Question 1

Consider the state-space system

$$x = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 1 & 0 & 0 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \\ -1 \end{bmatrix} u$$
$$y = \begin{bmatrix} 1 & -0.5 & -1 \end{bmatrix} x$$

- (a) Is this system internally asymptotically stable? Why or why not?
- (b) Obtain the transfer function from input u to output y.
- (c) Is this system BIBO stable? Why or why not?

(a)

The system is internally asymptotically stable if all eigenvalues of A have negative real parts. The eigenvalues of A are -0.5, -0.5, 1. The system is not internally asymptotically stable.

A =

V =

D =

>> real(D)

ans =

(b)

The transfer function is:

$$G(s) = C(sI - A)^{-1}B + D$$

Evaluating with Matlab:

$$>> A = [0 1 0; 0 0 1; 1 0 0]$$

A =

B =

1 0 -1

$$>> C = [1 -0.5 -1]$$

C =

$$>> D = 0$$

D =

0

>> syms s

>> C*inv(s*eye(3)-A)*B+D

ans =

$$(2*s^2)/(s^3 - 1) - 3/(2*(s^3 - 1)) - s/(2*(s^3 - 1))$$

Then,

$$G(s) = \frac{2s^2}{s^3 - 1} - \frac{3}{2(s^3 - 1)} - \frac{s}{2(s^3 - 1)}$$

(c)

The system is BIBO stable if all poles of G(s) have negative real parts. The poles of G(s) are 1, -0.5 + 0.8660i, -0.5 - 0.8660i. The system is not BIBO stable.

Question 2

Check the BIBO stability of each of the following transfer functions:

- (a) $\frac{s-1}{s+1}$
- (b) $\frac{s+1}{s(s+2)^2}$
- (c) $\frac{s}{s^2+4}$
- (a)

The pole is at s = -1. The system is BIBO stable.

(b)

The poles are at s=0 and s=-2. The system is not BIBO stable.

(c)

The poles are at $s=0\pm 2j$. The system is not BIBO stable.

Question 3

Consider the following closed-loop system: The controller's state-space matrices are

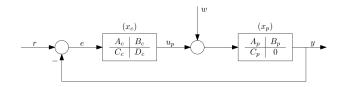


Figure 1: Block diagram of the closed-loop system

$$A_c = -4$$
, $B_c = 1$, $C_c = -5$, $D_c = 1$

and the plant's are

$$A_p = \begin{bmatrix} 0 & 1 \\ -2 & -3 \end{bmatrix}, \quad B_p = \begin{bmatrix} 1 \\ 0 \end{bmatrix}, \quad C_p = \begin{bmatrix} 1 & 0 \end{bmatrix}$$

- (a) Form the closed-loop system dynamics matrix A_{cl} . Verify the internal stability of the system.
- (b) Convert the state-space control and plant blocks into transfer functions $G_c(s)$ and $G_p(s)$, then form the characteristic polynomial. Verify the input-output stability of the system (note you can either compute the roots numerically, or use the Routh-Hurwitz criterion)
- (c) What's the relationship between the results in (a) and (b)?

(a)

To be able to use the developments in the notes, two assumptions have to be satisfied:

- Both controller (A_c, B_c, C_c, D_c) and plant (A_p, B_p, C_p, D_p) are minimal realizations.
- $D_c = 0$ or $D_p = [D_u, D_w] = 0$, such that $D_c D_u = D_u D_c = 0$ and $D_c D_w = D_w D_c = 0$. This is satisfied in this case, since $D_p = 0$.

Then, the state-space form of the closed loop system is calculated by Matlab:

>> Gc

Gc =

$$1 - 5/(s + 4)$$

>> Gp

Gp =

$$(s + 3)/(s^2 + 3*s + 2)$$

Both $G_c(s)$ and $G_p(s)$ are minimal realizations, so the derivations in the notes can be used.

$$A_c l = \begin{bmatrix} A_c & -B_c C_p \\ B_u C_c & A_p - B_u D_c C_p \end{bmatrix}$$

Employing Matlab again,

Acl =

D =

Real part of eigenvalues of Acl:

Since all the eigenvalues of A_{cl} have negative real parts, the closed-loop system is internally stable.

All Matlab calculations can be found in the script:

```
clc; clear; close all;
Ac = -4;
Bc = 1;
Cc = -5;
Dc = 1;

Ap = [0 1; -2 -3];
Bp = [1; 0];
Cp = [1 0];
Dp = 0;

syms s
Gc = Cc*inv(s*eye(1)-Ac)*Bc + Dc
Gp = Cp*inv(s*eye(2)-Ap)*Bp + Dp

Acl = [Ac -Bc*Cp; Bp*Cc Ap-Bp*Dc*Cp]
[V, D] = eig(Acl)
```

```
disp('Real part of eigenvalues of Acl: ')
real(D)
```

Question 4

Recall the analog circuit from Assignment #1

As we saw in Assignment #1, the state-space form of the system is:

$$A = \begin{bmatrix} -1/(R_1C) & 0 \\ 0 & -R_2/L \end{bmatrix}, \quad B = \begin{bmatrix} 1/(R_1C) \\ 1/L \end{bmatrix}$$
$$C = [0, R_2], \quad D = 0$$

Computing the transfer function in Matlab,

G =

R2/(R2 + L*s)

Since
$$size(A) = 2, deg(G(s)) = 1$$
, the system is not minimal as $2 > 1$

The code used to generate this is:

```
clc; clear; close all;
syms s R1 R2 C L
A = [-1/(R1*C) 0; 0 -R2/L];
B = [1/(R1*C); 1/L];
c = [0 R2];
D = 0;
% Transfer function
G = c*inv(s*eye(2)-A)*B + D
```