

DIGITAL LOGIC

Chapter 2: Introduction to logic circuits

Ru Han

2.11 MINIMIZATION AND KARNAUGH MAPS

Karnaugh Maps (K-Map): Part I

- What are K-Maps?
- Minimization by Using K-Maps: SOP
- Minimization in POS form
- Don't Care Conditions
- Networks with Multiple Outputs
- Design Examples

KARNAUGH MAPS (K-MAP)

- Recall the problems you may confront in manipulation:
 - It is not obvious and tedious
 - Whether it is the simplest expression
- Distributive property and 7a/7b is mostly used
- Shown in Example 2.48

KARNAUGH MAPS (K-MAP)

Row number	x_1	x_2	x_3	f
0	0	0	0	1
1	0	0	1	0
2	0	1	0	1
3	0	1	1	0
4	1	0	0	1
5	1	0	1	1
6	1	1	0	1
7	1	1	1	0

Figure 2.48 The function $f(x_1, x_2, x_3) = \sum m(0, 2, 4, 5, 6)$.

KARNAUGH MAPS (K-MAP)

- A graphical way to represent Boolean functions
- Useful for minimizing functions
- Can generate SOP or POS forms
- Grid of squares:
 - Each square represents a different minterm (or maxterm).
 - Adjacent squares differ in only one variable
- Makes it easy to perform steps like
 - $e.g., x\bar{y}\bar{z} \leftrightarrow x y \bar{z}$
 - $e.g., x\bar{y}\bar{z} + x y \bar{z} = x(\bar{y} + y)\bar{z} = x\bar{z}$

2-VARIABLE MAP: $f(A, B)$

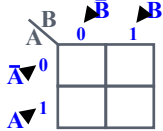
- 4 minterms:

$A \backslash B$	0	1
0	m_0	m_2
1	m_1	m_3

A	B	minterms
0	0	$\bar{A}\bar{B} \quad m_0$
0	1	$\bar{A}B \quad m_1$
1	0	$AB \quad m_2$
1	1	$AB \quad m_3$

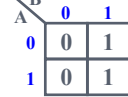
2-VARIABLE MAP: $f(A, B)$

- 4 minterms:

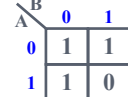


- To represent $f(A, B)$, put 1's in the squares for minterms that are included in f , and 0's in squares that are not included.

$$\text{Ex: } f = \bar{A}B + AB = (\bar{A} + A)B = B$$



$$\text{Ex: } f = \bar{A}\bar{B} + \bar{A}B + A\bar{B} = \bar{A} + \bar{B}$$



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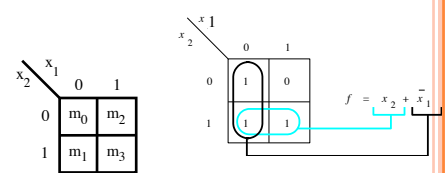
EXAMPLE: HOW TO SIMPLIFY THE 2-VARIABLE MAP:

$f(x_1, x_2)$

$$\text{Ex: } f = \bar{x}_1\bar{x}_2 + \bar{x}_1x_2 + x_1x_2$$

x_1	x_2	
0	0	m_0
0	1	m_1
1	0	m_2
1	1	m_3

(a) Truth table



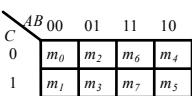
(b) Karnaugh map

Figure 2.49. Location of two-variable minterms.

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3-VARIABLE MAP: $f(A, B, C)$

- 8 minterms:

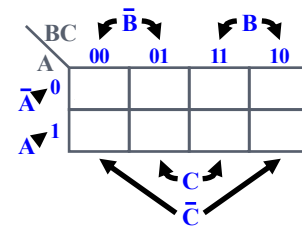


A	B	C	minterms
0	0	0	$\bar{A}\bar{B}\bar{C}$ m_0
0	0	1	$\bar{A}\bar{B}C$ m_1
0	1	0	$\bar{A}B\bar{C}$ m_2
0	1	1	$\bar{A}BC$ m_3
1	0	0	$A\bar{B}\bar{C}$ m_4
1	0	1	$AB\bar{C}$ m_5
1	1	0	ABC m_6
1	1	1	ABC m_7

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3-VARIABLE MAP: $f(A, B, C)$

- 8 minterms:

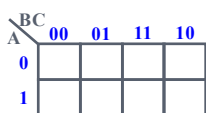


Note the unusual order!!
Adjacent squares must differ in only one variable

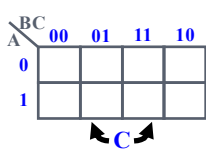
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3-VARIABLE MAP: $f(A, B, C)$

$$\text{Ex: } f = \bar{A}\bar{B}\bar{C} + \bar{A}\bar{B}C + \bar{A}B\bar{C} + A\bar{B}\bar{C}$$



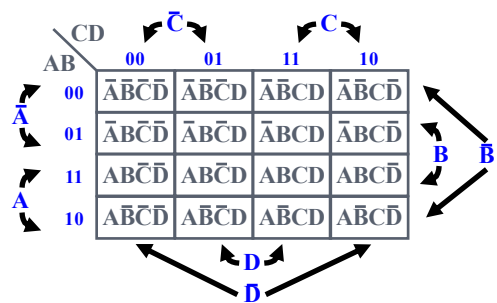
$$\text{Ex: } f = \bar{A}\bar{B}\bar{C} + \bar{A}\bar{B}C + \bar{A}B\bar{C} + ABC = \bar{A}\bar{C} + AC = C$$



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4-VARIABLE MAP: $f(A, B, C, D)$

- 16 minterms:



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4-VARIABLE MAP: $f(A, B, C, D)$

Ex:

$$f = \bar{A}\bar{B}CD + \bar{A}B\bar{C}\bar{D} + \bar{A}B\bar{C}D + \bar{A}BCD + ABCD + A\bar{B}\bar{C}D$$

CD \ AB	00	01	11	10
00				
01				
11				
10				

Ex:

$$\begin{aligned} f &= AB\bar{C}\bar{D} + AB\bar{C}D \\ &\quad + ABC\bar{D} + ABCD \\ &= AB\bar{C} + ABC = AB \end{aligned}$$

CD \ AB	00	01	11	10
00				
01				
11				
10				

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$$F = \underbrace{ABC}_{(110 \times)} + \underbrace{CD}_{(\times \times 11)} + \underbrace{A}_{(1 \times \times \times)}$$

CD \ AB	00	01	11	10
00	0	0	1	1
01	0	0	1	1
11	1	1	1	1
10	0	0	1	1

$$F = (A + \bar{B} + D)(\bar{A} + C)$$

CD \ AB	00	01	11	10
00	1	0	0	0
01	1	1	0	0
11	1	1	1	1
10	1	0	1	1

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2.12 MINIMIZATION USING K-MAPS: SOP

Principle

- Combine rectangles containing 2, 4, 8, or 16 adjacent 1's
- More bigger the circle is, more simpler the expression is
- Note: Squares at opposite edges are considered adjacent

Ex: $\bar{A}\bar{B}\bar{C} \leftrightarrow \bar{A}\bar{B}\bar{C}$

BC \ A	0	1	1	0
0	$\bar{A}\bar{B}\bar{C}$	$\bar{A}\bar{B}C$	$\bar{A}B\bar{C}$	$\bar{A}BC$
1	$A\bar{B}\bar{C}$	$A\bar{B}C$	$AB\bar{C}$	ABC

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MINIMIZATION USING K-MAPS: JARGON

- Implicant:** Any rectangle containing 2^n adjacent ones that represents a product term.
- Prime Implicant:** An implicant that is not part of a larger implicant containing fewer literals

Ex: $f(x, y, z) = \sum m(2, 3, 4, 6, 7)$

Implicants:

yz \ x	0	1	1	0
0	0	0	1	1
1	1	0	1	1

Prime Implicants:

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MINIMIZATION USING K-MAPS: JARGON

- Essential Prime Implicant:** A prime implicant that includes a square which is not part of any other prime implicant.
- Cover:** Any collection of implicants that includes all squares where $f = 1$. Each cover corresponds to a SOP form for f .

yz \ x	0	1	1	0
0	0	0	1	1
1	1	0	1	1

$$f = x\bar{z} + yz + \bar{x}y$$

yz \ x	0	1	1	0
0	0	0	1	1
1	1	0	1	1

$$f = y + x\bar{z}$$

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MINIMIZATION USING K-MAPS: JARGON

- Cost:** Cost = # Gates + # Gate Inputs
- Except complements of input variables are available at zero cost (for simplicity, we don't count inverters.)

(i) minterms: $f = \bar{x}y\bar{z} + \bar{x}yz + x\bar{y}\bar{z} + x\bar{y}z + xyz$

$$\text{Cost}_i = 6 + (3 \times 5 + 1 \times 5) = 6 + 20 = 26$$

(ii) $f = x\bar{z} + yz + \bar{x}y$

yz \ x	0	1	1	0
0	0	0	1	1
1	1	0	1	1

$$\text{Cost}_{ii} = 4 + (2 \times 3 + 1 \times 3) = 4 + 9 = 13$$

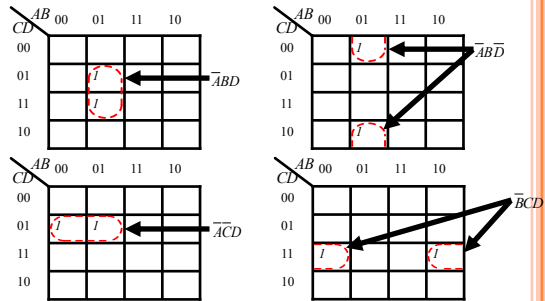
(iii) $f = y + x\bar{z}$

yz \ x	0	1	1	0
0	0	0	1	1
1	1	0	1	1

$$\text{Cost}_{iii} = 2 + (1 \times 2 + 1 \times 2) = 2 + 4 = 6$$

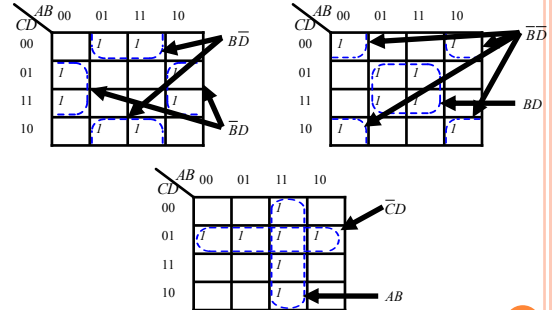
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PRINCIPLE OF MINIMIZATION USING K-MAPS



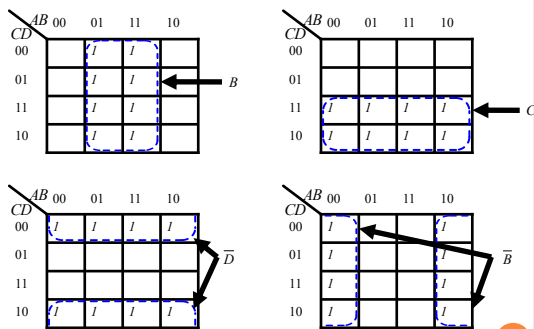
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PRINCIPLE OF MINIMIZATION USING K-MAPS



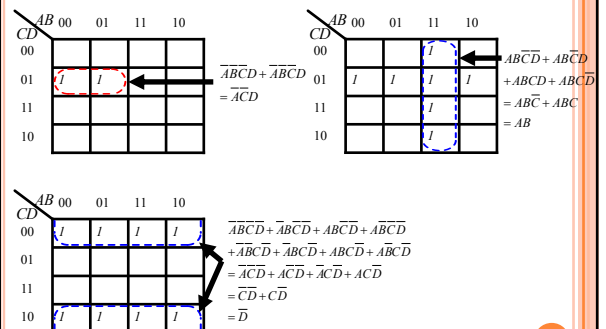
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PRINCIPLE OF MINIMIZATION USING K-MAPS



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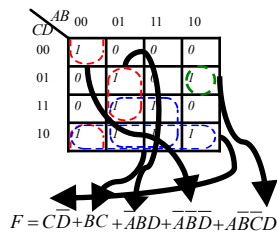
PRINCIPLE OF MINIMIZATION USING K-MAPS



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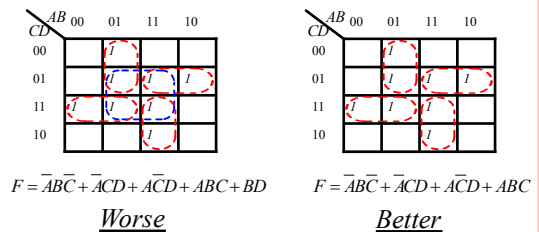
PRINCIPLE OF MINIMIZATION USING K-MAPS

- Steps of the minimization using K-map



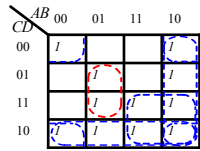
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MINIMIZATION USING K-MAPS: EXAMPLE 2



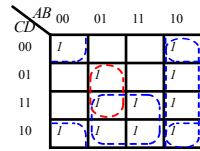
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MINIMIZATION USING K-MAPS: EXAMPLE 3



$$F = \overline{B}\overline{D} + \overline{A}\overline{B} + \overline{C}\overline{D} + \overline{A}C + \overline{A}BD$$

Worse

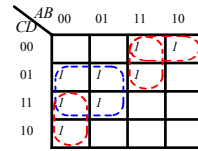


$$F = \overline{B}\overline{D} + \overline{A}\overline{B} + \overline{B}C + \overline{A}BD$$

Better

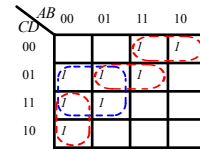
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MINIMIZATION USING K-MAPS: EXAMPLE 4



$$F = \overline{A}D + \overline{A}\overline{B}C + \overline{A}\overline{C}\overline{D} + \overline{A}BC$$

Better



$$F = \overline{A}D + \overline{A}\overline{B}C + \overline{A}\overline{C}\overline{D} + \overline{B}\overline{C}D$$

Better

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MINIMIZATION USING K-MAPS: SOP

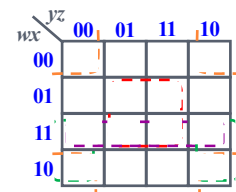
Procedure:

- Build K-map for function f
- Combine adjacent squares containing 1's into rectangles of size 2, 4, 8, or 16
- Determine prime implicants, essential P.I.'s
- Find the largest circle with one different minterm
- Find all essential prime implicants
- Add the minimum number of non-essential prime implicants needed to cover f
- Form the sum of the selected implicants

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MINIMIZATION USING K-MAPS: SOP

Example: $f(w,x,y,z) = \sum m(0,2,5,7,8,10,12,13,14,15)$



Prime Implicants:

Essential Prime Implicants:

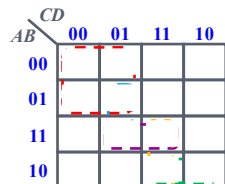
$$\left. \begin{aligned} f &= xz + \overline{x}z + wx \\ f &= xz + \overline{x}z + w\overline{z} \end{aligned} \right\} \text{Cost} = 4 + 9 = 13$$

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MINIMIZATION USING K-MAPS: SOP

Example:

$$f(A,B,C,D) = \overline{A}\overline{B}\overline{C} + \overline{A}\overline{C}D + ABD + ACD + \overline{A}B\overline{C}\overline{D} + \overline{A}BC$$



Prime Implicants:

$$\overline{A}C, \overline{B}CD, ABD, ACD, \overline{A}BC$$

Essential: $\overline{A}C, \overline{A}BC$

Cover 1:

$$f = \overline{A}C + \overline{A}BC + \overline{B}CD + ACD$$

$$\text{cost} = 5 + 15 = 20$$

Cover 2:

$$f = \overline{A}C + \overline{A}BC + ABD$$

$$\text{cost} = 4 + 11 = 15$$

Rule of Thumb:
Minimize overlap among prime implicants!

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2.13 MINIMIZATION USING K-MAPS: POS

Procedure:

- Build K-map for function f
- Combine adjacent squares containing 0's into rectangles of size 2, 4, 8, or 16
- Determine prime implicants, essential P.I.'s
- Find minimum cover of all squares where $f = 0$
- Express in terms of sum terms that correspond to selected prime implicants
- Compare cost to SOP realization

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MINIMIZATION USING K-MAPS: POS

Example: $f(w,x,y,z) = \sum m(0,2,5,7,8,10,12,13,14,15)$

yz	00	01	11	10
wx	1	0	0	1
01	0	1	1	0
11	1	1	1	1
10	1	0	0	1

Prime Implicants: $\bar{x}z, \bar{w}x\bar{z}$

Essential Prime Implicants: $\bar{x}z, \bar{w}x\bar{z}$

$$\bar{f} = \bar{x}z + \bar{w}x\bar{z}$$

$$f = \overline{\bar{x}z + \bar{w}x\bar{z}}$$

$$= (\bar{x}z)(\bar{w}x\bar{z})$$

$$f = (x + \bar{z})(w + \bar{x} + z)$$

$$\text{Cost of POS} = 3 + 7 = 10$$

$$\left. \begin{aligned} f &= xz + \bar{x}\bar{z} + wx \\ f &= xz + \bar{x}\bar{z} + w\bar{z} \end{aligned} \right\} \text{Cost} = 4 + 9 = 13$$

Cost??

Less than SOP!

MINIMIZATION USING K-MAPS: POS

Example: $f(w,x,y,z) = \sum m(0,2,5,7,8,10,12,13,14,15)$

yz	00	01	11	10
wx	1	0	0	1
01	0	1	1	0
11	1	1	1	1
10	1	0	0	1

Prime Implicants: $\bar{x}z, \bar{w}x\bar{z}$

Essential Prime Implicants: $\bar{x}z, \bar{w}x\bar{z}$

How to obtain the same result directly from 2 P.I.s ?

What sum term is 0 for a given P.I.?

$$\begin{aligned} x = 0, z = 1 & \quad w = 0, x = 1, z = 0 \\ (x + \bar{z}) & \quad (w + \bar{x} + z) \end{aligned}$$

$$f = (x + \bar{z})(w + \bar{x} + z)$$

More examples:

$x_1 x_2$	00	01	11	10
x_3	1	1	0	0
1	1	1	1	0

$$(\bar{x}_1 + x_3)$$

$$(x_1 + x_2)$$

Figure 2.60. POS minimization of $f(x_1, x_2, x_3) = \prod M(4, 5, 6)$.

More examples:

$x_1 x_2$	00	01	11	10
$x_3 x_4$	0	0	0	0
01	0	1	1	0
11	1	1	0	1
10	1	1	1	1

$$(x_3 + x_4)$$

$$(x_2 + x_3)$$

$$(\bar{x}_1 + \bar{x}_2 + \bar{x}_3 + \bar{x}_4)$$

Figure 2.61. POS minimization of $f(x_1, \dots, x_4) = \prod M(0, 1, 4, 8, 9, 12, 15)$.

CONCLUSION

- K-map is an effective way to find the minimal-cost circuit
- Learn the way to simplify the 3/4 variables logic function
- Textbook Reading: Chapter 2.11 ~ 2.14
- Assignment: None

Karnaugh Maps (K-Map): Part II

- What are K-Maps?
- Minimization by Using K-Maps: SOP
- Minimization in POS form
- Don't Care Conditions
- Networks with Multiple Outputs
- Design Examples

2.14 INCOMPLETELY SPECIFIED FUNCTIONS DON'T CARE CONDITIONS

- In some cases, particular input combinations can never occur. These are **Don't Care Conditions**.

- Example: Buffer with Set/Reset inputs

Inputs: Set (S), Reset (R), Data (D)

Output: f

If $S = R = 0$, $f = D$
 If $S = 0, R = 1$, $f = 0$
 If $S = 1, R = 0$, $f = 1$
 $S = R = 1$, *Cannot occur*

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DON'T CARE CONDITIONS

- When drawing K-maps, place a **d** in squares for don't care conditions.
- Use these **d**'s as either 0's or 1's, in whatever way to create the largest prime implicants.
- Can count some as 1's and others as 0's. $S=R=0, f=D$
 $S=0, R=1, f=0$
 $S=1, R=0, f=1$
 $S=R=1$, *Cannot occur*
- Example:

SR	00	01	11	10
D	0			
	1			

$$\text{SOP: } f = S + \bar{R}D$$

SR	00	01	11	10
D	0			
	1			

$$\text{POS: } f = \bar{R} (S + D)$$

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DON'T CARE CONDITIONS

Example: Let $wxyz$ represent a 4-bit number n

$$f = \begin{cases} 1 & \text{if } 2 \leq n \leq 5 \\ 0 & \text{otherwise} \end{cases}$$

yz	00	01	11	10
wx	00			
	01			
	11			
	10			

$$\text{SOP: } f = \bar{w}\bar{x}y + \bar{w}x\bar{y}$$

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DON'T CARE CONDITIONS

Example: Let $wxyz$ represent a 4-bit number n

$$f = \begin{cases} 1 & \text{if } 2 \leq n \leq 5 \\ 0 & \text{otherwise} \end{cases}$$

yz	00	01	11	10
wx	00	0	0	1
	01	1	1	0
	11	0	0	0
	10	0	0	0

$$\text{SOP: } f = \bar{w}\bar{x}y + \bar{w}x\bar{y}$$

$$\text{POS: } f = \bar{w} (\bar{x} + \bar{y}) (x + y)$$

$$f = \bar{w} (x \oplus y)$$

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DON'T CARE CONDITIONS

Example: Now assume that n must have a decimal value 0-9. Other cases \Rightarrow don't care

It is a BCD (Binary Coded Decimal)

yz	00	01	11	10
wx	00			
	01			
	11			
	10			

$$\text{SOP: } f = \bar{x}y + x\bar{y}$$

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DON'T CARE CONDITIONS

Example: Now assume that n must have a decimal value 0-9. Other cases \Rightarrow don't care

It is a BCD (Binary Coded Decimal)

yz	00	01	11	10
wx	00	0	0	1
	01	1	1	0
	11	d	d	d
	10	0	0	d

$$\text{SOP: } f = \bar{x}y + x\bar{y}$$

$$\text{POS: } f = (\bar{x} + \bar{y}) (x + y)$$

$$f = x \oplus y$$

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DON'T CARE CONDITIONS

SO:

- To Form the largest groups, you can assign proper value to the Don't care Cell.
- The freedom in choosing the value of don't care leads to greatly simplified realizations

- Example in Figure 2.62
- Example 2.15

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DON'T CARE CONDITIONS

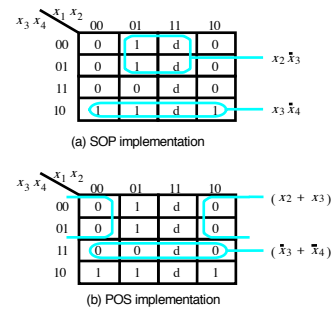
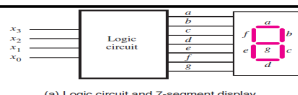


Figure 2.62. Two implementations of the function $f(x_1, \dots, x_4) = \Sigma m(2, 4, 5, 6, 10) + D(12, 13, 14, 15)$.

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(a) Logic circuit and 7-segment display

x_3	x_2	x_1	x_0	a	b	c	d	e	f	g
0	0	0	0	1	1	1	1	1	0	
0	0	0	1	0	1	1	0	0	0	
0	0	1	0	1	1	0	1	1	0	
0	0	1	1	1	1	1	0	0	1	
0	1	0	0	0	1	1	0	0	1	
0	1	0	1	1	0	1	1	0	1	
0	1	1	0	1	0	1	1	1	1	
0	1	1	1	1	1	1	1	1	1	
1	0	0	0	1	1	1	1	1	1	
1	0	0	1	1	1	1	1	0	1	

(b) Truth table

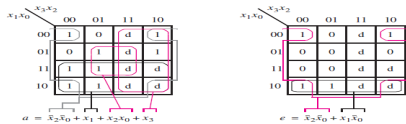


Figure 2.63. Using don't-care minterms when displaying BCD numbers.

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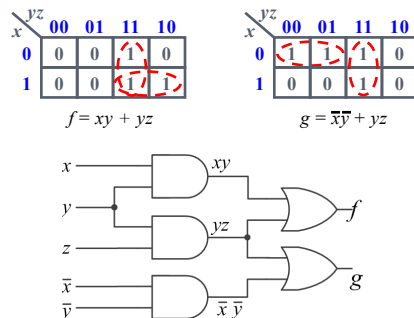
2.15 NETWORKS WITH MULTIPLE OUTPUTS

- Often, it is necessary to implement two or more functions of the same inputs
Example: arithmetic
- In many cases, some gates can be shared between the two functions to keep the total cost down.
- Such cases can be recognized by finding prime implicants that are common to both K-maps.

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NETWORKS WITH MULTIPLE OUTPUTS

Example 1:



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NETWORKS WITH MULTIPLE OUTPUTS

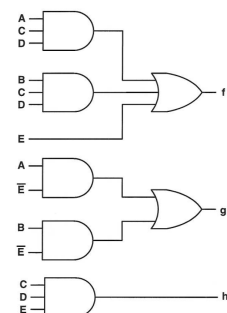
Example 2: Minimal SOP – Separate circuits

$$f = ACD + BCD + E$$

$$g = AE + B\bar{E}$$

$$h = CDE$$

Inverter is 0 cost.
cost = 7 + 19 = 26

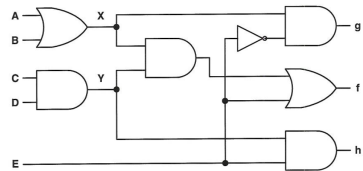


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NETWORKS WITH MULTIPLE OUTPUTS

$$\begin{aligned} f &= ACD + BCD + E = (A + B)CD + E & f &= XY + E \\ g &= A\bar{E} + B\bar{E} = (A + B)\bar{E} & \Rightarrow g &= X\bar{E} \\ h &= CDE & h &= YE \end{aligned}$$

Let $X = A + B$, $Y = CD$



Inverter is 0 cost.
cost = 6 + 12 = 18

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Example 2.16

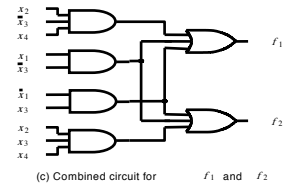
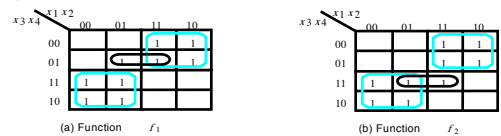


Figure 2.64. An example of multiple-output synthesis.

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Example 2.17

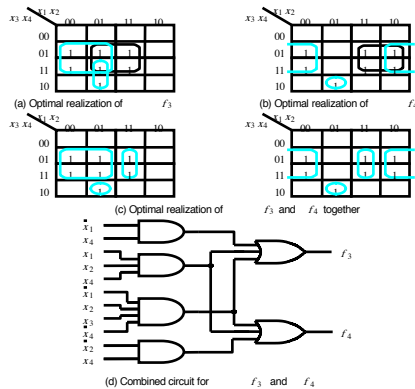


Figure 2.65. An example of multiple-output synthesis.

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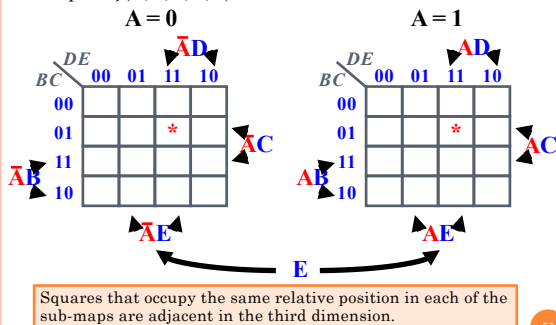
FIVE-VARIABLE K-MAP

- A K-map must have the property that two minterms which differ in only one variable lie in adjacent squares.
- For a two-dimensional map, this is not possible with more than 4 variables.
- However, we can create a 5-variable map as two 4-variable maps that lie on top of another.

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FIVE-VARIABLE K-MAP

- Map for $f(A,B,C,D,E)$



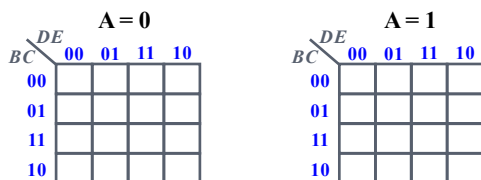
Squares that occupy the same relative position in each of the sub-maps are adjacent in the third dimension.

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FIVE-VARIABLE K-MAP

- Example:

$$f(A,B,C,D,E) = \sum(2,5,7,8,10,13,15,17,19,21,23,24,29,31)$$

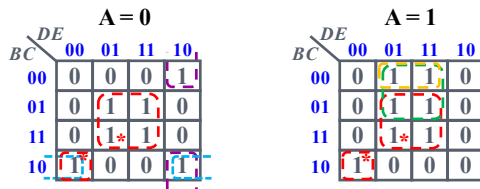


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FIVE-VARIABLE K-MAP

Example:

$$f(A,B,C,D,E) = \sum(2,5,7,8,10,13,15,17,19,21,23,24,29,31)$$



Prime Implicants: CE, $\overline{A}CDE$, ABE, BCDE, $\overline{A}BCE$

$$f = CE + \overline{A}CDE + ABE + BCDE$$

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2.17 DESIGN EXAMPLE

BCD to Gray Code Converter

k	A	B	C	D	w	x	y	z
0	0	0	0	0	0	0	0	0
1	0	0	0	1	0	0	0	1
2	0	0	1	0	0	0	1	1
3	0	0	1	1	0	0	1	0
4	0	1	0	0	0	1	1	0
5	0	1	0	1	1	1	1	0
6	0	1	1	0	1	0	1	0
7	0	1	1	1	1	0	1	1
8	1	0	0	0	1	0	0	1
9	1	0	0	1	1	0	0	0
10	1	0	1	0	X	X	X	X
11	1	0	1	1	X	X	X	X
12	1	1	0	0	X	X	X	X
13	1	1	0	1	X	X	X	X
14	1	1	1	0	X	X	X	X
15	1	1	1	1	X	X	X	X

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DESIGN EXAMPLE



$$w = A + BD + BC$$

$$x = B\overline{C}$$

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DESIGN EXAMPLE



$$y = B + C$$

$$z = \overline{A}\overline{B}\overline{C}D + BCD + A\overline{D} + \overline{B}\overline{C}D$$

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DESIGN EXAMPLE

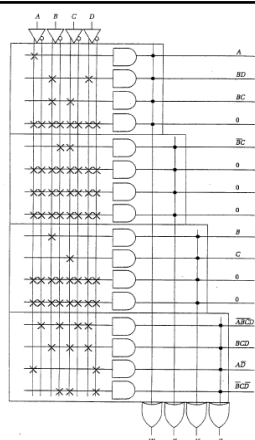
PAL Implementation

$$w = A + BD + BC$$

$$x = B\overline{C}$$

$$y = B + C$$

$$z = \overline{A}\overline{B}\overline{C}D + BCD + A\overline{D} + \overline{B}\overline{C}D$$



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MORE EXAMPLE FOR CHAPTER 2 Pages: 102-111

CONCLUSIONS OF THIS CHAPTER

Chapter 2 is the key of the book

2.16 Remarks:

- The concept of logic circuit
- Boolean algebra
- Synthesis
- CAD tool and Verilog
- K-map

Textbook Reading: Chapter 2.15 ~ 2.17

Assignment : 2.37, 2.38, 2.40, 2.45, 2.48, 2.69

Note: using K-map method