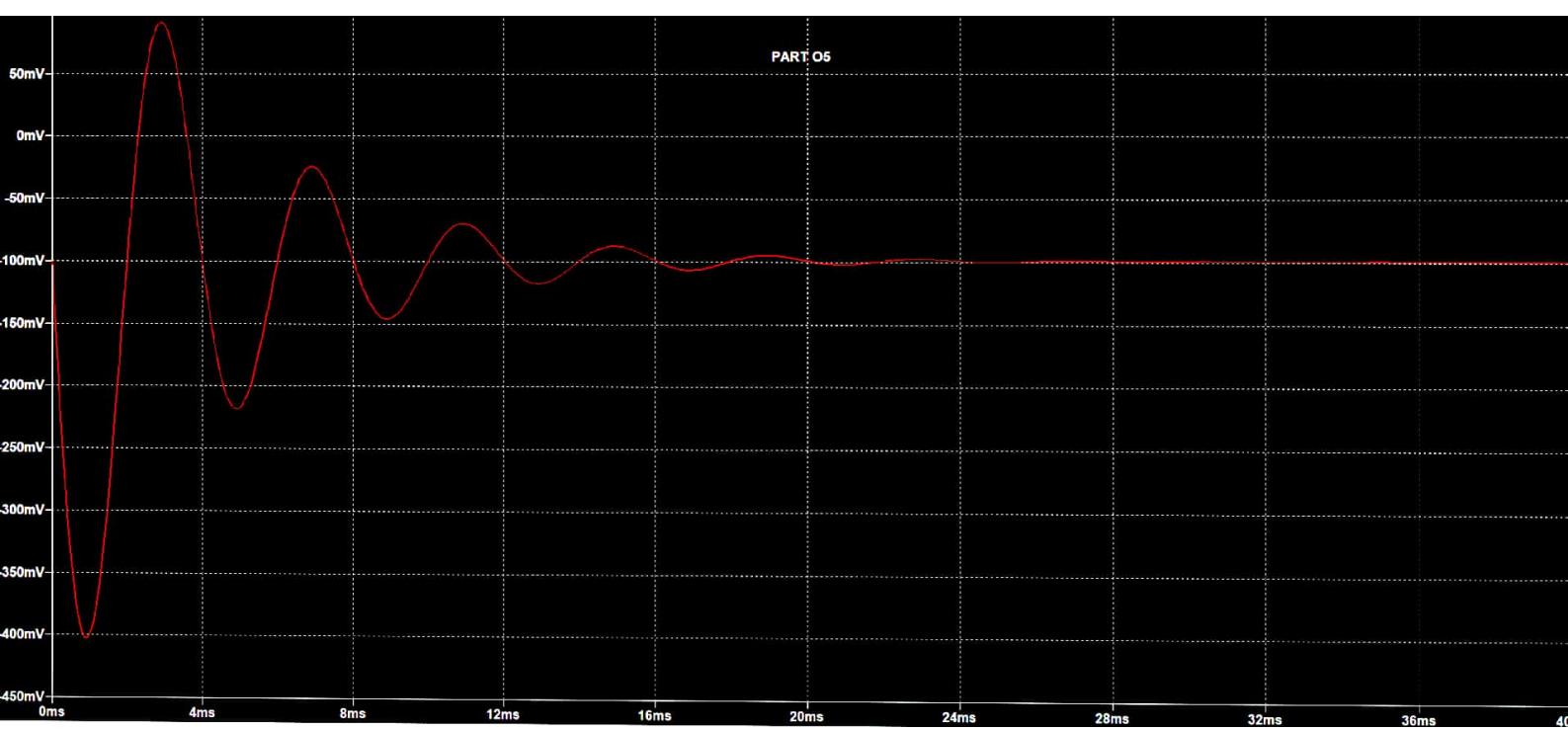
Part 05 The voltage across the shunt resistance $R_1 = 1\Omega$ was captured for different values of $i_L(0^-)$. ($C = 20\mu F$)

$i_L(0^-)$	R_2	M.O.	f	Time Constant	$v_{R_1}(0^-)$	$v_{R_1}(\infty)$
(A)	(Ω)	(mV)	(Hz)		(mV)	(mV)
0.10	120	176.27	248.76	4.47ms	99.75	99.17
0.30	40	128.96	213.67	9.69ms	299.74	292.67
0.60	20	103.99	198.81	14.88ms	599.72	571.40

* The plots obtained on LTSpice for the first entries in table 3, 4 and 5 follow from the next page.







Simple Calculations:① Analytical determination of $v_c(t)$.

By KVL, $V_1 - L \frac{di}{dt} - v_c - R_1 i = 0$

By KCL, $i = C \frac{dv_c}{dt} + \frac{v_c}{R_2}$

$$\therefore LC \frac{d^2 v_c}{dt^2} + \left(\frac{L}{R_2} + R_1 C \right) \frac{dv_c}{dt} + v_c = V_1$$

Now $v_c(0^-) = 5V$, $\dot{v}_c(0^-) = 0 \text{ V/s.}$

$$L = 20 \text{ mH}, C = 20 \text{ nF}$$

$$R_2 = 86 \Omega, R_1 = 1 \Omega, V_1 = 12 \text{ V}$$

$$\Rightarrow 4 \times 10^{-7} \frac{d^2 v_c}{dt^2} + 2.52 \times 10^{-4} \frac{dv_c}{dt} + v_c = 12$$

* Taking one sided Laplace transform on both sides.

$$4 \times 10^{-7} (\delta^2 V_c(\delta) - 5\delta) + 2.52 \times 10^{-4} (\delta V_c(\delta) - 5) + V_c(\delta) = 12/\delta.$$

where $V_c(\delta)$ is $\mathcal{L}(v_c(t))$.

* upon rearranging; (and apply partial fractions)

$$V_c(\delta) = \frac{12}{\delta} - \frac{(7(\delta+315) + 2205)}{(\delta+315)^2 + 24 \times 10^5}$$

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-315t

$$L^{-1}(V_c(s)) = (12 - 7e^{-315t} \cos(1549t) - 1.423 e^{-315t} \sin(1549t)) \cdot u(t).$$

$$\Rightarrow V_c(t) = 12 - e^{-315t} (7 \cos(1549t) + 1.423 \sin(1549t))$$

$$\lim_{t \rightarrow \infty} (V_c(t)) = 12 \text{ V. } (V_{c,\max})$$

\therefore The transient part of the output is the term added to 12V.

$$V_c^T(t) = e^{-315t} (7 \cos(1549t) + 1.423 \sin(1549t))$$

upon comparing $V_c^T(t)$ with

$$K_1 e^{-\sigma t} \cos \omega t + K_2 e^{-\sigma t} \sin \omega t.$$

$$K_1 = 7V \quad K_2 = 1.423V$$

$$\omega = 1549 \text{ rad/s}$$

$$\sigma = 315 \text{ sec}^{-1}$$

$$\text{Now } \sigma = \zeta \omega_n \text{ and } \omega = \omega_n \sqrt{1 - \zeta^2}.$$

$$\sqrt{1 - \zeta^2} = \omega \zeta / \sigma$$

Put the values and get

$$\zeta = 0.199.$$

$$\omega_n = \sigma / \zeta = 1580.70 \text{ rad/s.}$$

② Experimental verification of analytical solⁿ of $v_c(t)$.

$$v_c(t) = V_{max} + e^{-\sigma t} (K_1 \cos \omega t + K_2 \sin \omega t)$$

In the observation table, the following values deduced from the waveform are mentioned

$$v_c(\infty) = 11.86 \text{ V}$$

$$\tau = 2.91 \text{ ms}$$

$$f = 248.14 \text{ Hz}$$

$$V_{max} = \lim_{t \rightarrow \infty} v_c(t) = 11.86 \text{ V}$$

$$\sigma = 1/\tau = 343.64$$

$$\omega = 2\pi f = 1559 \text{ rad/s.}$$

$$v_c(0) = V_{max} + K_1 = 5$$

$$\Rightarrow K_1 = 5 - 11.86 = -6.86 \text{ V.}$$

$$v_c(0) = 0$$

$$\Rightarrow K_2 \omega - \sigma K_1 = 0$$

$$\Rightarrow K_2 = -1.51 \text{ V.}$$

$$\therefore v_c(t) = 11.86 - e^{-343.64t} (6.86 \cos(1559t) + 1.51 \sin(1559t))$$

(very similar to analytical solⁿ)

On solving $\sqrt{1-\zeta^2} = \omega \zeta / \sigma$
we get $\zeta = 0.215$

$$\omega_n = \sigma / \zeta = 1596.42 \text{ rad/s.}$$

③ Analytical determination of $i_L(t)$.

$$\text{By KVL, } V_1 - i_L R_1 - L \frac{di_L}{dt} - U_C = 0$$

$$\text{By KCL, } C \frac{dU_C}{dt} + \frac{U_C}{R_2} = i_L - (1)$$

$$U_C = V_1 - i_L R_1 - L \frac{di_L}{dt} - (2)$$

$$\frac{dU_C}{dt} = -R_1 \frac{di_L}{dt} - L \frac{d^2 i_L}{dt^2} - (3)$$

Put (2) and (3) in (1).

$$V_1 = R_2 C \frac{d^2 i_L}{dt^2} + \frac{di_L}{dt} (L + R_1 R_2 C) + (R_1 + R_2) i_L$$

$$\text{Now, } i_L(0^-) = 0.1 \text{ A}, \dot{i}_L(0^-) = 0 \text{ A/s}.$$

$$L = 20 \text{ mH}, C = 20 \mu\text{F}, R_2 = 120 \Omega, \\ R_1 = 1 \Omega, V_1 = 12 \text{ V}.$$

$$4.8 \times 10^{-5} \frac{d^2 i_L}{dt^2} + 2.24 \times 10^{-2} \frac{di_L}{dt} + 12 i_L = 12.$$

Take one-sided Laplace transform on both sides.

$$4.8 \times 10^{-5} (\delta^2 I_L(\delta) - 0.1 \delta) + 2.24 \times 10^{-2} (\delta I_L(\delta) - 0.1) + 12 I_L(\delta) = 12 / \delta.$$

where $I_L(s) = L(i_L(t)u(t))$

- * Upon rearranging and forming partial fractions, we get.

$$I_L(s) = \frac{12}{121s} + \frac{1.23 \times 10^{-4} \times 1570}{(s + 233.33)^2 + (1570)^2} + \frac{8.26 \times 10^{-4}(s + 233.33)}{(s + 233.33)^2 + (1570)^2}$$

- * Taking laplace inverse both sides.

$$\checkmark i_L(t) = \frac{12}{121} + 10^{-4} e^{-233.33t} (1.23 \sin(1570t) + 8.26 \cos(1570t))$$

$$\text{At } t \rightarrow \infty, i_L(t) = 99.17 \text{ mA} = i_{L,\max}$$

$$i_L^T(t) = 10^{-4} e^{-233.33t} (1.23 \sin(1570t) + 8.26 \cos(1570t))$$

(transient)

On comparing $i_L^T(t)$ with
 $K_1 e^{-\sigma t} \cos(\omega t) + K_2 e^{-\sigma t} \sin(\omega t)$.

$$K_1 = 8.26 \times 10^{-4} \text{ A}, K_2 = 1.23 \times 10^{-4} \text{ A}$$

$$\sigma = 233.33 \text{ sec}^{-1}$$

$$\omega = 1570 \text{ rad/s.}$$

$$\text{On solving } J1 - \xi_p^2 = \omega \xi_p / \sigma$$

$$\xi_p = 0.147; \omega_n = \sigma / \xi_p = 1587 \text{ rad/sec}$$

(Q) Experimental verification of the analytically determined $i_L(t)$.

$$i_L(t) = I_{\max} + e^{-\sigma t} (K_1 \cos \omega t + K_2 \sin \omega t)$$

The magnitude of current flowing through inductor in A is same as the magnitude of voltage drop across shunt resistor R_1 in V because its resistance is 1Ω .

* All the values used below are in obs. table

$$i_L(0) = 0.1A$$

$$\text{At } t \rightarrow \infty, i_L(t) = 99.17 \times 10^{-3} A.$$

$$t \rightarrow \infty$$

$$\therefore I_{\max} = 9.917 \times 10^{-2} A.$$

$$i_L(0) = I_{\max} + K_1 = 0.1$$

$$\Rightarrow K_1 = 8.30 \times 10^{-4} A$$

$$\tau = 4.47 \times 10^{-3} \text{ sec (time constant)}$$

$$\sigma = 1/\tau = 223.71 \text{ sec}^{-1}$$

$$f = 248.76 \text{ Hz}$$

$$\omega = 2\pi f = 1563 \text{ rad/sec}$$

$$i_L(0) = 0$$

$$\Rightarrow K_2 \omega = \sigma K_1$$

$$K_2 = \sigma K_1 / \omega = 1.18 \times 10^{-4} A.$$

Put all the values in the eqⁿ of $i_L(t)$.

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 $i_L(t) = 9.917 \times 10^{-2} + e^{-223.7t} (8.30 \cos \omega t + 1.18 \sin \omega t) \times 10^{-4}$

put $\omega = 1563$

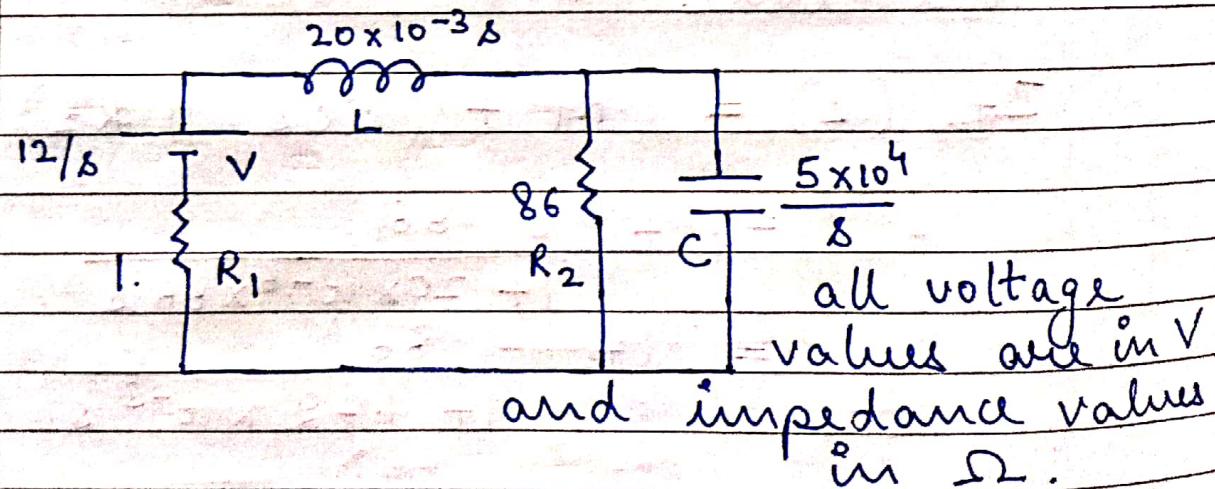
$$i_L(t) = 9.917 \times 10^{-2} + e^{-223.7t} \times 10^{-4} (8.30 \cos(1563t) + 1.18 \sin(1563t))$$

(very similar to the analytical sol^u)

On solving, $\sqrt{1 - \zeta^2} = \omega \zeta / \sigma$

$$\zeta = 0.142, \omega_n = \sigma / \zeta = 1579 \text{ rad/sec}$$

⑤ Finding transform-domain network.



$$Z_i(s) = R_1 \sim (R_2 || C) \sim L$$

$$= R_1 + \frac{R_2 / Cs}{R_2 + 1/Cs} + Ls$$

$$= 1 + \frac{86}{1 + 20 \times 86 \times 10^{-6} s} + 20 \times 10^{-3} s$$

$$Z_i(s) = \frac{3.44 \times 10^{-5} s^2 + 2.17 \times 10^{-2} s + 87}{1 + 1.72 \times 10^{-3} s}$$

$$Z_o(s) = (R_1 + Ls) \parallel (R_2) \parallel (1/Cs)$$

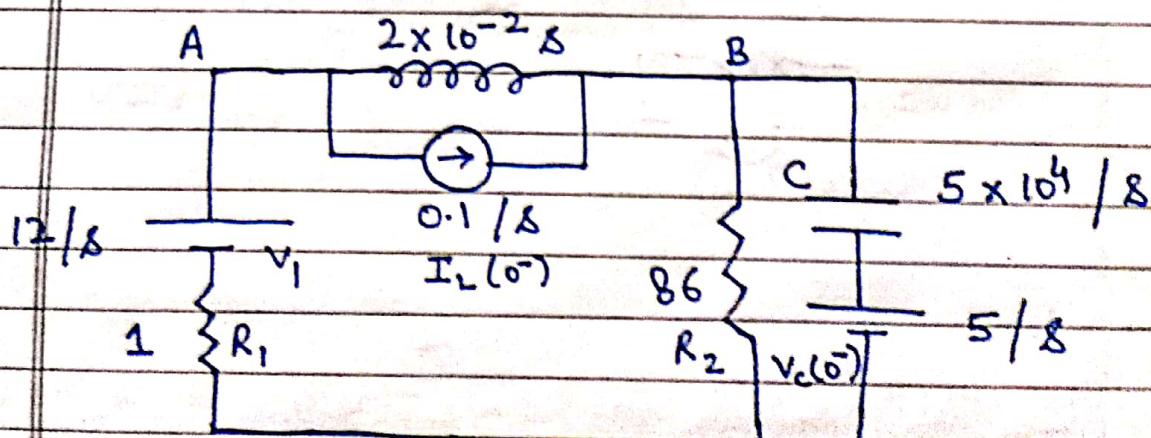
$$= \left[\left(\frac{1}{R_1 + Ls} \right) + \frac{1}{R_2} + Cs \right]^{-1}$$

$$= \left(\frac{1}{1 + 20 \times 10^{-3} s} + \frac{1}{86} + 2 \times 10^{-5} s \right)^{-1}$$

$$Z_o(s) = \frac{86 + 1.72 \times 10^{-3} s}{3.44 \times 10^{-8} s^2 + 2.17 \times 10^{-2} s + 87}$$

⑥ Determining $i_L(t)$ through transfer domain network and Thevenin's theorem

$$I_L(0^-) = 0.1A, V_C(0^-) = 5V$$



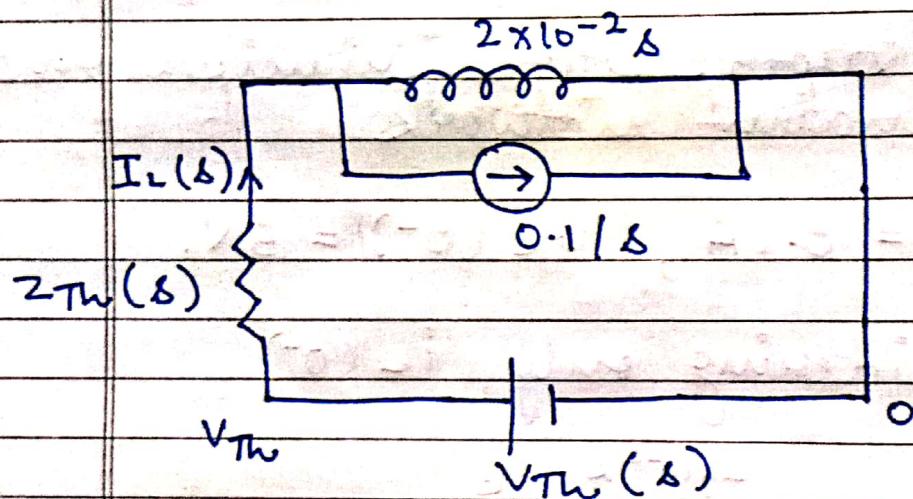
$$Z_{Th}(s) (b/w A, B) \\ = R_1 \sim (R_2 \parallel C)$$

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$$Z_{Th}(\delta) = 1 + \frac{86 \times 5 \times 10^4 / \delta}{86 + 5 \times 10^4 / \delta}$$

$$\begin{aligned} V_{Th}(\delta) (V_A - V_B) &= \frac{12 - 5}{\delta} = \frac{7}{\delta} \\ &= \frac{12 - \frac{5/\delta}{5 \times 10^4 / \delta + 86}}{\delta} \times 86 \\ &= \frac{12}{\delta} - \frac{430}{86\delta + 5 \times 10^4} \end{aligned}$$



$$I_L(\delta) = \frac{0.1 + \frac{V_{Th} - I_L Z_{Th}}{2 \times 10^{-2} \delta}}{\delta}$$

$$I_L(\delta) = \frac{0.1 + 50 V_{Th}}{\delta + 50 Z_{Th}}$$

$$I_L(\delta) = \frac{0.1 \delta^2 + 408.14 \delta + 3.49 \times 10^5}{\delta (\delta^2 + 631.39 \delta + 2.53 \times 10^6)}$$

(put the values of V_{Th} and Z_{Th})

Now expand $I_L(\delta)$ into partial fractions.

$$I_L(s) = \frac{0.108 - 0.0379}{s} \left(\frac{s + 315.69}{(s + 315.69)^2 + 2.43 \times 10^6} \right)$$

$$= 333.02 \left(\frac{1}{(s + 315.69)^2 + 2.43 \times 10^6} \right)$$

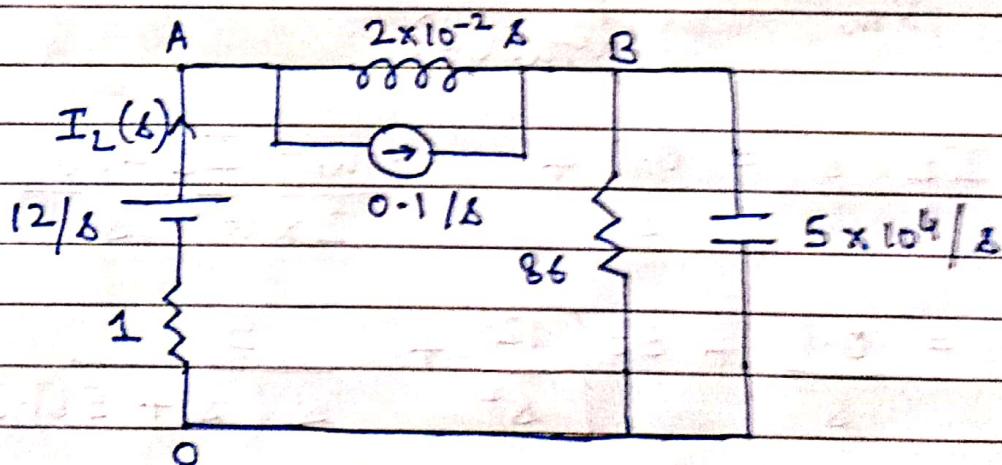
Take Laplace inverse on both sides.

$$i_L(t) = 0.1 - 0.0379 e^{-315.69t} \cos(1559t) + 0.214 e^{-315.69t} \sin(1559t)$$

- ⑦ Applying superposition principle on the s-domain network.

$$i_L(0^-) = 0.1A, v_C(0^-) = 5V$$

(i) considering only $i_L(0^-)$



$$V_A = \frac{12 - I_L(s)}{s}$$

$$V_B = \frac{I_L \times 5 \times 10^4 / s \times 86}{\frac{5 \times 10^4}{s} + 86} = \frac{5 \times 10^4 I_L}{s + 581.39}$$

$$I_L(\delta) = \frac{V_A - V_B}{2 \times 10^{-2} \delta} + \frac{0.1}{\delta}$$

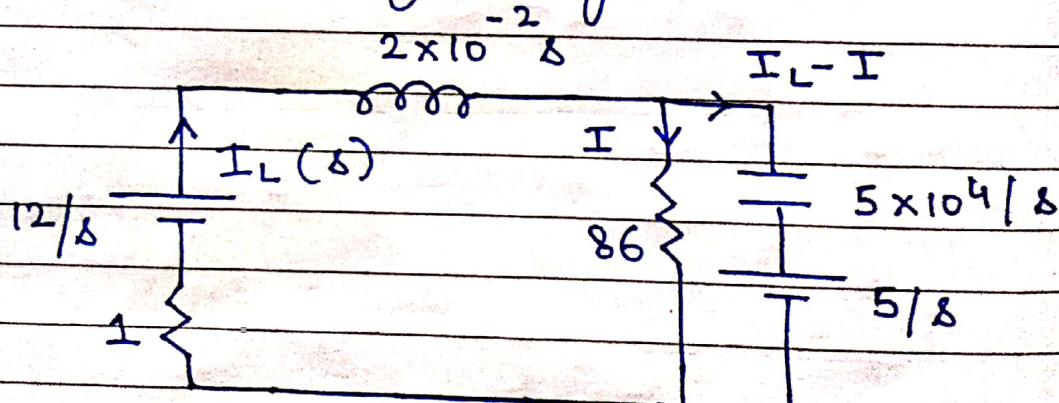
$$\delta I_L = 0.1 + 50 \left(\frac{12 - I_L}{\delta} - \frac{5 \times 10^4 I_L}{\delta + 581.39} \right)$$

$$I_L = \frac{(0.1 + 600/\delta)(\delta + 581.39)}{25 \times 10^5 + (\delta + 50)(\delta + 581.39)}$$

Break into partial fractions and take Laplace inverse on both sides.

$$i_L(t) = 0.138 - 0.038 e^{-315.69t} \cos(1558t) + 0.374 e^{-315.69t} \sin(1558t)$$

(ii) considering only $V_C(0^-)$



$$86 I = \frac{5}{\delta} + \frac{5 \times 10^4}{\delta} (I_L - I) \quad (1)$$

$$\frac{12}{\delta} - I_L - 2 \times 10^{-2} \delta I_L = 86 I \quad (2)$$

Put I from (1) in (2)

$$I_L(\delta) = \frac{12(\delta + 581.39) - 5\delta}{\delta(\delta + 581.39)(2 \times 10^{-2} \delta + 1) + 5 \times 10^4 \delta}$$

Break into partial fractions and take Laplace inverse on both sides

$$i_{L_2}(t) = 0.138 - 0.138 e^{-315.69t} \cos(1558t) + 0.197 e^{-315.69t} \sin(1558t).$$

By superposition theorem, add the two eq's together.

$$i_L(t) = i_{L_1}(t) + i_{L_2}(t)$$

$$i_L(t) = 0.276 - 0.176 e^{-315.69t} \cos(1558t) + 0.571 e^{-315.69t} \sin(1558t)$$

Discussion: The transient response of RLC network with initial conditions was successfully studied. Various parameters like σ , w , w_n , ξ , $f_c(\infty)$, $v_c(0^+)$ obtained from the practical were extremely close to the values determined theoretically. Differential eq's were formulated theoretically to confirm the plots obtained from LT Spice. The plots followed the behaviour that was predicted by the theoretical eq's. The values of experimentally determined T , f , w , w_n and σ were used to find the closed form eq's of $i_L(t)$ and $v_c(t)$. These eq's

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were very similar to the eqⁿs derived from scratch using the initial value conditions. Besides the eqⁿs of $i_L(t)$ (or $v_c(t)$) obtained from superposition theorem were also obeying the practical observations.

The analytical and experimental results differ slightly and that might be because of the number of decimal places we take in our calculations. Moreover the values of time constant and frequency that were obtained from the practical LTSpice plots might not be 100% correct. An explicit eqⁿ gives a more correct estimate of these values than just a plot.