

ECE310 Recitation Fall 2023

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1. The frequency response of a generalized linear phase (GLP) filter can be expressed as $H(w) = A(w)e^{j(\alpha w + \beta)}$ where $A(w)$ is a real function, and α, β are real constants. For each of the following filters, determine whether it is a GLP filter. If it is, find $A(w)$, α , and β , and indicate whether it is also a linear phase filter.

a)

$$\{h_n\}_{n=0}^2 = \{-1, 4, 5\}$$

b)

$$\{h_n\}_{n=0}^2 = \{2, 0, -2\}$$

c)

$$\{h_n\}_{n=0}^3 = \{0, 1, 1, 0\}$$

a) **Solution:** The filter doesn't display any form of symmetry. Hence, this is not a GLP filter.

b) **Solution:** The filter displays an odd symmetry about the midpoint as $h[0] = -h[2]$ and the midpoint $h[1]$ is zero. Therefore, this is a GLP filter.

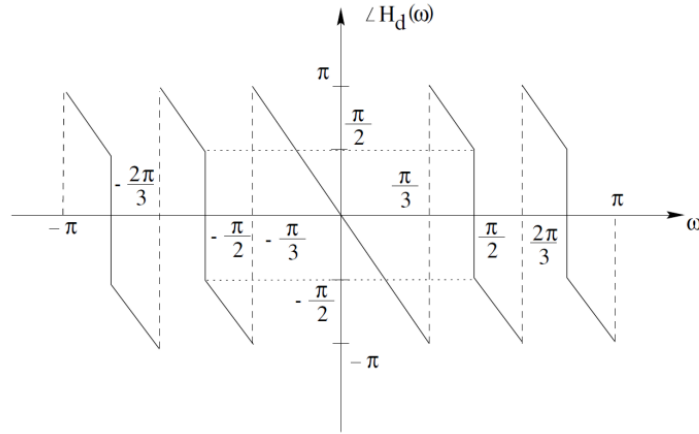
$H_d(\omega) = 2 - 2e^{-j2\omega} = 2e^{-j\omega}(e^{j\omega} - e^{-j\omega}) = 2e^{-j\omega}(2j\sin(\omega)) = 4j\sin(\omega)e^{j(-\omega + \frac{\pi}{2})}$. $A(\omega) = 4\sin(\omega)$, $\alpha = 1$ and $\beta = \frac{\pi}{2}$. $A(\omega)$ has a zero crossing at $\omega = 0$ and thus there will be a π jump in the phase plot. Hence, this is not a strict linear phase filter.

c) **Solution:** The filter displays an even symmetry and an odd order. Therefore, this filter is a Type II GLP filter.

$H_d(\omega) = e^{-j\omega} + e^{-j2\omega} = e^{-j1.5\omega}(e^{-j0.5\omega} + e^{j0.5\omega}) = 2\cos(0.5\omega)e^{-j1.5\omega}$. $A(\omega) = 2\cos(0.5\omega)$, $\alpha = 1.5$ and $\beta = 0$.

For $-\pi < \omega < \pi$, $A(\omega) > 0$. There won't be π jumps in the phase plot. Hence, this is a strict linear phase filter.

2. . Given the following phase response $\angle H_d(\omega)$ of a generalized linear-phase FIR filter, answer the following questions. Explain your answers.



Determine the system output $y[n]$ for the following input:

- Is the filter (i) type-1 GLP, (ii) type-2 GLP, or (iii) neither type-1 GLP nor type-2 GLP?
- Determine the filter length from the phase plot.
- Can you characterize the filter as (i) possibly low-pass, (ii) possibly high-pass, (iii) neither high-pass nor low-pass, or (iv) the given information is insufficient to make any of the preceding statements? (Specify all correct answers).
- Determine $H_d(\frac{\pi}{2})$

- Is the filter (i) type-1 GLP, (ii) type-2 GLP, or (iii) neither type-1 GLP nor type-2 GLP?

Solution: For any GLP filter, $H_d(\omega) = A(\omega)e^{j(\beta - \alpha\omega)}$, where $A(\omega)$ is real. Since $\angle H_d(\omega)$ has only π or 2π jumps and constant slopes, this is a GLP filter. Since the phase is zero when $\omega = 0$, this is a type-1 GLP.

- Determine the filter length from the phase plot.

Solution: The slope of the phase plot is -3. Therefore, the filter length is $2 \times 3 + 1 = 7$.

- Can you characterize the filter as (i) possibly low-pass, (ii) possibly high-pass, (iii) neither high-pass nor low-pass, or (iv) the given information is insufficient to make any of the preceding statements? (Specify **all** correct answers).

Solution: Since the GLP filter is type-1 and it has an odd length, both $H_d(0)$ and $H_d(\pi)$ are not necessarily zero. This means there are no restrictions on this filter. (i), (ii) and (iii) are all correct. (Selecting iv with sufficient reasoning and explanations is also acceptable.)

- Determine $H_d(\frac{\pi}{2})$.

Solution: From the phase plot, we can see that there is a π jump at $\omega = \frac{\pi}{2}$. From this, we can infer that $A(\omega)$ has a zero crossing at $\omega = \frac{\pi}{2}$ and thus $A(\frac{\pi}{2}) = 0$. Therefore, $H_d(\frac{\pi}{2}) = 0$.

3. Design a length-24 FIR lowpass filter with cutoff frequency $\omega_c = \frac{\pi}{5}$ radians using the window design method.
- a) Find an expression for the filter coefficients $\{h_n\}_{n=0}^{23}$ if the rectangular window is used for the design.
- b) Find an expression for the filter coefficients $\{h_n\}_{n=0}^{23}$ if the Hanning window is used for the design.
- a) **Solution:** Find an expression for the filter coefficients $\{h_n\}_{n=0}^{23}$ if the rectangular window is used for the design. The impulse response of an ideal LPF is given by:

$$h_{lp}[n] = \frac{\sin(\omega_c n)}{\pi n}$$

Before applying the window, we need to shift this filter by a factor of $\frac{M}{2}$ so that our filter will be causal and exhibit generalized linear phase. This lines up with the definition of a LPF given in the textbook, where we set $\omega_c = \frac{\pi}{5}$ and $\alpha = \frac{23}{2}$. Our shifted ideal LPF is then:

$$\begin{aligned} h_{lp}[n] &= \frac{\sin(\frac{\pi}{5}(n - \frac{23}{2}))}{n - \frac{23}{2}} \\ &= \frac{1}{5} \text{sinc}(\frac{\pi}{5}(n - \frac{23}{2})) \end{aligned}$$

Here, and in following problems, we define the sinc function as $\text{sinc}(x) = \frac{\sin(x)}{x}$. Applying the rectangular window gives the solution:

$$\{h[n]\}_{n=0}^{23} = \begin{cases} \frac{1}{5} \text{sinc}(\frac{\pi}{5}(n - \frac{23}{2})) & 0 \leq n \leq 23 \\ 0 & \text{else} \end{cases}$$

- b) **Solution:** Find an expression for the filter coefficients $\{h_n\}_{n=0}^{23}$ if the Hann window is used for the design. The definition of the Hann window has been given in the textbook. We simply apply that to $h_{lp}[n]$ in place of the rectangular window to get the answer:

$$\{h[n]\}_{n=0}^{23} = \begin{cases} (0.5 - 0.5 \cos(\frac{2\pi n}{23})) \frac{1}{5} \text{sinc}(\frac{\pi}{5}(n - \frac{23}{2})) & 0 \leq n \leq 23 \\ 0 & \text{else} \end{cases}$$