Source and Channel Polarization over Finite Fields and Reed-Solomon Matrices

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Abstract—Polarization phenomenon over any finite field \mathbb{F}_q with size q being a power of a prime is considered. This problem is a generalization of the original proposal of channel polarization by Arıkan for the binary field, as well as its extension to a prime field by Şaşoğlu, Telatar, and Arıkan. In this paper, a necessary and sufficient condition of a matrix over a finite field \mathbb{F}_q is shown under which any source and channel are polarized. Furthermore, the result of the speed of polarization for the binary alphabet obtained by Arıkan and Telatar is generalized to arbitrary finite field. It is also shown that the asymptotic error probability of polar codes is improved by using the Reed-Solomon matrices, which can be regarded as a natural generalization of the 2×2 binary matrix used in the original proposal by Arıkan.

Index Terms—Polar code, channel polarization, source polarization, Reed-Solomon code, Reed-Muller code.

I. Introduction

RIKAN introduced the method of source and channel A polarization which gives efficient capacity-achieving binary source and channel codes, respectively [3]. Şaşoğlu et al. generalized the polarization phenomenon to non-binary alphabets whose size is a prime [4]. They showed an example of a quaternary channel which is not polarized by Arıkan's 2×2 matrix. Although there are channels not polarized by Arikan's 2×2 matrix for non-prime alphabets, one can argue that any channel is polarized in a weaker sense, as discussed in [5]. From this observation, the symmetric capacity of any nonbinary channel is efficiently achievable by directly using the channel polarization phenomenon [4], [5], [6], [7]. In [8], a sufficient condition for a matrix over a ring $\mathbb{Z}/q\mathbb{Z}$ is shown on which any q-ary channel is polarized. In this paper, we study the polarization phenomenon caused by matrices over finite fields.

The contributions of this paper are threefold. The first contribution is that we give a complete characterization as to whether an $\ell \times \ell$ matrix over a finite field gives rise to polarization. This extends the result on the binary field by Korada et al. [9] to a general finite field. The second contribution is that we characterize the asymptotic speed of polarization in terms of the matrix used. This is again an

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extension of the result on the binary field by Korada et al. [9] to a general finite field. The third contribution of this paper is that we provide an explicit construction of an $\ell \times \ell$ matrix, which is based on the Reed-Solomon matrix, with asymptotically the fastest polarization for $\ell < q$.

The organization of this paper is as follows. In Section II, notations and definitions used in this paper are introduced. In Section III, the basic transform of a source and polarization phenomenon by an $\ell \times \ell$ matrix over a finite field are introduced. In Section IV, an equivalence relation of qary source is defined for showing equivalence among several polarization problems. On the concept of equivalence among sources, equivalence of matrices is considered as well. Using the equivalence of matrices, the main theorem of this paper is stated, which is a necessary and sufficient condition of matrix under which any source or channel is polarized. In Section V, the Bhattacharyya parameter and its properties are shown. They are useful for proving the main theorem in Section VI and speed of the polarization in Section VII. In Section VI, a proof of the main theorem is shown. In Section VII, the speed of the polarization for a general $\ell \times \ell$ matrix is proved similarly to the binary case. In Section VIII, the Reed-Solomon matrices are introduced, which yield asymptotically the fastest polarization in the sense discussed in Section VII. In Section IX, the quaternary polar codes using a Reed-Solomon matrix are compared numerically with the original binary polar codes. Finally, Section X summarizes the paper.

II. PRELIMINARIES

Let p be a prime number and $q:=p^m$ where m is a natural number. Let \mathbb{F}_q be a finite field of size q. Let \mathbb{F}_q^{\times} be $\mathbb{F}_q\setminus\{0\}$ and $\mathbb{F}_p(\gamma)$ be the simple extension of \mathbb{F}_p generated by the adjunction of $\gamma\in\mathbb{F}_q$. Similarly, for $A\subseteq\mathbb{F}_q$ and a matrix G over \mathbb{F}_q , $\mathbb{F}_p(A)$ and $\mathbb{F}_p(G)$ denote the field extensions of \mathbb{F}_p generated by the adjunction of all elements of A and G, respectively. Let $\Delta_q:=\{[p_1,\ldots,p_q]\in\mathbb{R}_{\geq 0}^q\mid p_1+\cdots+p_q=1\}$ denote the set of all q-dimensional probability vectors. For random variables X on a finite set \mathcal{X} of size q and Y on a discrete set \mathcal{Y} , entropy H(X) of X and conditional entropy $H(X\mid Y)$ of X conditioned on Y are defined as

$$H(X) := -\sum_{x \in \mathcal{X}} P_X(x) \log P_X(x)$$

$$H(X \mid Y) := -\sum_{x \in \mathcal{X}, y \in \mathcal{Y}} P_{X,Y}(x, y) \log P_{X|Y}(x \mid y).$$

In this paper, the base of the logarithm is assumed to be q unless otherwise stated, and hence H(X) and $H(X \mid Y)$ are

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in [0,1]. If a quantity A((X,Y)) determined from $P_{X,Y}$ has the form $\mathbb{E}[f([P_{X|Y}(x\mid Y)]_{x\in\mathbb{F}_q})]$ for some $f:\Delta_q\to\mathbb{R}$, where \mathbb{E} denotes the expectation, we write it as $A(X\mid Y)$ (Here, $P_{X|Y}(x\mid Y)$ means the random variable g(x,Y) where $g(x,y):=P_{X|Y}(x\mid y)$). It should be noted that the arguments in this paper are directly applicable to the case where $\mathcal Y$ is a continuous alphabet such as $\mathbb R$, by replacing the summation $\sum_{y\in\mathcal Y}$ with the integral $\int_{-\infty}^{+\infty}\mathrm{d}y$. The notation $u_0^{\ell-1}$ denotes the row vector $[u_0,u_1,\ldots,u_{\ell-1}]$.

III. SOURCE AND CHANNEL POLARIZATION

A. Source and channel polarization phenomenon

In this paper, we consider source polarization on an $\ell \times \ell$ invertible matrix G over \mathbb{F}_q . Let a q-ary source (X,Y) be defined as a pair of random variables on $\mathbb{F}_q \times \mathcal{Y}$. We first introduce a *basic transform* of source, $(X,Y) \to \{(X^{(i)},Y^{(i)})\}_{i=0,\dots,\ell-1}$.

Definition 1 (Basic transform). Let $\{(X_i,Y_i)\}_{i=0,\dots,\ell-1}$ be ℓ independent drawings of (X,Y). Let $U_0^{\ell-1}$ be a random vector defined by the equation $X_0^{\ell-1} = U_0^{\ell-1}G$. Letting $(X^{(i)},Y^{(i)}):=(U_i,(U_0^{i-1},Y_0^{\ell-1}))$ for $i=0,\dots,\ell-1$ defines the basic transform $(X,Y)\to\{(X^{(i)},Y^{(i)})\}_{i=0,\dots,\ell-1}$ where the $\sigma(X_0^{\ell-1},Y_0^{\ell-1})$ -measurable random pair $(X^{(i)},Y^{(i)})$ takes values in $\mathbb{F}_q\times(\mathbb{F}_q^i\times\mathcal{Y}^\ell)$.

From the chain rule for the entropy, one has

$$\ell H(X \mid Y) = H(X_0^{\ell-1} \mid Y_0^{\ell-1}) = H(U_0^{\ell-1} \mid Y_0^{\ell-1})$$

$$= \sum_{i=0}^{\ell-1} H(U_i \mid U_0^{i-1}, Y_0^{\ell-1}) = \sum_{i=0}^{\ell-1} H(X^{(i)} \mid Y^{(i)}). \quad (1)$$

By starting with a source (X,Y) and recursively applying the basic transform to depth n, we obtain ℓ^n random pairs $\{(X^{(b_1)\cdots(b_n)},Y^{(b_1)\cdots(b_n)})\}_{(b_1,\dots,b_n)\in\{0,\dots,\ell-1\}^n}^1$. Let B_1,\dots,B_n,\dots be independent uniform random variables on $\{0,\dots,\ell-1\}$. The random process $\{(X_n,Y_n)\}_{n=0,1,\dots}$ defined via the recursive applications of the basic transform as follows puts the foundation of whatever will be discussed in this paper.

Definition 2. Let $(X_n, Y_n) := (X^{(B_1)\cdots(B_n)}, Y^{(B_1)\cdots(B_n)})$ be a $\sigma(X_0^{\ell^n-1}, Y_0^{\ell^n-1}, B_1, \ldots, B_n)$ -measurable random variable for $n \in \{0, 1, \ldots\}$.

A random sequence $\{H_n\colon \sigma(B_1,\dots,B_n)$ -measurable $\}_{n=0,1,1}$ is defined as $H_n:=H(\mathsf{X}_n\mid\mathsf{Y}_n)$ where the conditional entropy does not take account of randomness of (B_1,\dots,B_n) . From the chain rule (1) for the entropy, the random sequence $\{H_n\}_{n=0,1,\dots}$ is shown to be a martingale i.e., $\mathbb{E}[H_n\mid B_1,\dots,B_{n-1}]=H_{n-1}$. Then, noting that the sequence $\{H_n\}_{n=0,1,\dots}$ is bounded in the interval [0,1], from the martingale convergence theorem, there exists a random variable H_∞ such that H_n converges to H_∞ almost surely. The source polarization is defined in terms of H_∞ as in the following definition.

Definition 3 (Polarization). A source (X, Y) is said to be polarized by G if and only if

$$H_{\infty} = \begin{cases} 0, & \text{with probability } 1 - H(X \mid Y) \\ 1, & \text{with probability } H(X \mid Y). \end{cases}$$

It should be noted that if H_{∞} is $\{0,1\}$ -valued, the probability of $H_{\infty}=1$ is necessarily equal to $H(X\mid Y)$ because of the martingale property $\mathbb{E}[H_n\mid H_0]=H_0=H(X\mid Y)$. Note also that Park and Barg [6] have adopted a different, weaker definition of polarization, in which H_{∞} may take more than two values. In this paper, such cases are regarded as not being polarized.

When the marginal distribution of X is uniform, the source polarization is called the channel polarization. As shown in Section IV, the source polarization problem is also translated into the channel polarization problem. We therefore use the terms "source" and "channel" almost interchangeably, unless otherwise stated. As the first and main contribution of this paper, we show a necessary and sufficient condition of G under which any source or channel is polarized. Let $G_{\gamma} := \begin{bmatrix} 1 & 0 \\ 1 & \gamma \end{bmatrix}$ over \mathbb{F}_q where $\gamma \in \mathbb{F}_q^{\times}$. Arikan proved for the case q=2 that the matrix G_1 polarizes any source/channel [3], [10]. Şaşoğlu et al. generalized the result for prime fields [4]. They also showed that for the matrix G_1 over the ring $\mathbb{Z}/q\mathbb{Z}$ where q is not a prime, there is a counterexample of non-polarizing q-ary channel. Their counterexample also works for \mathbb{F}_q whose size q is not a prime. A purpose of this paper is to generalize these results to any matrix over any finite field.

B. Construction of source and channel codes

The polar code for source/channel coding is based on the polarization phenomenon. In this subsection, a rough sketch of construction of the polar code for channel coding is described. Given an $\ell \times \ell$ invertible matrix G which appears in the previous section, we first consider an $\ell^n \times \ell^n$ matrix $G^{\otimes n}$ where ℓ^n denotes the Kronecker power. For $i \in \{0,1,\ldots,\ell^n-1\}$, $i_n i_{n-1} \cdots i_1$ denotes the ℓ -ary expansion of i. Then, the generator matrix of a polar code is, roughly speaking, obtained from $\ell^{\otimes n}$ by choosing rows with indices in the set

$$\{i \in \{0, \dots, \ell^n - 1\} \mid H(X^{(i_1)\cdots(i_n)} \mid Y^{(i_1)\cdots(i_n)}) < \epsilon\}$$

with some threshold $\epsilon > 0$. If a channel (X,Y) is polarized by G, the ratio of chosen rows is asymptotically $1-H(X\mid Y)$ for any fixed $\epsilon \in (0,1)$. For detailed descriptions of encoding and decoding algorithms, see [3] for the channel coding and [10] and [11] for the source coding.

IV. EQUIVALENCE RELATION ON SOURCES AND MAIN THEOREM

In order to deal with a source (X,Y) in terms of polarization phenomenon, it is useful to define an equivalence relation up to which we do not have to distinguish sources.

¹ Joint distribution of these random pairs is not considered in this paper.

²Row and column indices of matrices start with 0 rather than 1.

An equivalence relation $(X, Y) \sim (X', Y')$ which is desirable for our purpose has to satisfy the following two conditions.

$$(X,Y) \sim (X',Y') \Longrightarrow H(X \mid Y) = H(X' \mid Y')$$
 (2)
 $(X,Y) \sim (X',Y') \Longrightarrow (X^{(i)},Y^{(i)}) \sim (X'^{(i)},Y'^{(i)})$
for $i = 0,1,\dots,\ell-1$ (3)

The second condition (3) should be satisfied for any $\ell \times \ell$ invertible matrix G. The significance of these two conditions is that sources which are equivalent in the above sense yield the same random sequence $\{H_n\}_{n=0,1,\ldots}$, thereby behaving exactly the same as for the polarization phenomenon.

Given a source (X,Y), the a posteriori distribution $[P_{X|Y}(x\mid y)]_{x\in\mathbb{F}_q}\in\Delta_q$ plays a fundamental role, in particular in determining the conditional entropy $H(X\mid Y)$ and other relevant quantities. We first introduce two equivalence relations on probability vectors.

Definition 4. For $p_0^{q-1} \in \Delta_q$ and $p'_0^{q-1} \in \Delta_q$, we say $p_0^{q-1} \stackrel{\text{p}}{\sim} p'_0^{q-1}$ if and only if there exists a permutation matrix σ such that $p_0^{q-1} = p'_0^{q-1} \sigma$. For any $s \in \mathbb{N}$, $[p_x]_{x \in \mathbb{F}_q^s} \in \Delta_{q^s}$ and $[p'_x]_{x \in \mathbb{F}_q^s} \in \Delta_{q^s}$, we say $[p_x]_{x \in \mathbb{F}_q^s} \stackrel{\text{q}(s)}{\sim} [p'_x]_{x \in \mathbb{F}_q^s}$ if and only if there exists $z \in \mathbb{F}_q^s$ such that $p_x = p'_{x+z}$ for all $x \in \mathbb{F}_q^s$.

It is straightforward to see that

$$[p_{\boldsymbol{x}}]_{\boldsymbol{x}\in\mathbb{F}_q^s} \overset{\mathrm{q}(s)}{\sim} [p_{\boldsymbol{x}}']_{\boldsymbol{x}\in\mathbb{F}_q^s} \iff [p_{\boldsymbol{x}H}]_{\boldsymbol{x}\in\mathbb{F}_q^s} \overset{\mathrm{q}(s)}{\sim} [p_{\boldsymbol{x}H}']_{\boldsymbol{x}\in\mathbb{F}_q^s} \tag{4}$$

holds for any $s \times s$ invertible matrix H since $p_x = p'_{x+z} \iff p_{xH} = p'_{xH+zH}$ for any $z \in \mathbb{F}_q^s$.

The q-dimensional random vector $[P_{X|Y}(x \mid Y)]_{x \in \mathbb{F}_q} \in \Delta_q$ induces a probability measure on Δ_q . If two random vectors $[P_{X|Y}(x \mid Y)]_{x \in \mathbb{F}_q}$ and $[P_{X'|Y'}(x \mid Y')]_{x \in \mathbb{F}_q}$ defined from sources (X,Y) on $\mathbb{F}_q \times \mathcal{Y}$ and (X',Y') on $\mathbb{F}_q \times \mathcal{Y}'$, respectively, induce the same probability measure on Δ_q , we say $(X,Y) \stackrel{!}{\sim} (X',Y')$. In this case, $A(X \mid Y) = A(X' \mid Y')$ holds for any quantity of the form $A(X \mid Y) = \mathbb{E}[f([P_{X|Y}(x \mid Y)]_{x \in \mathbb{F}_q})]$, and hence the condition (2) is satisfied. Furthermore, the equivalence relation $\stackrel{!}{\sim}$ obviously satisfies (3). However, a weaker equivalence relation than $\stackrel{!}{\sim}$ exists which satisfies both of the conditions (2) and (3). First, a weak equivalence relation which only satisfies the condition (2) is defined as follows.

Definition 5. For sources (X,Y) on $\mathbb{F}_q \times \mathcal{Y}$ and (X',Y') on $\mathbb{F}_q \times \mathcal{Y}'$, we say $(X,Y) \stackrel{\circ}{\sim} (X',Y')$ if and only if the q-dimensional random vector $[P_{X|Y}(x \mid Y)]_{x \in \mathbb{F}_q}$ induces the same distribution on $\Delta_q / \stackrel{\circ}{\sim}$ as the random vector $[P_{X'|Y'}(x \mid Y')]_{x \in \mathbb{F}_q}$. For a function $f: \Delta_q \to \mathbb{R}$ which is invariant under any permutation of its arguments, a quantity $\mathbb{E}[f([P_{X|Y}(x \mid Y)]_{x \in \mathbb{F}_q})]$ is said to be invariant under any permutation of symbols in the a posteriori distribution.

The equivalence $(X,Y) \stackrel{\circ}{\sim} (X',Y')$ implies $A(X \mid Y) = A(X' \mid Y')$ for any quantity $A(X \mid Y)$ invariant under any permutation of symbols in the a posteriori distribution, including the conditional entropy $H(X \mid Y)$. Hence, the equivalence relation $\stackrel{\circ}{\sim}$ satisfies the first condition (2). However, the equivalence relation $\stackrel{\circ}{\sim}$ does not satisfy the second condition (3).

The equivalence relation $\stackrel{a(s)}{\sim}$ defined in the following is weaker than $\stackrel{i}{\sim}$ and satisfies both of the conditions (2) and (3). It plays an essential role in the following argument.

Definition 6. Let $s \in \mathbb{N}$. For pairs of random variables (X,Y) on $\mathbb{F}_q^s \times \mathcal{Y}$ and (X',Y') on $\mathbb{F}_q^s \times \mathcal{Y}'$, we say $(X,Y) \overset{\mathrm{a}(s)}{\sim} (X',Y')$ if and only if there exists $r \in \mathbb{F}_q^\times$ such that the q^s -dimensional random vector $[P_{X|Y}(r\boldsymbol{x} \mid Y)]_{\boldsymbol{x} \in \mathbb{F}_q^s}$ induces the same distribution on $\Delta_{q^s}/\overset{\mathrm{q}(s)}{\sim}$ as $[P_{X'|Y'}(\boldsymbol{x} \mid Y')]_{\boldsymbol{x} \in \mathbb{F}_q^s}$.

It is not hard to confirm the properties (X,Y) $\stackrel{\circ}{\sim}$ (X',Y') \Longrightarrow (X,Y) $\stackrel{\circ}{\sim}$ (X',Y') and (X,Y) $\stackrel{\circ}{\sim}$ (X',Y') \Longrightarrow (X,Y) $\stackrel{\circ}{\sim}$ (X',Y'). From the latter property, it holds that (X,Y) $\stackrel{\circ}{\sim}$ (X',Y') \Longrightarrow $H(X\mid Y)=H(X'\mid Y')$, implying that the equivalence relation $\stackrel{\circ}{\sim}$ satisfies the first condition (2). The equivalence relation $\stackrel{\circ}{\sim}$ also satisfies the second condition (3).

Lemma 7.

$$(X,Y) \stackrel{\mathrm{a}(1)}{\sim} (X',Y') \Longrightarrow (X^{(i)},Y^{(i)}) \stackrel{\mathrm{a}(1)}{\sim} (X'^{(i)},Y'^{(i)})$$

for $i=0,1,\ldots,\ell-1$ and for an arbitrary $\ell \times \ell$ invertible matrix G.

Proof: For a source (X,Y), let $X_0^{\ell-1}, Y_0^{\ell-1}$ and $U_0^{\ell-1}$ be what appear in the definition of the basic transform of it. The random variables $X_0'^{\ell-1}, Y_0'^{\ell-1}$ and $U_0'^{\ell-1}$ are defined in the same way for (X',Y'). The equivalence relation $(X,Y) \stackrel{\text{a}(1)}{\sim} (X',Y')$ between sources (X,Y) and (X',Y') immediately leads to the equivalence $(X_0^{\ell-1},Y_0^{\ell-1}) \stackrel{\text{a}(\ell)}{\sim} (X'_0^{\ell-1},Y'_0^{\ell-1})$ between their ℓ th-order extensions. From (4) and the identity $(r\boldsymbol{x})G^{-1} = r(\boldsymbol{x}G^{-1})$ for any $r \in \mathbb{F}_q^{\times}$ and $\boldsymbol{x} \in \mathbb{F}_q^{\ell}$, it holds that $(X_0^{\ell-1}G^{-1},Y_0^{\ell-1}) \stackrel{\text{a}(\ell)}{\sim} (X'_0^{\ell-1}G^{-1},Y'_0^{\ell-1})$, or equivalently, $(U_0^{\ell-1},Y_0^{\ell-1}) \stackrel{\text{a}(\ell)}{\sim} (U'_0^{\ell-1},Y'_0^{\ell-1})$. One therefore obtains $(U_i,(U_0^{i-1},Y_0^{\ell-1})) \stackrel{\text{a}(1)}{\sim} (U'_i,(U'_0^{i-1},Y'_0^{\ell-1}))$.

The equivalence relation $\stackrel{a(1)}{\sim}$ gives rise to the following several useful lemmas.

Lemma 8 (Source-channel equivalence [12]). Let (N, Z) be a random pair on $\mathbb{F}_q \times \mathcal{Y}$ and X be a uniform random variable on \mathbb{F}_q which is independent of (N, Z). Then, it holds that $(N, Z) \stackrel{\mathrm{a}(1)}{\sim} (X, (X + N, Z))$.

Proof: One has $(X,(X+N,Z)) \stackrel{\text{a}(1)}{\sim} (-X+(X+N),(X+N,Z)) = (N,(X+N,Z)) \stackrel{\text{i}}{\sim} (N,Z)$, where the last equivalence relation is due to the assumptions on X.

The channel (X,(X+N,Z)) in Lemma 8 is a symmetric channel in the following sense.

Definition 9 (Symmetric channel). A channel (X,Y) on $\mathbb{F}_q \times \mathcal{Y}$ is said to be symmetric if and only if there exists a permutation σ_x on \mathcal{Y} for each $x \in \mathbb{F}_q$ such that $P_{Y|X}(y \mid x) = P_{Y|X}(\sigma_{x'-x}(y) \mid x')$ for any $y \in \mathcal{Y}$ and $x, x' \in \mathbb{F}_q$.

The symmetricity is preserved under the basic transform.

Lemma 10. For a symmetric channel (X,Y), $(X^{(i)},Y^{(i)})$ is symmetric for any $i \in \{0,\ldots,\ell-1\}$.

$$\begin{array}{lll} \textit{Proof:} & \text{The statement holds} & \text{since} \\ P_{U_0^{\ell-1},Y_0^{\ell-1}}((u_0^{i-1},u_i,u_{i+1}^{\ell-1}),y_0^{\ell-1}) & & = \\ P_{U_0^{\ell-1},Y_0^{\ell-1}}((u_0^{i-1},u_i',u_{i+1}^{\ell-1}),w_0^{\ell-1}) & \text{where} & w_j & = \\ \sigma_{G_{i,j}(u_i'-u_i)}(y_j). & & & & = \\ \end{array}$$

The following technical lemma implies that one can ignore effects of addition of a known constant to input of symmetric channels. It will be used in the proof of the main theorem.

Lemma 11. For any channel (X,Y) and any symmetric channel (X',Y'), let (Z,(Y,Y')) and (Z',(Y,Y')) be the channels defined by letting Z=X=X' and Z'=X=X'+a for any fixed $a \in \mathbb{F}_q$, respectively. For these channels, it holds that $(Z,(Y,Y')) \stackrel{\sim}{\sim} (Z',(Y,Y'))$.

$$\begin{array}{lll} \textit{Proof:} & \text{The} & \text{equality} & P_{Z,(Y,Y')}(z,(y,y')) & = \\ P_{Z',(Y,Y')}(z,(y,\sigma_a^{-1}(y'))) & \text{implies} & (Z,(Y,Y')) & & \\ (Z',(Y,Y')). & & & \\ \end{array}$$

We next introduce an equivalence relation on matrices on the basis of the equivalence relation $\stackrel{\mathrm{a}(1)}{\sim}$ on sources/channels. We say that $\ell \times \ell$ invertible matrices \hat{G} and \bar{G} are equivalent when $(\hat{X}^{(i)}, \hat{Y}^{(i)}) \stackrel{\mathrm{a}(1)}{\sim} (\bar{X}^{(i)}, \bar{Y}^{(i)})$ for $i = 0, \ldots, \ell-1$ where $\{(\hat{X}^{(i)}, \hat{Y}^{(i)})\}_{i=0,\ldots,\ell-1}$ and $\{(\bar{X}^{(i)}, \bar{Y}^{(i)})\}_{i=0,\ldots,\ell-1}$ are two sets of ℓ random pairs generated from an arbitrary common source (X,Y) via the basic transform using matrices \hat{G} and \bar{G} , respectively.

Lemma 12. Let G and V be an $\ell \times \ell$ invertible matrix and an $\ell \times \ell$ invertible upper triangular matrix, respectively. Then, G and VG are equivalent.

Obviously, a permutation of columns of G does not change $(X^{(i)},Y^{(i)})$ up to the equivalence $\stackrel{\mathrm{i}}{\sim}$ for $i=0,\ldots,\ell-1$, so that G and its column permutation are equivalent. Hence, without loss of generality, one can assume that G is a lower triangular matrix.

Definition 13 (Standard form). Lower triangular matrices with unit diagonal elements equivalent to G are called standard forms of G.

A standard form of G is not generally unique. For example, the standard forms of G_{γ} are $\begin{bmatrix} 1 & 0 \\ \gamma^{-1} & 1 \end{bmatrix}$ and $\begin{bmatrix} 1 & 0 \\ \gamma & 1 \end{bmatrix}$. If there exists the identity matrix as a standard form of G, it is the unique standard form of G. In this case, one obviously has the identity $(X^{(i)},Y^{(i)})\stackrel{\mathrm{a}(1)}{\sim}(X,Y)$ for all $i\in\{0,\dots,\ell-1\}$, implying that G does not polarize any source. For other cases, the following main theorem shows necessary and sufficient conditions of G under which any source is polarized.

Theorem 14. The followings are equivalent for an $\ell \times \ell$ invertible matrix G over \mathbb{F}_q with a non-identity standard form.

- Any q-ary source is polarized by G.
- It holds $\mathbb{F}_p(\bar{G}) = \mathbb{F}_q$ for any standard form \bar{G} of G.
- It holds $\mathbb{F}_p(\bar{G}) = \mathbb{F}_q$ for one of the standard forms \bar{G} of G.

Corollary 15. Any q-ary source is polarized by the 2×2 matrix G_{γ} over \mathbb{F}_q with $\gamma \in \mathbb{F}_q^{\times}$ if and only if $\mathbb{F}_p(\gamma) = \mathbb{F}_q$.

Note that the identity matrix is the standard form of an invertible matrix G if and only if there exists an upper triangular matrix as a column permutation of G. Thus, Theorem 14 includes the known results that an invertible matrix G is polarizing if and only if any column permutation of G is not upper triangular for G = 2 [9, Lemma 1] and for G prime [4].

V. BHATTACHARYYA PARAMETER

Bhattacharyya parameter is useful both for proving the polarization phenomenon, and for evaluating asymptotic speed of polarization. In this section, it is shown that polarization of Bhattacharyya parameter and polarization of the conditional entropy are equivalent. Let $(\Omega := \{1, \dots, q\}, 2^{\Omega}, P)$ be a probability space. The probability measure P can be represented by the vector $[\sqrt{P(1)}, \dots, \sqrt{P(q)}] \in \mathcal{S}_q$ where $\mathcal{S}_q := \{ [p_1, \dots, p_q] \in \mathbb{R}^q_{\geq 0} \mid p_1^2 + \dots + p_q^2 = 1 \}.$ The L_p norm of $x \in \mathbb{C}^q$ is defined as $L_p(x) := (|x_1|^p + \cdots + |x_q|^p)^{1/p}$ for any $p \ge 1$. The L_1 norm of $p \in \mathcal{S}_q$ attains the minimum 1 at the deterministic distributions i.e., the distributions of the form $[0,\ldots,0,1,0,\ldots,0]$, and the maximum \sqrt{q} at the uniform distribution, represented by $\boldsymbol{u} := [1/\sqrt{q}, \dots, 1/\sqrt{q}] \in \mathcal{S}_q$. On the other hand, the deterministic and uniform distributions also minimize and maximize the entropy $H(\mathbf{p}) := -\sum_i p_i^2 \log p_i^2$ of $p \in S_q$, respectively.

The following lemma states that closeness of a probability distribution to determinism or uniformity measured in terms of its entropy value is equivalent to that measured in terms of its L_1 -norm value.

Lemma 16. For any $\epsilon > 0$, there exists $\delta > 0$ such that

$$\{\boldsymbol{p} \in \mathcal{S}_{q} \mid H(\boldsymbol{p}) < \delta\} \subseteq \{\boldsymbol{p} \in \mathcal{S}_{q} \mid L_{1}(\boldsymbol{p}) - 1 < \epsilon\}$$

$$\{\boldsymbol{p} \in \mathcal{S}_{q} \mid L_{1}(\boldsymbol{p}) - 1 < \delta\} \subseteq \{\boldsymbol{p} \in \mathcal{S}_{q} \mid H(\boldsymbol{p}) < \epsilon\}$$

$$\{\boldsymbol{p} \in \mathcal{S}_{q} \mid 1 - H(\boldsymbol{p}) < \delta\} \subseteq \{\boldsymbol{p} \in \mathcal{S}_{q} \mid \sqrt{q} - L_{1}(\boldsymbol{p}) < \epsilon\}$$

$$\{\boldsymbol{p} \in \mathcal{S}_{q} \mid \sqrt{q} - L_{1}(\boldsymbol{p}) < \delta\} \subseteq \{\boldsymbol{p} \in \mathcal{S}_{q} \mid 1 - H(\boldsymbol{p}) < \epsilon\}.$$

$$\{\boldsymbol{p} \in \mathcal{S}_{q} \mid \sqrt{q} - L_{1}(\boldsymbol{p}) < \delta\} \subseteq \{\boldsymbol{p} \in \mathcal{S}_{q} \mid 1 - H(\boldsymbol{p}) < \epsilon\}.$$

$$(8)$$

Proof: Since

$$L_2(\mathbf{u} - \mathbf{p})^2 = \sum_{i=1}^q \left(\frac{1}{\sqrt{q}} - p_i\right)^2 = 2 - \frac{2}{\sqrt{q}} \sum_{i=1}^q p_i$$
$$= \frac{2}{\sqrt{q}} \left(\sqrt{q} - L_1(\mathbf{p})\right)$$

(8) is a consequence of continuity of $H(\mathbf{p})$. The relationship (7) follows from

$$1 - H(\mathbf{p}) = 1 + \sum_{i=1}^{q} p_i^2 \log p_i^2 = -2 \sum_{i=1}^{q} p_i^2 \log \frac{1}{\sqrt{q} p_i}$$
$$\geq \frac{2}{\log_e q} \sum_{i=1}^{q} p_i^2 \left(1 - \frac{1}{\sqrt{q} p_i} \right) = \frac{2}{\sqrt{q} \log_e q} \left(\sqrt{q} - L_1(\mathbf{p}) \right).$$

Since $H(\mathbf{p}) = 2\sum_i p_i^2 \log(1/p_i) \le 2\log\sum_i p_i = 2\log L_1(\mathbf{p})$, the relationship (6) holds. Since $H(\mathbf{p})\log_e q =$

 $-\sum_{i} p_i^2 \log_e p_i^2 \ge -\log_e \max_{i} p_i^2 \ge 1 - \max_{i} p_i^2 \ge (L_1(\boldsymbol{p}) - 1)^2/(q-1)$ (see (18) for the last inequality), the relationship (5) holds.

Hence, the entropy is close to 0 and 1 if and only if the L_1 norm is close to 1 and \sqrt{q} , respectively.

The above argument is applied to random pairs (X,Y) to establish the relationship between the conditional entropy and Bhattacharyya parameter. The expectation of the squared L_1 norm of the a posteriori probability vector $[\sqrt{P_{X|Y}(x\mid Y)}]_{x\in\mathbb{F}_q}\in\mathcal{S}_q$ satisfies

$$1 \leq \sum_{y \in \mathcal{Y}} P_Y(y) \left[\sum_{x \in \mathbb{F}_q} \sqrt{P_{X|Y}(x \mid y)} \right]^2 \leq q$$

$$\iff 0 \leq \frac{1}{q-1} \sum_{\substack{x \in \mathbb{F}_q, x' \in \mathbb{F}_q \\ x \neq x'}} \sum_{y \in \mathcal{Y}} P_Y(y) \sqrt{P_{X|Y}(x \mid y)} P_{X|Y}(x' \mid y)$$

$$\leq 1 \tag{9}$$

for any random pair (X,Y). From Lemma 16 and (9), the conditional entropy $H(X \mid Y)$ is close to 0 and 1 if and only if the *Bhattacharyya parameter* $Z(X \mid Y) \in [0,1]$ for (X,Y), defined as follows, is close to 0 and 1, respectively.

Definition 17 (Bhattacharyya parameter).

$$Z(X \mid Y) := \frac{1}{q-1} \sum_{\substack{x \in \mathbb{F}_q, x' \in \mathbb{F}_q \\ x \neq x'}} \sum_{y \in \mathcal{Y}} P_Y(y) \sqrt{P_{X|Y}(x \mid y) P_{X|Y}(x' \mid y)}.$$

Obviously, $Z(X \mid Y)$ is invariant under any permutation of symbols in the a posteriori distribution of (X,Y). For $d \in \mathbb{F}_q^{\times}$, we define $Z_d(X \mid Y) \in [0,1]$ as

$$Z_d(X\mid Y):=\sum_{x\in\mathbb{F}_q}\sum_{y\in\mathcal{Y}}P_Y(y)\sqrt{P_{X\mid Y}(x\mid y)P_{X\mid Y}(x+d\mid y)}.$$

The Bhattacharyya parameter $Z(X \mid Y)$ can be expressed as the average of $Z_d(X \mid Y)$

$$Z(X \mid Y) = \frac{1}{q-1} \sum_{d \in \mathbb{F}_a^{\times}} Z_d(X \mid Y).$$

Hence, $Z(X \mid Y)$ is close to 0 and 1 if and only if $Z_d(X \mid Y)$ is simultaneously close to 0 and 1 for all $d \in \mathbb{F}_q^{\times}$, respectively.

VI. PROOF OF THE MAIN THEOREM

A. Sketch

In this section, the proof of Theorem 14 is shown. In Section VI-B, it is proved that if there exists a standard form \bar{G} of G such that $\mathbb{F}_p(\bar{G}) \neq \mathbb{F}_q$, there exists a source which is not polarized by G. It means that if any source is polarized by G, any standard form \bar{G} of G satisfies $\mathbb{F}_p(\bar{G}) = \mathbb{F}_q$. In Section VI-C, it is proved that if there exists a standard form \bar{G} of G such that $\mathbb{F}_p(\bar{G}) = \mathbb{F}_q$, any source is polarized by G. This completes the proof of Theorem 14.

B. Necessity

Let \bar{G} be an arbitrary standard form of G. Assume $\mathbb{F}_p(\bar{G}) \neq$ \mathbb{F}_q . Let $M:=[\mathbb{F}_q:\mathbb{F}_p(\bar{G})]$ be a degree of a field extension $\mathbb{F}_q/\mathbb{F}_p(\bar{G})$. Since $\mathbb{F}_q/\mathbb{F}_p(\bar{G})$ is an M-dimensional linear space over $\mathbb{F}_p(G)$, there is an isomorphism $\psi \colon \mathbb{F}_q/\mathbb{F}_p(G) \to$ $\mathbb{F}_p(\bar{G})^M$. Let $[V_0,\ldots,V_{M-1}]\in\mathbb{F}_p(\bar{G})^M$ be the random vector $\psi(X)$ for $X \in \mathbb{F}_q$. If one takes a source (X,Y)for which V_0, \ldots, V_{M-1} are independent conditioned on Y, recursive application of the basic transform to the source (X,Y) affects V_i separately for $i \in \{0,\ldots,M-1\}$, i.e., one can regard the polarization process of the source (X, Y) as a collection of M distinct polarization processes $\{(V_{i,n}, Y_n) :=$ $(V_i^{(B_1)\cdots(B_n)}, Y^{(B_1)\cdots(B_n)})\}_{n=0,1,\dots}, i = 0,\dots,M-1.$ In this case, if $H(V_i \mid Y)$ is not constant among all $i \in$ $\{0, \ldots, M-1\}$, the source $([V_0, \ldots, V_{M-1}], Y)$ cannot be polarized in principle, in the sense defined in Definition 3. Note that the situation is essentially equivalent to the polar coding for the M-user multiple access channel [5].

C. Sufficiency

In the proof of sufficiency, (X,Y) is assumed to be a symmetric channel. From Lemma 8 we do not lose generality by this assumption. For any $j \in \{0, \dots, \ell-1\}$, it holds via the chain rule for the entropy that

$$\sum_{i=j}^{\ell-1} H(X^{(i)} \mid Y^{(i)}) = H(U_j^{\ell-1} \mid U_0^{j-1}, Y_0^{\ell-1})$$

$$= \sum_{i=j}^{\ell-1} H(U_i \mid U_0^{j-1}, U_{i+1}^{\ell-1}, Y_0^{\ell-1}) \quad (10)$$

for any (X,Y). Let \bar{G} be an arbitrary standard form of G, and assume that $U_0^{\ell-1}$ and $(X^{(i)},Y^{(i)})$ for $i\in\{0,\dots,\ell-1\}$ are defined with \bar{G} . All the terms in the rightmost side of (10) are at most $H(X\mid Y)$ for any standard form \bar{G} . It also holds that $|H(X_n^{(i)}\mid Y_n^{(i)})-H(X_n\mid Y_n)|\to 0$ with probability 1 as $n\to\infty$ for all $i\in\{0,\dots,\ell-1\}$ since $\{H(X_n\mid Y_n)\}_{n=0,1,\dots}$ converges almost surely. Combining these two facts, one observes that each of the terms in the sum on the rightmost side of (10) evaluated with $(X,Y)=(X_n,Y_n)$ must be close to $H(X_n\mid Y_n)$ with probability 1 as $n\to\infty$. In particular,

$$H(X_n \mid Y_n) - H(U_j \mid U_0^{j-1}, U_{j+1}^{\ell-1}, Y_0^{\ell-1}) \Big|_{(X,Y) = (X_n, Y_n)} \to 0$$

holds with probability 1. Hence, it also holds

$$H(X_n \mid Y_n) - H(U_j \mid U_0^{j-1}, U_{j+1}^{\ell-1}, Y_k, Y_j) \Big|_{(X,Y) = (X_n, Y_n)} \to 0$$

for any $0 \leq k < j \leq \ell-1$. From Lemmas 10 and 11, the effects of U_0^{j-1} and $U_{j+1}^{\ell-1}$ can be ignored, i.e., it holds $(U_j, (U_0^{j-1}, U_{j+1}^{\ell-1}, Y_k, Y_j)) \stackrel{\sim}{\sim} (U_j, (Y_k, Y_j))$ where the channel on the right-hand side is defined from $(U_j, (U_0^{j-1}, U_{j+1}^{\ell-1}, Y_k, Y_j))$ by fixing U_0^{j-1} and $U_{j+1}^{\ell-1}$ to the all-zero vectors. Assume that the (j,k)-element of G is $\gamma \neq 0$. Let $\{(\bar{\mathsf{X}}_n^{(0)}, \bar{\mathsf{Y}}_n^{(0)}), (\bar{\mathsf{X}}_n^{(1)}, \bar{\mathsf{Y}}_n^{(1)})\}$ be the random pairs obtained from $(\mathsf{X}_n, \mathsf{Y}_n)$ via the basic transform with the 2×2 matrix

$$\begin{split} & (\mathsf{X}_0, \mathsf{Y}_0) \overset{G_{\gamma}}{\to} \{ (\bar{\mathsf{X}}_0^{(0)}, \bar{\mathsf{Y}}_0^{(0)}), (\bar{\mathsf{X}}_0^{(1)}, \bar{\mathsf{Y}}_0^{(1)}) \} \\ & \qquad G \downarrow \\ & (\mathsf{X}_1, \mathsf{Y}_1) \overset{G_{\gamma}}{\to} \{ (\bar{\mathsf{X}}_1^{(0)}, \bar{\mathsf{Y}}_1^{(0)}), (\bar{\mathsf{X}}_1^{(1)}, \bar{\mathsf{Y}}_1^{(1)}) \} \\ & \qquad \vdots \\ & \qquad \vdots \\ & (\mathsf{X}_n, \mathsf{Y}_n) \overset{G_{\gamma}}{\to} \{ (\bar{\mathsf{X}}_n^{(0)}, \bar{\mathsf{Y}}_n^{(0)}), (\bar{\mathsf{X}}_n^{(1)}, \bar{\mathsf{Y}}_n^{(1)}) \} \\ & \qquad \vdots \\ \end{split}$$

Fig. 1. The relationships of (X_n, Y_n) and $(X_n^{(1)}, Y_n^{(1)})$. In the vertical arrows, the basic transform defined in Section III based on the matrix G is applied. In the horizontal arrows, the basic transform based on the matrix G_{γ} is applied.

 $\begin{bmatrix} 1 & 0 \\ \gamma & 1 \end{bmatrix}$, which is a standard form of G_{γ} . Then, from (11), it holds that $H(X_n \mid Y_n) - H(\bar{X}_n^{(1)} \mid \bar{Y}_n^{(1)}) \to 0$ with probability 1. The relationships of random variables are described in Fig. 1. In the rest of the proof, we do not use the relationship between (X_n, Y_n) and (X_{n-1}, Y_{n-1}) , and only use the fact that $H(X_n \mid Y_n) - H(\bar{X}_n^{(1)} \mid \bar{Y}_n^{(1)}) \to 0$ with probability 1 for G_{γ} where γ is an arbitrary off-diagonal non-zero element of \bar{G} . The following proposition implies the sufficiency of the main theorem.

Proposition 18. Let A be a non-empty subset of \mathbb{F}_q^{\times} . Let $\{(X_{(n)},Y_{(n)})\}_{n=0,1,\dots}$ be a sequence of random pairs. Assume $H(X_{(n)} \mid Y_{(n)}) - H(X_{(n)}^{(1)} \mid Y_{(n)}^{(1)}) \rightarrow 0$ for all G_{γ} where $\gamma \in A$. Then, for any $\epsilon > 0$, there exists $n_0 \in \mathbb{N}$ such that

$$Z_{td}(X_{(n)} \mid Y_{(n)}) < \epsilon, \quad \text{for all} \quad t \in \mathbb{F}_p(A)^{\times}$$
or
$$Z_{td}(X_{(n)} \mid Y_{(n)}) > 1 - \epsilon, \quad \text{for all} \quad t \in \mathbb{F}_p(A)^{\times}$$

for any $n \geq n_0$ and any $d \in \mathbb{F}_q^{\times}$.

When $\mathbb{F}_p(\bar{G}) = \mathbb{F}_q$, Proposition 18 states that the random sequence $H_n = H(X_n \mid Y_n)$ is close to 0 or 1 for sufficiently large n with probability 1. Hence, H_{∞} must be $\{0,1\}$ -valued, i.e., any source (X,Y) is polarized by G.

What remains is to prove Proposition 18. It is equivalent to the following proposition, which will be proved in the rest of this section.

Proposition 19. Let A be a non-empty subset of \mathbb{F}_q^{\times} . Let $\{(X_{(n)},Y_{(n)})\}_{n=0,1,...}$ be a sequence of random pairs. Assume $H(X_{(n)} \mid Y_{(n)}) - H(X_{(n)}^{(1)} \mid Y_{(n)}^{(1)}) \rightarrow 0$ for all G_{γ} where $\gamma \in A$. Then, for any $\epsilon > 0$, there exists $n_0 \in \mathbb{N}$ such that

(p0)
$$Z_d(X_{(n)} | Y_{(n)}) < \epsilon \text{ or } Z_d(X_{(n)} | Y_{(n)}) > 1 - \epsilon$$
,

$$\begin{array}{ll} \text{(p0)} & Z_d(X_{(n)} \mid Y_{(n)}) < \epsilon \ \textit{or} \ Z_d(X_{(n)} \mid Y_{(n)}) > 1 - \epsilon, \\ \text{(p1)} & \left(Z_d(X_{(n)} \mid Y_{(n)}) > 1 - \epsilon \Rightarrow Z_{\gamma d}(X_{(n)} \mid Y_{(n)}) > 1 - \epsilon \right) \\ & \textit{for any } \gamma \in A, \end{array}$$

(p2)
$$((Z_d(X_{(n)} \mid Y_{(n)}) > 1 - \epsilon \text{ and } Z_{d'}(X_{(n)} \mid Y_{(n)}) > 1 - \epsilon) \Rightarrow Z_{d+d'}(X_{(n)} \mid Y_{(n)}) > 1 - \epsilon) \text{ for any } d' \in \mathbb{F}_q^{\times},$$
 for any $n \geq n_0$ and any $d \in \mathbb{F}_q^{\times}$.

Note that from (p0),
$$\gamma^{q-1} = 1$$
 and $d' + d' + \cdots + d' = 0$, the conditions (p1) and (p2) imply

(p'1)
$$\left(Z_d(X_{(n)}\mid Y_{(n)})<\epsilon\Rightarrow Z_{\gamma d}(X_{(n)}\mid Y_{(n)})<\epsilon\right)$$
 for any $\gamma\in A$,

respectively, for any $n \geq n_0$ and any $d \in \mathbb{F}_q^{\times}$. It is easy to confirm that Proposition 18 implies Proposition 19. The other direction also holds since $\mathbb{F}_p(A) = \{\gamma_1^{i_1} + \gamma_2^{i_2} + \dots + \gamma_m^{i_m} \mid m \in \mathbb{N}, i_j = 0, 1, \dots, q - 2, \gamma_j \in A, \forall j = 1, 2, \dots, m\}.$

Remark 1. Note that among the three conditions (p0), (p1) and (p2), only (p1) uses the set A. Indeed, (p0) and (p2) hold for any matrix as shown in [4]. When q is a prime, the conditions (p0) and (p2) are sufficient to prove Proposition 18

since $\mathbb{F}_p^{\times} = \{1, 1+1, 1+1+1, \dots, 1+\dots+1\}$ [4]. When A includes a primitive element γ of $\mathbb{F}_p(A)$, i.e., $\mathbb{F}_p(A)^{\times} =$ $\{1, \gamma, \gamma^2, \dots, \gamma^{q-2}\}\$, the conditions (p0) and (p1) are also sufficient to prove Proposition 18 [1]. However, generally, we need all of (p0), (p1) and (p2) for proving Proposition 18.

The following lemma implies (p0) and (p1) to hold under the assumptions of Proposition 19.

Lemma 20. Let $\{(X_{(n)}, Y_{(n)})\}_{n=0,1,...}$ be a sequence of random pairs. Assume $H(X_{(n)} | Y_{(n)}) - H(X_{(n)}^{(1)} | Y_{(n)}^{(1)}) \to 0$ for G_{γ} where $\gamma \in \mathbb{F}_{q}^{\times}$. Then, for any $\epsilon > 0$ there exists $n_0 \in \mathbb{N}$ such that

$$\begin{split} Z_{\gamma^i d}(X_{(n)} \mid Y_{(n)}) < \epsilon, \quad \text{for all} \quad i = 0, \dots, q-2 \\ \text{or} \quad Z_{\gamma^i d}(X_{(n)} \mid Y_{(n)}) > 1 - \epsilon, \quad \text{for all} \quad i = 0, \dots, q-2 \\ \text{for any } n \geq n_0 \text{ and any } d \in \mathbb{F}_q^\times. \end{split}$$

The proof of Lemma 20 is in Appendix A. The following lemma and (p0) imply (p2) to hold under the assumptions of Proposition 19, completing the proof of sufficiency of the main theorem.

Lemma 21 ([4]). For any d_1 and d_2 in \mathbb{F}_q^{\times} satisfying $d_2 \neq$

$$\sqrt{1 - Z_{d_1 + d_2}(X \mid Y)}
\leq \sqrt{1 - Z_{d_1}(X \mid Y)} + \sqrt{1 - Z_{d_2}(X \mid Y)}.$$

Proof: Since

$$1 - Z_d(X \mid Y)$$

$$= \frac{1}{2} \sum_{x \in \mathbb{F}_q} \sum_{y \in \mathcal{Y}} \left(\sqrt{P_{X,Y}(x,y)} - \sqrt{P_{X,Y}(x+d,y)} \right)^2$$

the statement is obtained from the triangle inequality of the Euclidean distance.

VII. ERROR PROBABILITY, TOTAL VARIATION DISTANCE TO THE UNIFORM DISTRIBUTION AND SPEED OF POLARIZATION

A. Preliminaries

In this section, we consider speed of polarization by an $\ell \times \ell$ invertible matrix G over \mathbb{F}_q . Let

$$P_{\mathbf{e}}(X \mid Y) := 1 - \sum_{y \in \mathcal{Y}} P_Y(y) \max_{x \in \mathbb{F}_q} P_{X|Y}(x \mid y).$$

This is the average error probability of the maximum a posteriori estimator $\hat{x}(y) := \arg\max_{x \in \mathbb{F}_q} P_{X|Y}(x \mid y)$ of X given Y. The random quantity $P_{\mathbf{e}}(\mathsf{X}_n \mid \mathsf{Y}_n)$ plays a key role in studying speed of polarization. It provides a bound of the block error probability of polar codes with successive cancellation decoding applied to channel coding [3]. More precisely, if one has

$$\Pr(P_{\mathbf{e}}(\mathsf{X}_n \mid \mathsf{Y}_n) < \epsilon) \ge R$$

then it implies existence of a polar code for channel coding with blocklength ℓ^n , rate R, and the block error probability at most $\ell^n R \epsilon$. Obviously, $P_{\rm e}(X \mid Y)$ is invariant under any permutation of symbols in the a posteriori distribution of (X,Y). The average error probability $P_{\rm e}(X \mid Y)$ takes a value in [0,(q-1)/q]. As it has been the case in the study of the binary case [9], the Bhattacharyya parameter is useful for bounding the error probability.

Lemma 22.

$$\begin{split} & \frac{q-1}{q^2} \left(\sqrt{1 + (q-1)Z(X \mid Y)} - \sqrt{1 - Z(X \mid Y)} \right)^2 \\ & \leq P_{\mathrm{e}}(X \mid Y) \\ & \leq \min_{k=1,2,\dots,q-1} \left\{ \frac{(q-1)Z(X \mid Y) + k(k-1)}{k(k+1)} \right\}. \end{split}$$

Proof of Lemma 22 is in Appendix B.

Another quantity which we study in this section is the expected total variation distance $T(X \mid Y)$ between the a posteriori probability and the uniform distribution, defined as

$$T(X \mid Y) := \sum_{y \in \mathcal{Y}} P_Y(y) \sum_{x \in \mathbb{F}_q} \left| P_{X|Y}(x \mid y) - \frac{1}{q} \right|.$$

Properties of the random quantity $T(X_n | Y_n)$ is important in polar codes for lossy source coding [13], [14]. More precisely, if one has

$$\Pr(T(X_n \mid Y_n) < \epsilon) > R$$

for the test channel $(X_0,Y_0)=(X,Y)$, then there exists a polar code for source coding with blocklength $\ell^n(1-R)$, rate 1-R and the average distortion at most $\mathcal{D}+d_{\max}\ell^nR\epsilon$ where \mathcal{D} denotes the average distortion for the test channel and where d_{\max} is the maximum value of the distortion function [13], [14]. Note that $T(X\mid Y)$ is invariant under any permutation of symbols in the a posteriori distribution. The total variation distance $T(X\mid Y)$ takes a value in [0,2(q-1)/q]. The following lemma establishes a relationship between the total variation distance $T(X\mid Y)$ and the average error probability $P_{\mathrm{e}}(X\mid Y)$.

Lemma 23.

$$\begin{split} 2\left(\frac{q-1}{q}-P_{\mathrm{e}}(X\mid Y)\right) &\leq T(X\mid Y) \leq \frac{2(q-1)}{q} \\ &-\frac{2}{q}\max_{k=1,\ldots,q-1}\bigg\{k(k+1)P_{\mathrm{e}}(X\mid Y)-k(k-1)\bigg\}. \end{split}$$

The proof is in Appendix C.

The Fourier transform of the a posteriori probability is defined for analyzing $T(X \mid Y)$.

Definition 24 (Character). Let $\omega_p \in \mathbb{C}$ be a primitive complex p-th root of unity. Define $\chi(x) := \omega_p^{\operatorname{Tr}(x)}$ for any $x \in \mathbb{F}_q$ where $\operatorname{Tr}: \mathbb{F}_q \to \mathbb{F}_p$ is defined as $x \mapsto \sum_{j=0}^{m-1} x^{p^j}$. Here, $\operatorname{Tr}(x) \in \mathbb{F}_p$ appearing in the exponent should be regarded as an integer via the natural correspondence between \mathbb{F}_p and $\mathbb{Z}/p\mathbb{Z}$.

From the definition of $\chi(x)$, it satisfies the following properties.

$$\begin{split} \chi(0) &= 1 \\ |\chi(x)| &= 1, \qquad \text{for any } x \in \mathbb{F}_q \\ \chi(x+z) &= \chi(x)\chi(z), \qquad \text{for any } x,z \in \mathbb{F}_q \\ \sum_{x \in \mathbb{F}_q} \chi(x) &= 0. \end{split}$$

In this paper, we only use $\chi(x)$ through these properties.

Definition 25 (Fourier transform). For any fixed $y \in \mathcal{Y}$, the Fourier transform of the a posteriori probability $P_{X|Y}$ of a source (X,Y) is defined as

$$P_{X\mid Y}^*(w\mid y) := \sum_{z\in\mathbb{F}_q} P_{X\mid Y}(z\mid y)\chi(wz)$$

for $w \in \mathbb{F}_q$.

Note that $P^*_{X|Y}(0 \mid y) = 1$ for any $y \in \mathcal{Y}$. Like the role of $Z(X \mid Y)$ in studying $P_{\mathrm{e}}(X \mid Y)$, the auxiliary quantity $S(X \mid Y)$, defined as

$$S(X \mid Y) := \frac{1}{q-1} \sum_{w \in \mathbb{F}_{a}^{\times}} \sum_{y \in \mathcal{Y}} P_{Y}(y) \left| P_{X|Y}^{*}(w \mid y) \right|$$

can be used for analyzing $T(X \mid Y)$. The quantity $S(X \mid Y)$ takes a value in [0,1]. Note that, although $S(X \mid Y)$ is identical to $T(X \mid Y)$ (and $1-2P_{\rm e}(X \mid Y)$) when q=2, $S(X \mid Y)$ is in general different from $T(X \mid Y)$. In this regard, consideration of the quantity $S(X \mid Y)$ is a novel idea that comes into play when one considers non-binary cases. Although $S(X \mid Y)$ is *not* invariant under arbitrary permutations of symbols in the a posteriori distribution, $S(X \mid Y)$ is invariant under a permutation of symbols in the a posteriori distribution when the permutation is addition or multiplication on the finite field i.e., $S(X \mid Y) = S(r(Y)X + d(Y) \mid Y)$ for any $d: \mathcal{Y} \to \mathbb{F}_q$ and $r: \mathcal{Y} \to \mathbb{F}_q^{\times}$. Hence, if $(X,Y) \stackrel{\mathrm{a}(1)}{\sim} (X',Y')$, it holds that $S(X \mid Y) = S(X' \mid Y')$.

The following lemma relates the quantity $S(X \mid Y)$ with the average error probability $P_{e}(X \mid Y)$.

Lemma 26.

$$\begin{split} 1 - \frac{q}{q-1} P_{e}(X \mid Y) &\leq S(X \mid Y) \\ &\leq \min_{k=1,...,q-1} \left\{ k(k+1) \right. \\ &\cdot \left[\left(\frac{k}{k+1} - P_{e}(X \mid Y) \right) \sqrt{1 - \frac{q}{q-1} \frac{k-1}{k}} \right. \\ &+ \left(P_{e}(X \mid Y) - \frac{k-1}{k} \right) \sqrt{1 - \frac{q}{q-1} \frac{k}{k+1}} \right] \right\}. \end{split}$$

The proof is in Appendix D.

We now define the following equivalence relation for establishing relationship among several quantities for a source (X,Y) defined so far.

Definition 27. For $A(X \mid Y) \in [0,1]$ and $B(X \mid Y) \in [0,1]$, we say $A(X \mid Y) \stackrel{e}{\sim} B(X \mid Y)$ if and only if there exists $\epsilon > 0$ and $c \in (0,1]$ such that if $B(X \mid Y) < \epsilon$,

$$B(X \mid Y)^{\frac{1}{c}} \le A(X \mid Y) \le B(X \mid Y)^{c}$$

and if $1 - B(X \mid Y) < \epsilon$,

$$(1 - B(X \mid Y))^{\frac{1}{c}} \le 1 - A(X \mid Y) \le (1 - B(X \mid Y))^{c}$$

for any source (X, Y).

From Lemmas 22, 23 and 26, the following corollary is obtained.

Corollary 28.
$$(q/(q-1))P_{e}(X \mid Y) \stackrel{e}{\sim} Z(X \mid Y) \stackrel{e}{\sim} 1 - (q/(2(q-1)))T(X \mid Y) \stackrel{e}{\sim} 1 - S(X \mid Y).$$

The following four quantities are used in the derivation of the speed of polarization in the next subsection.

Definition 29. For any channel (X,Y), $Z_{\max}(X,Y)$ and $Z_{\min}(X,Y)$ are defined as

$$Z_{\max}(X,Y) := \max_{\substack{x \in \mathbb{F}_q, x' \in \mathbb{F}_q \\ x \neq x'}} \sum_{y \in \mathcal{Y}} \sqrt{P_{Y|X}(y \mid x) P_{Y|X}(y \mid x')}$$

$$Z_{\min}(X,Y) := \min_{x \in \mathbb{F}_q, x' \in \mathbb{F}_q} \sum_{y \in \mathcal{Y}} \sqrt{P_{Y\mid X}(y\mid x) P_{Y\mid X}(y\mid x')}.$$

For any source (X,Y), $S_{\max}(X,Y)$ and $S_{\min}(X,Y)$ are defined as

$$S_{\max}(X,Y) := \max_{w \in \mathbb{F}_q^{\times}} \sum_{y \in \mathcal{Y}} P_Y(y) \left| P_{X\mid Y}^*(w \mid y) \right|$$

$$S_{\min}(X,Y) := \min_{w \in \mathbb{F}_q^{\times}} \sum_{y \in \mathcal{V}} P_Y(y) \left| P_{X|Y}^*(w \mid y) \right|.$$

The quantities $Z_{\max}(X,Y)$ and $Z_{\min}(X,Y)$ are invariant under any permutation of symbols in the a posteriori distribution. Although $S_{\max}(X,Y)$ and $S_{\min}(X,Y)$ are not invariant under any permutation of symbols in the a posteriori distribution, it holds that $S_{\max/\min}(X,Y) = S_{\max/\min}(rX + d(Y),Y)$ for any $d: \mathcal{Y} \to \mathbb{F}_q$ and $r \in \mathbb{F}_q^{\times}$. Hence, if $(X,Y) \overset{\text{a(1)}}{\sim} (X',Y')$, it holds that $S_{\max/\min}(X,Y) = S_{\max/\min}(X',Y')$. It is also straightforward to see the inequalities $Z_{\min}(X,Y) \leq Z(X \mid Y) \leq Z_{\max}(X,Y)$ and $S_{\min}(X,Y) \leq S(X \mid Y) \leq S_{\max}(X,Y)$ to hold.

B. Speed of polarization

In this subsection, we assume that $H(X \mid Y) \in (0,1)$, and also assume in view of Lemma 8, without loss of generality, that (X,Y) is a channel. The exponents for channel coding and source coding are introduced in [9], [15] for expressing the speed of polarization.

Definition 30. The exponent of G for channel coding is defined as

$$E_{c}(G) := \frac{1}{\ell \log \ell} \sum_{i=0}^{\ell-1} \log D_{c}^{(i)}(G)$$

where $D_{\mathrm{c}}^{(i)}(G)$ denotes the Hamming distance between the i-th row of G and the linear space spanned by (i+1)-th row to $(\ell-1)$ -th row of G. The exponent of G for source coding is defined as

$$E_{\mathbf{s}}(G) := \frac{1}{\ell \log \ell} \sum_{i=0}^{\ell-1} \log D_{\mathbf{s}}^{(i)}(G)$$

where $D_{\rm s}^{(i)}(G)$ denotes the Hamming distance between the *i*-th column of G^{-1} and the linear space spanned by 0-th column to (i-1)-th column of G^{-1} .

The following theorem holds, which was shown by Arıkan and Telatar [16], Korada et al. [9] and Korada [15] for the binary case with an additional condition.

Theorem 31. If a channel (X,Y) is polarized by G, it holds that for any $\epsilon > 0$,

$$\lim_{n \to \infty} \Pr\left(P_{e}(\mathsf{X}_{n} \mid \mathsf{Y}_{n}) < 2^{-\ell^{(E_{c}(G) - \epsilon)n}}\right) = 1 - H(X \mid Y)$$

$$\lim_{n \to \infty} \Pr\left(P_{e}(\mathsf{X}_{n} \mid \mathsf{Y}_{n}) < 2^{-\ell^{(E_{c}(G) + \epsilon)n}}\right) = 0.$$
(12)

Furthermore, it holds that for any $\epsilon > 0$,

$$\lim_{n \to \infty} \Pr\left(T(\mathsf{X}_n \mid \mathsf{Y}_n) < 2^{-\ell^{(E_s(G) - \epsilon)n}}\right) = H(X \mid Y)$$

$$\lim_{n \to \infty} \Pr\left(T(\mathsf{X}_n \mid \mathsf{Y}_n) < 2^{-\ell^{(E_s(G) + \epsilon)n}}\right) = 0.$$
(13)

Remark 2. Korada proved (13) for the binary case with the aid of the condition $D_{\rm s}^{(i)}(G) \geq D_{\rm s}^{(i+1)}(G)$ for $i=0,\ldots,\ell-2$ [15]. In this paper, (13) is proved without any additional condition for both binary and non-binary cases.

From Theorem 31, the error probability of polar codes as channel codes of rate smaller than I(W) and the distortion gap to the optimal distortion of polar codes as source codes are asymptotically bounded by $2^{-\ell^{(E_c(G)-\epsilon)n}}$ and $2^{-\ell^{(E_s(G)-\epsilon)n}}$, respectively [15]. From Corollary 28, it is sufficient to prove (12) and (13) for $Z(\mathsf{X}_n \mid \mathsf{Y}_n)$ and $S(\mathsf{X}_n \mid \mathsf{Y}_n)$ instead of $P_e(\mathsf{X}_n \mid \mathsf{Y}_n)$ and $T(\mathsf{X}_n \mid \mathsf{Y}_n)$, respectively. The general proof shown in [17], [18] can be used for our purpose.

Lemma 32 ([17], [18]). Let $\{Z_n\}_{n=0,1,...}$ be a random process ranging in [0,1] and $\{D_n\}_{n=0,1,...}$ be i.i.d. random variables ranging in $[1,\infty)$. Assume that the expectation of $\log D_0$ exists. Four conditions (c0)–(c3) are defined as follows.

- (c0) $Z_n \in (0,1]$ with probability 1.
- (c1) There exists a random variable Z_{∞} such that $Z_n \to Z_{\infty}$ almost surely.
- (c2) There exists a positive constant c_0 such that $Z_{n+1} \le c_0 Z_n^{D_n}$ with probability 1.
- (c3) $Z_n^{D_n} \leq Z_{n+1}$ with probability 1.

If (c0), (c1) and (c2) are satisfied, it holds that

$$\lim_{n \to \infty} \Pr\left(Z_n < 2^{-\ell^{(\mathbb{E}[\log_{\ell} D_0] - \epsilon)n}}\right) = \Pr(Z_\infty = 0).$$

If (c0), (c1) and (c3) are satisfied, it holds that

$$\lim_{n \to \infty} \Pr\left(Z_n < 2^{-\ell^{\left(\mathbb{E}[\log_{\ell} D_0] + \epsilon\right)n}} \right) = 0.$$

In the above, ℓ is any constant greater than 1.

Remark 3. We do not assume the condition $Z_n < 1$ to hold in Lemma 32, although it was assumed to hold with probability 1 in the arguments in [17] and [18]. The condition is not needed in our argument here because we have only to deal with the case $Z_n \to 0$.

From the assumption of Theorem 31, the channel is polarized by G. From Lemma 16 and Corollary 28, $Z_{\max}(X_n \mid Y_n)$, $Z_{\min}(X_n \mid Y_n)$, $S_{\max}(X_n \mid Y_n)$ and $S_{\min}(X_n \mid Y_n)$ converge almost surely to $\{0,1\}$ -valued random variables. From this observation, Lemma 32 implies:

- If the pair of $\{Z_n=Z_{\max}(\mathsf{X}_n,\mathsf{Y}_n)\}_{n=0,1,\dots}$ and $\{D_n=D_{\mathsf{c}}^{(B_n)}(G)\}_{n=0,1,\dots}$ satisfies (c0) and (c2), then the first equation of (12) holds.
- If the pair of $\{Z_n=Z_{\min}(\mathsf{X}_n,\mathsf{Y}_n)\}_{n=0,1,\dots}$ and $\{D_n=D_{\mathsf{c}}^{(B_n)}(G)\}_{n=0,1,\dots}$ satisfies (c0) and (c3), then the second equation of (12) holds.
- If the pair of $\{Z_n = S_{\max}(\mathsf{X}_n, \mathsf{Y}_n)\}_{n=0,1,\dots}$ and $\{D_n = D_{\mathrm{s}}^{(B_n)}(G)\}_{n=0,1,\dots}$ satisfies (c0) and (c2), then the first equation of (13) holds.
- If the pair of $\{Z_n = S_{\min}(\mathsf{X}_n, \mathsf{Y}_n)\}_{n=0,1,\dots}$ and $\{D_n = D_{\mathrm{s}}^{(B_n)}(G)\}_{n=0,1,\dots}$ satisfies (c0) and (c3), then the second equation of (13) holds.

The following lemma shows that the pair of $\{Z_n = Z_{\max}(\mathsf{X}_n,\mathsf{Y}_n)\}_{n=0,1,\dots}$ and $\{D_n = D_{\mathsf{c}}^{(B_n)}(G)\}_{n=0,1,\dots}$ satisfies the condition (c2), and that the pair of $\{Z_n = Z_{\min}(\mathsf{X}_n,\mathsf{Y}_n)\}_{n=0,1,\dots}$ and $\{D_n = D_{\mathsf{c}}^{(B_n)}(G)\}_{n=0,1,\dots}$ satisfies the condition (c3).

Lemma 33 ([9]). For $i \in \{0, ..., \ell - 1\}$, it holds for any channel (X, Y) that

$$Z_{\max}(X^{(i)}, Y^{(i)}) \le q^{\ell - 1 - i} Z_{\max}(X, Y)^{D_{c}^{(i)}(G)}$$
$$Z_{\min}(X, Y)^{D_{c}^{(i)}(G)} \le Z_{\min}(X^{(i)}, Y^{(i)}).$$

The proof is omitted since the same proof for the binary alphabet in [9] applies to the non-binary cases as well. The following lemma shows that the pair of $\{Z_n = S_{\max}(\mathsf{X}_n,\mathsf{Y}_n)\}_{n=0,1,\dots}$ and $\{D_n = D_{\mathrm{s}}^{(B_n)}(G)\}_{n=0,1,\dots}$ satisfies the condition (c2), and that the pair of $\{Z_n = S_{\min}(\mathsf{X}_n,\mathsf{Y}_n)\}_{n=0,1,\dots}$ and $\{D_n = D_{\mathrm{s}}^{(B_n)}(G)\}_{n=0,1,\dots}$ satisfies the condition (c3).

Lemma 34. For $i \in \{0, ..., \ell - 1\}$, it holds for any source (X, Y) that

$$\begin{split} S_{\max}(X^{(i)},Y^{(i)}) &\leq q^{i} S_{\max}(X,Y)^{D_{\mathrm{s}}^{(i)}(G)} \\ S_{\min}(X,Y)^{D_{\mathrm{s}}^{(i)}(G)} &\leq S_{\min}(X^{(i)},Y^{(i)}). \end{split}$$

The proof is in Appendix E.

Finally, we should prove that all the four processes satisfy (c0). If the channel (X,Y) satisfies the two inequalities $Z_{\min}(X,Y)>0$ and $S_{\min}(X,Y)>0$, (c0) obviously holds for the four processes since the property (c0) is inherited in the

processes i.e., if $Z_{\min}(X,Y) > 0$, then $Z_{\min}(X^{(i)},Y^{(i)}) > 0$ for $i=0,\ldots,\ell-1$. In the following, we deal with the other cases. It is sufficient to prove the following lemma.

Lemma 35. Assume that (X,Y) is polarized by G. Then,

$$\lim_{n \to \infty} \Pr \left(Z_{\min}(\mathsf{X}_n, \mathsf{Y}_n) > 0 \right) = 1$$

$$\lim_{n \to \infty} \Pr \left(S_{\min}(\mathsf{X}_n, \mathsf{Y}_n) > 0 \right) = 1.$$

The proof is in Appendix F. Lemma 35 implies Theorem 31 for the cases $Z_{\min}(X,Y)=0$ or $S_{\min}(X,Y)=0$ due to the following reason. For any $\delta>0$, there exists n_0 such that

$$\Pr (Z_{\min}(\mathsf{X}_n, \mathsf{Y}_n) > 0) \ge 1 - \delta$$

$$\Pr (S_{\min}(\mathsf{X}_n, \mathsf{Y}_n) > 0) \ge 1 - \delta$$

for any $n \geq n_0$. Theorem 31 can be applied to each of the channels $(X^{(b_1)\cdots(b_{n_0})},Y^{(b_1)\cdots(b_{n_0})})$ satisfying the inequalities $Z_{\min}(X^{(b_1)\cdots(b_{n_0})},Y^{(b_1)\cdots(b_{n_0})})>0$ and $S_{\min}(X^{(b_1)\cdots(b_{n_0})},Y^{(b_1)\cdots(b_{n_0})})>0$. As a consequence, it holds that for any $\delta>0$ and $\epsilon>0$

$$\begin{split} &(1 - H(X \mid Y)) \, (1 - \delta) \\ &\leq \liminf_{n \to \infty} \Pr \left(P_{\mathbf{e}}(\mathsf{X}_n \mid \mathsf{Y}_n) < 2^{-\ell^{(E_{\mathbf{c}}(G) - \epsilon)n}} \right), \\ &\limsup_{n \to \infty} \Pr \left(P_{\mathbf{e}}(\mathsf{X}_n \mid \mathsf{Y}_n) < 2^{-\ell^{(E_{\mathbf{c}}(G) - \epsilon)n}} \right) \\ &\leq \left(1 - H(X \mid Y) \right) (1 - \delta) + \delta, \\ &\limsup \Pr \left(P_{\mathbf{e}}(\mathsf{X}_n \mid \mathsf{Y}_n) < 2^{-\ell^{(E_{\mathbf{c}}(G) + \epsilon)n}} \right) \leq \delta. \end{split}$$

Similar inequalities corresponding to (13) also hold. By letting $\delta \to 0$, Theorem 31 is obtained.

A more detailed asymptotic analysis depending on the rate can also be performed as shown in [19], [20], [17], [18] for the binary case. For example, under the condition that G polarizes (X,Y), one can prove that for $R \in (0,1-H(X \mid Y))$,

$$\lim_{n \to \infty} \Pr\left(P_{\mathbf{e}}(\mathsf{X}_n \mid \mathsf{Y}_n)\right) < 2^{-\ell^{E_{\mathbf{e}}(G)n} + \sqrt{V_{\mathbf{e}}(G)n}Q^{-1}\left(\frac{R}{1 - H(X|Y)}\right) + f(n)}}\right) = R$$

holds for an arbitrary function satisfying $f(n) = o(\sqrt{n})$,

$$V_{c}(G) := \frac{1}{\ell} \sum_{i=0}^{\ell-1} (\log_{\ell} D_{c}^{(i)}(G) - E_{c}(G))^{2}$$

and where $Q^{-1}(\cdot)$ is the inverse function of the error function $Q(t) := \int_{t}^{\infty} e^{-z^{2}/2} dz / \sqrt{2\pi}$.

In the binary case, any source is polarized by G if and only if $E_{\rm c}(G)>0$ [9]. The property also holds when q is a prime since the condition $E_{\rm c}(G)>0$ is equivalent to the condition that a standard form of G is not the identity matrix. However, it no longer holds when q is not a prime, in which case there may be sources which are not polarized by G even if $E_{\rm c}(G)>0$, as shown in Section VI-B. Since non-zero scalar multiplication of a column does not change the exponent $E_{\rm c}(G)$, even if there are non-polarizing sources for G satisfying $E_{\rm c}(G)>0$, appropriate scalar multiplication of a column of G gives a matrix with the same exponent $E_{\rm c}(G)$ which polarizes any source.

VIII. REED-SOLOMON MATRIX AND ITS EXPONENT

Let $\mathbb{F}_q=\{x_0,\dots,x_{q-1}\}$. Let $a=[a_0,\dots,a_{k-1}]\in\mathbb{F}_q^k$ and $p_a(X)=a_0+a_1X+\dots+a_{k-1}X^{k-1}$. The encoder of the q-ary extended Reed-Solomon code is defined as $\varphi(a):=[p_a(x_0),p_a(x_1),\dots,p_a(x_{q-1})]$. Let α be a primitive element of \mathbb{F}_q . When $x_{q-1}=0$ and $x_i=\alpha^{-i}$ for $i=0,\dots,q-2$, the generator matrix of the q-ary extended Reed-Solomon code is a lower submatrix of the $q\times q$ matrix $G_{\mathrm{RS}}(q)$ over \mathbb{F}_q which we call the Reed-Solomon matrix

$$G_{\mathrm{RS}}(q) := \begin{bmatrix} 1 & 1 & 1 & \cdots & 1 & 0 \\ 1 & \alpha & \alpha^2 & \cdots & \alpha^{q-2} & 0 \\ 1 & \alpha^2 & \alpha^4 & \cdots & \alpha^{2(q-2)} & 0 \\ \vdots & \vdots & \vdots & \cdots & \vdots & \vdots \\ 1 & \alpha^{q-2} & \alpha^{2(q-2)} & \cdots & \alpha^{(q-2)(q-2)} & 0 \\ 1 & 1 & 1 & \cdots & 1 & 1 \end{bmatrix}.$$

From Theorem 14, any source is polarized by the Reed-Solomon matrix. Since extended Reed-Solomon codes are maximum distance separable (MDS) codes, one has $D_{\rm c}^{(i)}=i+1$ for $i=0,\ldots,q-1$, and therefore the exponent of the Reed-Solomon matrix for channel coding is $E_{\rm c}(G_{\rm RS}(q))=\log(q!)/q$. The inverse matrix of the Reed-Solomon matrix $G_{\rm RS}(q)$ is

$$\begin{bmatrix} 1 & 1 & 1 & \cdots & 1 & 0 \\ 1 & \alpha^{-1} & \alpha^{-2} & \cdots & \alpha^{-(q-2)} & 0 \\ 1 & \alpha^{-2} & \alpha^{-4} & \cdots & \alpha^{-2(q-2)} & 0 \\ \vdots & \vdots & \vdots & \cdots & \vdots & \vdots \\ 1 & \alpha^{-(q-2)} & \alpha^{-2(q-2)} & \cdots & \alpha^{-(q-2)(q-2)} & 0 \\ 1 & 0 & 0 & \cdots & 0 & -1 \end{bmatrix}.$$

Hence, the exponent of the Reed-Solomon matrix for source coding is also $E_{\rm s}(G_{\rm RS}(q)) = \log(q!)/q$. Note that both of the exponents $\log(q!)/q$ monotonically increase in q and converge to 1 as $q \to \infty$.

For $i \in \{0,1,\ldots,q^n-1\}$, $i_n i_{n-1} \cdots i_1$ denotes the q-ary expansion of i. For polar codes constructed on the basis of the matrix $G_{\mathrm{RS}}(q)$, rows of $G_{\mathrm{RS}}(q)^{\otimes n}$ whose indices are in the set

$$\{i \in \{0, \dots, q^n - 1\} \mid H(X^{(i_1)\cdots(i_n)} \mid Y^{(i_1)\cdots(i_n)}) < \epsilon\}$$

with some threshold $\epsilon>0$ are chosen, as mentioned in Section III-B. For the Reed-Muller codes, on the other hand, rows of $G_{\mathrm{RS}}(q)^{\otimes n}$ whose indices belong to

$$\{i \in \{0, \dots, q^n - 1\} \mid i_1 + \dots + i_n > n_0\}$$

are chosen for some threshold $n_0 \in \{0, 1, \dots, n(q-1)\}^3$. In order to maximize the minimum distance, rows of $G_{RS}(q)^{\otimes n}$ with indices in the set

$$\{i \in \{0, \dots, q^n - 1\} \mid (i_1 + 1) \cdots (i_n + 1) > n_0\}.$$
 (14)

with some threshold $n_0 \in \{1, 2, \dots, q^n\}$ should be chosen. Hence, unless q = 2, the selection rule for the Reed-Muller codes does not maximize the minimum distance. Codes based

on the selection rule (14) are sometimes called Massey-Costello-Justesen codes [21] and hyperbolic cascaded Reed-Solomon codes [22]. Note that the minimum distance of Reed-Muller codes grows like $q^{n/2+o(n)}$ while the minimum distance of polar codes and hyperbolic codes grows like $q^{E_c(G_{\rm RS}(q))n+o(n)}$. From the above observation, the Reed-Solomon matrices can be regarded as a natural generalization of the matrix $\begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix}$ in the binary case.

We now consider the maximum exponent $E_{\max}(q,\ell) :=$ $\max_{G \in \mathbb{F}_c^{\ell \times \ell}} E_{\mathrm{c}}(G)$ for channel coding on given size q of a finite field and size ℓ of a matrix. For q=2, Korada et al. [9] show that $E_{\rm max}(2,\ell) < 0.55$ for $\ell \leq 31$, and also show a method of construction of binary matrices with large exponents using the Bose-Chaudhuri-Hocquenghem (BCH) codes. For $q \geq 2$ and $\ell \leq q$, the $\ell \times \ell$ lower-right submatrix of the q-ary Reed-Solomon matrix gives the largest exponent so that $E_{\max}(q,\ell) = \log(\ell!)/(\ell \log \ell)$ for $\ell \leq q$ since the Reed-Solomon code is an MDS code [23]. Thus, the Reed-Solomon matrices with q > 2 can be regarded as providing a systematic means to construct polar codes with larger exponents for the case $\ell \leq q$. For example, for q=4, $E_{\rm max}(4,4) = E_{\rm c}(G_{\rm RS}(4)) \approx 0.57312$, which is larger than the upper bound 0.55 of $E_{\text{max}}(2, \ell)$ for $\ell \leq 31$. For $\ell > q > 2$, on the other hand, algebraic geometry codes are considered to be useful since they have a large minimum distance and the nested structure which are plausible in making $D_c^{(i)}$ s larger. The examples using the Hermitian codes are shown in [2], in which $q = p^m$ and $\ell = p^{3m/2}$ for an even integer m. The q-ary $\ell \times \ell$ matrix constructed on the basis of the Hermitian code has a yet larger exponent than the Reed-Solomon matrix $G_{RS}(q)$ for q > 4.

IX. NUMERICAL RESULTS

In Fig. 2, performance of the original binary polar codes with G_1 and quaternary polar codes using the Reed-Solomon matrix $G_{RS}(4)$ are compared on the binary-input additive-white-Gaussian-noise (AWGN) channel with capacity about 0.5. Instead of the actual error probability, the upper bound $\sum_{i\in\mathcal{A}}P_{\mathrm{e}}(X^{(i_1)\cdots(i_n)}\mid Y^{(i_1)\cdots(i_n)})$ is plotted where \mathcal{A} denotes the set of chosen row indices in constructing polar codes. This bound is accurate for rates not close to the capacity [24]. A significant improvement by the quaternary polar codes over the binary counterparts is observed in terms of the block error probability, although the error probability of the quaternary polar codes is still larger than that of (3,6)-regular low-density-parity-check (LDPC) codes except in a low-rate region.

X. SUMMARY

We have shown that a necessary and sufficient condition for a q-ary $\ell \times \ell$ invertible matrix G over \mathbb{F}_q with a non-identity standard form \bar{G} to polarize any source/channel is $\mathbb{F}_p(\bar{G}) = \mathbb{F}_q$. The result about speed of polarization for the binary alphabet has been generalized to non-binary cases. We have also explicitly given q-ary $\ell \times \ell$ matrices with $\ell \leq q$ on the basis of the q-ary Reed-Solomon matrices, which have the largest exponent $E_{\max}(q,\ell) = \log(\ell!)/(\ell \log \ell)$ among all

³Here, i_1, \ldots, i_n are treated as integers in the additions.

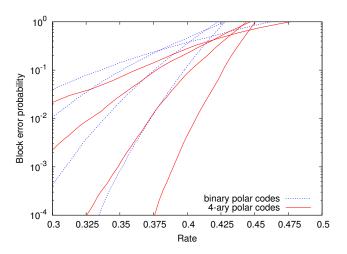


Fig. 2. Numerical results on the upper bound of the block error probability of polar codes over an AWGN channel, for which the standard deviation of noise is set equal to 0.978 65. The capacity of the AWGN channel is about 0.5. Results of binary polar codes and quaternary polar codes using $G_{\rm RS}(4)$ are shown by dotted curves and solid curves, respectively. Blocklengths are $2^7, 2^9, 2^{11}$, and 2^{13} viewed as binary codes.

 $\ell \times \ell$ matrices. Performance of non-binary polar codes based on Reed-Solomon matrices are found via numerical evaluation to be significantly better than the performance of the original binary polar codes.

APPENDIX A PROOF OF LEMMA 20

In order to relate the entropy and the Bhattacharyya parameter, the following lemma is useful.

Lemma 36 ([25, Sec. 5.6]). For any random variables X, Y and Z on sets X, Y and Z, respectively,

$$\sum_{x \in \mathcal{X}, y \in \mathcal{Y}} P_{X,Y}(x, y) \log \frac{P_{X,Y}(x, y)}{P_X(x) P_Y(y)}$$

$$\geq -\log \sum_{y \in \mathcal{Y}} \left(\sum_{x \in \mathcal{X}} P_X(x) \sqrt{P_{Y|X}(y \mid x)} \right)^2$$

$$\sum_{x \in \mathcal{X}, y \in \mathcal{Y}, z \in \mathcal{Z}} P_{X,Y,Z}(x, y, z) \log \frac{P_{X,Y|Z}(x, y \mid z)}{P_{X|Z}(x \mid z) P_{Y|Z}(y \mid z)}$$

$$\geq -\log \sum_{y \in \mathcal{Y}, z \in \mathcal{Z}} P_Z(z)$$

$$\cdot \left(\sum_{x \in \mathcal{X}} P_{X|Z}(x \mid z) \sqrt{P_{Y|X,Z}(y \mid x, z)} \right)^2.$$

The second inequality is an immediate consequence of the first inequality and Jensen's inequality. The first inequality is obtained in [25, Sec. 5.6]. In Lemma 36, the quantities on the left-hand sides are the mutual information between X and Y, and the conditional mutual information between X and Y given Z, respectively. The quantities on the right-hand sides are the cutoff rate and the conditional cutoff rate, respectively.

Given a source (X, Y), let $(U_0, U_1, X_0, X_1, Y_0, Y_1)$ be the random variables defined by applying the basic transform with

 G_{γ} to the source (X,Y), as described in Section III. Then, one obtains

$$\begin{split} &H(U_1\mid Y_1) - H(U_1\mid U_0, Y_0, Y_1) \\ &= \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}, y_1 \in \mathcal{Y}} P_{U_0, U_1, Y_0, Y_1}(u_0, u_1, y_0, y_1) \\ &\cdot \log \frac{P_{U_0, U_1, Y_0\mid Y_1}(u_0, u_1, y_0\mid y_1)}{P_{U_1\mid Y_1}(u_1\mid y_1) P_{U_0, Y_0\mid Y_1}(u_0, y_0\mid y_1)} \\ &\geq -\log \left[\sum_{y_1 \in \mathcal{Y}} P_Y(y_1) \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \left[\sum_{u_1 \in \mathbb{F}_q} P_{U_1\mid Y_1}(u_1\mid y_1) \sqrt{P_{U_0, Y_0\mid U_1, Y_1}(u_0, y_0\mid u_1, y_1)} \right]^2 \right] \\ &= -\log \left[\sum_{y_1 \in \mathcal{Y}} P_Y(y_1) \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \left[\sum_{u_1 \in \mathbb{F}_q} P_{U_1\mid Y_1}(u_1\mid y_1) \sqrt{P_{X_0, Y_0}(u_0 + u_1, y_0)} \right]^2 \right] \\ &= -\log \left[\sum_{y_1 \in \mathcal{Y}} P_Y(y_1) \sum_{u_1 \in \mathbb{F}_q, u_1' \in \mathbb{F}_q} P_{X\mid Y}(\gamma u_1\mid y_1) P_{X\mid Y}(\gamma u_1'\mid y_1) \\ \cdot \sum_{u_1 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0 + u_1, y_0)} \sqrt{P_{X,Y}(u_0 + u_1', y_0)} \right] \\ &= -\log \left[1 - \sum_{y_1 \in \mathcal{Y}} P_Y(y_1) \right] \\ &\cdot \sum_{u_1 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0 + u_1, y_0)} \sqrt{P_{X,Y}(u_0 + u_1', y_0)} \right] \\ &= -\log \left[1 - q \sum_{d \in \mathbb{F}_q, y_0 \in \mathcal{Y}} P_Y(y_1) \right] \\ &\cdot \sum_{u_1 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0 + u_1, y_0)} \sqrt{P_{X,Y}(u_0 + u_1', y_0)} \right] \\ &\geq -\log \left[1 - q \sum_{d \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0 + u_1, y_0)} \right] \\ &\geq -\log \left[1 - q \sum_{d \in \mathbb{F}_q'} \sum_{u_1 \in \mathcal{Y}} \sum_{u_1 \in \mathcal{Y}} \frac{1}{q} P_Y(y_1) \right] \\ &\cdot \sqrt{P_{X,Y}(u_0 + u_1 + d, y_0)} \right] \\ &\cdot \left(1 - \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0, y_0)} \sqrt{P_{X,Y}(u_0 + d, y_0)} \right) \right] \\ &\cdot \left(1 - \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0, y_0)} \sqrt{P_{X,Y}(u_0 + d, y_0)} \right) \\ &\cdot \left(1 - \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0, y_0)} \sqrt{P_{X,Y}(u_0 + d, y_0)} \right) \right] \\ &\cdot \left(1 - \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0, y_0)} \sqrt{P_{X,Y}(u_0 + d, y_0)} \right) \\ &\cdot \left(1 - \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0, y_0)} \sqrt{P_{X,Y}(u_0 + d, y_0)} \right) \right] \\ &\cdot \left(1 - \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0, y_0)} \sqrt{P_{X,Y}(u_0 + d, y_0)} \right) \\ &\cdot \left(1 - \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0, y_0)} \sqrt{P_{X,Y}(u_0 + d, y_0)} \right) \right] \\ &\cdot \left(1 - \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0, y_0)} \sqrt{P_{X,Y}(u_0 + d, y_0)} \right) \\ &\cdot \left(1 - \sum_{u_0 \in \mathbb{F}_q, y_0 \in \mathcal{Y}} \sqrt{P_{X,Y}(u_0, y_0)} \sqrt{P_{X,Y}(u_0 + d, y_0)}$$

$$= -\log \left[1 - \frac{1}{q} \sum_{d \in \mathbb{F}_q^{\times}} Z_{\gamma d}(X \mid Y)^2 (1 - Z_d(X \mid Y)) \right].$$

The first and second inequalities are obtained by Lemma 36 and Jensen's inequality, respectively.

The assumption of Lemma 20 implies that the above formula evaluated for $(X,Y)=(X_{(n)},Y_{(n)})$ approaches 0 as $n\to\infty$, or equivalently, that for any $\epsilon>0$, there exists n_0 such that

$$Z_{\gamma d}(X_{(n)} \mid Y_{(n)})(1 - Z_d(X_{(n)} \mid Y_{(n)})) < \epsilon$$

for any $n \geq n_0$ and any $d \in \mathbb{F}_q^{\times}$. Fix $\epsilon \in (0, 1/2)$. Then, there exists n_0 such that

$$Z_{\gamma d}(X_{(n)} \mid Y_{(n)})(1 - Z_d(X_{(n)} \mid Y_{(n)})) < \epsilon^2$$

for any $n \geq n_0$ and any $d \in \mathbb{F}_q^{\times}$, which in turn implies

$$Z_{\gamma d}(X_{(n)} \mid Y_{(n)}) < \epsilon$$
 or $1 - Z_d(X_{(n)} \mid Y_{(n)}) < \epsilon$

for any $n \geq n_0$ and any $d \in \mathbb{F}_q^{\times}$. Assume $1 - Z_{d'}(X_{(n')}) < \epsilon$ for fixed $n' \geq n_0$ and fixed $d' \in \mathbb{F}_q^{\times}$. Then, from

$$Z_{d'}(X_{(n')} \mid Y_{(n')})(1 - Z_{\gamma^{-1}d'}(X_{(n')} \mid Y_{(n')})) < \epsilon^2$$

one obtains $1-Z_{\gamma^{-1}d'}(X_{(n')}\mid Y_{(n')}))<\epsilon^2/(1-\epsilon)<\epsilon$. By iterating this procedure, one proves that $1-Z_{\gamma^id'}(X_{(n')}\mid Y_{(n')}))<\epsilon$ holds for all $i\in\{0,\ldots,q-2\}$. In the same way, when $Z_{d'}(X_{(n')}\mid Y_{(n')})<\epsilon$ is assumed for fixed $n'\geq n_0$ and fixed $d'\in\mathbb{F}_q^\times$, one can prove that $Z_{\gamma^id'}(X_{(n')}\mid Y_{(n')}))<\epsilon$ holds for all $i\in\{0,\ldots,q-2\}$. This completes the proof of Lemma 20.

APPENDIX B

BHATTACHARYYA PARAMETER AND ERROR PROBABILITY

In this appendix, an unconditional version of Lemma 22 is proved. Lemma 22 itself is then proved straightforwardly by Jensen's inequality. For the proof of the unconditional version, one can regard $\mathcal X$ as any finite set whose size q is not necessarily a power of a prime. Let X be a random variable on $\mathcal X$. The optimum estimator for X minimizing the probability of error is given by $\hat x := \arg\max_x P_X(x)$, with the error probability

$$P_{e}(X) := 1 - \max_{x \in \mathcal{X}} P_X(x).$$

The Bhattacharyya parameter is defined as

$$Z(X) := \frac{1}{q-1} \sum_{\substack{x \in \mathcal{X}, x' \in \mathcal{X}, \\ x' \neq x}} \sqrt{P_X(x)P_X(x')}.$$

The following lemma gives an upper bound of the error probability in terms of the Bhattacharyya parameter.

Lemma 37.

$$P_{e}(X) \le \min_{k=1,2,\dots,q-1} \left\{ \frac{(q-1)Z(X) + k(k-1)}{k(k+1)} \right\}.$$

Proof: Noting that $P_X(\hat{x}) = 1 - P_{\rm e}(X)$ holds by the definition, one has

$$\sum_{x} \sqrt{P_X(x)} = \sqrt{1 - P_{e}(X)} + \sum_{x \neq \hat{x}} \sqrt{P_X(x)}.$$

In order to prove the lemma, we first find the extremal distribution of X for which Z(X) is minimized with $P_{\rm e}(X)$ fixed. As we will show, this amounts to minimizing the second term on the right-hand side with respect to $P_X(x)$ under the constraint that the error probability is $P_{\rm e}(X)$. We thus consider the following minimization problem for $\{p_i\}_{i=0,1,\ldots,q-2}$.

minimize:
$$\sum_i \sqrt{p_i}$$
 subject to:
$$\sum_i p_i = P_{\rm e}(X)$$

$$0 < p_i < 1 - P_{\rm e}(X).$$

Let $\{p_i^*\}_{i=0,1,\dots,q-2}$ be the optimum solution of the minimization problem. Since \sqrt{x} is a concave function, p_i^* is 0 or $1-P_{\mathrm{e}}(X)$ except for at most one i [26]. Let t-1 be the number of p_i^* s which are equal to $1-P_{\mathrm{e}}(X)$, then $t=\lfloor 1/(1-P_{\mathrm{e}}(X)) \rfloor$ holds. The value of p_i^* which is not 0 or $1-P_{\mathrm{e}}(X)$ is equal to $1-t(1-P_{\mathrm{e}}(X))$. Hence,

$$\sum_{T} \sqrt{P_X(x)} \ge t\sqrt{1 - P_{\rm e}(X)} + \sqrt{1 - t(1 - P_{\rm e}(X))}.$$
 (15)

By squaring both sides of (15), one obtains the inequality

$$1 + (q - 1)Z(X) \ge 1 + t(t - 1)(1 - P_{e}(X)) + 2t\sqrt{(1 - P_{e}(X))(1 - t(1 - P_{e}(X)))}$$

which implies the minimum achievable value of the Bhattacharyya parameter for a given error probability. The right-hand side of the above inequality is further lower bounded by applying the inequality $1-P_{\rm e}(X)\geq 1-t(1-P_{\rm e}(X))\Leftrightarrow t\geq 1/(1-P_{\rm e}(X))-1$ to the last term, yielding

$$(q-1)Z(X) \ge t(t-1)(1-P_{e}(X)) + 2t(1-t(1-P_{e}(X)))$$

= $-(1-P_{e}(X))t^{2} + (1+P_{e}(X))t$. (16)

Since the quadratic function $-(1-P_{\rm e}(X))x^2+(1+P_{\rm e}(X))x$ is concave and takes a maximum at $x=(1+P_{\rm e}(X))/(2(1-P_{\rm e}(X)))$, which is the center of the unit interval $[P_{\rm e}(X)/(1-P_{\rm e}(X)),1/(1-P_{\rm e}(X))]$ containing t, the inequality (16) still holds even if t is replaced by any integer $k=1,2,\ldots,q-1$.

By replacing t by $1/(1-P_{\rm e}(X))$ in (16), one obtains a looser but smooth bound

$$P_{\rm e}(X) \le \frac{(q-1)Z(X)}{(q-1)Z(X)+1}.$$
 (17)

This bound is also obtained from the monotonicity of the Rényi entropy i.e., $H_{1/2}(X) \geq H_{\infty}(X)$ where $H_{\alpha}(X) := (1-\alpha)^{-1} \log \sum_{x} P_{X}(x)^{\alpha}$. These upper bounds are plotted in Fig. 3 for q=5.

The next lemma provides a lower bound of the error probability in terms of the Bhattacharyya parameter.

Lemma 38.

$$P_{\rm e}(X) \ge \frac{q-1}{g^2} \left(\sqrt{1 + (q-1)Z(X)} - \sqrt{1 - Z(X)} \right)^2.$$

Proof: We start with the same formula as that used as the starting point of the proof of Lemma 37.

$$\sum_{x} \sqrt{P_X(x)} = \sqrt{1 - P_e(X)} + \sum_{x \neq \hat{x}} \sqrt{P_X(x)}$$

$$= \sqrt{1 - P_e(X)} + (q - 1) \sum_{x \neq \hat{x}} \frac{1}{q - 1} \sqrt{P_X(x)}$$

$$\leq \sqrt{1 - P_e(X)} + (q - 1) \sqrt{\frac{1}{q - 1} P_e(X)}$$

$$= \sqrt{1 - P_e(X)} + \sqrt{(q - 1) P_e(X)}. \tag{18}$$

The above inequality is obtained from Jensen's inequality. This proof is the same as the proof of Fano's inequality for the Rényi entropy [27]. By squaring both sides of the above inequality, one has

$$1 + (q - 1)Z(X) \le \left[\sqrt{1 - P_{e}(X)} + \sqrt{q - 1}\sqrt{P_{e}(X)}\right]^{2}$$

$$\iff Z(X)$$

$$\le \frac{(q - 2)P_{e}(X) + 2\sqrt{q - 1}\sqrt{P_{e}(X)(1 - P_{e}(X))}}{q - 1}.$$

The function

$$f(x) := \frac{(q-2)x + 2\sqrt{q-1}\sqrt{x(1-x)}}{q-1}$$

defined for $x \in [0,(q-1)/q]$ is continuous and strictly increasing since

$$f'(x) = \frac{q-2}{q-1} + \frac{1-2x}{\sqrt{q-1}\sqrt{x(1-x)}}$$
$$f''(x) = -\frac{1}{2\sqrt{q-1}(x(1-x))^{3/2}} < 0$$

and f'((q-1)/q)=0. Hence, $f^{-1}(Z(X))\leq P_{\rm e}(X)$ where the inverse function $f^{-1}(x)$ of f(x) is

$$f^{-1}(x) = \frac{q-1}{q^2} \left(\sqrt{1 + (q-1)x} - \sqrt{1-x} \right)^2.$$

Lemma 22 is obtained from Lemmas 37 and 38 by applying Jensen's inequality. The lower and upper bounds are plotted in Fig. 3 for q=5. The bounds given in Lemma 22 are the tightest among those which are given in terms of the Bhattacharyya parameter only. Tight examples are shown below. The lower bound in Lemma 22 is tight for the q-ary symmetric channel, defined by $\mathcal{X}=\mathcal{Y}=\{0,\ldots,q-1\}$ and

$$P_{Y|X}(y \mid x) = \begin{cases} 1 - \epsilon, & \text{if } y = x \\ \epsilon/(q - 1), & \text{if } y \neq x \end{cases}$$

for $\epsilon \in [0, (q-1)/q]$. In this case,

$$P_{e}(X \mid Y) = \epsilon$$

$$Z(X \mid Y) = \frac{q-2}{q-1}\epsilon + 2\sqrt{\frac{\epsilon(1-\epsilon)}{q-1}}$$

which satisfies the lower bound with equality. The upper bound in Lemma 22 is tight for the following channel. Let $\mathcal{X} = \{0, \dots, q-1\}$. For fixed $k \in \{1, \dots, q-1\}$, let

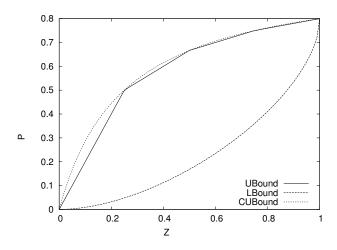


Fig. 3. The upper and lower bounds of error probability for q = 5.

 $\mathcal{Y}=A_k\cup A_{k+1}$ where $A_k:=\{\mathcal{A}\subseteq\mathcal{X}\mid |\mathcal{A}|=k\},$ and let

$$P_{Y|X}(y\mid x) = \begin{cases} (1-\epsilon)/\binom{q-1}{k-1}, & \text{if } |y| = k \text{ and } x \in y \\ \epsilon/\binom{q-1}{k}, & \text{if } |y| = k+1 \text{ and } x \in y \\ 0, & \text{otherwise} \end{cases}$$

for $\epsilon \in [0,1]$. That is, the output of the channel is a subset of $\mathcal X$ containing the input x and with size k or k+1. This channel satisfies the upper bound with equality since it holds

$$P_{e}(X \mid Y) = \frac{k^{2} - 1 + \epsilon}{k(k+1)}$$
$$Z(X \mid Y) = \frac{k - 1 + \epsilon}{a - 1}.$$

APPENDIX C PROOF OF LEMMA 23

Similarly to Appendix B, it is sufficient to prove an unconditional version of the inequalities in Lemma 23. Let \mathcal{X} be a finite set of size q, and let X be a random variable on \mathcal{X} . Let

$$T(X) := \sum_{x \in \mathcal{X}} \left| P_X(x) - \frac{1}{q} \right|$$

be the total variation distance between P_X and the uniform distribution over \mathcal{X} .

Let $t := \lfloor 1/(1 - P_e(X)) \rfloor$. The same argument as that of minimizing the concave function in Appendix B applies to minimizing -T(X) given $P_e(X)$, yielding the upper bound

$$T(X) \le t \left(1 - P_{e}(X) - \frac{1}{q} \right) + \left| 1 - t(1 - P_{e}(X)) - \frac{1}{q} \right|$$

$$+ (q - t - 1) \frac{1}{q}$$

$$= \frac{q - 1}{q} + t \left(1 - P_{e}(X) - \frac{2}{q} \right) + \left| \frac{q - 1}{q} - t(1 - P_{e}(X)) \right|$$

$$=: f_{T}(P_{e}(X)).$$

We now derive the concave hull of $f_T(x)$ for obtaining the upper bound of $T(X \mid Y)$. Let k be a positive integer smaller

than q. When x satisfies $(k-1)/k \le x < k/(k+1)$, one has $k \le 1/(1-x) < k+1$, so that the value of $t = \lfloor 1/(1-x) \rfloor$ is equal to the constant k. The function $f_T(x)$ is hence a convex function of x in the interval $(k-1)/k \le x < k/(k+1)$, and the linear interpolation of the values of $f_T(x)$ at the two endpoints x = (k-1)/k and $x \uparrow k/(k+1)$ thus gives the concave hull of $f_T(x)$ for $(k-1)/k \le x < k/(k+1)$. One therefore obtains the inequality

$$f_T(x) \le (k+1)k \left[(k/(k+1) - x) f_T((k-1)/k) + (x - (k-1)/k) \lim_{x \uparrow k/(k+1)} f_T(x) \right]$$

for x satisfying $(k-1)/k \le x < k/(k+1)$. By substituting

$$f_T((k-1)/k) = \frac{2}{q}(q-k)$$

$$\lim_{x \uparrow k/(k+1)} f_T(x) = \frac{2}{q}(q-k-1)$$

one obtains

$$f_T(x) \le \frac{2}{q}(q-k-1) + \frac{2}{q}k\left[\left(k-(k+1)x\right)\right]$$
$$= \frac{2(q-1)}{q} - \frac{2}{q}[-(1-x)k^2 + (1+x)k]$$

and therefore

$$T(X) \le \frac{2(q-1)}{q} - \frac{2}{q} [-(1-P_{e}(X))k^{2} + (1+P_{e}(X))k]$$
 (19)

for $P_{\rm e}(X)$ satisfying $(k-1)/k \leq P_{\rm e}(X) < k/(k+1)$. As shown in the proof of Lemma 37, the inequality (19) is correct for any $P_{\rm e}(X) \in [0, (q-1)/q]$. Note that by replacing k by $1/(1-P_{\rm e}(X))$, one obtains a looser but smooth upper bound

$$f_T(P_{e}(X)) \le \frac{2}{1 - P_{e}(X)} \left(\frac{q - 1}{q} - P_{e}(X)\right).$$

The unconditional version of the other inequality in Lemma 23 is obtained by applying the triangle inequality, as

$$T(X) = \left(1 - P_{e}(X) - \frac{1}{q}\right) + \sum_{x \neq \hat{x}} \left| P_{X}(x) - \frac{1}{q} \right|$$

$$\geq \left(1 - P_{e}(X) - \frac{1}{q}\right) + \left| \sum_{x \neq \hat{x}} \left(P_{X}(x) - \frac{1}{q} \right) \right|$$

$$= \left(1 - P_{e}(X) - \frac{1}{q}\right) + \frac{q - 1}{q} - P_{e}(X)$$

$$= 2\left(\frac{q - 1}{q} - P_{e}(X)\right).$$

APPENDIX D PROOF OF LEMMA 26

As before, it is again sufficient to prove an unconditional version of the inequalities in Lemma 26. The unconditional version S(X) of $S(X \mid Y)$ is defined as

$$S(X) := \frac{1}{q-1} \sum_{w \in \mathbb{F}_a^\times} |P_X^*(w)|$$

where $P_X^*(w)$ denotes the unconditional version of $P_{X\mid Y}^*(w\mid y)$ defined as

$$P_X^*(w) := \sum_{z \in \mathbb{F}_q} P_X(z) \chi(wz).$$

For the upper bound, one obtains

$$(q-1)S(X) = \sum_{w \in \mathbb{F}_q^{\times}} |P_X^*(w)| \le \sqrt{q-1} \sqrt{\sum_{w \in \mathbb{F}_q^{\times}} |P_X^*(w)|^2}$$
$$= \sqrt{q(q-1)} \sqrt{\sum_{x \in \mathbb{F}_q} |P_X(x) - \frac{1}{q}|^2}.$$

Here, the inequality is obtained from the Cauchy-Schwarz inequality $\|p_0^{q-1}\|_1 \leq \sqrt{q}\|p_0^{q-1}\|_2$ which holds for $p_0^{q-1} \in \mathbb{C}^q$. The last equality holds via Perseval's identity, i.e., since the Fourier transform is unitary up to the constant factor \sqrt{q} . Let $t:=|1/(1-P_{\mathrm{e}}(X))|$.

$$\sqrt{\sum_{x \in \mathbb{F}_q} \left| P_X(x) - \frac{1}{q} \right|^2} \le \left(t \left| 1 - P_{e}(X) - \frac{1}{q} \right|^2 + \left| 1 - t(1 - P_{e}(X)) - \frac{1}{q} \right|^2 + (q - t - 1) \frac{1}{q^2} \right)^{\frac{1}{2}}$$

$$= \left((1 - P_{e}(X))t \left((1 - P_{e}(X))t - P_{e}(X) \right) - t(1 - P_{e}(X)) + \frac{q - 1}{q} \right)^{\frac{1}{2}} \tag{20}$$

Since (20) is piecewise convex with respect to $P_{\rm e}(X)$, its concave hull is

$$t(t+1) \left[(t/(t+1) - P_{e}(X)) \sqrt{\frac{q-1}{q} - \frac{t-1}{t}} + (P_{e}(X) - (t-1)/t) \sqrt{\frac{q-1}{q} - \frac{t}{t+1}} \right]$$

for $P_{\rm e}(X) \in [0,(q-1)/q)$. Since this is piecewise linear and convex, t can be replaced by any $k=1,\ldots,q-1$. Note that the following smooth upper bound is obtained by replacing the first $(1-P_{\rm e}(X))t$ in (20) by 1.

$$S(X) \le \sqrt{1 - \frac{q}{q - 1} P_{\mathbf{e}}(X)}.$$

The unconditional version of the lower bound in Lemma 26 is obtained via the triangle inequality, as

$$(q-1)S(X) + 1 = \sum_{w \in \mathbb{F}_q} |P_X^*(w)| = \sum_{w \in \mathbb{F}_q} \left| \sum_{z \in \mathbb{F}_q} P_X(z) \chi(wz) \right|$$

$$= \max_{a \in \mathbb{F}_q} \sum_{w \in \mathbb{F}_q} \left| \sum_{z \in \mathbb{F}_q} P_X(z) \chi(w(z-a)) \right|$$

$$\geq \max_{a \in \mathbb{F}_q} \left| \sum_{z \in \mathbb{F}_q} P_X(z) \sum_{w \in \mathbb{F}_q} \chi(w(z-a)) \right|$$

$$= q \max_{a \in \mathbb{F}} P_X(a) = q(1 - P_e(X)).$$

APPENDIX E PROOF OF LEMMA 34

As in the argument for the binary case in [15, Chapter 5], MacWilliams identity is useful for the proof. Let $H:=G^{-1}$ and $H_{\langle i \rangle}:=[h_0,\ldots,h_i]$ where h_i is the i-th column of H. Furthermore, we let the Fourier transform of the joint probability $P_{X,Y}$ be defined as $P_{X,Y}^*(w,y):=P_Y(y)P_{X\mid Y}^*(w\mid y)$. The generalized MacWilliams identity is obtained as follows.

$$\begin{split} &P_{X^{(i)},Y^{(i)}}(u_i,(u_0^{i-1},y_0^{\ell-1}))\\ &= \sum_{x_0^{\ell-1}\in\mathbb{F}_q^\ell} \mathbb{I}\{x_0^{\ell-1}H_{\langle i\rangle} = u_0^i\} \prod_{j=0}^{\ell-1} P_{X,Y}(x_j,y_j)\\ &= \sum_{x_0^{\ell-1}\in\mathbb{F}_q^\ell} \prod_{j=0}^i \left[\frac{1}{q} \sum_{w_j\in\mathbb{F}_q} \chi\left(w_j\left(\sum_{k=0}^{\ell-1} H_{kj}x_k - u_j\right)\right)\right]\\ &\cdot \prod_{j=0}^{\ell-1} \left[\frac{1}{q} \sum_{z_j\in\mathbb{F}_q} \chi(-z_jx_j) P_{X,Y}^*(z_j,y_j)\right]\\ &= \frac{1}{q^{\ell+i+1}}\\ &\cdot \sum_{z_0^{\ell-1}\in\mathbb{F}_q^\ell, w_0^i\in\mathbb{F}_q^{i+1}} \prod_{j=0}^{\ell-1} \left[\sum_{x_j\in\mathbb{F}_q} \chi\left(x_j\left(\sum_{k=0}^i H_{jk}w_k - z_j\right)\right)\right]\\ &\cdot \prod_{j=0}^{\ell-1} P_{X,Y}^*(z_j,y_j) \prod_{j=0}^i \chi(-w_ju_j)\\ &= \frac{1}{q^{i+1}} \sum_{z_0^{\ell-1}\in\mathbb{F}_q^\ell, w_0^i\in\mathbb{F}_q^{i+1}} \mathbb{I}\{w_0^i H_{\langle i\rangle}^t = z_0^{\ell-1}\}\\ &\cdot \prod_{j=0}^{\ell-1} P_{X,Y}^*(z_j,y_j) \prod_{j=0}^i \chi(-w_ju_j). \end{split}$$

Hence, the Fourier transform $P_{X^{(i)},Y^{(i)}}^*$ of the joint probability $P_{X^{(i)},Y^{(i)}}$ is given by

$$\begin{split} P_{X^{(i)},Y^{(i)}}^*(w_i,(u_0^{i-1},y_0^{\ell-1})) &= \frac{1}{q^i} \\ &\cdot \sum_{z_0^{\ell-1} \in \mathbb{F}_q^\ell, w_0^{i-1} \in \mathbb{F}_q^i} \mathbb{I} \big\{ w_0^{i-1} H_{\langle i-1 \rangle}^t + w_i h_i^t = z_0^{\ell-1} \big\} \\ &\cdot \prod_{i=0}^{\ell-1} P_{X,Y}^*(z_j,y_j) \prod_{i=0}^{i-1} \chi(-w_j u_j). \end{split}$$

Then, one can derive the first inequality in Lemma 34 as

$$\begin{split} S_{\max}(X^{(i)}, Y^{(i)}) &= \max_{w_i \in \mathbb{F}_q^{\times}} \sum_{y_0^{\ell-1} \in \mathcal{Y}^{\ell}, u_0^{i-1} \in \mathbb{F}_q^{i}} |P_{X^{(i)}, Y^{(i)}}^{*}(w_i, (u_0^{i-1}, y_0^{\ell-1}))| \\ &= \max_{w_i \in \mathbb{F}_q^{\times}} \sum_{y_0^{\ell-1} \in \mathcal{Y}^{\ell}, u_0^{i-1} \in \mathbb{F}_q^{i}} \left| \frac{1}{q^i} \right. \\ &\cdot \sum_{z_0^{\ell-1} \in \mathbb{F}_q^{\ell}, w_0^{i-1} \in \mathbb{F}_q^{i}} \mathbb{I} \left\{ w_0^{i-1} H_{\langle i-1 \rangle}^t + w_i h_i^t = z_0^{\ell-1} \right\} \\ &\cdot \prod_{i=0}^{\ell-1} P_{X,Y}^{*}(z_j, y_j) \prod_{i=0}^{i-1} \chi(-w_j u_j) \right| \end{split}$$

$$\leq \max_{w_{i} \in \mathbb{F}_{q}^{\times}} \sum_{z_{0}^{\ell-1} \in \mathbb{F}_{q}^{\ell}, w_{0}^{i-1} \in \mathbb{F}_{q}^{i}} \mathbb{I}\{w_{0}^{i-1} H_{\langle i-1 \rangle}^{t} + w_{i} h_{i}^{t} = z_{0}^{\ell-1}\}
\cdot \prod_{j=0}^{\ell-1} \sum_{y \in \mathcal{Y}} |P_{X,Y}^{*}(z_{j}, y)|
\leq q^{i} S_{\max}(X, Y)^{D_{s}^{(i)}(G)}.$$
(21)

The last inequality in the above is obtained by observing that $z_0^{\ell-1}$ satisfying $w_0^{i-1}H^t_{\langle i-1\rangle}+w_ih^t_i=z_0^{\ell-1}$ should contain at least $D_{\mathbf{s}}^{(i)}(G)$ nonzero elements, and that $\sum_{u\in\mathcal{V}}|P^*_{X,Y}(0,y)|=1$ holds.

As for the second inequality in Lemma 34, one has

$$\begin{split} S_{\min}(X^{(t)}, Y^{(t)}) &= \min_{w_{i} \in \mathbb{F}_{q}^{\times}} \sum_{y_{0}^{\ell-1} \in \mathcal{Y}^{\ell}, u_{0}^{i-1} \in \mathbb{F}_{q}^{i}} |P_{X^{(i)}, Y^{(i)}}^{*}(w_{i}, (u_{0}^{i-1}, y_{0}^{\ell-1}))| \\ &= \min_{w_{i} \in \mathbb{F}_{q}^{\times}} \sum_{y_{0}^{\ell-1} \in \mathcal{Y}^{\ell}, u_{0}^{i-1} \in \mathbb{F}_{q}^{i}} \left| \frac{1}{q^{i}} \right| \\ &\cdot \sum_{z_{0}^{\ell-1} \in \mathbb{F}_{q}^{\ell}, w_{0}^{i-1} \in \mathbb{F}_{q}^{i}} \mathbb{E}\left\{ w_{0}^{i-1} H_{\langle i-1 \rangle}^{t} + w_{i} h_{i}^{t} = z_{0}^{\ell-1} \right\} \\ &\cdot \prod_{j=0}^{\ell-1} P_{X, Y}^{*}(z_{j}, y_{j}) \prod_{j=0}^{i-1} \chi(-w_{j} u_{j}) \right| \\ &= \min_{w_{i} \in \mathbb{F}_{q}^{\times}} \max_{a_{0}^{i-1} \in \mathbb{F}_{q}^{i}} \sum_{y_{0}^{\ell-1} \in \mathcal{Y}^{\ell}, u_{0}^{i-1} \in \mathbb{F}_{q}^{i}} \left| \frac{1}{q^{i}} \right| \\ &\cdot \sum_{z_{0}^{\ell-1} \in \mathbb{F}_{q}^{\ell}, w_{0}^{i-1} \in \mathbb{F}_{q}^{i}} \mathbb{E}\left\{ w_{0}^{i-1} H_{\langle i-1 \rangle}^{t} + w_{i} h_{i}^{t} = z_{0}^{\ell-1} \right\} \\ &\cdot \prod_{j=0}^{\ell-1} P_{X, Y}^{*}(z_{j}, y_{j}) \prod_{j=0}^{i-1} \chi((a_{j} - w_{j}) u_{j}) \right| \\ &\geq \min_{w_{i} \in \mathbb{F}_{q}^{\times}} \max_{a_{0}^{i-1} \in \mathbb{F}_{q}^{i}} \sum_{y_{0}^{\ell-1} \in \mathcal{Y}^{\ell}} \mathbb{E}\left\{ w_{0}^{i-1} H_{\langle i-1 \rangle}^{t} + w_{i} h_{i}^{t} = z_{0}^{\ell-1} \right\} \\ &\cdot \prod_{j=0}^{\ell-1} P_{X, Y}^{*}(z_{j}, y_{j}) \prod_{j=0}^{i-1} \left(\frac{1}{q} \sum_{u \in \mathbb{F}_{q}} \chi((a_{j} - w_{j}) u) \right) \right| \\ &= \min_{w_{i} \in \mathbb{F}_{q}^{\times}} \max_{a_{0}^{i-1} \in \mathbb{F}_{q}^{i}} \prod_{j=0}^{\ell-1} \sum_{y \in \mathcal{Y}} \left| P_{X, Y}^{*}((a_{0}^{i-1} H_{\langle i-1 \rangle}^{t} + w_{i} h_{i}^{t})_{j} \neq 0) \right| \\ &\geq \min_{w_{i} \in \mathbb{F}_{q}^{\times}} \max_{a_{0}^{i-1} \in \mathbb{F}_{q}^{i}} \prod_{j=0}^{\ell-1} S_{\min}(X, Y)^{\mathbb{I}\left\{ (a_{0}^{i-1} H_{\langle i-1 \rangle}^{t} + w_{i} h_{i}^{t})_{j} \neq 0 \right\}} \\ &= S_{\min}(X, Y)^{D_{s}^{(i)}(G)} \end{split}$$

where the last equality in the above is obtained by noting that the maximization with respect to a_0^{i-1} amounts to making the number of nonzero elements in $a_0^{i-1}H_{\langle i-1\rangle}^t+w_ih_i^t$ to be as small as possible.

APPENDIX F PROOF OF LEMMA 35

For the first equation, let us consider a $\sigma(B_1, \dots, B_n)$ measurable random process $\{\xi_n := \xi(X_n, Y_n)\}_{n=0,1,\dots}$ where

$$\xi(X,Y) := \{(x,x') \in \mathbb{F}_q^2 \mid Z_{x,x'}(X \mid Y) = 0\}$$

and where

$$Z_{x,x'}(X \mid Y) := \sum_{y \in \mathcal{V}} \sqrt{P_{Y|X}(y \mid x)P_{Y|X}(y \mid x')}.$$

Then, $\{\xi_n\}_{n=0,1,\dots}$ is obviously a Markov chain. The Markov chain $\{\xi_n\}_{n=0,1,\dots}$ has the empty set ϕ as the absorbing state, i.e., $\Pr(\xi_n=\phi\mid\xi_{n-1}=\phi)=1$. Since any source accessible from the original source (X,Y) by G is also polarized by G, ϕ is the unique accessible absorbing state. Hence, $\lim_{n\to\infty}\Pr(\xi_n=\phi)=1$, proving the first equation of the lemma.

The second equation is obtained in the same way. Let us define

$$S_w(X \mid Y) := \sum_{y \in \mathcal{Y}} P_Y(y) \left| P_{X|Y}^*(w \mid y) \right|$$

and let $\{\eta_n := \eta(X_n, Y_n)\}_{n=0,1,...}$ be a $\sigma(B_1, \ldots, B_n)$ -measurable random process where

$$\eta(X,Y) := \{ w \in \mathbb{F}_q \mid S_w(X \mid Y) = 0 \}.$$

Then, $\{\eta_n\}_{n=0,1,...}$ is a Markov chain since one obtains from the derivations of (21) and (22) in Appendix E that

$$\max_{z_0^{\ell-1} \in \mathcal{C}_i(w)} \prod_{j=0}^{\ell-1} S_{z_j}(X \mid Y) \le S_w(X^{(i)} \mid Y^{(i)})$$

$$\le q^i \max_{z_0^{\ell-1} \in \mathcal{C}_i(w)} \prod_{j=0}^{\ell-1} S_{z_j}(X \mid Y)$$

for any $w\in\mathbb{F}_q^{\times}$ and $i=0,\dots,\ell-1$ where $\mathcal{C}_i(w)$ is the affine space $\{\sum_{j=0}^{i-1}a_jh_j^t+wh_i^t\mid a_0^{i-1}\in\mathbb{F}_q^i\}$ defined on the basis of the columns of $G^{-1}:=[h_0,h_1,\dots,h_{\ell-1}]$. The superscript t here denotes transpose of a vector. Then, it holds that $\lim_{n\to\infty}\Pr(\eta_n=\phi)=1$ due to the same reason as that for $\{\xi_n\}_{n=0,1,\dots}$, which proves the second equation of the lemma.

REFERENCES

- R. Mori and T. Tanaka, "Channel polarization on q-ary discrete memoryless channels by arbitrary kernels," in *Proc. 2010 IEEE Int. Symp. Inf. Theory, Austin, TX*, Jul. 13–18, 2010, pp. 894–898.
- [2] ——, "Non-binary polar codes using Reed-Solomon codes and algebraic geometry codes," in *Proc. 2010 IEEE Information Theory Workshop*, *Dublin Ireland*, Aug. 30–Sep. 3, 2010.
- [3] E. Arıkan, "Channel polarization: A method for constructing capacity-achieving codes for symmetric binary-input memoryless channels," *IEEE Trans. Inf. Theory*, vol. 55, no. 7, pp. 3051–3073, Jul. 2009.
- [4] E. Şaşoğlu, E. Telatar, and E. Arıkan, "Polarization for arbitrary discrete memoryless channels," 2009. [Online]. Available: http://arxiv.org/abs/0908.0302v1
- [5] E. Abbe and E. Telatar, "Polar codes for the m-user multiple access channel," *IEEE Trans. Inf. Theory*, vol. 58, no. 8, pp. 5437–5448, Aug. 2012.
- [6] W. Park and A. Barg, "Polar codes for q-ary channels, $q = 2^r$," *IEEE Trans. Inf. Theory*, vol. 59, no. 2, pp. 955–969, Feb. 2013.

- [7] A. G. Sahebi and S. S. Pradhan, "Multilevel polarization of polar codes over arbitrary discrete memoryless channels," in 2011 49th Annual Allerton Conference on Communication, Control, and Computing (Allerton), Sep. 28–30, 2011, pp. 1718–1725.
- [8] E. Şaşoğlu, "Polar codes for discrete alphabets," in Proc. 2012 IEEE Int. Symp. Inf. Theory, Cambridge, MA, Jul. 1–6, 2012, pp. 2137–2141.
- [9] S. Korada, E. Sasoglu, and R. Urbanke, "Polar codes: Characterization of exponent, bounds, and constructions," *IEEE Trans. Inf. Theory*, vol. 56, no. 12, pp. 6253–6264, 2010.
- [10] E. Arikan, "Source polarization," in *Proc. 2010 IEEE Int. Symp. Inf. Theory, Austin, TX*, Jul. 13–18, 2010, pp. 899–903.
- [11] H. Cronie and S. Korada, "Lossless source coding with polar codes," in Proc. 2010 IEEE Int. Symp. Inf. Theory, Austin, TX, Jul. 13–18, 2010, pp. 904–908.
- [12] N. Hussami, R. Urbanke, and S. Korada, "Performance of polar codes for channel and source coding," in *Proc. IEEE Int. Symp. Inf. Theory*, *Seoul, South Korea*, Jun. 28–Jul. 3 2009, pp. 1488–1492.
- [13] S. Korada and R. Urbanke, "Polar codes are optimal for lossy source coding," *IEEE Trans. Inf. Theory*, vol. 56, no. 4, pp. 1751–1768, 2010.
- [14] M. Karzand and E. Telatar, "Polar codes for q-ary source coding," in Proc. 2010 IEEE Int. Symp. Inf. Theory, Austin, TX, Jun. 13–18, 2010, pp. 909–912.
- [15] S. Korada, "Polar codes for channel and source coding," Ph.D. dissertation, Ecole Polytechnique Federale de Lausanne, 2009. [Online]. Available: http://library.epfl.ch/theses/?nr=4461
- [16] E. Arıkan and E. Telatar, "On the rate of channel polarization," in *Proc. IEEE Int. Symp. Inf. Theory, Seoul, South Korea*, Jun. 28–Jul. 3 2009, pp. 1493–1495.
- [17] R. Mori, "Properties and construction of polar codes," Master's thesis, Kyoto University, 2010. [Online]. Available: http://arxiv.org/abs/1002. 3521
- [18] S. Hassani, R. Mori, T. Tanaka, and R. Urbanke, "Rate-dependent analysis of the asymptotic behavior of channel polarization," *IEEE Trans. Inf. Theory*, vol. 59, no. 4, pp. 2267–2276, Apr. 2013. [Online]. Available: http://arxiv.org/abs/1110.0194
- [19] T. Tanaka and R. Mori, "Refined rate of channel polarization," in *Proc.* 2010 IEEE Int. Symp. Inf. Theory, Austin, TX, Jun. 13–18, 2010, pp. 880–893
- [20] S. Hassani and R. Urbanke, "On the scaling of polar codes: I. the behavior of polarized channels," in *Proc. 2010 IEEE Int. Symp. Inf. Theory, Austin, TX*, Jun. 13–18, 2010, pp. 874–888.
- [21] J. Massey, D. Costello, and J. Justesen, "Polynomial weights and code constructions," *IEEE Trans. Inf. Theory*, vol. 19, no. 1, pp. 101–110, Jan. 1973.
- [22] K. Saints and C. Heegard, "On hyperbolic cascaded Reed-Solomon codes," Applied Algebra, Algebraic Algorithms and Error-Correcting Codes, vol. 673, pp. 291–303, May 1993.
- [23] F. MacWilliams and N. Sloane, The Theory of Error-Correcting Codes. North-Holland Amsterdam, 1977.
- [24] R. Mori and T. Tanaka, "Performance and construction of polar codes on symmetric binary-input memoryless channels," in *Proc.* 2009 IEEE Int. Symp. Inf. Theory, Seoul, South Korea, Jun. 28–Jul. 3, 2009, pp. 1496–1500.
- [25] R. Gallager, Information Theory and Reliable Communication. John Wiley & Sons, Inc. New York, NY, USA, 1968.
- [26] M. Feder and N. Merhav, "Relations between entropy and error probability," *IEEE Trans. Inf. Theory*, vol. 40, no. 1, pp. 259–266, Jan. 1994.
- [27] M. Ben-Bassat and J. Raviv, "Renyi's entropy and the probability of error," *IEEE Trans. Inf. Theory*, vol. 24, no. 3, pp. 324–331, May 1978.

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17

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