

ISO1042-Q1 Automotive Isolated CAN Transceiver With 70-V Bus Fault Protection and Flexible Data Rate

1 Features

- AEC Q100: Qualified for automotive applications
- Grade 1: -40°C to 125°C ambient temperature
- Functional Safety-Capable
 - Documentation available to aid functional safety system design
- Meets the ISO 11898-2:2016 physical layer standard
- Supports classic CAN up to 1 Mbps and FD (Flexible Data Rate) up to 5 Mbps
- Low loop delay: 152 ns
- Protection features
 - DC bus fault protection voltage: ±70 V
 - HBM ESD tolerance on bus pins: ±16 kV
 - Driver Dominant Time Out (TXD DTO)
 - Undervoltage protection on V_{CC1} and V_{CC2}
- Common-Mode Voltage Range: ±30 V
- Ideal passive, high impedance bus terminals when unpowered
- High CMTI: 100 kV/µs
- V_{CC1} voltage range: 1.71 V to 5.5 V
 - Supports 1.8-V, 2.5-V, 3.3-V and 5.0-V logic interface to the CAN controller
- V_{CC2} Voltage Range: 4.5 V to 5.5 V
- Robust Electromagnetic Compatibility (EMC)
 - System-level ESD, EFT, and surge immunity
 - Low emissions
- 16-SOIC and 8-SOIC package options
- Industrial version available: ISO1042
- Safety-related certifications:
 - 7071-V_{PK} V_{IOTM} and 1500-V_{PK} V_{IORM} (Reinforced and Basic Options) per DIN VDE V 0884-11:2017-01
 - 5000-V_{RMS} Isolation for 1 Minute per UL 1577
 - CQC, TUV and CSA certifications

2 Applications

- Starter/generator
- Battery Management System (BMS)
- DC/DC converter
- On-Board (OBC) & wireless charger
- Inverter & motor control

3 Description

The ISO1042-Q1 device is a galvanically-isolated controller area network (CAN) transceiver that meets the specifications of the ISO11898-2 (2016) standard. The ISO1042-Q1 device offers ±70-V DC bus fault protection and ±30-V common-mode voltage range. The device supports up to 5-Mbps data rate in CAN FD mode allowing much faster transfer of payload compared to classic CAN. This device uses a silicon dioxide (SiO 2) insulation barrier with a withstand voltage of 5000 V_{RMS} and a working voltage of 1060 V Electromagnetic compatibility significantly enhanced to enable system-level ESD, EFT, surge, and emissions compliance. Used in conjunction with isolated power supplies, the device protects against high voltage, and prevents noise currents from the bus from entering the local ground. The ISO1042-Q1 device is available for both basic and reinforced isolation (see Reinforced and Basic Isolation Options). The ISO1042-Q1 device supports a wide ambient temperature range of -40°C to +125°C. The device is available in the SOIC-16 (DW) package and a smaller SOIC-8 (DWV) package.

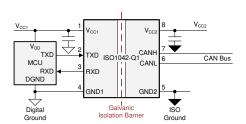
Device Information

PART NUMBER ⁽¹⁾	PACKAGE	BODY SIZE (NOM)
ISO1042-Q1	SOIC (8)	5.85 mm × 7.50 mm
	SOIC (16)	10.30 mm × 7.50 mm

For all available packages, see the orderable addendum at the end of the data sheet.

Reinforced and Basic Isolation Options

FEATURE	ISO1042x-Q1	ISO1042Bx-Q1
Protection Level	Reinforced	Basic
Surge Test Voltage	10000 V _{PK}	6000 V _{PK}
Isolation Rating	5000 V _{RMS}	5000 V _{RMS}
Working Voltage	1060 V _{RMS} / 1500 V _{PK}	1060 V _{RMS} / 1500 V _{PK}



Application Diagram



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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision A (January 2020) to Revision B (October 2020)	Page
Added Functional Safety Bullets	1
Changes from Revision * (October 2018) to Revision A (January 2020)	Page
Changed new safety certification	1



5 Pin Configuration and Functions

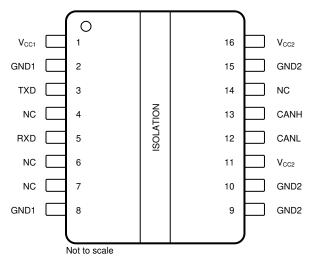


Figure 5-1. DW Package 16-Pin SOIC Top View

Table 5-1. Pin Functions—16 Pins

	PIN	I/O	DESCRIPTION
NO.	NAME	1/0	DESCRIPTION
1	V _{CC1}	_	Digital-side supply voltage, Side 1
2	GND1	_	Digital-side ground connection, Side 1
3	TXD	I	CAN transmit data input (LOW for dominant and HIGH for recessive bus states)
4	NC	_	Not connected
5	RXD	0	CAN receive data output (LOW for dominant and HIGH for recessive bus states)
6	NC	_	Not connected
7	NC	_	Not connected
8	GND1	_	Digital-side ground connection, Side 1
9	GND2		Transceiver side ground connection Side 2
10	GND2	_	Transceiver-side ground connection, Side 2
11	V _{CC2}	_	Transceiver-side supply voltage, Side 2. Must be externally connected to pin 16.
12	CANL	I/O	Low-level CAN bus line
13	CANH	I/O	High-level CAN bus line
14	NC	_	Not connected
15	GND2	_	Transceiver-side ground connection, Side 2
16	V _{CC2}	_	Transceiver-side supply voltage, Side 2. Must be externally connected to pin 11.



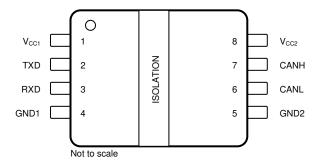


Figure 5-2. DWV Package 8-Pin SOIC Top View

Table 5-2. Pin Functions—8 Pins

	PIN		DESCRIPTION
NO.	NAME	· I/O	DESCRIPTION
1	V _{CC1}	_	Digital-side supply voltage, Side 1
2	TXD	I	CAN transmit data input (LOW for dominant and HIGH for recessive bus states)
3	RXD	0	CAN receive data output (LOW for dominant and HIGH for recessive bus states)
4	GND1	_	Digital-side ground connection, Side 1
5	GND2	_	Transceiver-side ground connection, Side 2
6	CANL	I/O	Low-level CAN bus line
7	CANH	I/O	High-level CAN bus line
8	V _{CC2}	_	Transceiver-side supply voltage, Side 2

6 Specifications

6.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)(1) (2)

		MIN	MAX	UNIT
V _{CC1}	Supply voltage, side 1	-0.5	6	V
V _{CC2}	Supply voltage, side 2	-0.5	6	V
V _{IO}	Logic input and output voltage range (TXD and RXD)	-0.5	V _{CC1} +0.5 ⁽³⁾	V
Io	Output current on RXD pin	-15	15	mA
V _{BUS}	Voltage on bus pins (CANH, CANL)	-70	70	V
V _{BUS_DIFF}	Differential voltage on bus pins (CANH-CANL)	-70	70	V
TJ	Junction temperature	-40	150	°C
T _{STG}	Storage temperature	-65	150	°C

- (1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values except differential I/O bus voltages are with respect to the local ground terminal (GND1 or GND2) and are peak voltage values.
- (3) Maximum voltage must not exceed 6 V

6.2 ESD Ratings

			VALUE	UNIT
	Electrostatic discharge	All pins ⁽¹⁾	±6000	V
V	Human body model (HBM), per ANSI/ ESDA/JEDEC JS-001	CANH and CANL to GND2 ⁽¹⁾	±16000	
V _(ESD)	Électrostatic discharge	All pins ⁽²⁾	±1500	V

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Transient Immunity

	PARAMETER	TEST CONDITIONS	VALUE	UNIT
		Pulse 1; CAN bus terminals (CANH, CANL) to GND2	-100	V
\		Pulse 2; CAN bus terminals (CANH, CANL) to GND2	75	V
V _{PULSE}	CAN EMC test specification	Pulse 3a; CAN bus terminals (CANH, CANL) to GND2	-150	V
		Pulse 3b; CAN bus terminals (CANH, CANL) to GND2	100	V

6.4 Recommended Operating Conditions

		MIN	MAX	UNIT
V	Supply Voltage, Side 1, 1.8-V operation	1.71	1.89	V
V _{CC1}	Supply Voltage, Side 1, 2.5-V, 3.3-V and 5.5-V operation	2.25	5.5	V
V _{CC2}	Supply Voltage, Side 2	4.5	5.5	V
T _A	Operating ambient temperature	-40	125	°C

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6.5 Thermal Information

		ISO10	ISO1042-Q1			
	THERMAL METRIC(1)	DW (SOIC)	DWV (SOIC)	UNIT		
		16 PINS	8 PINS			
R _{⊝JA}	Junction-to-ambient thermal resistance	69.9	100	°C/W		
R _{OJC(top)}	Junction-to-case (top) thermal resistance	31.8	40.8	°C/W		
R _{⊝JB}	Junction-to-board thermal resistance	29.0	51.8	°C/W		
Ψ_{JT}	Junction-to-top characterization parameter	13.2	16.8	°C/W		
Ψ_{JB}	Junction-to-board characterization parameter	28.6	49.8	°C/W		
R _{OJC(bot)}	Junction-to-case (bottom) thermal resistance	-	-	°C/W		

⁽¹⁾ For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

6.6 Power Ratings

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
P _D	Maximum power dissipation (both sides)	See Figure 7-3, $V_{CC1} = V_{CC2} = 5.5$ V, $T_J = 150^{\circ}$ C, $R_L = 50~\Omega$, A repetitive pattern on TXD with 1 ms time period, 990 μ s LOW time, and 10 μ s HIGH time.			385	mW
P _{D1}	Maximum power dissipation (side-1)	See Figure 7-5, $V_{CC1} = V_{CC2} = 5.5 \text{ V}$, $T_J = 150^{\circ}\text{C}$, $R_L = 50 \Omega$, Input a 2-V pk-pk 2.5-MHz 50% duty cycle differential square wave on CANH-CANL			25	mW
P _{D2}	Maximum power dissipation (side-2)	See Figure 7-3, $V_{CC1} = V_{CC2} = 5.5 \text{ V}$, $T_J = 150^{\circ}\text{C}$, $R_L = 50 \Omega$, A repetitive pattern on TXD with 1 ms time period, 990 μ s LOW time, and 10 μ s HIGH time.			360	mW

Product Folder Links: ISO1042-Q1



6.7 Insulation Specifications

	24244		SPECIFICATIONS		
PARAMETER IEC 60664-1		TEST CONDITIONS	DW-16	DWV-8	UNIT
IEC 6066	64-1			•	
CLR	External clearance ⁽¹⁾	Side 1 to side 2 distance through air	>8	>8.5	mm
CPG	External Creepage ⁽¹⁾	Side 1 to side 2 distance across package surface	>8	>8.5	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	>17	>17	μm
CTI	Comparative tracking index	IEC 60112; UL 746A	>600	>600	V
	Material Group	According to IEC 60664-1	I	I	
	0 11 1	Rated mains voltage ≤ 600 V _{RMS}	I-IV	I-IV	
	Overvoltage category	Rated mains voltage ≤ 1000 V _{RMS}	1-111	I-III	
DIN VDE	V 0884-11:2017-01 ⁽²⁾				
V _{IORM}	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	1500	1500	V _{PK}
V _{IOWM}	Maximum isolation working voltage	AC voltage (sine wave); time-dependent dielectric breakdown (TDDB) test;	1060	1060	V _{RMS}
1011111		DC voltage	1500	1500	V _{DC}
V _{IOTM}	Maximum transient isolation voltage	V _{TEST} = V _{IOTM} , t = 60 s (qualification); V _{TEST} = 1.2 × V _{IOTM} , t = 1 s (100% production)	7071	7071	V _{PK}
.,	Maximum surge isolation voltage ISO1042-Q1 (3)	Test method per IEC 62368-1, 1.2/50 µs waveform, V _{TEST} = 1.6 × V _{IOSM} = 10000 V _{PK} (qualification)	6250	6250	V _{PK}
V _{IOSM}	Maximum surge isolation voltage ISO1042B-Q1 (3)	Test method per IEC 62368-1, 1.2/50 µs waveform, V _{TEST} = 1.3 × V _{IOSM} = 6000 V _{PK} (qualification)	4615 4	4615	V _{PK}
		Method a: After I/O safety test subgroup 2/3, $V_{ini} = V_{IOTM}, t_{ini} = 60 \text{ s}; V_{pd(m)} = 1.2 \times V_{IORM}, t_{m} = 10 \text{ s}$	≤ 5	≤ 5	
q_{pd}	Apparent charge ⁽⁴⁾	Method a: After environmental tests subgroup 1, $V_{ini} = V_{IOTM}$, $t_{ini} = 60$ s; ISO1042-Q1: $V_{pd(m)} = 1.6 \times V_{IORM}$, $t_m = 10$ s ISO1042B-Q1: $V_{pd(m)} = 1.2 \times V_{IORM}$, $t_m = 10$ s	≤ 5	≤ 5	pC
		Method b1: At routine test (100% production) and preconditioning (type test), $V_{ini} = V_{IOTM}$, $t_{ini} = 1$ s; ISO1042-Q1: $V_{pd(m)} = 1.875 \times V_{IORM}$, $t_m = 1$ s ISO1042B-Q1: $V_{pd(m)} = 1.5 \times V_{IORM}$, $t_m = 1$ s	≤ 5	≤ 5	
C _{IO}	Barrier capacitance, input to output ⁽⁵⁾	$V_{IO} = 0.4 \times \sin(2 \pi ft), f = 1 MHz$	1	1	pF
		V _{IO} = 500 V, T _A = 25°C	> 10 ¹²	> 10 ¹²	
R _{IO}	Insulation resistance, input to output ⁽⁵⁾	V _{IO} = 500 V, 100°C ≤ T _A ≤ 150°C	> 10 ¹¹	> 10 ¹¹	Ω
		V _{IO} = 500 V at T _S = 150°C	> 10 ⁹	> 10 ⁹	
	Pollution degree		2	2	
	Climatic category		40/125/ 21	40/125/ 21	
UL 1577	1		1	1	I.
V _{ISO}	Withstand isolation voltage	$V_{TEST} = V_{ISO}$, $t = 60$ s (qualification); $V_{TEST} = 1.2 \times V_{ISO}$, $t = 1$ s (100% production)	5000	5000	V _{RMS}

⁽¹⁾ Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves, ribs, or both on a printed circuit board are used to help increase these specifications.



- (2) ISO1042-Q1 is suitable for *safe electrical insulation* and ISO1042B-Q1 is suitable for *basic electrical insulation* only within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.
- (3) Testing is carried out in air or oil to determine the intrinsic surge immunity of the isolation barrier.
- (4) Apparent charge is electrical discharge caused by a partial discharge (pd).
- (5) All pins on each side of the barrier tied together creating a two-pin device.

6.8 Safety-Related Certifications

VDE	CSA	UL	CQC	TUV
Certified according to DIN VDE V 0884-11:2017- 01	Certified according to IEC 60950-1, IEC 62368-1 and IEC 60601-1	Recognized under UL 1577 Component Recognition Program	Certified according to GB4943.1-2011	Certified according to EN 61010-1:2010/A1:2019, EN 60950-1:2006/A2:2013 and EN 62368-1:2014
Maximum transient isolation voltage, 7071 V _{PK} ; Maximum repetitive peak isolation voltage, 1500 V _{PK} ; Maximum surge isolation voltage, ISO1042-Q1: 6250 V _{PK} (Reinforced) ISO1042B-Q1: 4615 V _{PK} (Basic)	CSA 60950-1-07+A1+A2, IEC 60950-1 2 nd Ed.+A1+A2 and IEC 62368-1 2 nd Ed., for pollution degree 2, material group I ISO1042-Q1: 800 V _{RMS} reinforced isolation ISO1042B-Q1: 1060 V _{RMS} basic isolation	Single protection, 5000 V _{RMS}	Reinforced insulation, Altitude ≤ 5000 m, Tropical Climate, 700 V _{RMS} maximum working voltage	EN 61010-1:2010 / A1:2019 ISO1042-Q1: 600 V _{RMS} reinforced isolation ISO1042B-Q1: 1000 V _{RMS} basic isolation EN 60950-1:2006/A2:2013 and EN 62368-1:2014 ISO1042-Q1: 800 V _{RMS} reinforced isolation ISO1042B-Q1: 1060 V _{RMS} basic isolation
Certificates: Reinforced: 40040142 Basic: 40047657	Master contract number: 220991	File number: E181974	Certificate: CQC15001121716 (DW-16) CQC18001199096 (DWV-8)	Client ID number: 77311

6.9 Safety Limiting Values

Safety limiting⁽¹⁾ intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry.

	PARAMETER	TEST CONDITIONS	MIN TYP MAX	UNIT
DW-	16 PACKAGE			
		R _{0JA} = 69.9°C/W, V _I = 5.5 V, T _J = 150°C, T _A = 25°C, see Figure 6-1	325	
١,	Safety input, output, or supply	R _{0JA} = 69.9°C/W, V _I = 3.6 V, T _J = 150°C, T _A = 25°C, see Figure 6-1	496	
I _S	current	R _{0JA} = 69.9°C/W, V _I = 2.75 V, T _J = 150°C, T _A = 25°C, see Figure 6-1	650	- mA
		R _{0JA} = 69.9°C/W, V _I = 1.89 V, T _J = 150°C, T _A = 25°C, see Figure 6-1	946	
Ps	Safety input, output, or total power	R _{0JA} = 69.9°C/W, T _J = 150°C, T _A = 25°C, see Figure 6-3	1788	mW
T _S	Maximum safety temperature		150	°C
DWV	/-8 PACKAGE			
		R _{0JA} = 100°C/W, V _I = 5.5 V, T _J = 150°C, T _A = 25°C, see Figure 6-2	227	
١,	Safety input, output, or supply	R _{0JA} = 100°C/W, V _I = 3.6 V, T _J = 150°C, T _A = 25°C, see Figure 6-2	347	
IS	current	R _{0JA} = 100°C/W, V _I = 2.75 V, T _J = 150°C, T _A = 25°C, see Figure 6-2	454	- mA
		R _{0JA} = 100°C/W, V _I = 1.89 V, T _J = 150°C, T _A = 25°C, see Figure 6-2	661	
Ps	Safety input, output, or total power	R _{0JA} = 100°C/W, T _J = 150°C, T _A = 25°C, see Figure 6-4	1250	mW
T _S	Maximum safety temperature		150	°C

(1) The maximum safety temperature, T_S, has the same value as the maximum junction temperature, T_J, specified for the device. The I_S and P_S parameters represent the safety current and safety power respectively. The maximum limits of I_S and P_S should not be exceeded. These limits vary with the ambient temperature, T_A.

Product Folder Links: ISO1042-Q1



The junction-to-air thermal resistance, $R_{\theta JA}$, in the table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:

 $T_J = T_A + R_{\theta JA} \times P$, where P is the power dissipated in the device. $T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S$, where $T_{J(max)}$ is the maximum allowed junction temperature. $P_S = I_S \times V_I$, where V_I is the maximum input voltage.



6.10 Electrical Characteristics - DC Specification

Over recommended operating conditions (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SUPPLY	CHARACTERISTICS					
		V _{CC1} =1.71 V to 1.89 V, TXD = 0 V, bus dominant		2.3	3.5	mA
l	Supply current Side 4	V _{CC1} = 2.25 V to 5.5 V, TXD = 0 V, bus dominant		2.4	3.5	mA
I _{CC1}	Supply current Side 1	V_{CC1} = 1.71 V to 1.89 V, TXD = V_{CC1} , bus recessive		1.2	2.1	mA
		V_{CC1} = 2.25 V to 5.5 V, TXD = V_{CC1} , bus recessive		1.3	2.1	mA
	Supply ourrent Side 2	TXD = 0 V, bus dominant, $R_L = 60 \Omega$		43	73.4	mA
I _{CC2}	Supply current Side 2	TXD = V_{CC1} , bus recessive, R_L = 60 Ω		2.8	4.1	mA
UV _{VCC1}	Rising under voltage detection, Side 1				1.7	V
UV _{VCC1}	Falling under voltage detection, Side 1		1.0			V
V HYS(UVCC 1)	Hysterisis voltage on V _{CC1} undervoltage lock-out		75	125		mV
UV _{VCC2}	Rising under voltage detection, side 2			4.2	4.45	V
UV _{VCC2}	Falling under voltage detection, side 2		3.8	4.0	4.25	V
V HYS(UVCC 2)	Hysterisis voltage on V _{CC2} undervoltage lock-out			200		mV
TXD TER	MINAL					
V _{IH}	High level input voltage		0.7×V _{CC1}			V
V _{IL}	Low level input voltage			(0.3×V _{CC1}	V
I _{IH}	High level input leakage current	TXD = V _{CC1}		,	1	uA
I _{IL}	Low level input leakage current	TXD = 0V	-20			uA
Cı	Input capacitance	VIN = $0.4 \times \sin(2 \times \pi \times 1E+6 \times t) + 2.5 \text{ V}$, $V_{CC1} = 5 \text{ V}$		3		pF
RXD TER	MINAL					
		See Figure 7-4, I_O = -4 mA for 4.5 V \leq V $_{CC1} \leq$ 5.5 V	-0.4	-0.2		V
V _{OH} - V	High level output voltage	See Figure 7-4, I_O = -2 mA for 3.0 V \leq V $_{CC1} \leq$ 3.6 V	-0.2	-0.07		V
CC1	High level output voltage	See Figure 7-4, $I_O = -1$ mA for 2.25 V \leq V $_{CC1} \leq$ 2.75 V	-0.1	-0.04		V
		See Figure 7-4, $I_0 = -1$ mA for 1.71 V \leq V $_{CC1} \leq$ 1.89 V	-0.1	-0.045		V
		See Figure 7-4, $I_O = 4$ mA for 4.5 V \leq V $_{CC1} \leq 5.5$ V		0.2	0.4	V
Vo	Low level output voltage	See Figure 7-4, I_0 = 2 mA for 3.0 V ≤ V $_{CC1}$ ≤ 3.6 V		0.07	0.2	V
V _{OL}	Low level output voltage	See Figure 7-4, $I_0 = 1$ mA for 2.25 V \leq V $_{CC1} \leq$ 2.75 V		0.035	0.1	V
		See Figure 7-4, $I_0 = 1$ mA for 1.71 V \leq V $_{\text{CC1}} \leq$ 1.89 V		0.04	0.1	V
DRIVER I	ELECTRICAL CHARACTERISTICS					
V	Bus output voltage(Dominant), CANH	See Figure 7-1 and Figure 7-2, TXD = 0 V, $50 \Omega \le R_L \le 65 \Omega$, $C_L = open$	2.75		4.5	V
$V_{O(DOM)}$	Bus output voltage(Dominant), CANL	See Figure 7-1 and Figure 7-2, TXD = 0 V, $50 \Omega \le R_L \le 65 \Omega$, $C_L = open$	0.5		2.25	V
	•	The state of the s				

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{O(REC)}	Bus output voltage(recessive), CANH and CANL	See Figure 7-1 and Figure 7-2, TXD = V CC1, R _L = open	2.0	0.5 x VCC2	3.0	V
	Differential output voltage, CANH-CANL (dominant)	See Figure 7-1 and Figure 7-2, TXD = 0 V, $45 \Omega \le R_L \le 50 \Omega$, $C_L = open$	1.4		3.0	V
V _{OD(DOM)}	Differential output voltage, CANH-CANL (dominant)	See Figure 7-1 and Figure 7-2, TXD = 0 V, $50 \Omega \le R_L \le 65 \Omega$, $C_L = open$	1.5		3.0	V
	Differential output voltage, CANH-CANL (dominant)	See Figure 7-1 and Figure 7-2, TXD = 0 V, $R_L = 2240 \Omega$, $C_L = open$	1.5		5.0	V
\ <u>/</u>	Differential output voltage, CANH-CANL (recessive)	See Figure 7-1 and Figure 7-2, TXD = V $_{\text{CC1}}$, R_{L} = 60 Ω , C_{L} = open	-120.0		12.0	mV
V _{OD(REC)}	Differential output voltage, CANH-CANL (recessive)	See Figure 7-1 and Figure 7-2, TXD = V CC1, R _L = open, C _L = open	-50.0		50.0	mV
V _{SYM_DC}	DC Output symmetry (V _{CC2} - V _{O(CANH)} - V _{O(CANL)})	See Figure 7-1 and Figure 7-2, $R_L = 60$ Ω , $C_L = open$, $TXD = V_{CC1}$ or 0 V	-400.0		400.0	mV
I	Short circuit current steady state output	See Figure 7-9, VCANH = -5 V to 40 V, CANL = open, TXD = 0 V	-100.0			mA
SO(SS_DO M)	Current, dominant See Figure 7-9, VCANL = -5 V to 40 V, CANH = open, TXD = 0 V				100.0	mA
SO(SS_RE	Short circuit current steady state output current, recessive	See Figure 7-9, -27 V ≤ VBUS ≤ 32 V, VBUS = CANH = CANL, TXD = V _{CC1}	-5.0		5.0	mA
RECEIVE	R ELECTRICAL CHARACTERISTICS					
.,	Differential input threshold voltage	See Figure 7-4 and Table 7-1, VCM ≤ 20 V	500.0		900.0	
V _{IT}	Differential input threshold voltage	See Figure 7-4 and Table 7-1, 20 V ≤ VCM ≤ 30 V	400.0		1000.0	mV
V _{HYS}	Hysteresis voltage for differential input threshold	See Figure 7-4 and Table 7-1		120		
V _{CM}	Input common mode range	See Figure 7-4 and Table 7-1	-30.0		30.0	V
I _{OFF(LKG)}	Power-off bus input leakage current	CANH = CANL = 5 V, V_{CC2} to GND via 0 Ω and 47 k Ω resistor			4.8	uA
Cı	Input capacitance to ground (CANH or CANL)	TXD = V _{CC1}		24.0	30	pF
C _{ID}	Differential input capacitance (CANH-CANL)	TXD = V _{CC1}		12.0	15	pF
R _{ID}	Differential input resistance	TXD = V _{CC1} ; -30 V ≤ VCM ≤ +30 V	30.0		80.0	kΩ
R _{IN}	Input resistance (CANH or CANL)	TXD = V _{CC1} ; -30 V ≤ VCM ≤ +30 V	15.0		40.0	kΩ
R _{IN(M)}	Input resistance matching: (1 - R IN(CANH)/RIN(CANL)) x 100%	V _{CANH} = V _{CANL} = 5 V	-2.0		2.0	%
THERMAI	L SHUTDOWN					
T _{TSD}	Thermal shutdown temperature			170		°C
T TSD_HYST	Thermal shutdown hysteresis			5		°C



6.11 Switching Characteristics

Over recommended operating conditions (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DEVICE S	SWITCHING CHARACTERISTICS					
t ppopulos	Total loop delay, driver input TXD to	See Figure 7-6, $R_L = 60 \Omega$, $C_L = 100 pF$, $C_{L(RXD)} = 15 pF$; input rise/fall time (10% to 90%) on TXD =1 ns; 1.71 V \leq V $_{CC1} \leq$ 1.89 V	70	125	198.0	ns
PROP(LOO P1)	receiver RXD, recessive to dominant	See Figure 7-6, $R_L = 60 \Omega$, $C_L = 100 pF$, $C_{L(RXD)} = 15 pF$; input rise/fall time (10% to 90%) on TXD =1 ns; 2.25 V \leq V $CC1 \leq 5.5 V$	70	122	192.0	ns
t	Total loop delay, driver input TXD to	See Figure 7-6, R _L = 60 Ω, C _L = 100 pF, $C_{L(RXD)}$ = 15 pF; input rise/fall time (10% to 90%) on TXD =1 ns; 1.71 V ≤ V _{CC1} ≤ 1.89 V	70	155	215.0	ns
PROP(LOO P2)	receiver RXD, dominant to recessive	See Figure 7-6, $R_L = 60 \Omega$, $C_L = 100 pF$, $C_{L(RXD)} = 15 pF$; input rise/fall time (10% to 90%) on TXD =1 ns; 2.25 V \leq V $CC1 \leq 5.5 V$	70	152	215.0	ns
t UV_RE_EN ABLE	Re-enable time after Undervoltage event	Time for device to return to normal operation from V _{CC1} or V _{CC2} under voltage event			300.0	μs
CMTI	Common mode transient immunity	V _{CM} = 1200 V _{PK} , See Figure 7-10	85	100		kV/µs
DRIVER S	SWITCHING CHARACTERISTICS					
t _{pHR}	Propagation delay time, HIGH TXD to driver recessive			76	120	
t _{pLD}	Propagation delay time, LOW TXD to driver dominant	See Figure 7-3, $R_L = 60 \Omega$ and $C_L = 100$ pF; input rise/fall time (10% to 90%) on		61	120	ns
t _{sk(p)}	Pulse skew (tpHR - tpLD)	TXD =1 ns		14		
t_R	Differential output signal rise time			45		
t_{F}	Differential output signal fall time			45		
V _{SYM}	Output symmetry (dominant or recessive) (V _{O(CANH)} + V _{O(CANL)}) / V _{CC2}	See Figure 7-3 and Figure 9-4 , R_{TERM} = 60 Ω , C_{SPLIT} = 4.7 nF, C_L = open, R_L = open, TXD = 250 kHz, 1 MHz	0.9		1.1	V/V
t _{TXD_DTO}	Dominant time out	See Figure 7-8, R_L = 60 Ω and C_L = open	1.2		3.8	ms
RECEIVE	R SWITCHING CHARACTERISTICS					
t _{pRH}	Propagation delay time, bus recessive input to RXD high output			75	130	ns
t _{pDL}	Propogation delay time, bus dominant input to RXD low output	See Figure 7-5, C _{L(RXD)} = 15 pF		63	130	ns
t _R	Output signal rise time(RXD)			1.4		ns
t _F	Output signal fall time(RXD)			1.8		ns
FD TIMIN	G PARAMETERS	·		,		
	Bit time on CAN bus output pins with t BIT(TXD) = 500 ns	See Figure 7-7, R_L = 60 Ω , C_L = 100 pF, $C_{L(RXD)}$ = 15 pF; input rise/fall time (10% to 90%) on TXD =1 ns	435.0		530.0	ns
t _{BIT(BUS)}	Bit time on CAN bus output pins with t BIT(TXD) = 200 ns	See Figure 7-7, R_L = 60 Ω , C_L = 100 pF, $C_{L(RXD)}$ = 15 pF; input rise/fall time (10% to 90%) on TXD =1 ns	155.0		210.0	ns

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Over recommended operating conditions (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP MAX	UNIT
4	Bit time on RXD output pins with t _{BIT(TXD)} = 500 ns	See Figure 7-7, R_L = 60 Ω , C_L = 100 pF, $C_{L(RXD)}$ = 15 pF; input rise/fall time (10% to 90%) on TXD =1 ns	400	550.0	ns
t _{BIT(RXD)}	Bit time on RXD output pins with t _{BIT(TXD)} = 200 ns	See Figure 7-7, R_L = 60 Ω , C_L = 100 pF, $C_{L(RXD)}$ = 15 pF; input rise/fall time (10% to 90%) on TXD =1 ns	120.0	120.0 220.0	ns
ΔtREC	Receiver timing symmetry with t _{BIT(TXD)} = 500 ns	See Figure 7-7, R_L = 60 Ω , C_L = 100 pF, $C_{L(RXD)}$ = 15 pF; input rise/fall time (10% to 90%) on TXD =1 ns; $\Delta tREC = t_{BIT(RXD)}$ - $t_{BIT(BUS)}$	-65.0	40.0	ns
Διιίο	Receiver timing symmetry with t _{BIT(TXD)} = 200 ns	See Figure 7-7, R_L = 60 Ω , C_L = 100 pF, $C_{L(RXD)}$ = 15 pF; input rise/fall time (10% to 90%) on TXD =1 ns; $\Delta tREC = t_{BIT(RXD)}$ - $t_{BIT(BUS)}$	-45.0	15.0	ns

6.12 Insulation Characteristics Curves

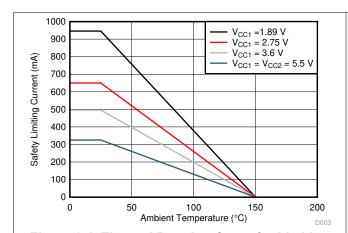


Figure 6-1. Thermal Derating Curve for Limiting **Current per VDE for DW-16 Package**

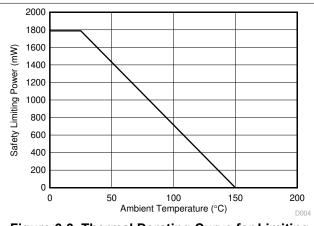


Figure 6-3. Thermal Derating Curve for Limiting Power per VDE for DW-16 Package

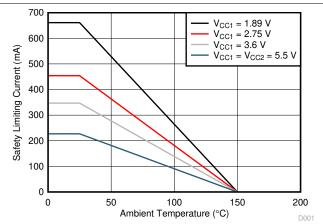


Figure 6-2. Thermal Derating Curve for Limiting **Current per VDE for DWV-8 Package**

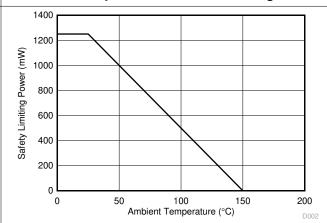


Figure 6-4. Thermal Derating Curve for Limiting Power per VDE for DWV-8 Package



6.13 Typical Characteristics

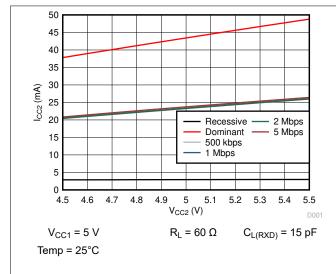


Figure 6-5. I_{CC2} vs V_{CC2} for Recessive, Dominant and Different CAN Datarates

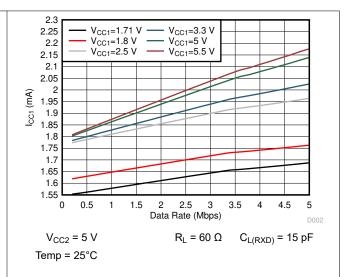


Figure 6-6. I_{CC1} vs Datarate

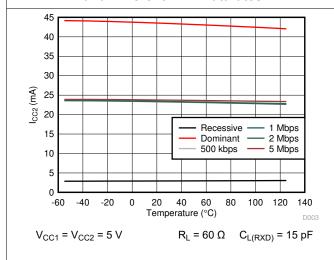


Figure 6-7. I_{CC2} vs Ambient Temperature for Recessive, Dominant and Different CAN Datarates

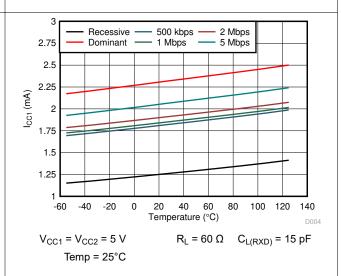


Figure 6-8. : ICC1 vs Ambient Temperature for Recessive, Dominant and Different CAN Datarates.

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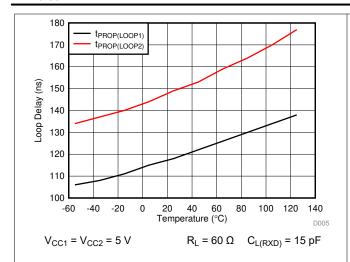


Figure 6-9. Loop Delay vs Ambient Temperature

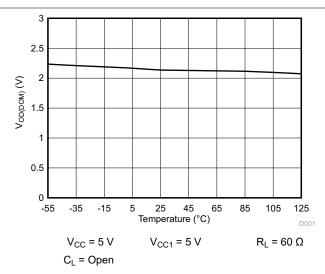


Figure 6-10. V_{OD(DOM)} Over Temperature

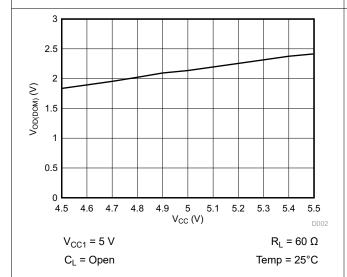


Figure 6-11. V_{OD(DOM)} Over V_{CC}

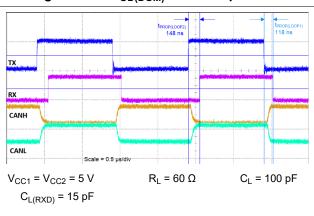


Figure 6-12. Typical TXD, RXD, CANH and CANL Waveforms at 1 Mbps

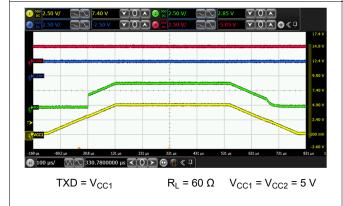


Figure 6-13. Glitch Free Power Up on V_{CC1} – CAN Bus Remains Recessive

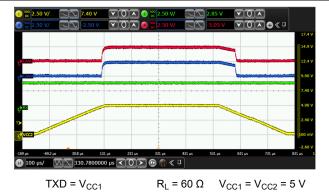


Figure 6-14. Glitch Free Power Up on V_{CC2} – CAN Bus Remains Recessive



7 Parameter Measurement Information

7.1 Test Circuits

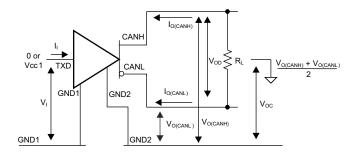


Figure 7-1. Driver Voltage, Current and Test Definitions

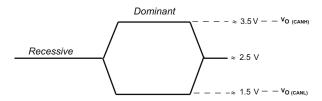
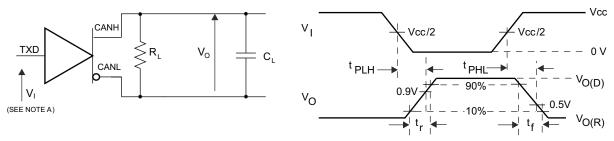


Figure 7-2. Bus Logic State Voltage Definitions



A. The input pulse is supplied by a generator having the following characteristics: PRR \leq 125 kHz, 50% duty cycle, $t_r \leq$ 6 ns, $t_f \leq$ 6 ns, $t_G =$ 50 Ω .

Figure 7-3. Driver Test Circuit and Voltage Waveforms

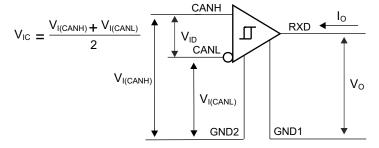
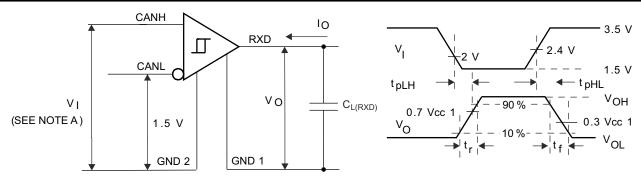


Figure 7-4. Receiver Voltage and Current Definitions

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A. The input pulse is supplied by a generator having the following characteristics: PRR \leq 125 kHz, 50% duty cycle, $t_r \leq$ 6 ns, $t_f \leq$ 6 ns, $Z_O =$ 50 Ω .

Figure 7-5. Receiver Test Circuit and Voltage Waveforms

Table 7-1. Receiver Differential Input Voltage Threshold Test

INPUT			OUTPUT		
V _{CANH}	V _{CANL}	V _{ID}	RXD		
-29.5 V	-30.5 V	1000 mV	L		
30.5 V	29.5 V	1000 mV	L	,,	
-19.55 V	-20.45 V	900 mV	L	V _{OL}	
20.45 V	19.55 V	900 mV	L		
-19.75 V	-20.25 V	500 mV	Н		
20.25 V	19.75 V	500 mV	Н		
-29.8 V	-30.2 V	400 mV	Н	V _{OH}	
30.2 V	29.8 V	400 mV	Н		
Open	Open	X	Н		

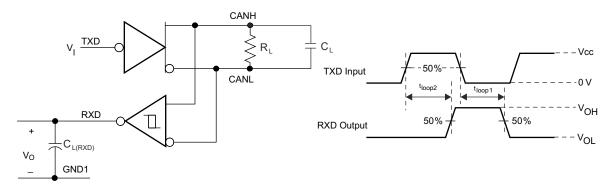


Figure 7-6. t_{LOOP} Test Circuit and Voltage Waveforms



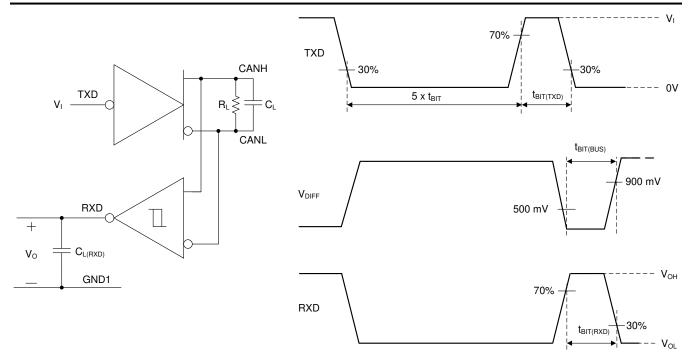
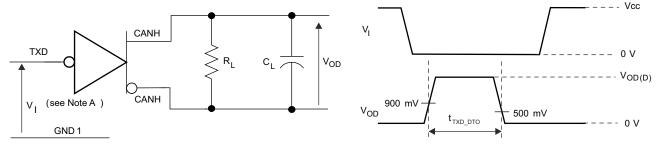


Figure 7-7. CAN FD Timing Parameter Measurement



A. The input pulse is supplied by a generator having the following characteristics: $t_r \le 6$ ns, $t_f \le 6$ ns, $Z_O = 50$ Ω .

Figure 7-8. Dominant Time-out Test Circuit and Voltage Waveforms

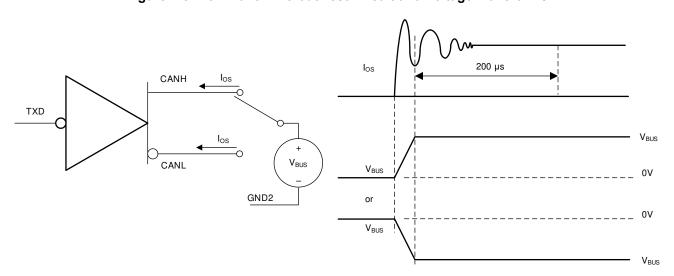


Figure 7-9. Driver Short-Circuit Current Test Circuit and Waveforms

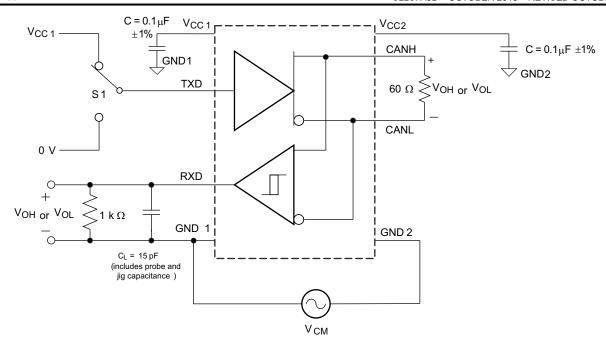


Figure 7-10. Common-Mode Transient Immunity Test Circuit

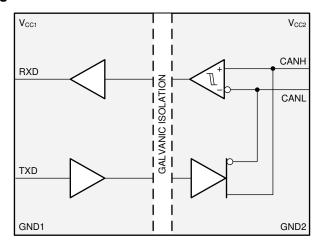


8 Detailed Description

8.1 Overview

The ISO1042-Q1 device is a digitally isolated CAN transceiver that offers $\pm 70\text{-V}$ DC bus fault protection and $\pm 30\text{-V}$ common-mode voltage range. The device supports up to 5-Mbps data rate in CAN FD mode allowing much faster transfer of payload compared to classic CAN. The ISO1042-Q1 device has an isolation withstand voltage of 5000 V_{RMS} and is available in basic and reinforced isolation with a surge test voltage of 6 kV_{PK} and 10 kV_{PK} respectively. The device can operate from 1.8-V, 2.5-V, 3.3-V, and 5-V supplies on side 1 and a 5-V supply on side 2. This supply range is of particular advantage for applications operating in harsh industrial environments because the low voltage on side 1 enables the connection to low-voltage microcontrollers for power conservation, whereas the 5 V on side 2 maintains a high signal-to-noise ratio of the bus signals.

8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 CAN Bus States

The CAN bus has two states during operation: *dominant* and *recessive*. A dominant bus state, equivalent to logic low, is when the bus is driven differentially by a driver. A recessive bus state is when the bus is biased to a common mode of V_{CC} / 2 through the high-resistance internal input resistors of the receiver, equivalent to a logic high. The host microprocessor of the CAN node uses the TXD pin to drive the bus and receives data from the bus on the RXD pin. See Figure 8-1 and Figure 8-2.

Product Folder Links: ISO1042-Q1

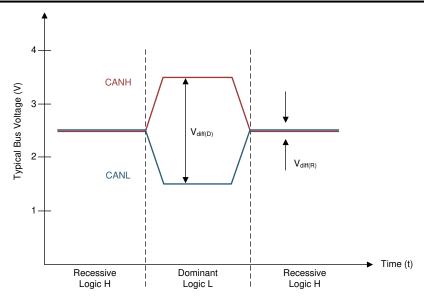


Figure 8-1. Bus States (Physical Bit Representation)

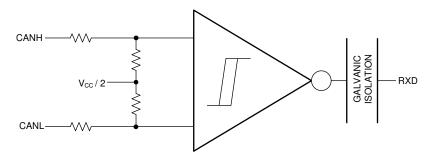


Figure 8-2. Simplified Recessive Common Mode Bias and Receiver

8.3.2 Digital Inputs and Outputs: TXD (Input) and RXD (Output)

The V_{CC1} supply for the isolated digital input and output side of the device can be supplied by 1.8-V, 2.5-V, 3.3-V, and 5-V supplies and therefore the digital inputs and outputs are 1.8-V, 2.5-V, 3.3-V, and 5-V compatible.

Note

The TXD pin is very weakly internally pulled up to V_{CC1} . An external pullup resistor should be used to make sure that the TXD pin is biased to recessive (high) level to avoid issues on the bus if the microprocessor does not control the pin and the TXD pin floats. The TXD pullup strength and CAN bit timing require special consideration when the device is used with an open-drain TXD output on the CAN controller of the microprocessor. An adequate external pullup resistor must be used to make sure that the TXD output of the microprocessor maintains adequate bit timing input to the input on the transceiver.

8.3.3 Protection Features

8.3.3.1 TXD Dominant Timeout (DTO)

The TXD DTO circuit prevents the transceiver from blocking network communication in the event of a hardware or software failure where the TXD pin is held dominant longer than the timeout period, t_{TXD_DTO} . The DTO circuit timer starts on a falling edge on the TXD pin. The DTO circuit disables the CAN bus driver if no rising edge occurs before the timeout period expires, which frees the bus for communication between other nodes on the network. The CAN driver is activated again when a recessive signal occurs on the TXD pin, clearing the TXD DTO condition. The receiver and RXD pin still reflect activity on the CAN bus, and the bus terminals are biased to the recessive level during a TXD dominant timeout.



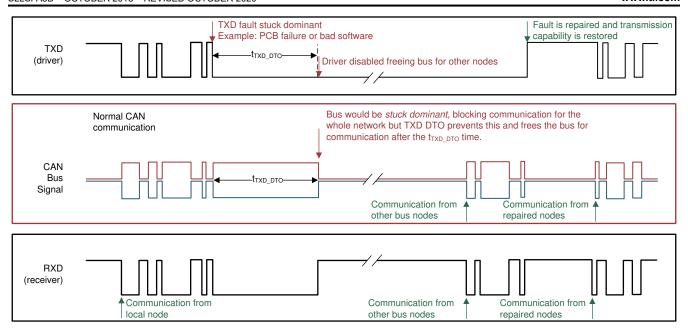


Figure 8-3. Example Timing Diagram for TXD DTO

Note

The minimum dominant TXD time (t_{TXD_DTO}) allowed by the TXD DTO circuit limits the minimum possible transmitted data rate of the device. The CAN protocol allows a maximum of eleven successive dominant bits (on TXD) for the worst case, where five successive dominant bits are followed immediately by an error frame. This, along with the t_{TXD_DTO} minimum, limits the minimum data rate. Calculate the minimum transmitted data rate with Equation 1.

Minimum Data Rate =
$$11 / t_{TXD DTO}$$
 (1)

8.3.3.2 Thermal Shutdown (TSD)

If the junction temperature of the device exceeds the thermal shutdown threshold (T_{TSD}), the device turns off the CAN driver circuits, blocking the TXD-to-bus transmission path. The CAN bus terminals are biased to the recessive level during a thermal shutdown, and the receiver-to-RXD path remains operational. The shutdown condition is cleared when the junction temperature drops at least the thermal shutdown hysteresis temperature (T_{TSD} HYST) below the thermal shutdown temperature (T_{TSD}) of the device.

8.3.3.3 Undervoltage Lockout and Default State

The supply pins have undervoltage detection that places the device in protected or default mode which protects the bus during an undervoltage event on the V_{CC1} or V_{CC2} supply pins. If the bus-side power supply, V_{CC2} , is less than about 4 V, the power shutdown circuits in the ISO1042-Q1 device disable the transceiver to prevent false transmissions because of an unstable supply. If the V_{CC1} supply is still active when this occurs, the receiver output (RXD) goes to a default HIGH (recessive) value. Table 8-1 summarizes the undervoltage lockout and fail-safe behavior.

Table 8-1. Undervoltage Lockout and Default State

V _{CC1}	V _{CC2}	DEVICE STATE	BUS OUTPUT	RXD		
> UV _{VCC1}	> UV _{VCC2}	Functional	Per Device State and TXD	Mirrors Bus		
<uv<sub>VCC1</uv<sub>	> UV _{VCC2}	Protected	Recessive	Undetermined		
>UV _{VCC1}	< UV _{VCC2}	Protected	High Impedance	Recessive (Default High)		

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Note

After an undervoltage condition is cleared and the supplies have returned to valid levels, the device typically resumes normal operation in $300 \mu s$.

8.3.3.4 Floating Pins

Pullup and pulldown resistors should be used on critical pins to place the device into known states if the pins float. The TXD pin should be pulled up through a resistor to the V_{CC1} pin to force a recessive input level if the microprocessor output to the pin floats.

8.3.3.5 Unpowered Device

The device is designed to be *ideal passive* or *no load* to the CAN bus if it is unpowered. The bus pins (CANH, CANL) have extremely low leakage currents when the device is unpowered to avoid loading down the bus which is critical if some nodes of the network are unpowered while the rest of the of network remains in operation.

8.3.3.6 CAN Bus Short Circuit Current Limiting

The device has two protection features that limit the short circuit current when a CAN bus line has a short-circuit fault condition. The first protection feature is driver current limiting (both dominant and recessive states) and the second feature is TXD dominant state time out to prevent permanent higher short circuit current of the dominant state during a system fault. During CAN communication the bus switches between dominant and recessive states, therefore the short circuit current may be viewed either as the instantaneous current during each bus state or as an average current of the two states. For system current (power supply) and power considerations in the termination resistors and common-mode choke ratings, use the average short circuit current. Determine the ratio of dominant and recessive bits by the data in the CAN frame plus the following factors of the protocol and PHY that force either recessive or dominant at certain times:

- · Control fields with set bits
- Bit stuffing
- · Interframe space
- TXD dominant time out (fault case limiting)

These factors ensure a minimum recessive amount of time on the bus even if the data field contains a high percentage of dominant bits. The short circuit current of the bus depends on the ratio of recessive to dominant bits and their respective short circuit currents. Use Equation 2 to calculate the average short circuit current.

$$I_{OS(AVG)} = \text{\%Transmit} \times [(\text{\%REC_Bits} \times I_{OS(SS)_REC}) + (\text{\%DOM_Bits} \times I_{OS(SS)_DOM})] + [\text{\%Receive} \times I$$
os(ss) REC] (2)

where

- I_{OS(AVG)} is the average short circuit current
- %Transmit is the percentage the node is transmitting CAN messages
- %Receive is the percentage the node is receiving CAN messages
- %REC_Bits is the percentage of recessive bits in the transmitted CAN messages
- %DOM Bits is the percentage of dominant bits in the transmitted CAN messages
- I_{OS(SS)} REC is the recessive steady state short circuit current
- I_{OS(SS)} DOM is the dominant steady state short circuit current

Note

Consider the short circuit current and possible fault cases of the network when sizing the power ratings of the termination resistance and other network components.

8.4 Device Functional Modes

Table 8-2 and Table 8-3 list the driver and receiver functions. Table 8-4 lists the functional modes for the ISO1042-Q1 device.



Table 8-2. Driver Function Table

INPUT	OUTI	DRIVEN BUS STATE	
TXD ⁽¹⁾	CANH ⁽¹⁾	CANL ⁽¹⁾	DRIVEN BUS STATE
L	Н	L	Dominant
Н	Z	Z	Recessive

(1) H = high level, L = low level, Z = common mode (recessive) bias to V_{CC} / 2. See Figure 8-1 and Figure 8-2 for bus state and common mode bias information.

Table 8-3. Receiver Function Table

DEVICE MODE	CAN DIFFERENTIAL INPUTS V _{ID} = V _{CANH} - V _{CANL} (3)	BUS STATE R	
Normal	$V_{ID} \ge V_{IT(MAX)}$	Dominant	L
	$V_{IT(MIN)} < V_{ID} < V_{IT(MAX)}$?	?
	$V_{ID} \le V_{IT(MIN)}$	Recessive	Н
	Open (V _{ID} ≈ 0 V)	Open	Н

(1) H = high level, L = low level, ? = indeterminate.

Table 8-4. Function Table

DRIVER				RECEIVER			
INPUTS ⁽¹⁾	OUTPUTS		BUS STATE	DIFFERENTIAL INPUTS V	OUTPUT	BUS STATE	
TXD	CANH	CANL	BUSSIAIE	ID = CANH-CANL ⁽³⁾	RXD	BUSSIAIE	
L ⁽²⁾	Н	L	DOMINANT	$V_{ID} \ge V_{IT(MAX)}$	L	DOMINANT	
Н	Z	Z	RECESSIVE	$V_{IT(MIN)} < V_{ID} < V_{IT(MAX)}$?	?	
Open	Z	Z	RECESSIVE	$V_{ID} \le V_{IT(MIN)}$	Н	RECESSIVE	
Х	Z	Z	RECESSIVE	Open (V _{ID} ≈ 0 V)	Н	RECESSIVE	

- (1) H = high level; L = low level; X = irrelevant; ? = indeterminate; Z = high impedance
- (2) Logic low pulses to prevent dominant time-out.
- (3) See Receiver Electrical Characteristics section for input thresholds.

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9 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The ISO1042-Q1 device can be used with other components from Texas Instruments such as a microcontroller, a transformer driver, and a linear voltage regulator to form a fully isolated CAN interface.

9.2 Typical Application

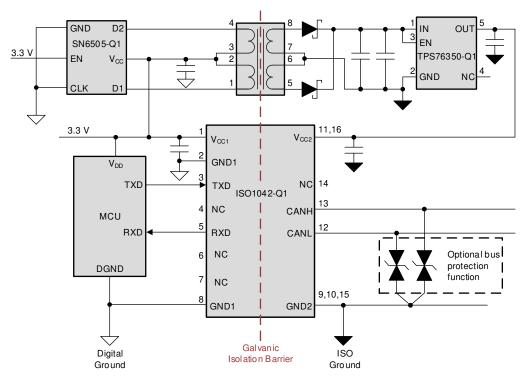


Figure 9-1. Application Circuit With ISO1042-Q1 in 16-SOIC Package



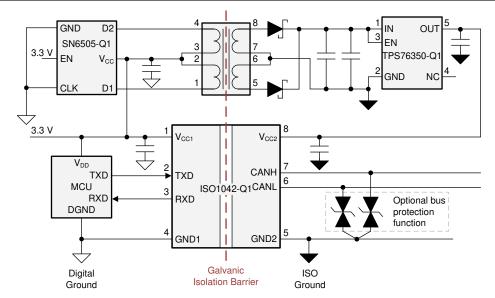


Figure 9-2. Application Circuit With ISO1042-Q1 in 8-SOIC Package

9.2.1 Design Requirements

Unlike an optocoupler-based solution, which requires several external components to improve performance, provide bias, or limit current, the ISO1042-Q1 device only requires external bypass capacitors to operate.

9.2.2 Detailed Design Procedure

9.2.2.1 Bus Loading, Length and Number of Nodes

The ISO 11898-2 Standard specifies a maximum bus length of 40 m and maximum stub length of 0.3 m. However, with careful design, users can have longer cables, longer stub lengths, and many more nodes to a bus. A large number of nodes requires transceivers with high input impedance such as the ISO1042-Q1 transceivers.

Many CAN organizations and standards have scaled the use of CAN for applications outside the original ISO 11898-2 Standard. These organizations and standards have made system-level trade-offs for data rate, cable length, and parasitic loading of the bus. Examples of some of these specifications are ARINC825, CANopen, DeviceNet, and NMEA2000.

The ISO1042-Q1 device is specified to meet the 1.5-V requirement with a 50- Ω load, incorporating the worst case including parallel transceivers. The differential input resistance of the ISO1042-Q1 device is a minimum of 30 k Ω . If 100 ISO1042-Q1 transceivers are in parallel on a bus, this requirement is equivalent to a 300- Ω differential load worst case. That transceiver load of 300 Ω in parallel with the 60 Ω gives an equivalent loading of 50 Ω . Therefore, the ISO1042-Q1 device theoretically supports up to 100 transceivers on a single bus segment. However, for CAN network design margin must be given for signal loss across the system and cabling, parasitic loadings, network imbalances, ground offsets and signal integrity, therefore a practical maximum number of nodes is typically much lower. Bus length may also be extended beyond the original ISO 11898 standard of 40 m by careful system design and data-rate tradeoffs. For example, CANopen network design guidelines allow the network to be up to 1 km with changes in the termination resistance, cabling, less than 64 nodes, and a significantly lowered data rate.

This flexibility in CAN network design is one of the key strengths of the various extensions and additional standards that have been built on the original ISO 11898-2 CAN standard. Using this flexibility requires the responsibility of good network design and balancing these tradeoffs.

9.2.2.2 CAN Termination

The ISO11898 standard specifies the interconnect to be a single twisted pair cable (shielded or unshielded) with $120-\Omega$ characteristic impedance ($Z_{\rm O}$). Resistors equal to the characteristic impedance of the line should be used to terminate both ends of the cable to prevent signal reflections. Unterminated drop-lines (stubs) connecting

nodes to the bus should be kept as short as possible to minimize signal reflections. The termination may be in a node, but if nodes are removed from the bus, the termination must be carefully placed so that it is not removed from the bus.

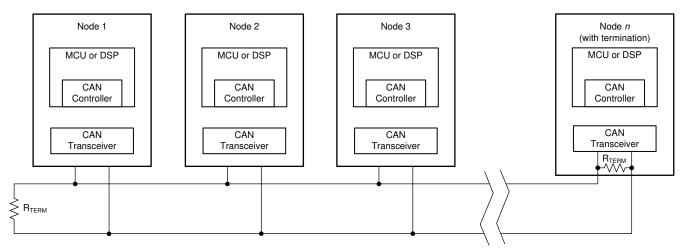


Figure 9-3. Typical CAN Bus

Termination may be a single $120-\Omega$ resistor at the end of the bus, either on the cable or in a terminating node. If filtering and stabilization of the common-mode voltage of the bus is desired, then split termination can be used. (See Figure 9-4). Split termination improves the electromagnetic emissions behavior of the network by eliminating fluctuations in the bus common-mode voltages at the start and end of message transmissions.

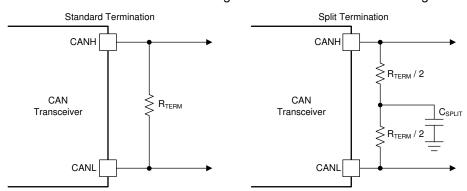
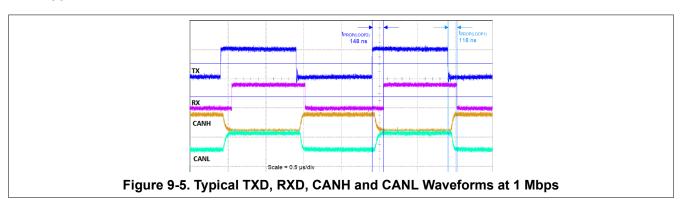


Figure 9-4. CAN Bus Termination Concepts

9.2.3 Application Curve



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10 Power Supply Recommendations

To make sure operation is reliable at all data rates and supply voltages, a 0.1- μ F bypass capacitor is recommended at the input and output supply pins (V_{CC1} and V_{CC2}). The capacitors should be placed as close to the supply pins as possible. In addition, a bulk capacitance, typically 4.7 μ F, should be placed near the V_{CC2} supply pin. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver such as TI's SN6505B. For such applications, detailed power supply design, and transformer selection recommendations are available in the SN6505 Low-Noise 1-A Transformer Drivers for Isolated Power Supplies data sheet.

11 Layout

11.1 Layout Guidelines

A minimum of four layers is required to accomplish a low EMI PCB design (see Section 11.2). Layer stacking should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/in².
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links
 usually have margin to tolerate discontinuities such as vias.

Suggested placement and routing of ISO1042-Q1 bypass capacitors and optional TVS diodes is shown in Figure 11-2 and Figure 11-3. In particular, place the V_{CC2} bypass capacitors on the top layer, as close to the device pins as possible, and complete the connection to the V_{CC2} and G_{ND2} pins without using vias. Note that the SOIC-16 variant needs two V_{CC2} bypass capacitor, one on each V_{CC2} pin.

If an additional supply voltage plane or signal layer is needed, add a second power or ground plane system to the stack to keep it symmetrical. This makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the highfrequency bypass capacitance significantly.

For detailed layout recommendations, refer to the Digital Isolator Design Guide.

11.1.1 PCB Material

For digital circuit boards operating at less than 150 Mbps, (or rise and fall times greater than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 UL94V-0 printed circuit board. This PCB is preferred over lower-cost alternatives because of lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and the self-extinguishing flammability-characteristics.

11.2 Layout Example

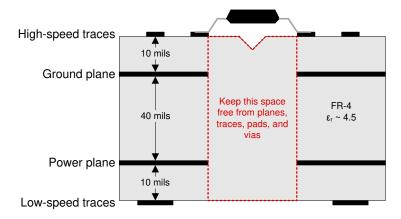


Figure 11-1. Recommended Layer Stack



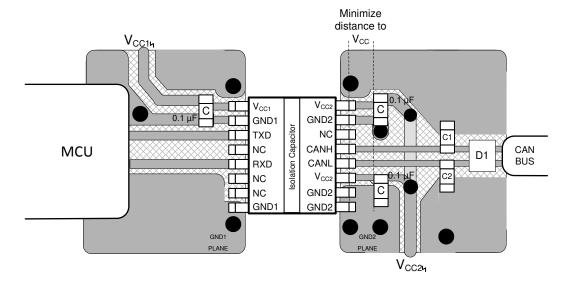


Figure 11-2. 16-DW Layout Example

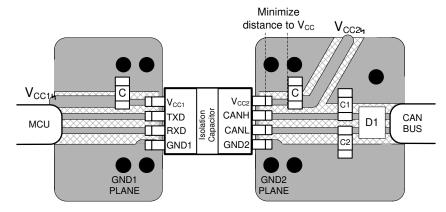


Figure 11-3. 8-DWV Layout Example

12 Device and Documentation Support

12.1 Documentation Support

12.1.1 Related Documentation

For related documentation see the following:

- Texas Instruments, Digital Isolator Design Guide
- Texas Instruments, ISO1042DW Isolated CAN Transceiver Evaluation Module User's Guide
- Texas Instruments, Isolate your CAN systems without compromising on performance or space TI TechNote
- Texas Instruments, Isolation Glossary
- Texas Instruments, High-voltage reinforced isolation: Definitions and test methodologies
- Texas Instruments, How to Isolate Signal and Power in Isolated CAN Systems TI TechNote
- Texas Instruments, How to Design Isolated CAN Systems With Correct Bus Protection Application Report

12.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

12.3 Support Resources

TI E2E™ support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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12.4 Trademarks

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12.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

Orderable part number	Status (1)	Material type	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
ISO1042BQDWQ1	Active	Production	SOIC (DW) 16	40 TUBE	Yes	NIPDAU	NIPDAU Level-3-260C-168 HR		ISO1042BQ1
ISO1042BQDWQ1.A	Active	Production	SOIC (DW) 16	40 TUBE	Yes	NIPDAU	NIPDAU Level-3-260C-168 HR		ISO1042BQ1
ISO1042BQDWQ1.B	Active	Production	SOIC (DW) 16	40 TUBE	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042BQ1
ISO1042BQDWRQ1	Active	Production	SOIC (DW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042BQ1
ISO1042BQDWRQ1.A	Active	Production	SOIC (DW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042BQ1
ISO1042BQDWRQ1.B	Active	Production	SOIC (DW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042BQ1
ISO1042BQDWVQ1	Active	Production	SOIC (DWV) 8	64 TUBE	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042B Q1
ISO1042BQDWVQ1.A	Active	Production	SOIC (DWV) 8	64 TUBE	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042B Q1
ISO1042BQDWVQ1.B	Active	Production	SOIC (DWV) 8	64 TUBE	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042B Q1
ISO1042BQDWVRQ1	Active	Production	SOIC (DWV) 8	1000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042B Q1
ISO1042BQDWVRQ1.A	Active	Production	SOIC (DWV) 8	1000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042B Q1
ISO1042BQDWVRQ1.B	Active	Production	SOIC (DWV) 8	1000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042B Q1
ISO1042QDWQ1	Active	Production	SOIC (DW) 16	40 TUBE	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042Q1
ISO1042QDWQ1.A	Active	Production	SOIC (DW) 16	40 TUBE	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042Q1
ISO1042QDWQ1.B	Active	Production	SOIC (DW) 16	40 TUBE	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042Q1
ISO1042QDWRQ1	Active	Production	SOIC (DW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042Q1
ISO1042QDWRQ1.A	Active	Production	SOIC (DW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042Q1
ISO1042QDWRQ1.B	Active	Production	SOIC (DW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042Q1
ISO1042QDWVQ1	Active	Production	SOIC (DWV) 8	64 TUBE	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042 Q1
ISO1042QDWVQ1.A	Active	Production	SOIC (DWV) 8	64 TUBE	Yes	NIPDAU	NIPDAU Level-3-260C-168 HR		ISO1042 Q1
ISO1042QDWVQ1.B	Active	Production	SOIC (DWV) 8	64 TUBE	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042 Q1
ISO1042QDWVRQ1	Active	Production	SOIC (DWV) 8	1000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042 Q1

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Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
						(4)	(5)		
ISO1042QDWVRQ1.A	Active	Production	SOIC (DWV) 8	1000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042 Q1
ISO1042QDWVRQ1.B	Active	Production	SOIC (DWV) 8	1000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	ISO1042 Q1

⁽¹⁾ Status: For more details on status, see our product life cycle.

- (2) Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.
- (3) RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.
- (4) Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.
- (5) MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.
- (6) Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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OTHER QUALIFIED VERSIONS OF ISO1042-Q1:

Catalog: ISO1042



PACKAGE OPTION ADDENDUM

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NOTE: Qualified Version Definitions:

 $_{\bullet}$ Catalog - TI's standard catalog product

PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO1042BQDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO1042BQDWVRQ1	SOIC	DWV	8	1000	330.0	16.4	12.05	6.15	3.3	16.0	16.0	Q1
ISO1042QDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO1042QDWVRQ1	SOIC	DWV	8	1000	330.0	16.4	12.05	6.15	3.3	16.0	16.0	Q1



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*All dimensions are nominal

7 111 41111011010110 41 0 11011111141							
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO1042BQDWRQ1	SOIC	DW	16	2000	350.0	350.0	43.0
ISO1042BQDWVRQ1	SOIC	DWV	8	1000	350.0	350.0	43.0
ISO1042QDWRQ1	SOIC	DW	16	2000	350.0	350.0	43.0
ISO1042QDWVRQ1	SOIC	DWV	8	1000	350.0	350.0	43.0

PACKAGE MATERIALS INFORMATION

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TUBE



*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
ISO1042BQDWQ1	DW	SOIC	16	40	506.98	12.7	4826	6.6
ISO1042BQDWQ1.A	DW	SOIC	16	40	506.98	12.7	4826	6.6
ISO1042BQDWQ1.B	DW	SOIC	16	40	506.98	12.7	4826	6.6
ISO1042BQDWVQ1	DWV	SOIC	8	64	505.46	13.94	4826	6.6
ISO1042BQDWVQ1.A	DWV	SOIC	8	64	505.46	13.94	4826	6.6
ISO1042BQDWVQ1.B	DWV	SOIC	8	64	505.46	13.94	4826	6.6
ISO1042QDWQ1	DW	SOIC	16	40	506.98	12.7	4826	6.6
ISO1042QDWQ1.A	DW	SOIC	16	40	506.98	12.7	4826	6.6
ISO1042QDWQ1.B	DW	SOIC	16	40	506.98	12.7	4826	6.6
ISO1042QDWVQ1	DWV	SOIC	8	64	505.46	13.94	4826	6.6
ISO1042QDWVQ1.A	DWV	SOIC	8	64	505.46	13.94	4826	6.6
ISO1042QDWVQ1.B	DWV	SOIC	8	64	505.46	13.94	4826	6.6





NOTES:

- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing
- per ASME Y14.5M.

 2. This drawing is subject to change without notice.

 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.





NOTES: (continued)

- 5. Publication IPC-7351 may have alternate designs.
- 6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.





NOTES: (continued)

- 7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 8. Board assembly site may have different recommendations for stencil design.



7.5 x 10.3, 1.27 mm pitch

SMALL OUTLINE INTEGRATED CIRCUIT

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.







NOTES:

- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing
- per ASME Y14.5M.

 2. This drawing is subject to change without notice.

 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
- 5. Reference JEDEC registration MS-013.





NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.





NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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