## Combinational Logic Technologies

- Standard gates
  - gate packages
  - cell libraries
- Regular logic
  - multiplexers
  - decoders
- Two-level programmable logic
  - PALs
  - PLAs
  - ROMs

## Random logic

- Transistors quickly integrated into logic gates (1960s)
- Catalog of common gates (1970s)
  - Texas Instruments Logic Data Book the yellow bible
  - all common packages listed and characterized (delays, power)
  - typical packages:
    - in 14-pin IC: 6-inverters, 4 NAND gates, 4 XOR gates
- Today, very few parts are still in use
- However, parts libraries exist for chip design
  - designers reuse already characterized logic gates on chips
  - same reasons as before
  - difference is that the parts don't exist in physical inventory created as needed

## Random logic

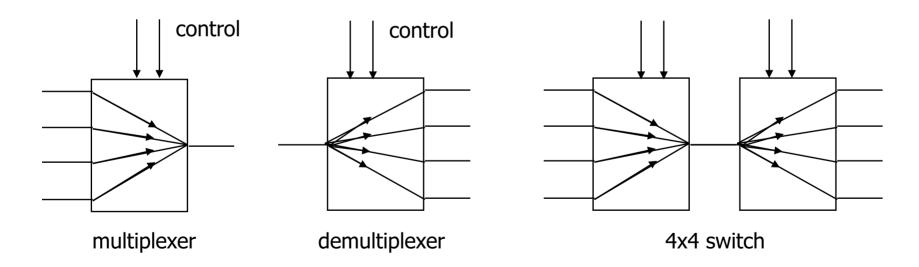
- Too hard to figure out exactly what gates to use
  - map from logic to NAND/NOR networks
  - determine minimum number of packages
    - slight changes to logic function could decrease cost
- Changes to difficult to realize
  - need to rewire parts
  - may need new parts
  - design with spares (few extra inverters and gates on every board)

## Regular logic

- Need to make design faster
- Need to make engineering changes easier to make
- Simpler for designers to understand and map to functionality
  - harder to think in terms of specific gates
  - better to think in terms of a large multi-purpose block

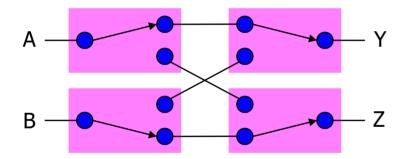
## Making connections

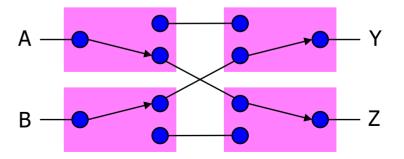
- Direct point-to-point connections between gates
  - wires we've seen so far
- Route one of many inputs to a single output --- multiplexer
- Route a single input to one of many outputs --- demultiplexer



## Mux and demux

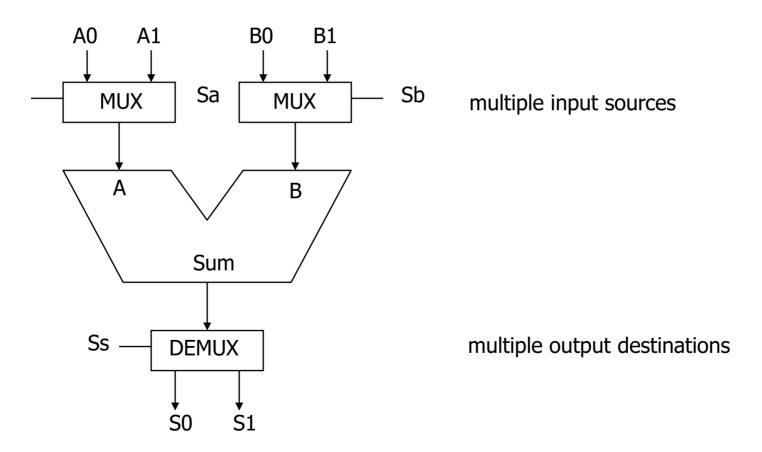
- Switch implementation of multiplexers and demultiplexers
  - can be composed to make arbitrary size switching networks
  - used to implement multiple-source/multiple-destination interconnections





## Mux and demux (cont'd)

Uses of multiplexers/demultiplexers in multi-point connections



## Multiplexers/selectors

- Multiplexers/selectors: general concept
  - □ 2<sup>n</sup> data inputs, n control inputs (called "selects"), 1 output
  - used to connect 2<sup>n</sup> points to a single point
  - control signal pattern forms binary index of input connected to

output

$$Z = A' I_0 + A I_1$$

 $\begin{array}{c|c} \hline 0 & I_0 \\ 1 & I_1 \\ \hline \end{array}$  functional form logical form

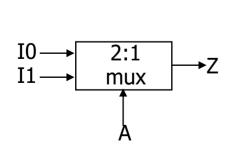
two alternative forms for a 2:1 Mux truth table

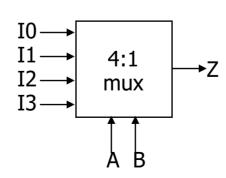
Ζ

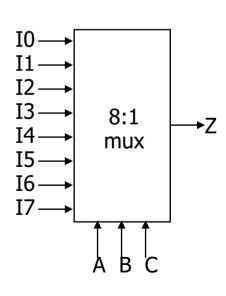
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## Multiplexers/selectors (cont'd)

- 2:1 mux:  $Z = A'I_0 + AI_1$
- 4:1 mux:  $Z = A'B'I_0 + A'BI_1 + AB'I_2 + ABI_3$
- 8:1 mux:  $Z = A'B'C'I_0 + A'B'CI_1 + A'BC'I_2 + A'BCI_3 + AB'C'I_4 + AB'CI_5 + ABC'I_6 + ABCI_7$
- In general:  $Z = \sum_{k=0}^{2^{n}-1} (m_k I_k)$ 
  - □ in minterm shorthand form for a 2<sup>n</sup>:1 Mux

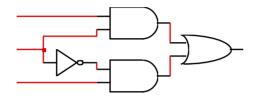


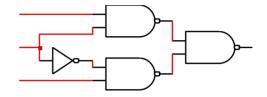




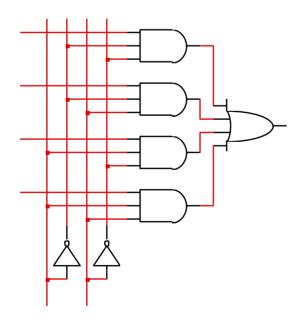
## Gate level implementation of muxes

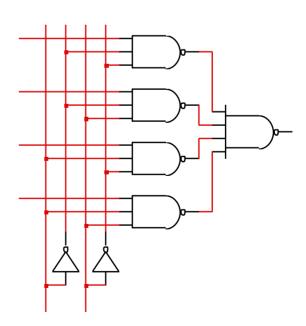
2:1 mux





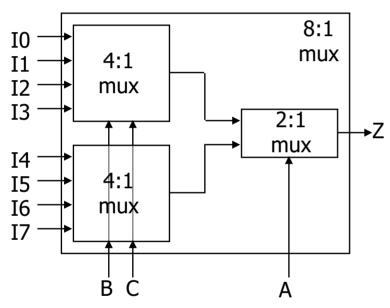
4:1 mux





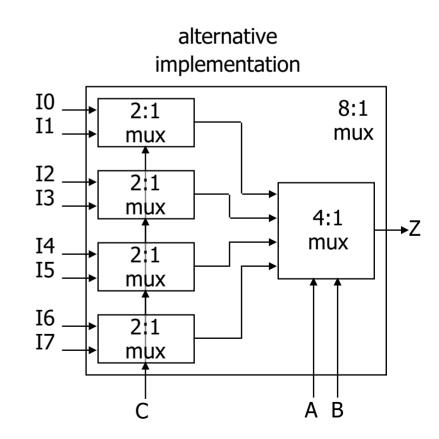
## Cascading multiplexers

Large multiplexers can be made by cascading smaller ones



control signals B and C simultaneously choose one of I0, I1, I2, I3 and one of I4, I5, I6, I7

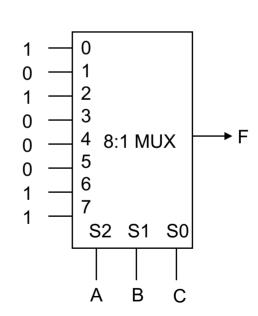
control signal A chooses which of the upper or lower mux's output to gate to Z



## Multiplexers as general-purpose logic

- A 2<sup>n</sup>:1 multiplexer can implement any function of n variables
  - with the variables used as control inputs and
  - the data inputs tied to 0 or 1
  - □ in essence, a lookup table
- Example:

□ 
$$F(A,B,C) = m0 + m2 + m6 + m7$$
  
=  $A'B'C' + A'BC' + ABC' + ABC$   
=  $A'B'C'(1) + A'B'C(0)$   
+  $A'BC'(1) + A'BC(0)$   
+  $AB'C'(0) + AB'C(0)$   
+  $ABC'(1) + ABC(1)$ 



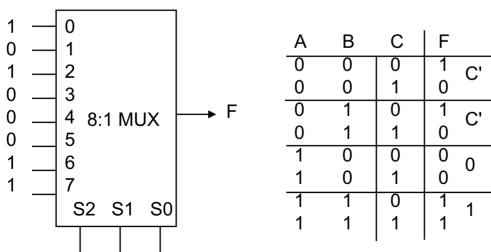
$$Z = A'B'C'I_0 + A'B'CI_1 + A'BC'I_2 + A'BCI_3 + AB'C'I_4 + AB'CI_5 + ABC'I_6 + ABCI_7$$

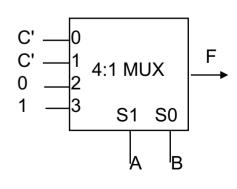
## Multiplexers as general-purpose logic (cont'd)

- A 2<sup>n-1</sup>:1 multiplexer can implement any function of n variables
  - with n-1 variables used as control inputs and
  - the data inputs tied to the last variable or its complement

### Example:

□ 
$$F(A,B,C) = m0 + m2 + m6 + m7$$
  
=  $A'B'C' + A'BC' + ABC' + ABC$   
=  $A'B'(C') + A'B(C') + AB'(0) + AB(1)$ 





## Multiplexers as general-purpose logic (cont'd)

Generalization

 n-1 mux control variables

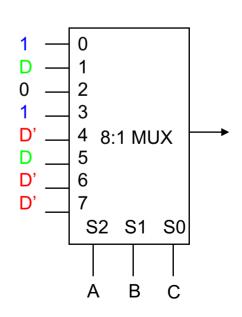
 single mux data variable

four possible configurations of truth table rows can be expressed as a function of I<sub>n</sub>

Example: G(A,B,C,D) can be realized by an 8:1 MUX

choose A,B,C as control variables

				_
Α	В	С	D	G
0	0	0	0	1 1
0	0	0	1	1 '
0	0	1	0	0
0 0 0 0 0	0	1	1	1 D
0	1	0	0	TO _
0	1	0	1	$\begin{bmatrix} 0 \end{bmatrix}^0$
0	1	1	0	1 1
0	1	1	1	1 '
1	0	0	0	1
1	0	0	1	0 D,
1	0	1	0	0
1	0	1	1	1 D
1	1	0	0	<u></u>
1	1	0	1	0
1	1	1	0	1 D'
1	1	1	1	
				•



# Activity

Realize F = B'CD' + ABC' with a 4:1 multiplexer and a minimum of other gates:

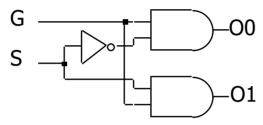
## Demultiplexers/decoders

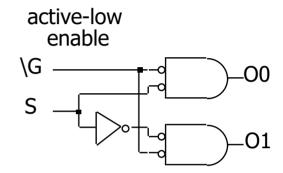
- Decoders/demultiplexers: general concept
  - single data input, n control inputs, 2<sup>n</sup> outputs
  - control inputs (called "selects" (S)) represent binary index of output to which the input is connected
  - data input usually called "enable" (G)

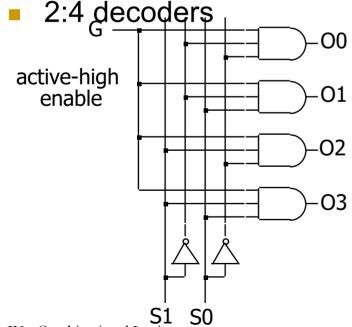
1:2 Decoder:	3:8 Decoder:					
$O0 = G \bullet S'$	$O0 = G \bullet S2' \bullet S1' \bullet S0'$					
$O1 = G \bullet S$	$O1 = G \bullet S2' \bullet S1' \bullet S0$					
	$O2 = G \bullet S2' \bullet S1 \bullet S0'$					
2:4 Decoder:	$O3 = G \bullet S2' \bullet S1 \bullet S0$					
$O0 = G \bullet S1' \bullet S0'$	$O4 = G \bullet S2 \bullet S1' \bullet S0'$					
$O1 = G \bullet S1' \bullet S0$	$O5 = G \bullet S2 \bullet S1' \bullet S0$					
$O2 = G \bullet S1 \bullet S0'$	$O6 = G \bullet S2 \bullet S1 \bullet S0'$					
$O3 = G \bullet S1 \bullet S0$	$O7 = G \bullet S2 \bullet S1 \bullet S0$					

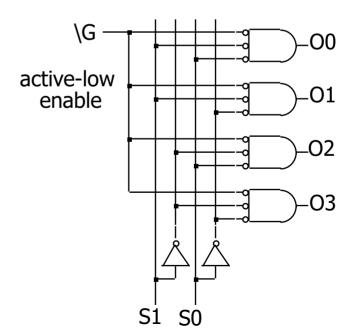
## Gate level implementation of demultiplexers

1:2 decoders active-high enable





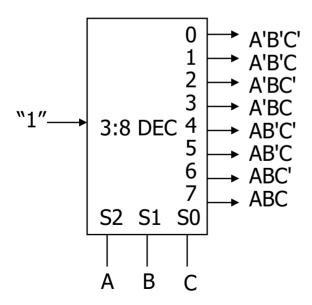




IV - Combinational Logic Technologies

## Demultiplexers as general-purpose logic

- A n:2<sup>n</sup> decoder can implement any function of n variables
  - with the variables used as control inputs
  - the enable inputs tied to 1 and
  - the appropriate minterms summed to form the function

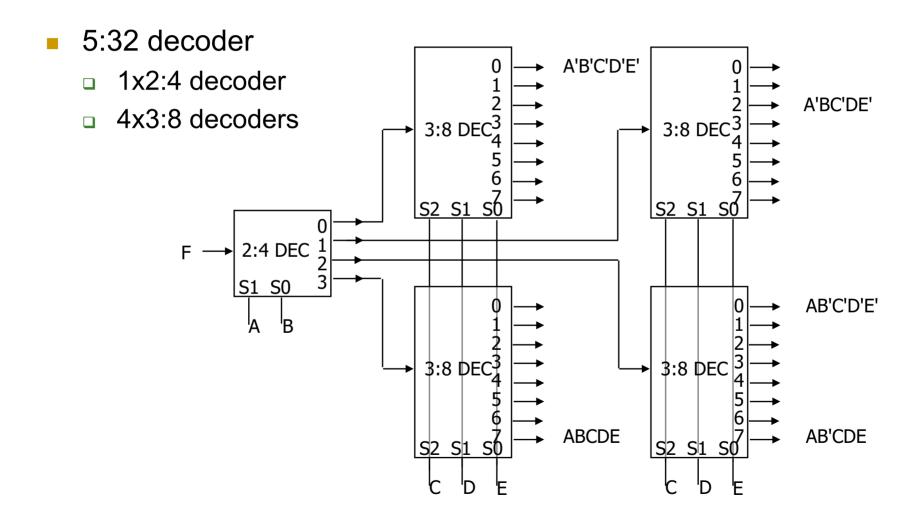


demultiplexer generates appropriate minterm based on control signals (it "decodes" control signals)

# Demultiplexers as general-purpose logic (cont'd)

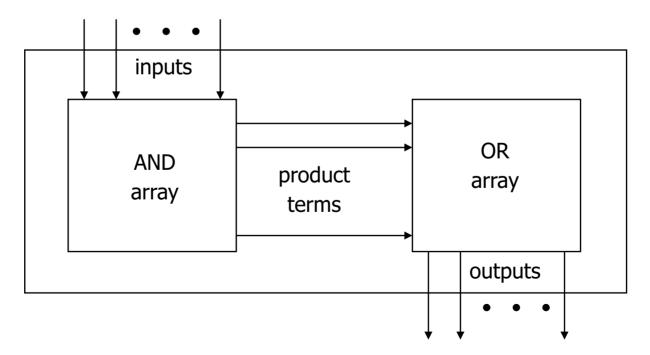
F1 = A'BC'D + A'B'CD + ABCDF2 = ABC'D' + ABC→A'B'C'D' F3 = (A' + B' + C' + D')→A'B'C'D 2 →A'B'CD' F1 3 → A'B'CD 4 →A'BC'D' 5 → A'BC'D 6 →A'BCD' → A'BCD 4:16 Enable →AB'C'D' **DEC** F2 9 →AB'C'D 10 →AB'CD' 11 →AB'CD 12 →ABC'D' 13 →ABC'D →ABCD¹ 14 15 →ABCD ABCD

## Cascading decoders



## Programmable logic arrays

- Pre-fabricated building block of many AND/OR gates
  - actually NOR or NAND
  - "personalized" by making/breaking connections among the gates
  - programmable array block diagram for sum of products form



## Enabling concept

### Shared product terms among outputs

example: 
$$F0 = A + B' C'$$
  
 $F1 = A C' + A B$   
 $F2 = B' C' + A B$   
 $F3 = B' C + A$ 

### personality matrix

inputs			outputs			
Α	В	С	F0	F1	F2	F3
1	1	_	0	1	1	0 🛌
_	0	1	0	0	0	1
1	_	0	0	1	0	0
_	0	0	1	0	1	0 👡
1	_	_	1	0	0	1
				.	.	.

### input side:

1 = uncomplemented in term

0 = complemented in term

– = does not participate

### output side:

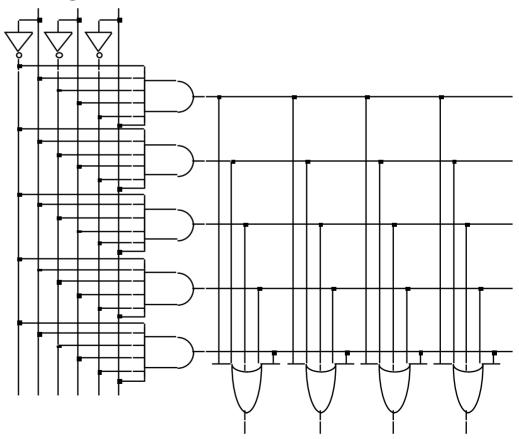
1 = term connected to output

0 = no connection to output

reuse of terms

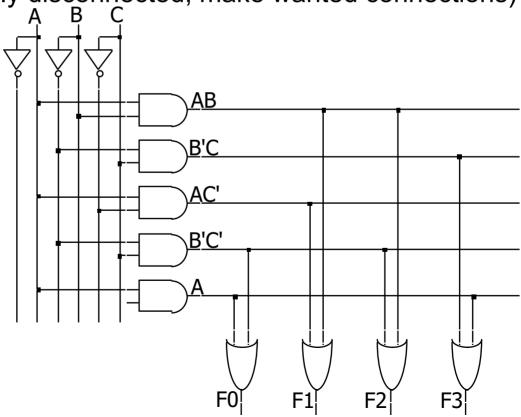
# Before programming

- All possible connections are available before "programming"
  - in reality, all AND and OR gates are NANDs



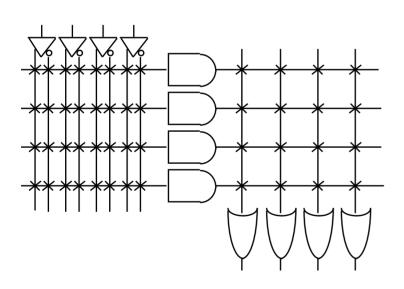
# After programming

- Unwanted connections are "blown"
  - fuse (normally connected, break unwanted ones)
  - anti-fuse (normally disconnected, make wanted connections)



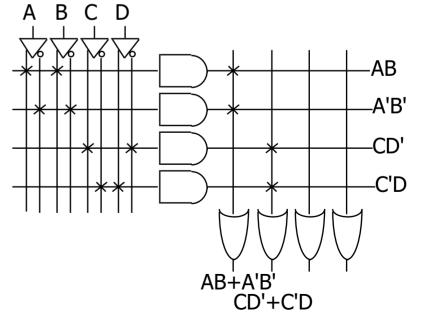
## Alternate representation for high fan-in structures

- Short-hand notation so we don't have to draw all the wires
  - × signifies a connection is present and perpendicular signal is an input to gate



notation for implementing

$$F0 = A B + A' B'$$
  
 $F1 = C D' + C' D$ 



## Programmable logic array example

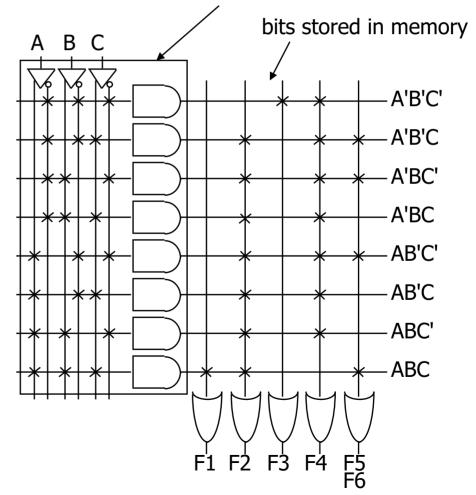
Multiple functions of A, B, C

$$\Rightarrow$$
 F1 = ABC

$$\Box$$
 F6 = A xnor B xnor C

Α	В	C	F1	F2	F3	F4	F5	F6
0	0	0	0 0 0 0 0 0	0	1	1	0	0
0	0	1	0	1	0	1	1	1
0	1	0	0	1	0	1	1	1
0	1	1	0	1	0	1	0	0
1	0	0	0	1	0	1	1	1
1	0	1	0	1	0	1	0	0
	Τ	U	U	T	U	T	U	U
1	1	1	1	1	0	0	1	1

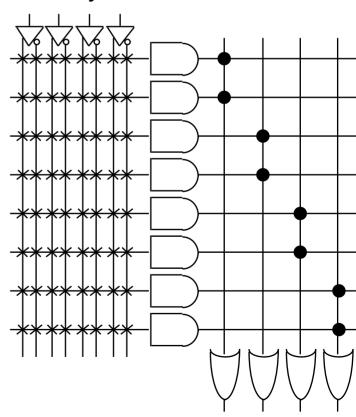
full decoder as for memory address



## PALs and PLAs

- Programmable logic array (PLA)
  - what we've seen so far
  - unconstrained fully-general AND and OR arrays
- Programmable array logic (PAL)
  - constrained topology of the OR array
  - innovation by Monolithic Memories
  - faster and smaller OR plane

a given column of the OR array has access to only a subset of the possible product terms



## PALs and PLAs: design example

### BCD to Gray code converter

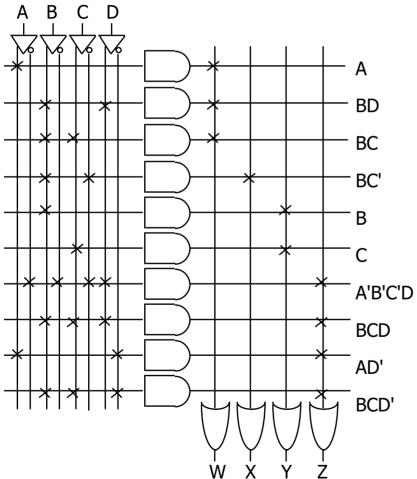
Α	В	C	D	W	X	Υ	Z
0	0	0	0	0	0	0	0
0	0	0	1	0	0	0	1
0	0	1	0	0	0	1	1
0	0	1	1	0	0	1	0
0	1	0	0	0	1	1	0
0	1	0	1	1	1	1	0
0	1	1	0	1	0	1	0
0	1	1	1	1	0	1	1
1	0	0	0	1	0	0	1
1	0	0	1	1	0	0	0
1	0	1	_	_	_	_	_
1	1	_	_	_	_	_	_
				•			

#### minimized functions:

$$W = A + BD + BC$$
  
 $X = BC'$   
 $Y = B + C$   
 $Z = A'B'C'D + BCD + AD' + B'CD'$ 

## PALs and PLAs: design example (cont'd)

### Code converter: programmed PLA



### minimized functions:

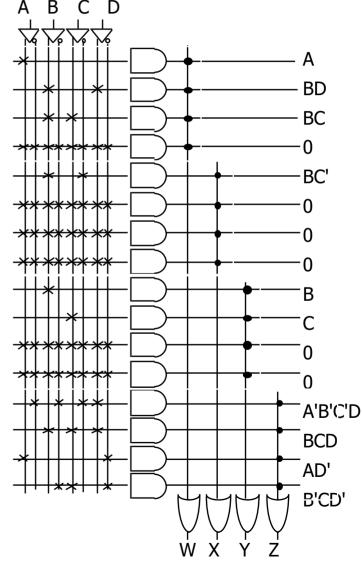
not a particularly good candidate for PAL/PLA implementation since no terms are shared among outputs

however, much more compact and regular implementation when compared with discrete AND and OR gates

## PALs and PLAs: design example (cont'd)

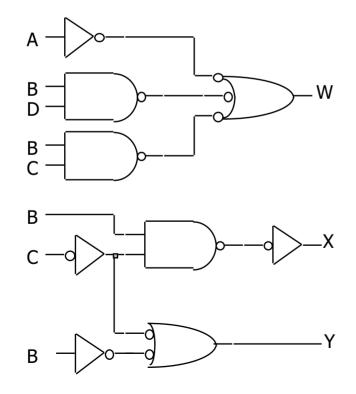
Code converter: programmed PAL

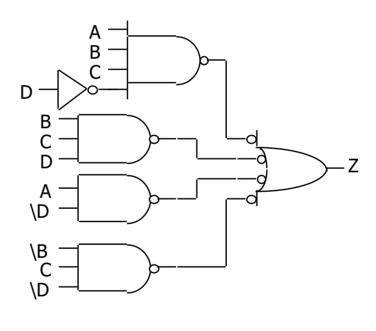
4 product terms per each OR gate



# PALs and PLAs: design example (cont'd)

- Code converter: NAND gate implementation
  - loss or regularity, harder to understand
  - harder to make changes

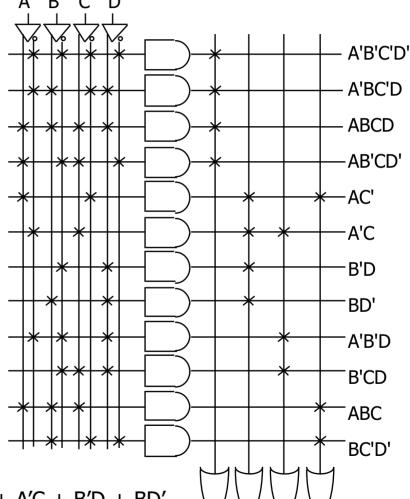




## PALs and PLAs: another design example

### Magnitude comparator

Α	В	C	D	EQ	NE	LT	GT
0	0	0	0	1	0	0	0
0	0	0	1	0	1	1	0
0	0	1	0	0	1	1	0
0	0	1	1	0	1	1	0
0	1	0	0	0	1	0	1
0	1	0	1	1	0	0	0
0	1	1	0	0	1	1	0
0	1	1	1	0	1	1	0
1	0	0	0	0	1	0	1
1	0	0	1	0	1	0	1
1	0	1	0	1	0	0	0
1	0	1	1	0	1	1	0
1	1	0	0	0	1	0	1
1	1	0	1	0	1	0	1
1	1	1	0	0	1	0	1
1	$\bar{1}$	$\overline{1}$	1	1	Ō	Ö	Ō
•	_	_	-		-	-	-



minimized functions:

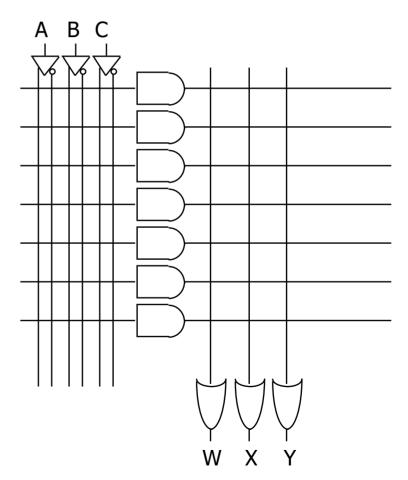
EQ = A'B'C'D' + A'BC'D + ABCD + AB'CD'

LT = A'C + A'B'D + B'CD
IV - Combinational Logic

$$NE = AC' + A'C + B'D + BD'$$
  
 $GT = AC' + ABC + BC'D'$ 

## Activity

- Map the following functions to the PLA below:
  - W = AB + A'C' + BC'
  - $\square$  X = ABC + AB' + A'B
  - Y = ABC' + BC + B'C'

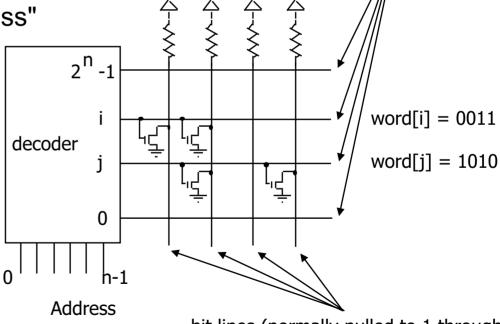


# Activity (cont'd)

## Read-only memories

- Two dimensional array of 1s and 0s
  - entry (row) is called a "word"
  - □ width of row = word-size
  - index is called an "address"
  - address is input
  - selected word is output

internal organization



bit lines (normally pulled to 1 through resistor – selectively connected to 0 by word line controlled switches)

word lines (only one

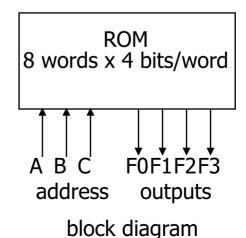
is active – decoder is just right for this)

## ROMs and combinational logic

Combinational logic implementation (two-level canonical form) using a ROM

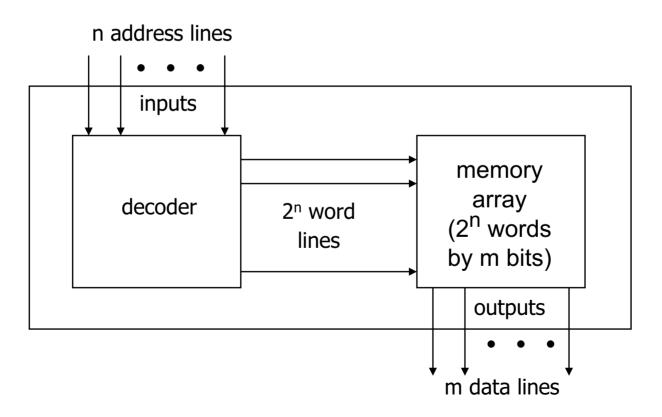
Α	В	C	F0	F1	F2	F3
0	0	0	0	0	1	0
0	0	1	1	1	1	0
0	1	0	0	1	0	0
0	1	1	0	0	1 1 0 0 1 0	1
1	0	0	1	0	1	1
1	0	1	1	0	0	0
1	1	0	0	0	0	1
1	1	1	0	0	0	0
			'			

truth table



## ROM structure

- Similar to a PLA structure but with a fully decoded AND array
  - completely flexible OR array (unlike PAL)



## ROM vs. PLA

- ROM approach advantageous when
  - design time is short (no need to minimize output functions)
  - most input combinations are needed (e.g., code converters)
  - little sharing of product terms among output functions
- ROM problems
  - size doubles for each additional input
  - can't exploit don't cares
- PLA approach advantageous when
  - design tools are available for multi-output minimization
  - there are relatively few unique minterm combinations
  - many minterms are shared among the output functions
- PAL problems
  - constrained fan-ins on OR plane

# Regular logic structures for two-level logic

- ROM full AND plane, general OR plane
  - cheap (high-volume component)
  - can implement any function of n inputs
  - medium speed
- PAL programmable AND plane, fixed OR plane
  - intermediate cost
  - can implement functions limited by number of terms
  - high speed (only one programmable plane that is much smaller than ROM's decoder)
- PLA programmable AND and OR planes
  - most expensive (most complex in design, need more sophisticated tools)
  - can implement any function up to a product term limit
  - slow (two programmable planes)

## Regular logic structures for multi-level logic

- Difficult to devise a regular structure for arbitrary connections between a large set of different types of gates
  - efficiency/speed concerns for such a structure
  - in 467 you'll learn about field programmable gate arrays (FPGAs)
     that are just such programmable multi-level structures
    - programmable multiplexers for wiring
    - lookup tables for logic functions (programming fills in the table)
    - multi-purpose cells (utilization is the big issue)
- Use multiple levels of PALs/PLAs/ROMs
  - output intermediate result
  - make it an input to be used in further logic

## Combinational logic technology summary

- Random logic
  - Single gates or in groups
  - conversion to NAND-NAND and NOR-NOR networks
  - transition from simple gates to more complex gate building blocks
  - reduced gate count, fan-ins, potentially faster
  - more levels, harder to design
- Time response in combinational networks
  - gate delays and timing waveforms
  - hazards/glitches (what they are and why they happen)
- Regular logic
  - multiplexers/decoders
  - ROMs
  - PLAs/PALs
  - advantages/disadvantages of each