

## Lecture 8

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## Frequency Response

- ⇒ The steady-state response of a linear time-invariant system to a complex exponential input of the form  $u(t) = e^{st}$  is

$$y_{ss}(t) = G(s) e^{st}$$

- ⇒ If we chose  $s = j\omega$  (real part of input is  $\cos(\omega t)$ ), and we assume that the system is stable, then the steady-state output will be given by

$$y_{ss}(t) = G(s) e^{j\omega t} = |G(j\omega)| e^{j\omega t + \angle G(j\omega)}$$

↓ (real part of output)

$$\Re[y_{ss}(t)] = |G(j\omega)| \cos(\omega t + \angle G(j\omega))$$

- ⇒ The steady-state response to a sinusoidal input of frequency  $\omega$  is a sinusoidal output of the same frequency such that:

- The amplitude of the output is  $|G(j\omega)|$  times the amplitude of the input.
- The phase of the output lags the phase of the input by  $\angle G(j\omega)$ .

## \* Frequency response plots

⇒ The frequency response  $G(j\omega) \in \mathbb{C}$  is a complex function of a single real argument  $\omega \in \mathbb{R}$ .

⇒ We basically have two options to plot the frequency response:

① A parametric curve showing  $G(j\omega)$  in the complex plane, in which  $\omega$  is implicit.

→ This leads to the polar plot and eventually to the **Nyquist plot**.

② Two separate plots for e.g. real and imaginary part of  $G(j\omega)$  or better the magnitude and phase of  $G(j\omega)$  as a function of  $\omega$ .

→ The latter choice leads to the **Bode plot**.

## \* The Bode Plot

⇒ The Bode plot is actually composed of two plots:

→ Magnitude plot

→ Phase plot



⇒ On the horizontal axis of both plots, we report the frequency  $\omega$  on a logarithmic scale (base 10).

⇒ On the vertical axis we report:

→  $|G(j\omega)|$  in dB (decibels)

$$|G(j\omega)| [\text{dB}] = 20 \log_{10} |G(j\omega)|$$

Note: One decade = 20dB

→ The phase  $\angle G(j\omega)$  in degree or radians.

⇒ Since magnitudes multiply (i.e. there log add) and phase add.

→ These choices of vertical coordinates makes it possible to just add Bode plots of serial connections.

⇒ Inverting the transfer function is equivalent to reflection about the horizontal axis, in both plots.

\* Drawing approximate buck plot

$$G(s) = \cancel{K} \frac{(s-z_1)(s-z_2)\dots(s-z_m)}{(s-p_1)(s-p_2)\dots(s-p_n)}$$

⇒ It can be broken down into product of following terms:

$$\textcircled{1} K \text{ (constant)}$$

$$\textcircled{2} \cancel{s} \propto s$$

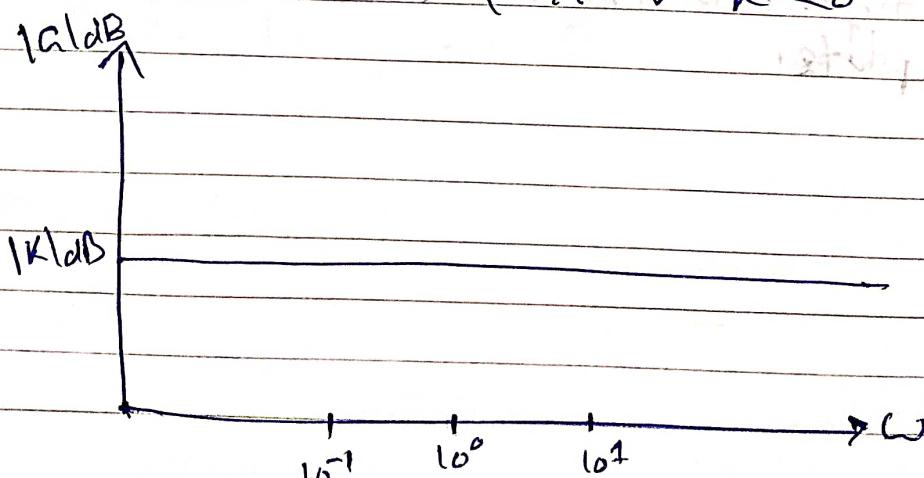
$$\textcircled{3} (1+zs) \propto (1+zs)$$

$$\textcircled{4} \frac{s^2}{\omega_n^2} + 2 \sum_{n=1}^{\infty} s + 1$$

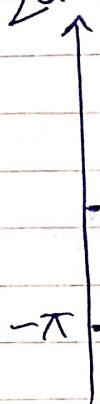
$$\textcircled{1} K \text{ (constant)}$$

$$|K|_{dB} = 20 \log_{10} |K|$$

$$\angle K = \tan^{-1} \left( \frac{\Omega}{K} \right) = \begin{cases} 0 & \forall K > 0 \\ -\pi & \forall K < 0 \end{cases}$$



$\angle G(j\omega)$



$K > 0 \rightarrow \omega$

$K < 0$

$S = 2\pi$   
 $-P_m$

Now

## ② $S, Y_S$

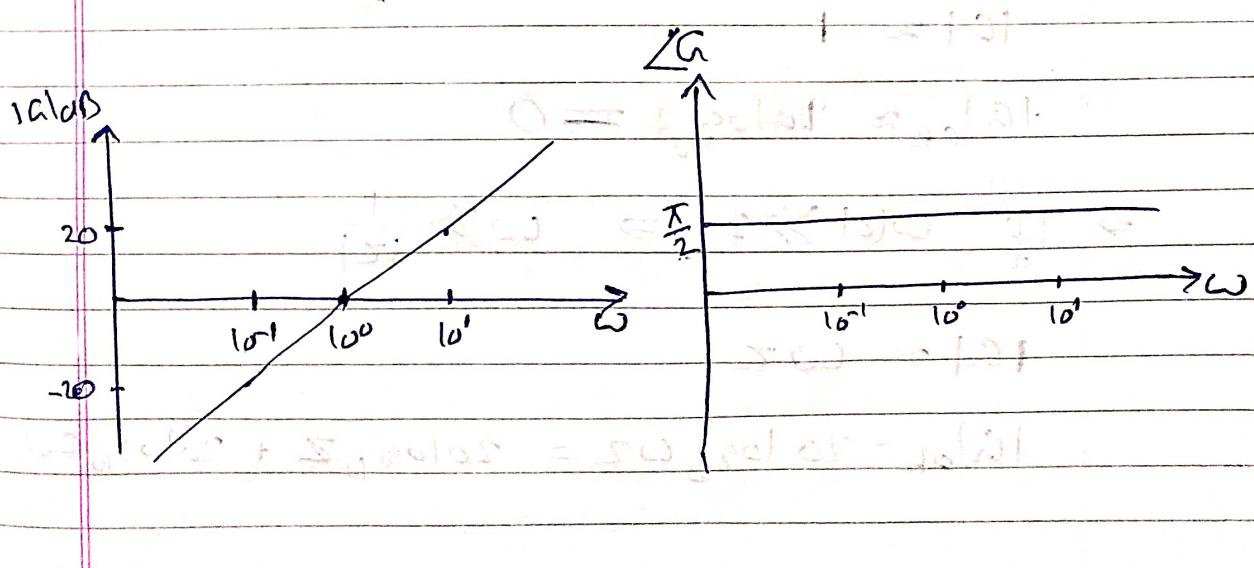
$\Rightarrow$  If we have bode plot of  $S$ , so bode plot of  $Y_S$  is negative of bode plot of  $S$ .

$\Rightarrow$  Or more generally if we have bode plot of  $f(s)$ , bode plot of  $Y_f(s)$  is negative of bode plot of  $f(s)$ .

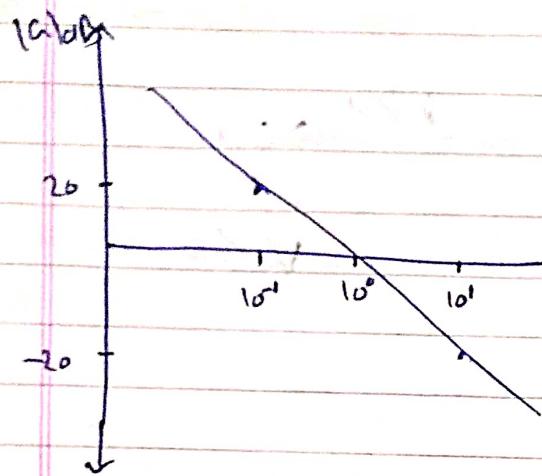
$$G(s) = S \quad G(j\omega) = j\omega \quad |G(j\omega)| = |\omega|$$

$$|G(j\omega)| = \omega \quad \angle G(j\omega) = \tan^{-1}(\frac{\omega}{0}) = \pi/2$$

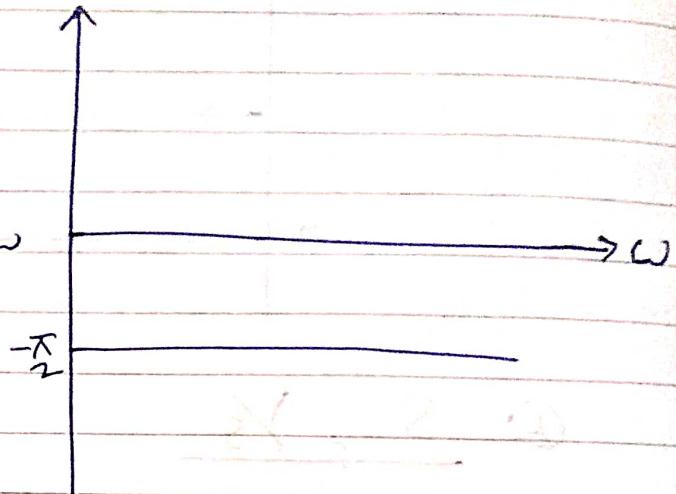
$$|G|_{dB} = 20 \log_{10} \omega$$



$\Rightarrow$  For  $Y_S$



$|G|$



③  $(1 + ZS) \cdot \frac{1}{(1 + ZS)}$

$$G(s) = 1 + Zs$$

$$G(j\omega) = 1 + j\omega Z$$

$$|G| = \sqrt{1^2 + \omega^2 Z^2} = \sqrt{1 + \omega^2 Z^2}$$

$$\Rightarrow \text{If } 1 \gg |\omega Z| \Rightarrow \omega \ll \frac{1}{|Z|}$$

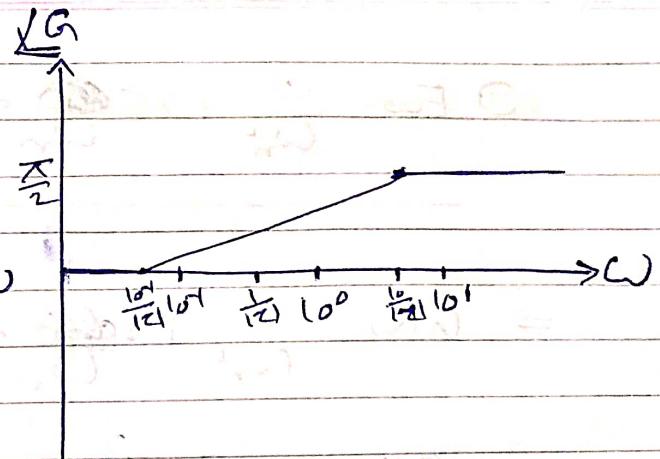
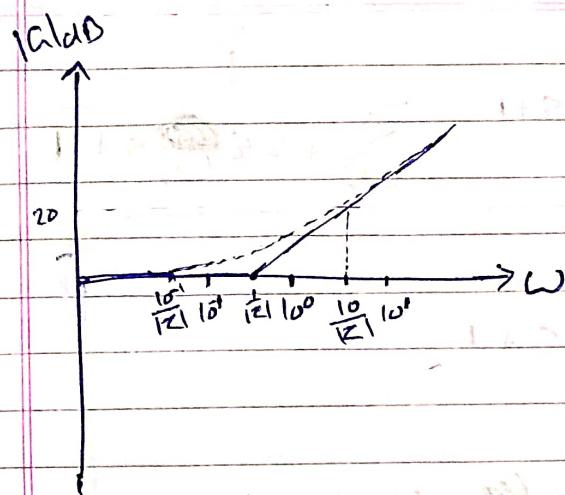
$$|G| \approx 1$$

$$|G|_{dB} \approx 20 \log_{10} 1 = 0$$

$$\Rightarrow \text{If } |\omega Z| \gg 1 \Rightarrow \omega \gg \frac{1}{|Z|}$$

$$|G| \approx \omega Z$$

$$|G|_{dB} = 20 \log_{10} \omega Z = 20 \log_{10} Z + 20 \log_{10} \omega$$



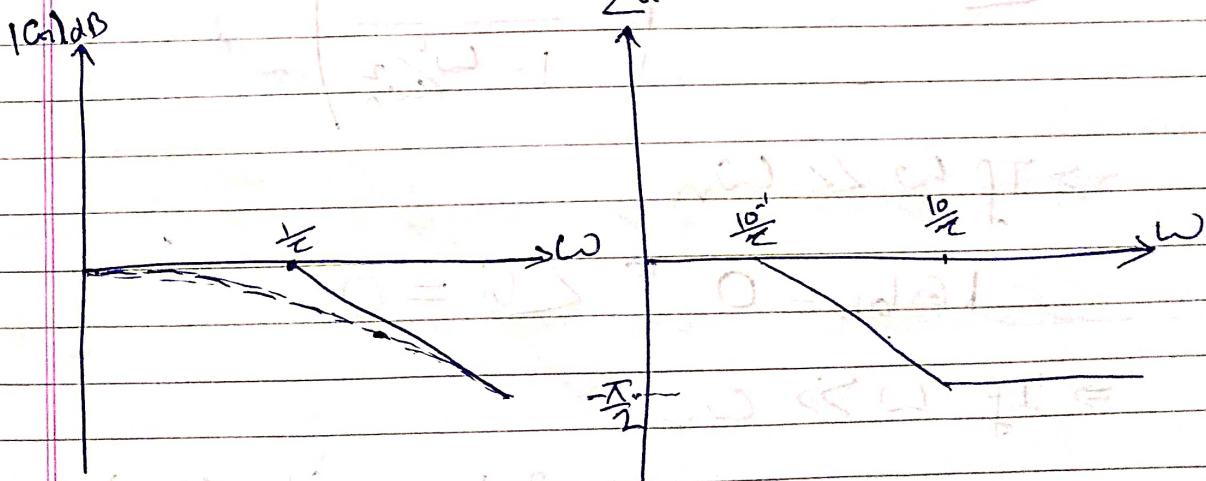
$$\angle G(j\omega) = \tan^{-1} \left( \frac{\omega z}{1} \right)$$

⇒ If  $\omega z \ll 1$ : If  $\omega z \gg 1$

$$\angle G(j\omega) \approx 0$$

$$\angle G(j\omega) \approx \pi/2$$

$$\Rightarrow F_{in} = \frac{1}{1 + zS}$$



$$(\text{Gain})_{\text{at } \omega = 10^0} = (\text{Gain})_{\text{at } \omega = 10^1}$$

$$(\text{Gain})_{\text{at } \omega = 10^0} - (\text{Gain})_{\text{at } \omega = 10^1}$$

④ For  $\frac{s^2}{\omega_n^2} + 2\varepsilon \frac{s}{\omega_n} s + 1$ ,  $\frac{s^2}{\omega_n^2} + 2\varepsilon \frac{s}{\omega_n} s + 1$

$$\Rightarrow G(s) = \frac{s^2}{\omega_n^2} + 2\varepsilon \frac{s}{\omega_n} s + 1$$

$$\Rightarrow G(j\omega) = \frac{(j\omega)^2}{\omega_n^2} + 2\varepsilon \frac{j\omega}{\omega_n} (j\omega) + 1$$

$$= \left(1 - \frac{\omega^2}{\omega_n^2}\right) + \frac{2\varepsilon \omega j}{\omega_n}$$

$$\Rightarrow |G(j\omega)|_{db} = 20 \log_{10} \left( \sqrt{\left(1 - \frac{\omega^2}{\omega_n^2}\right)^2 + \frac{4\varepsilon^2 \omega^2}{\omega_n^2}} \right)$$

$$\Rightarrow \angle G(j\omega) = \tan^{-1} \left( \frac{2\varepsilon \omega / \omega_n}{1 - \omega^2 / \omega_n^2} \right)$$

$\Rightarrow$  If  $\omega \ll \omega_n$

$$|G|_{db} = 0 \quad \angle G = 0$$

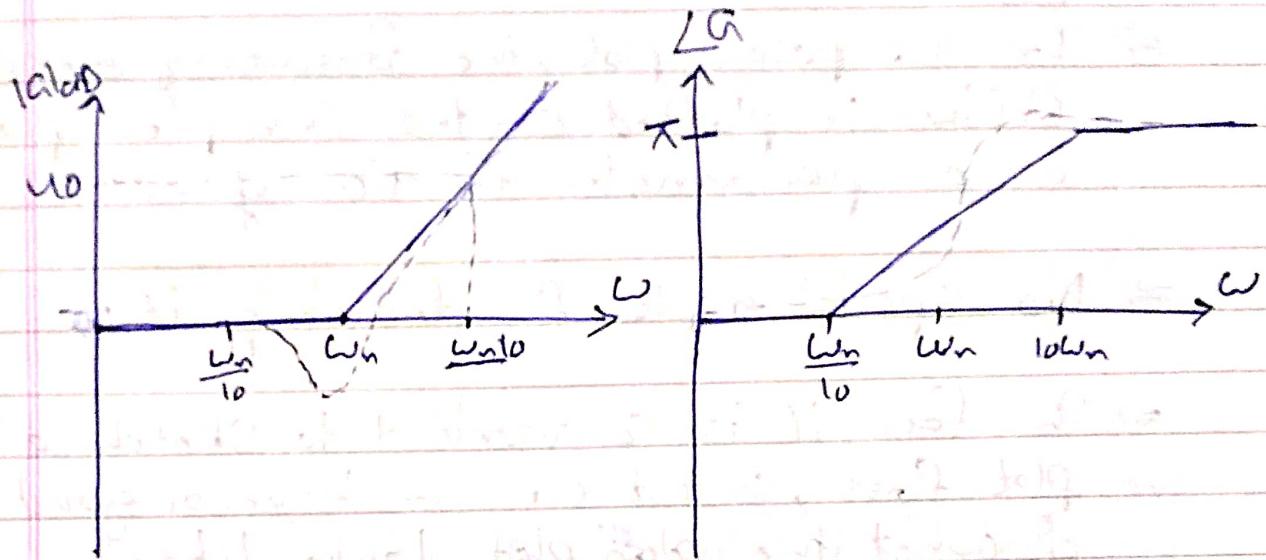
$\Rightarrow$  If  $\omega \gg \omega_n$

$$|G|_{db} = 20 \log_{10} \left( \frac{\omega^2}{\omega_n^2} \right) = 20 \log \left( \frac{\omega}{\omega_n} \right)$$

$$= 20 \log(\omega) - 20 \log(\omega_n)$$

~~EEG~~

$$\angle G \approx \pi$$



$$\Rightarrow \text{For } G(s) = \frac{1}{s^2 + 2\zeta s + \omega_n^2}$$

